

Wide Bandwidth Yaw Rate Gyroscope with SPI

ADIS16060

FEATURES

Complete angular rate digital gyroscope 14-bit resolution

Scalable measurement range

Initial range: ±80°/sec (typical)

Increase range with external resistor

Z-axis (yaw rate) response

SPI digital output interface

High vibration rejection over wide frequency

2000 g-powered shock survivability

1 kHz bandwidth

Selectable using external capacitor

Externally controlled self-test

Internal temperature sensor output

Dual auxiliary 14-bit ADC inputs

Absolute rate output for precision applications

5 V single-supply operation

8.2 mm \times 8.2 mm \times 5.2 mm package

-40°C to +105°C operation

RoHS compliant

APPLICATIONS

Platform stabilization

Image stabilization

Guidance and control

Inertia measurement units

Robotics

GENERAL DESCRIPTION

The ADIS16060 is a yaw rate gyroscope with an integrated serial peripheral interface (SPI). It features an externally selectable bandwidth response and scalable dynamic range.

The SPI port provides access to the rate sensor, an internal temperature sensor, and two external analog signals (using internal ADC). The digital data available at the SPI port is proportional to the angular rate about the axis that is normal to the top surface of the package.

An additional output pin provides a precision voltage reference. A digital self-test function electromechanically excites the sensor to test the operation of the sensor and the signal-conditioning circuits.

The ADIS16060 is available in an 8.2 mm \times 8.2 mm \times 5.2 mm, 16-terminal, peripheral land grid array (LGA) package.

FUNCTIONAL BLOCK DIAGRAM

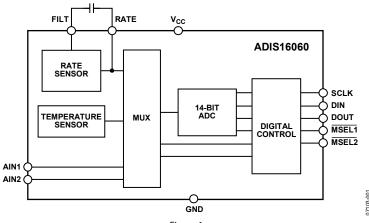


Figure 1.

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REVISION HISTORY

1/08—Revision 0: Initial Version

SPECIFICATIONS

 T_{A} = 25°C, V_{CC} = 5 V, angular rate = 0°/sec, C_{OUT} = 0.01 $\mu\text{F},$ ±1 g, unless otherwise noted.

Table 1.

