



# Audio Processor for Advanced TV

## ADAV4601

### FEATURES

- Fully programmable 28-bit audio processor for enhanced ATV sound—default audio processing flow loaded on reset
- Implements Analog Devices, Inc., and third-party branded audio algorithms
- Adjustable digital delay line for audio/video Synchronization for up to 200 ms stereo delay
- High performance 24-bit ADC and DAC
  - 94 dB DNR performance on DAC channels
  - 95 dB DNR performance on ADC channels
- Headphone output with integrated amplifiers
- High performance pulse-width modulation (PWM) digital outputs
- Multichannel digital baseband I/O
  - 4 stereo synchronous digital I<sup>2</sup>S input channels
  - One 6-channel sample rate converter (SRC) and one stereo SRC supporting input sample rates from 5 kHz to 50 kHz
  - One stereo synchronous digital I<sup>2</sup>S output
  - S/PDIF output with S/PDIF input mux capability
- Fast I<sup>2</sup>C control
- Operates from 3.3 V (analog), 1.8 V (digital core), and 3.3 V (digital interface)
- Available in 80-lead LQFP

### APPLICATIONS

- General-purpose consumer audio postprocessing
  - Home audio
  - DVD recorders
  - Home theater in a box (HTIB) systems and DVD receivers
- Audio processing subsystems for DTV-ready TVs
- Analog broadcast capability for iDTVs

### GENERAL DESCRIPTION

The ADAV4601 is an enhanced audio processor targeting advanced TV applications with full support for digital and analog baseband audio.

The audio processor, by default, loads a dedicated TV audio flow that incorporates full matrix switching (any input to any output), automatic volume control that compensates for volume changes during advertisements or when switching channels, dynamic bass, a multiband equalizer, and up to 200 ms of stereo delay memory for audio-video synchronization.

Alternatively, Analog Devices offers an award-winning graphical programming tool (SigmaStudio™) that allows custom flows to be quickly developed and evaluated. This allows the creation of customer-specific audio flows, including use of the Analog Devices library of third-party algorithms.

The analog I/O integrates Analog Devices proprietary continuous-time, multibit  $\Sigma$ - $\Delta$  architecture to bring a higher level of performance to ATV systems, required by third-party algorithm providers to meet system branding certification. The analog input is provided by 95 dB dynamic range (DNR) ADCs, and analog output is provided by 94 dB DNR DACs.

The main speaker outputs can be supplied as a digitally modulated PWM stream to support digital amplifiers.

The ADAV4601 includes multichannel digital inputs and outputs. In addition, digital input channels can be routed through integrated sample rate converters (SRC), which are capable of supporting any arbitrary sample rate from 5 kHz to 50 kHz.

Comprehensive documentation, which provides detailed operation guidelines and register map information, is available upon request from [AV.Products@analog.com](mailto:AV.Products@analog.com).