| Parameter | Conditions | Min ¹ | Тур | Max | Unit |
|---|--|-----------------------|--------|-----------------|-----------|
| SENSITIVITY | | | · | | |
| Dynamic Range ² | Full-scale range over specifications range | ±50 | ±80 | | °/sec |
| Initial | Clockwise rotation is positive output, $T_A = -40$ °C to $+85$ °C | 0.0110 | 0.0122 | 0.0134 | °/sec/LSB |
| Change Over Temperature ³ | $V_{CC} = 4.75 \text{ V to } 5.25 \text{ V}$ | | ±3 | | % |
| Nonlinearity | Best fit straight line | | 0.1 | | °/sec |
| NULL | | | | | |
| Initial | Nominal 0°/sec output is 8192 LSB | -44 | | +44 | °/sec |
| Change Over Temperature ³ | $V_{CC} = 4.75 \text{ V to } 5.25 \text{ V}$ | | ±0.11 | | °/sec/°C |
| Turn-On Time | Power on to ±0.5°/sec of final value | | 10 | | ms |
| Linear Acceleration Effect | Any axis | | ±0.1 | | °/sec/g |
| Voltage Sensitivity | $V_{CC} = 4.75 \text{ V to } 5.25 \text{ V}$ | | ±0.5 | | °/sec/V |
| NOISE PERFORMANCE | 100 | | | | ,,,,,, |
| Rate Noise Density | @ 25°C | | 0.04 | | °/sec/√Hz |
| FREQUENCY RESPONSE | | | | | |
| 3 dB Bandwidth (User-Selectable) ⁴ | $C_{OUT} = 0 \mu F$ | 1 | | 1000 | Hz |
| Sensor Resonant Frequency | C501 6 p. | ' | 14.5 | 1000 | kHz |
| SELF-TEST RESPONSE | | | | | - 111.12 |
| Positive Self-Test ⁵ | See Table 5 | | +6226 | | LSB |
| Negative Self-Test ⁵ | See Table 5 | | -6226 | | LSB |
| TEMPERATURE SENSOR | Jee lable 3 | | 0220 | | 235 |
| Reading at 298 K | | 7700 | 8192 | 8684 | LSB |
| Scale Factor | Proportional to absolute temperature | 7700 | 0.034 | 0001 | K/LSB |
| LOGIC INPUTS | Troportional to absolute temperature | | 0.03 1 | | 1,7233 |
| Input High Voltage, V _{INH} | | 0.7 × Vcc | | | V |
| Input Low Voltage, V _{INL} | | 0.7 × Vcc | | 0.8 | v |
| Input Current, I _{IN} | Typically 10 nA | -1 | | +1 | μA |
| Input Capacitance, C _{IN} (DIN) | Typically 10 11/1 | | 8 | | pF |
| Input Capacitance, C_{IN} (MSEL1, MSEL2) | | | 5 | | pF |
| ANALOG INPUTS | For V _{IN} < V _{CC} | | | | ρ. |
| Resolution | 1 OI VIN VCC | | 14 | | Bits |
| Integral Nonlinearity | Best fit straight line | -6 | 17 | +6 | LSB |
| Differential Nonlinearity | No missing codes to 13 bits | -1 | | +6 | LSB |
| Offset Error | 140 missing codes to 15 bits | -10 | | +10 | mV |
| Offset Error Temperature Drift | | -10 | ±0.3 | +10 | ppm/°C |
| Gain Error | | -40 | ±0.5 | +40 | mV |
| Gain Error Temperature Drift | | -40 | ±0.3 | +4 0 | ppm/°C |
| Input Voltage Range | | 0 | ±0.5 | Vcc | V Ppin/ C |
| | | 0 | 1 | VCC | |
| Leakage Current DIGITAL OUTPUTS | | | 1 | | nA |
| Output High Voltage, V _{OH} | Isoures = 500 HA | Vec - 0.2 | | | V |
| Output High Voltage, V _{OL} | $I_{SOURCE} = 500 \mu A$ $I_{SINK} = 500 \mu A$ | V _{CC} – 0.3 | | 0.4 | V |
| | 15ΙΝΚ — 300 μΑ | | | 0.4 | V |
| CONVERSION RATE | | | | 10 | |
| Conversion Time | | | | 10 | μs |
| Throughput Rate | | | | 100 | kSPS |

| Parameter | Conditions | Min ¹ | Тур | Max | Unit |
|------------------------------|---|------------------|-----|------|------|
| POWER SUPPLY | All at $T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$ | | | | |
| V_{CC} | | 4.75 | 5 | 5.25 | V |
| Vcc Quiescent Supply Current | V _{CC} @ 5 V, 50 kSPS sample rate | | 4.3 | 6.5 | mA |
| Power Dissipation | V _{CC} @ 5 V, 50 kSPS sample rate | | 22 | 33 | mW |
| TEMPERATURE RANGE | Operation | -40 | | +105 | °C |

¹ All minimum and maximum specifications are guaranteed. Typical specifications are neither tested nor guaranteed.
² Dynamic range is the maximum full-scale measurement range possible, including output swing range, initial offset, sensitivity, offset drift, and sensitivity drift at 5 V supply.
³ Defined as the output change from ambient to maximum temperature, or ambient to minimum temperature.
⁴ Frequency at which the response is 3 dB down from dc response. Bandwidth = 1/(2 × π × 200 kΩ × C_{OUT}). For C_{OUT} = 0.01 μF, bandwidth = 80 Hz.
⁵ Self-test response varies with temperature.

TIMING SPECIFICATIONS

 $T_A = 25$ °C, angular rate = 0°/sec, unless otherwise noted.