#### Rev. 0

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## REVISION HISTORY

3/08—Revision 0: Initial Version

# FUNCTIONAL BLOCK DIAGRAM

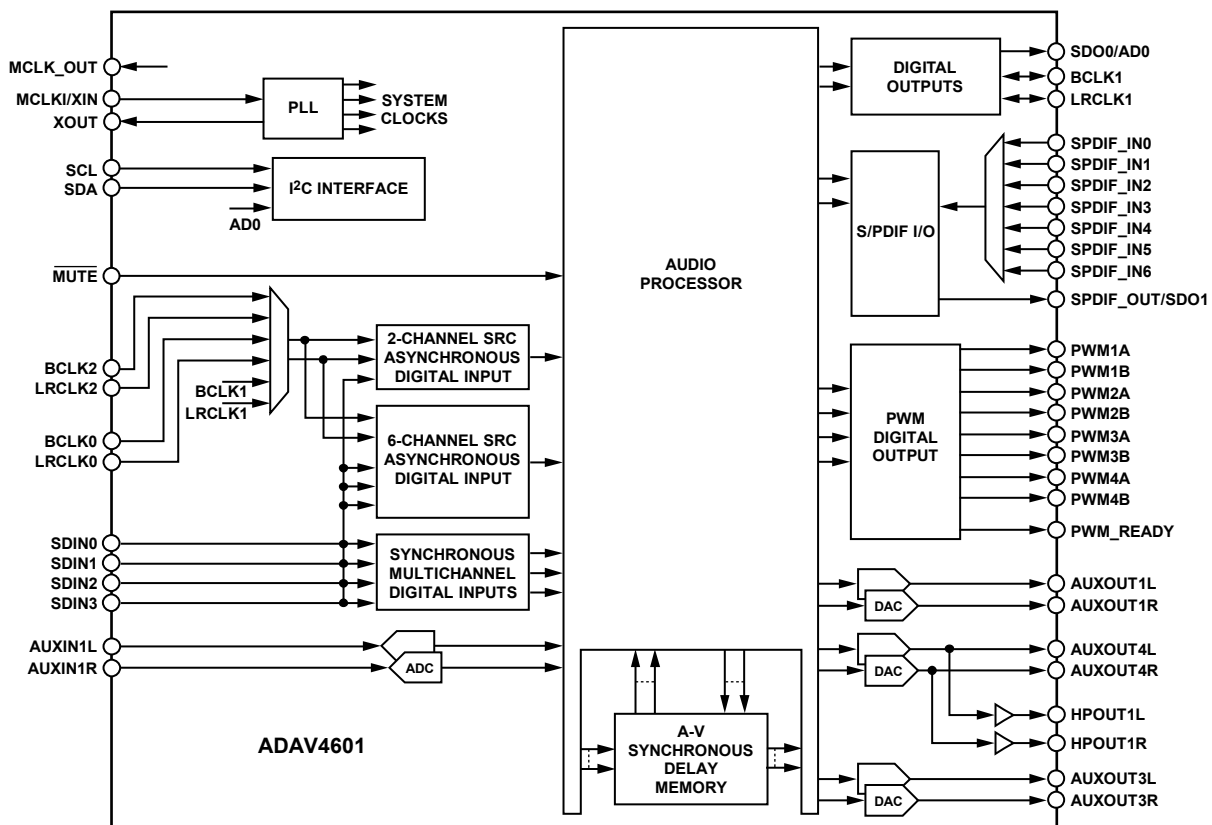


Figure 1. ADAV4601 with PWM-Based Speaker Outputs

07070-501

## SPECIFICATIONS

AVDD = 3.3 V, DVDD = 1.8 V, ODVDD = 3.3 V, operating temperature =  $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$ , master clock 24.576 MHz, measurement bandwidth = 20 Hz to 20 kHz, ADC input signal = DAC output signal = 1 kHz, unless otherwise noted.

### PERFORMANCE PARAMETERS

Table 1.

Parameter	Min	Typ	Max	Unit	Test Conditions/Comments
REFERENCE SECTION					
Absolute Voltage $V_{\text{REF}}$		1.53		V	
$V_{\text{REF}}$ Temperature Coefficient		100		ppm/ $^{\circ}\text{C}$	
ADC SECTION					
Number of Channels		2			One stereo channel
Full-Scale Input Level		100		$\mu\text{A rms}$	
Resolution		24		Bits	
Dynamic Range (Stereo Channel)					
A-Weighted		95		dB	$-60$ dBFS with respect to full-scale analog input
Total Harmonic Distortion + Noise (Stereo Channel)		$-90$		dB	$-3$ dBFS with respect to full-scale analog input
Gain Mismatch		0.2		dB	Left- and right-channel gain mismatch
Crosstalk (Left-to-Right, Right-to-Left)		$-110$		dB	
Gain Error		$-1$		dB	Input signal is 100 $\mu\text{A rms}$
Current Setting Resistor ( $R_{\text{ISET}}$ )		20		k $\Omega$	External resistor to set current input range of ADC for nominal 2.0 V rms input signal
Power Supply Rejection		$-87$		dB	1 kHz, 300 mV p-p signal at AVDD
ADC DIGITAL DECIMATOR FILTER CHARACTERISTICS					
Pass Band		22.5		kHz	At 48 kHz, guaranteed by design
Pass-Band Ripple		$\pm 0.0002$		dB	
Stop Band		26.5		kHz	
Stop-Band Attenuation		100		dB	
Group Delay		1040		$\mu\text{s}$	
PWM SECTION					
Frequency		384		kHz	Guaranteed by design
Modulation Index		0.976			Guaranteed by design
Dynamic Range					
A-Weighted		98		dB	$-60$ dBFS with respect to full-scale code input
Total Harmonic Distortion + Noise		$-80$		dB	$-3$ dBFS with respect to full-scale code input
DAC SECTION					
Number of Auxiliary Output Channels		6			Three stereo channels
Resolution		24		Bits	
Full-Scale Analog Output		1		V rms	
Dynamic Range					
A-Weighted		94		dB	$-60$ dBFS with respect to full-scale code input
Total Harmonic Distortion + Noise		$-86$		dB	$-3$ dBFS with respect to full-scale code input
Crosstalk (Left-to-Right, Right-to-Left)		$-102$		dB	
Interchannel Gain Mismatch		0.1		dB	Left- and right-channel gain mismatch
Gain Error		0.525		dB	1 V rms output
DC Bias		1.53		V	
Power Supply Rejection		$-90$		dB	1 kHz, 300 mV p-p signal at AVDD
Output Impedance		235		$\Omega$	
DAC DIGITAL INTERPOLATION FILTER CHARACTERISTICS					
Pass Band		21.769		kHz	At 48 kHz, guaranteed by design
Pass-Band Ripple		$\pm 0.01$		dB	
Transition Band		23.95		kHz	
Stop Band		26.122		kHz	
Stop-Band Attenuation		75		dB	
Group Delay		580		$\mu\text{s}$	