Table 2. Read/Output Sequence

| Parameter | Figure Reference | Symbol | Min | Тур | Max | Unit |
|--|------------------|-------------------|-----|-----|-----|------|
| Serial Clock Frequency | | | | | 2.9 | MHz |
| Throughput Rate | See Figure 2 | t _{CYC} | | | 100 | kHz |
| MSEL1 Falling to SCLK Low | See Figure 2 | t _{CSD} | | | 0 | μs |
| MSEL1 Falling to SCLK Rising | See Figure 2 | t _{sucs} | 20 | | | ns |
| SCLK Falling to Data Remains Valid | See Figure 2 | t _{HDO} | 5 | 16 | | ns |
| MSEL1 Rising Edge to Dout High Impedance | See Figure 2 | t _{DIS} | | 14 | 100 | ns |
| SCLK Falling to Data Valid | See Figure 2 | t _{EN} | | 16 | 50 | ns |
| Acquisition Time | See Figure 2 | t _{ACQ} | 400 | | | ns |
| D _{OUT} Fall Time | See Figure 2 | t _F | | 11 | 25 | ns |
| D _{OUT} Rise Time | See Figure 2 | t _R | | 11 | 25 | ns |
| Data Setup Time | See Figure 3 | t 5 | 5 | 5 | | ns |
| SCLK Falling Edge to MSEL2 Rising Edge | See Figure 3 | t ₇ | 0 | 0 | | ns |
| Data Hold Time | See Figure 3 | t ₆ | 4.5 | | | ns |

 $^{^{1}}$ Guaranteed by design. All input signals are specified with $t_R = t_F = 5$ ns (10% to 90% of V_{CC}) and timed from a voltage level of 1.6 V. The 5 V operating range spans from 4.75 V to 5.25 V.

Timing Diagrams

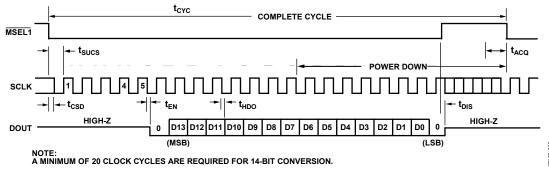


Figure 2. Serial Interface Timing Diagram-Read/Output Sequence (CPOL = 0, CPHA = 0)

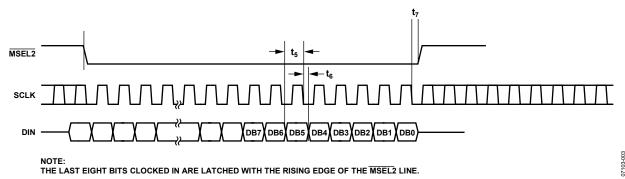


Figure 3. Serial interface Timing–Input/Configuration Sequence (CPOL = 0, CPHA = 1)

ABSOLUTE MAXIMUM RATINGS

Table 3.

| 1001001 | |
|--|--|
| Parameter | Rating |
| Acceleration (Any Axis, Unpowered, 0.5 ms) | 2000 g |
| Acceleration (Any Axis, Powered, 0.5 ms) | 2000 <i>g</i> |
| V _{CC} to GND | −0.3 V to +6.0 V |
| V _{CC} to GND | $-0.3 \text{ V to V}_{CC} + 0.3 \text{ V}$ |
| Analog Input Voltage to GND | $-0.3 \mathrm{V}$ to $\mathrm{V}_{\mathrm{CC}} + 0.3 \mathrm{V}$ |
| Digital Input Voltage to GND | -0.3 V to +7.0 V |
| Digital Output Voltage to GND | $-0.3 \mathrm{V}$ to $\mathrm{V}_{\mathrm{CC}} + 0.3 \mathrm{V}$ |
| Operating Temperature Range | −40°C to +105°C |
| Storage Temperature Range | −65°C to +150°C |

Stresses above those listed under the Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

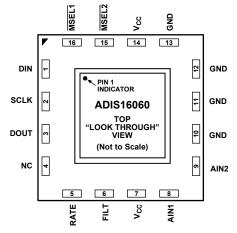
Drops onto hard surfaces can cause shocks of greater than 2000 *g* and exceed the absolute maximum rating of the device. Care should be exercised in handling the device to avoid damage.

ESD CAUTION



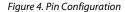
ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



NOTES

1. NC = NO CONNECT
2. THIS IS NOT AN ACTUAL "TOP VIEW," AS THE PINS ARE NOT VISIBLE FROM THE TOP. THIS IS A LAYOUT VIEW, WHICH REPRESENTS THE PIN CONFIGURATION, IF THE PACKAGE IS LOOKED THROUGH FROM THE TOP. THIS CONFIGURATION IS