Parameter	Min	Typ	Max	Unit	Test Conditions/Comments
<b>HEADPHONE AMPLIFIER</b>					
Number of Channels		2			Measured at headphone output with 32 $\Omega$ load
Full-Scale Output Power		31		mW rms	One stereo channel 1 V rms output
Dynamic Range					
A-Weighted		93		dB	–60 dBFS with respect to full-scale code input
Total Harmonic Distortion + Noise		–83		dB	–3 dBFS with respect to full-scale code input
Interchannel Gain Mismatch		0.1		dB	
DC Bias		1.53		V	
Power Supply Rejection		–85		dB	1 kHz, 300 mV p-p signal at AVDD
<b>SRC</b>					
Number of Channels		8			Two channels (SRC1), six channels (SRC2)
Dynamic Range					
A-Weighted		115		dB	–60 dBFS input (worst-case input $f_s = 50$ kHz)
Total Harmonic Distortion + Noise		–113		dB	–3 dBFS input (worst-case input $f_s = 50$ kHz)
Sample Rate	5		50	kHz	
<b>SRC DIGITAL INTERPOLATION FILTER CHARACTERISTICS</b>					
Pass Band		21.678		kHz	At 48 kHz, guaranteed by design
Pass-Band Ripple		0.005		dB	
Stop Band		26.232		kHz	
Stop-Band Attenuation		110		dB	
Group Delay		876		$\mu$ s	
<b>DIGITAL INPUT/OUTPUT</b>					
Input Voltage High ( $V_{IH}$ )	2.0		ODVDD	V	
Input Voltage Low ( $V_{IL}$ )			0.8	V	
Input Leakage					
$I_{IH}$ (SDIN0, SDIN1, SDIN2, SDIN3, LRCLK0, LRCLK1, LRCLK2, BCLK0, BCLK1, BCLK2, SPDIF_OUT, SPDIF_IN)		40		$\mu$ A	$V_{IH} = ODVDD$ , equivalent to a 90 k $\Omega$ pull-up resistor
$I_{IH}$ (RESET)		13.5		$\mu$ A	$V_{IH} = ODVDD$ , equivalent to a 266 k $\Omega$ pull-up resistor
$I_{IL}$ (SDO0, SCL, SDA)		–40		$\mu$ A	$V_{IL} = 0$ V, equivalent to a 90 k $\Omega$ pull-down resistor
Output Voltage High ( $V_{OH}$ )	2.4			V	$I_{OH} = 0.4$ mA
Output Voltage Low ( $V_{OL}$ )			0.4	V	$I_{OL} = -2$ mA
Output Voltage High ( $V_{OH}$ ) (MCLK_OUT)	1.4			V	$I_{OH} = 0.4$ mA
Output Voltage Low ( $V_{OL}$ ) (MCLK_OUT)			0.4	V	$I_{OL} = -3.2$ mA
Input Capacitance		10		pF	
<b>SUPPLIES</b>					
Analog Supplies (AVDD)	3.0	3.3	3.6	V	
Digital Supplies (DVDD)	1.65	1.8	2.0	V	
Interface Supply (ODVDD)	3.0	3.3	3.6	V	
Supply Currents					
Analog Current		115		mA	
Digital Current		160		mA	
Interface Current		2		mA	
Power Dissipation		0.674		W	
<b>Standby Currents</b>					
Analog Current		7		mA	ADC, DAC, and headphone outputs floating, RESET low, MCLK = 24 MHz
Digital Current		3		mA	
Interface Current		1.6		mA	
<b>TEMPERATURE RANGE</b>					
Operating Temperature	–40		+85	$^{\circ}$ C	
Storage Temperature	–65		+150	$^{\circ}$ C	