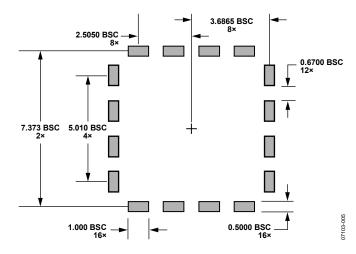


Figure 5. Second-Level Assembly Pad Layout

Table 4. Pin Function Descriptions

| Pin No. | Mnemonic | Type ¹ | Description |
|---------|-----------------|-------------------|---|
| 1 | DIN | 1 | SPI Data Input. |
| 2 | SCLK | 1 | SPI Serial Clock. |
| 3 | DOUT | Θ | SPI Data Output. |
| 4 | NC | | No Connect. |
| 5 | RATE | 0 | Buffered Analog Output. Represents the angular rate signal. |
| 6 | FILT | 1 | External Capacitor Connection to Control Bandwidth. |
| 7 | V _{CC} | S | Power Supply. |
| 8 | AIN1 | 1 | External Analog Input Channel 1. |
| 9 | AIN2 | 1 | External Analog Input Channel 2. |
| 10 | GND | S | Ground. |
| 11 | GND | S | Ground. |
| 12 | GND | S | Ground. |
| 13 | GND | S | Ground. |
| 14 | Vcc | S | Power Supply. |
| 15 | MSEL2 | 1 | SPI, Mode Select 2. Used for data input functions. |
| 16 | MSEL1 | 1 | SPI, Mode Select 1. Used for data output functions. |

 $^{^{1}}$ I = input; O = output; S = power supply.

TYPICAL PERFORMANCE CHARACTERISTICS

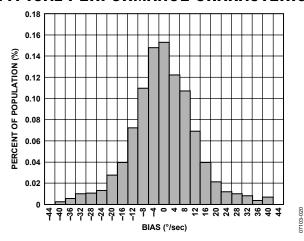


Figure 6. Initial Bias Error Distribution, 25°C, $V_{CC} = 5 V$

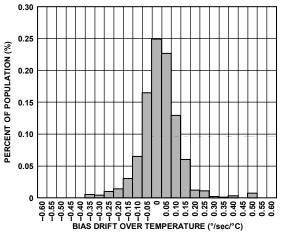


Figure 7. Bias Drift Over -40° C to $+85^{\circ}$ C, $V_{CC} = 5 \text{ V}$

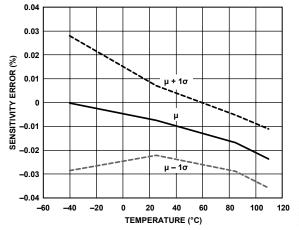


Figure 8. Sensitivity Drift vs. Temperature, $V_{CC} = 5 V$

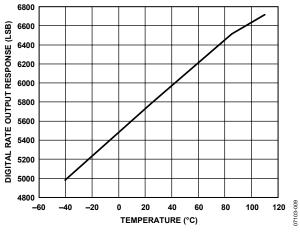


Figure 9. Positive Self-Test Response vs. Temperature, $V_{CC} = 5 V$

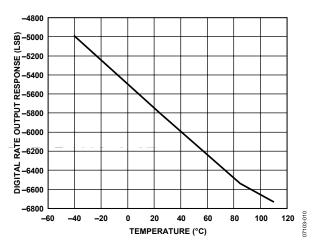


Figure 10. Negative Self-Test Output Response vs. Temperature, $V_{CC} = 5 V$

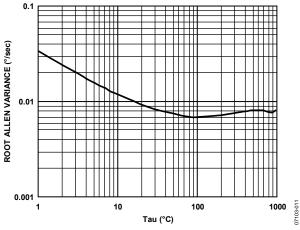


Figure 11. Allen Variance, 25°C, $V_{CC} = 5 V$

THEORY OF OPERATION

The ADIS16060 operates on the principle of a resonator gyroscope. Two polysilicon sensing structures each contain a dither frame that is electrostatically driven to resonance. This generates the necessary velocity element to produce a Coriolis force while rotating. At two of the outer extremes of each frame, orthogonal to the dither motion, are movable fingers that are placed between fixed pickoff fingers to form a capacitive pickoff structure that senses Coriolis motion.

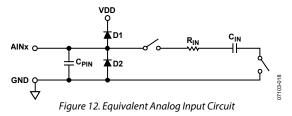
The resulting signal is fed to a series of gain and demodulation stages that produce the electrical rate signal output. The rate signal is then converted to a digital representation of the output on the SPI pins. The dual-sensor design provides linear acceleration (vibration, shock) rejection. Fabricating the sensor with the signal-conditioning electronics preserves signal integrity in noisy environments.

The electrostatic resonator requires 14 V to 16 V for operation. Because only 5 V is typically available in most applications, a charge pump is included on chip. After the demodulation stage, a single-pole, low-pass filter on the chip is used to limit high frequency artifacts before final amplification. The frequency response is dominated by the second low-pass filter, which is set by adding capacitance across RATE and FILT.

ANALOG-TO-DIGITAL CONVERTER INPUT

Figure 12 shows an equivalent circuit of the input structure of the ADIS16060 auxiliary ADC.

The two diodes, D1 and D2, provide ESD protection for the analog inputs, AINx (AIN1 and AIN2). Care must be taken to ensure that the analog input signal does not exceed the supply rails by more than 0.3 V, because exceeding this level causes these diodes to become forward-biased and to start conducting current. However, these diodes can handle a forward-biased current of 130 mA maximum. For instance, these conditions may eventually occur when the input signals exceed either $V_{\rm CC}$ or GND.



During the acquisition phase, the impedance model for AINx is a parallel combination of the capacitor C_{PIN} and the network formed by the series connection of R_{IN} and C_{IN} . C_{PIN} is primarily the pin capacitance. R_{IN} is typically 600 Ω and is a lumped component made up of some serial resistors and the on resistance of the switches. C_{IN} is typically 30 pF and mainly functions as the ADC sampling capacitor.

During the conversion phase, when the switches are open, the input impedance is limited to C_{PIN} . R_{IN} and C_{IN} make a 1-pole, low-pass filter that reduces undesirable aliasing effects and limits the noise.

When the source impedance of the driving circuit is low, the ADC input can be driven directly. Large source impedances significantly affect the ac performance, especially THD. The dc performances are less sensitive to the input impedance.

RATE SENSITIVE AXIS

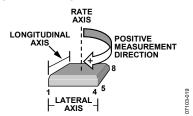


Figure 13. Rate Signal Increases with Clockwise Rotation

BASIC OPERATION

The ADIS16060 is designed for simple integration into industrial system designs, requiring only a 5 V power supply, two mode select lines, and three serial communications lines. The SPI handles all digital I/O communication in the ADIS16060.

SERIAL PERIPHERAL INTERFACE (SPI)

The ADIS16060 SPI port includes five signals: Mode Select 1 ($\overline{MSEL1}$), Mode Select 2 ($\overline{MSEL2}$), serial clock (\overline{SCLK}), data input (DIN), and data output (DOUT). The $\overline{MSEL1}$ line is used when reading data out of the sensor (DOUT), and the $\overline{MSEL2}$ line is used when configuring the sensor (DIN).

Selecting Output Data

Refer to Table 5 to determine the appropriate DIN bit sequence based on the required data source. Table 2 and Figure 3 provide the necessary timing details for the input configuration sequence. After the MSEL2 goes high, the last eight DIN bits are loaded into the internal control register, which represents DB0 to DB7 in Table 5.

Output Data Access

Use Table 2 and Figure 2 to determine the appropriate timing considerations for reading output data.

OUTPUT DATA FORMATTING

All of the output data is in an offset-binary format, which in this case, means that the ideal output for a zero rate condition is 8192 codes. If the sensitivity is equal to $+0.0122^{\circ}/\text{sec/LSB}$, a rate of $+10^{\circ}/\text{sec}$ results in a change of 820 codes, and a digital rate output of 9012 codes. If an offset error of $-20^{\circ}/\text{sec}$ is introduced, the output is reduced by 1639 codes (if typical sensitivity is assumed), resulting in a digital rate output of 6552 codes.

ADC CONVERSION

The internal successive approximation ADC begins the conversion process on the falling edge of $\overline{MSEL1}$ and starts to place data MSB first on the DOUT line at the 6th falling edge of SCLK, as shown in Figure 2. The entire conversion process takes 20 SCLK cycles. After $\overline{MSEL1}$ goes high, the acquisition process starts in preparation for the next conversion cycle.