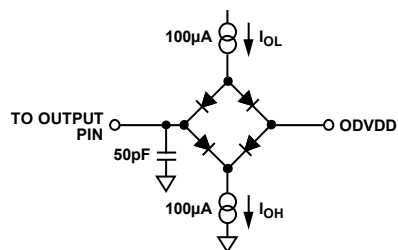
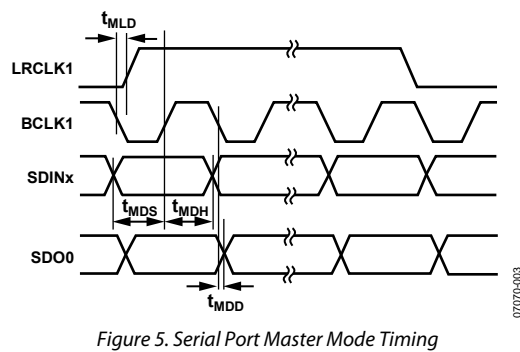
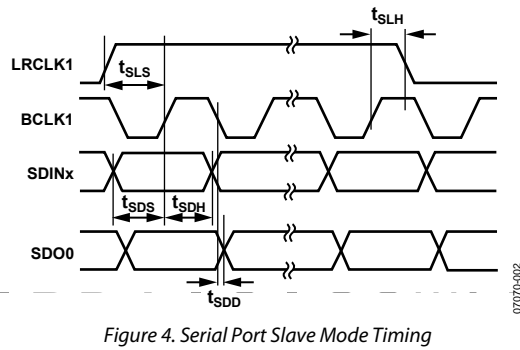
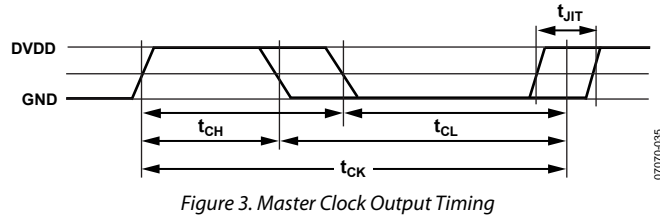
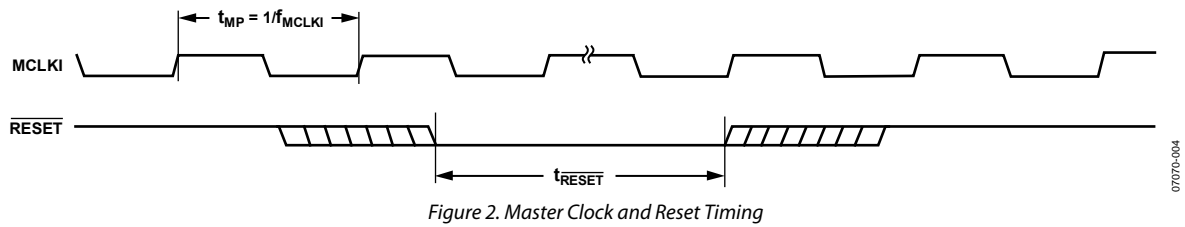
# ADAV4601

## TIMING SPECIFICATIONS

Table 2.