Table 5. DIN Configuration Bit Assignments

| Action | DB7 | DB6 | DB5 | DB4 | DB3 | DB2 | DB1 | DB0 |
|--|-----|-----|-----|-----|-----|-----|-----|-----|
| Measure Angular Rate (Gyro) | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 |
| Measure Temperature | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 |
| Measure AIN2 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Measure AIN1 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |
| Set Positive Self-Test and Output for Angular Rate | | 0 | 1 | 0 | 0 | 0 | 1 | 0 |
| Set Negative Self-Test and Output for Angular Rate | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 1 |

APPLICATIONS INFORMATION SUPPLY AND COMMON CONSIDERATIONS

Power supply noise and transient behaviors can influence the accuracy and stability of any sensor-based measurement system. The ADIS16060 provides 0.2 μF of decoupling capacitance on the $V_{\rm CC}$ pin. Depending on the level of noise present in the power supply of the system, the ADIS16060 may not require any additional decoupling capacitance for this supply.

SETTING BANDWIDTH

External Capacitor C_{OUT} is used in combination with the onchip R_{OUT} resistor to create a low-pass filter to limit the bandwidth of the ADIS16060 rate response. The -3 dB frequency set by R_{OUT} and C_{OUT} is

$$f_{OUT} = \frac{1}{\left(2 \times \pi \times R_{OUT} \times C_{OUT}\right)}$$

and can be well controlled because R_{OUT} has been trimmed during manufacturing to be 200 k Ω \pm 5%. Setting the range with an external resistor impacts R_{OUT} as follows:

$$R_{OUT} = \frac{\left(200 \text{ k}\Omega \times R_{EXT}\right)}{\left(200 \text{ k}\Omega + R_{EXT}\right)}$$

In general, an additional hardware or software filter is added to attenuate high frequency noise arising from demodulation spikes at the gyro's 14 kHz resonant frequency. The noise spikes at 14 kHz can be clearly seen in the power spectral density curve shown in Figure 14.

INCREASING MEASUREMENT RANGE

Scaling the measurement range requires the addition of a single resistor, connected across the RATE and FILT pins. The following equation provides the proper relationship for selecting the appropriate resistor:

$$R_{EXT} = \frac{200 \text{ k}\Omega}{\Delta - 1}$$

where Δ is the increase in range.

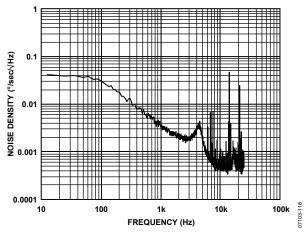


Figure 14. Noise Spectral Density with 2-Pole, Low-Pass Filter (40 Hz and 250 Hz)

DYNAMIC DIGITAL SENSITIVITY SCALING

This device supports in-system, dynamic, digital sensitivity scaling.

TEMPERATURE MEASUREMENTS

When using the temperature sensor, an acquisition time of greater than $40~\mu s$ helps to ensure proper setting and measurement accuracy. See Table 2 and Figure 2 for details on the definition of acquisition time.

SELF-TEST FUNCTION

- Exercising the self-test function is simple, as shown in this example.
 - 1. Configure using DIN = 00100010 (positive self-test, rate selected).
 - 2. Read output.
 - Configure using DIN = 00100000 (positive self-test off, rate selected)
 - Read output.
 - 5. Calculate the difference between Step 2 and Step 4, and compare this with the specified self-test output changes in the Specifications section.

Exercising the negative self-test requires changing the sequence in Step 1 to DIN = 00100001.

OUTLINE DIMENSIONS

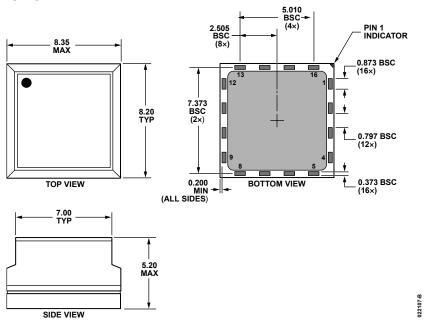


Figure 15. 16-Terminal Stacked Land Grid Array [LGA] (CC-16-1) Dimensions shown in millimeters

ORDERING GUIDE

| Model | Temperature Range | Package Description | Package Option |
|-----------------------------|-------------------|---|----------------|
| ADIS16060BCCZ ¹ | -40°C to +105°C | 16-Terminal Stacked Land Grid Array (LGA) | CC-16-1 |
| ADIS16060/PCBZ ¹ | | Evaluation Board | |

 $^{^{1}}$ Z = RoHS Compliant Part.