Parameter	Description	Min	Max	Unit	Comments
MASTER CLOCK AND RESET					
f <sub>MCLKI</sub>	MCLKI frequency	3.072	24.576	MHz	
t <sub>MCH</sub>	MCLKI high	10		ns	
t <sub>MCL</sub>	MCLKI low	10		ns	
t <sub>RESET</sub>	RESET low	200		ns	
MASTER CLOCK OUTPUT					
t <sub>JIT</sub>	Period jitter		800	ps	
t <sub>CH</sub>	MCLK_OUT high	45	55	%	
t <sub>CL</sub>	MCLK_OUT low	45	55	%	
I <sup>2</sup> C PORT					
f <sub>SCL</sub>	SCL clock frequency		400	kHz	Relevant for repeated start condition After this period, the first clock is generated
t <sub>SCLH</sub>	SCL high	600		ns	
t <sub>SCLL</sub>	SCL low	1.3		μs	
Start Condition					
t <sub>SCS</sub>	Setup time	600		ns	
t <sub>SCH</sub>	Hold time	600		ns	
t <sub>DS</sub>	Data setup time	100		ns	
t <sub>SCR</sub>	SCL rise time		300	ns	
t <sub>SCF</sub>	SCL fall time		300	ns	
t <sub>SDR</sub>	SDA rise time		300	ns	
t <sub>SDF</sub>	SDA fall time		300	ns	
Stop Condition					
t <sub>SCS</sub>	Setup time	0		ns	
SERIAL PORTS					
Slave Mode					
t <sub>SBH</sub>	BCLK high	40		ns	To BCLK rising edge From BCLK rising edge To BCLK rising edge From BCLK rising edge From BCLK falling edge
t <sub>SBL</sub>	BCLK low	40		ns	
f <sub>SBF</sub>	BCLK frequency	64 × f <sub>s</sub>			
t <sub>SLS</sub>	LRCLK setup	10		ns	
t <sub>SLH</sub>	LRCLK hold	10		ns	
t <sub>SDS</sub>	SDIN setup	10		ns	
t <sub>SDH</sub>	SDIN hold	10		ns	
t <sub>SDD</sub>	SDO delay		50	ns	
Master Mode					
t <sub>MLD</sub>	LRCLK delay		25	ns	From BCLK falling edge
t <sub>MDD</sub>	SDO delay		15	ns	From BCLK falling edge
t <sub>MDS</sub>	SDIN setup	10		ns	From BCLK rising edge
t <sub>MDH</sub>	SDIN hold	10		ns	From BCLK rising edge

## TIMING DIAGRAMS



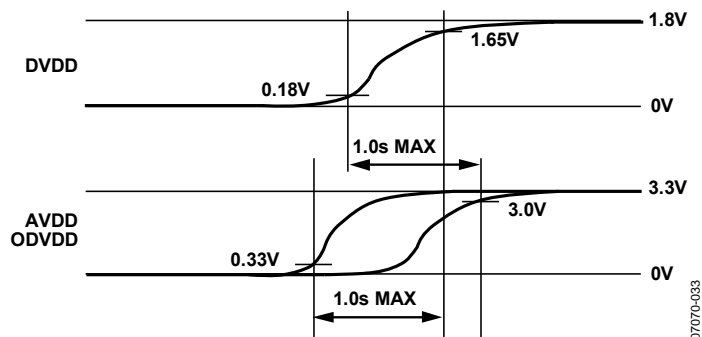


Figure 7. Power-Up Sequence Timing

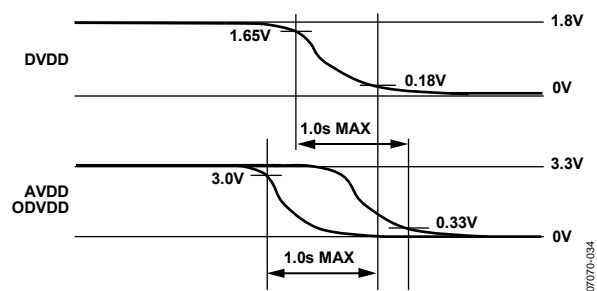


Figure 8. Power-Down Sequence Timing



## ABSOLUTE MAXIMUM RATINGS

Table 3.

Parameter	Rating
DVDD to DGND	0 V to 2.2 V
ODVDD to DGND	0 V to 4 V
AVDD to AGND	0 V to 4 V
AGND to DGND	–0.3 V to +0.3 V
Digital Inputs	DGND – 0.3 V to ODVDD + 0.3 V
Analog Inputs	AGND – 0.3 V to AVDD + 0.3 V
Reference Voltage	Indefinite short circuit to ground
Soldering (10 sec)	300°C

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## THERMAL RESISTANCE

$\theta_{JA}$  is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages.

Table 4. Thermal Resistance<sup>1</sup>

Package Type	$\theta_{JA}$	$\theta_{JC}$	Unit
80-Lead LQFP	38.1	7.6	°C/W

<sup>1</sup> Based on JEDEC 252P PCB.

## THERMAL CONDITIONS

To ensure correct operation of the device, the case temperature ( $T_{CASE}$ ) must be kept below 121°C to keep the junction temperature ( $T_J$ ) below the maximum allowed, 125°C.

## ESD CAUTION



**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

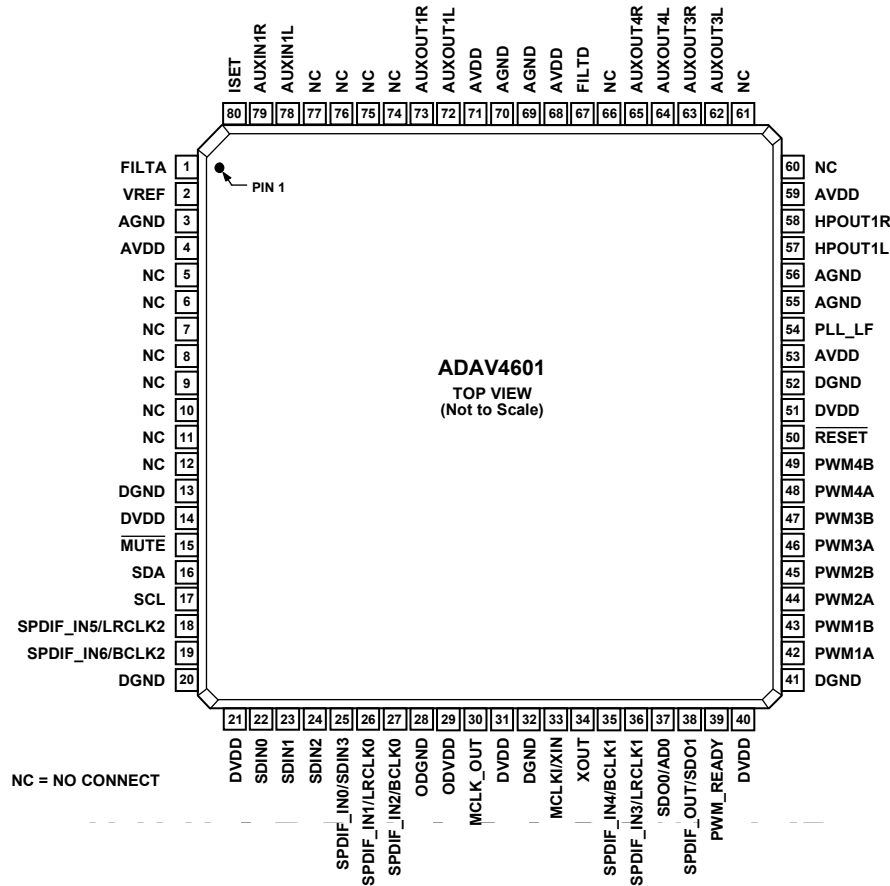


Figure 9. Pin Configuration

Table 5. Pin Function Descriptions

Pin No.	Mnemonic	Description
1	FILTA	ADC Filter Capacitor.
2	VREF	Reference Capacitor.
3	AGND	ADC Ground.
4	AVDD	ADC Supply (3.3 V).
5 to 12	NC	No Connection to this Pin Allowed.
13	DGND	Digital Ground.
14	DVDD	Digital Supply (1.8 V).
15	MUTE	Active-Low Mute Request Input Signal.
16	SDA	I <sup>2</sup> C Data.
17	SCL	I <sup>2</sup> C Clock.
18	SPDIF_IN5/LRCLK2	External Input to S/PDIF Mux/Left/Right Clock for SRC2 (Default).
19	SPDIF_IN6/BCLK2	External Input to S/PDIF Mux/Bit Clock for SRC2 (Default).
20	DGND	Digital Ground.
21	DVDD	Digital Supply (1.8 V).
22	SDIN0	Serial Data Input 0/SRC Data Input.
23	SDIN1	Serial Data Input 1/SRC Data Input.
24	SDIN2	Serial Data Input 2/SRC Data Input.
25	SPDIF_IN0/SDIN3	External Input to S/PDIF Mux/SRC Data Input/Serial Data Input 3 (Default).
26	SPDIF_IN1/LRCLK0	External Input to S/PDIF Mux/Left/Right Clock for SRC1 (Default).
27	SPDIF_IN2/BCLK0	External Input to S/PDIF Mux/Bit Clock for SRC1 (Default).
28	ODGND	Digital Ground.

Pin No.	Mnemonic	Description
29	ODVDD	Digital Interface Supply (3.3 V).
30	MCLK_OUT	Master Clock Output.
31	DVDD	Digital Supply (1.8 V).
32	DGND	Digital Ground.
33	MCLKI/XIN	Master Clock/Crystal Input.
34	XOUT	Crystal Output.
35	SPDIF_IN4/BCLK1	External Input to S/PDIF Mux/Bit Clock for Serial Data I/O (Default).
36	SPDIF_IN3/LRCLK1	External Input to S/PDIF Mux/Left/Right Clock for Serial Data I/O (Default).
37	SDO0/AD0	Serial Data Output. This pin acts as the I <sup>2</sup> C address select on reset. It has an internal pull-down resistor.
38	SPDIF_OUT/SDO1	Output of S/PDIF Mux/Serial Data Output.
39	PWM_READY	PWM Ready Flag.
40	DVDD	Digital Supply (1.8 V).
41	DGND	Digital Ground.
42	PWM1A	Pulse-Width Modulated Output 1A.
43	PWM1B	Pulse-Width Modulated Output 1B.
44	PWM2A	Pulse-Width Modulated Output 2A.
45	PWM2B	Pulse-Width Modulated Output 2B.
46	PWM3A	Pulse-Width Modulated Output 3A.
47	PWM3B	Pulse-Width Modulated Output 3B.
48	PWM4A	Pulse-Width Modulated Output 4A.
49	PWM4B	Pulse-Width Modulated Output 4B.
50	RESET	Reset Analog and Digital Cores.
51	DVDD	Digital Supply (1.8 V).
52	DGND	Digital Ground.
53	AVDD	PLL Supply (3.3 V).
54	PLL_LF	PLL Loop Filter.
55	AGND	PLL Ground.
56	AGND	Headphone Driver Ground.
57	HPOUT1L	Left Headphone Output.
58	HPOUT1R	Right Headphone Output.
59	AVDD	Headphone Driver Supply (3.3 V).
60, 61	NC	No Connection to this Pin Allowed.
62	AUXOUT3L	Left Auxiliary Output 3.
63	AUXOUT3R	Right Auxiliary Output 3.
64	AUXOUT4L	Left Auxiliary Output 4.
65	AUXOUT4R	Right Auxiliary Output 4.
66	NC	No Connection to this Pin Allowed.
67	FILTD	DAC Filter Capacitor.
68	AVDD	DAC Supply (3.3 V).
69, 70	AGND	DAC Ground.
71	AVDD	DAC Supply (3.3 V).
72	AUXOUT1L	Left Auxiliary Output 1.
73	AUXOUT1R	Right Auxiliary Output 1.
74 to 77	NC	No Connection to this Pin Allowed.
78	AUXIN1L	Left Auxiliary Input 1.
79	AUXIN1R	Right Auxiliary Input 1.
80	ISET	ADC Current Setting.

## TYPICAL PERFORMANCE CHARACTERISTICS

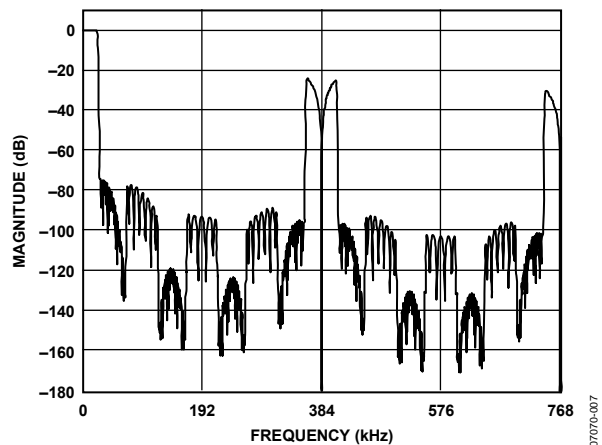


Figure 10. DAC Composite Filter Response (48 kHz)

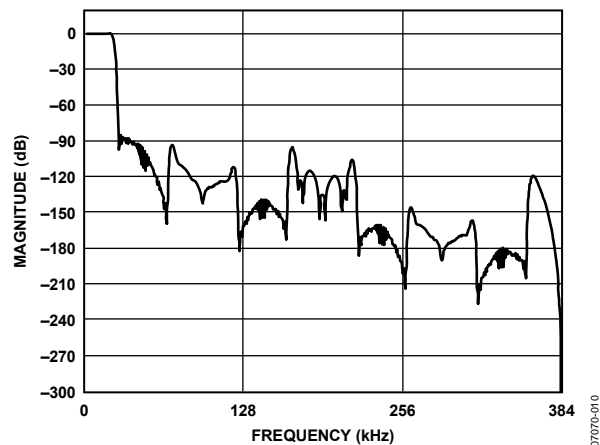


Figure 13. ADC Composite Filter Response (48 kHz)

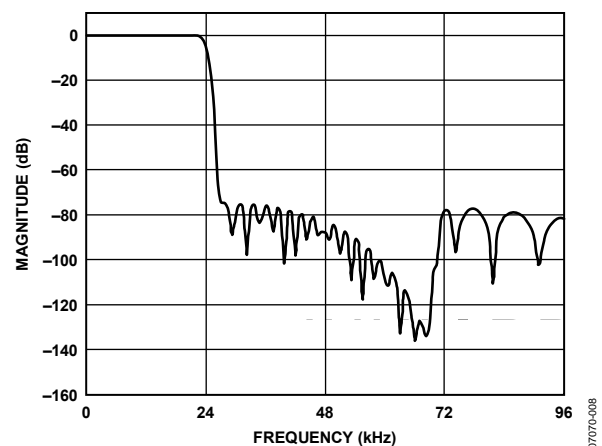


Figure 11. DAC Pass-Band Filter Response (48 kHz)

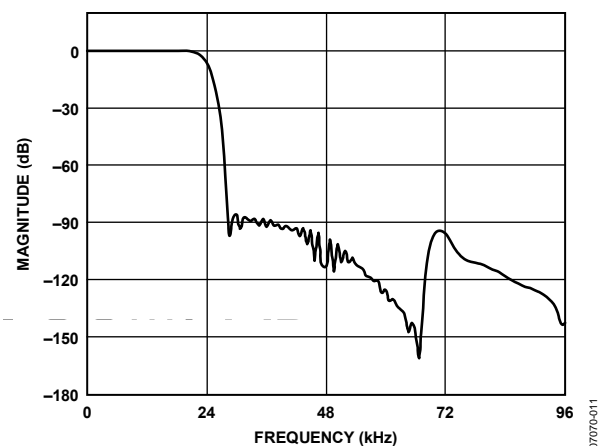


Figure 14. ADC Pass-Band Filter Response (48 kHz)

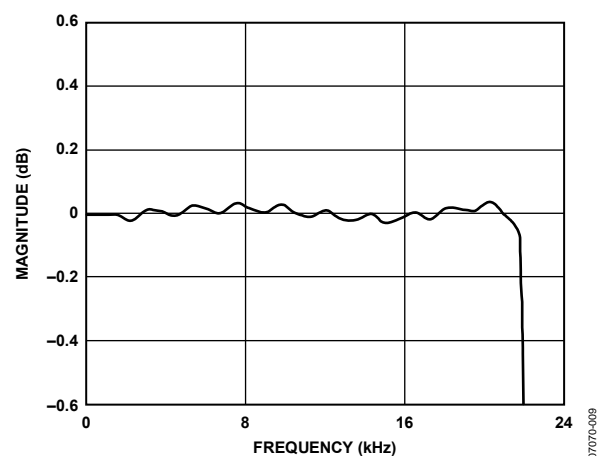


Figure 12. DAC Pass-Band Ripple (48 kHz)

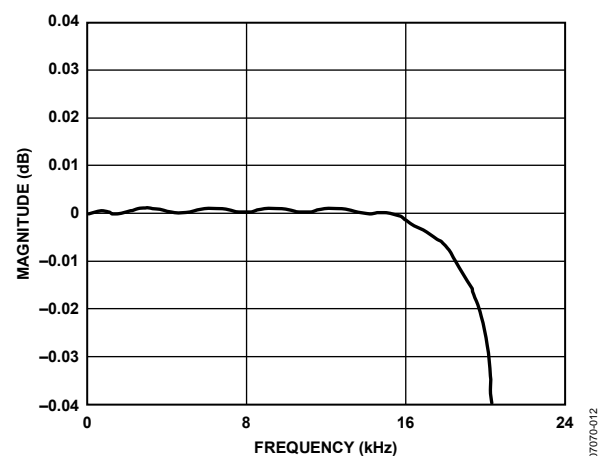


Figure 15. ADC Pass-Band Ripple (48 kHz)

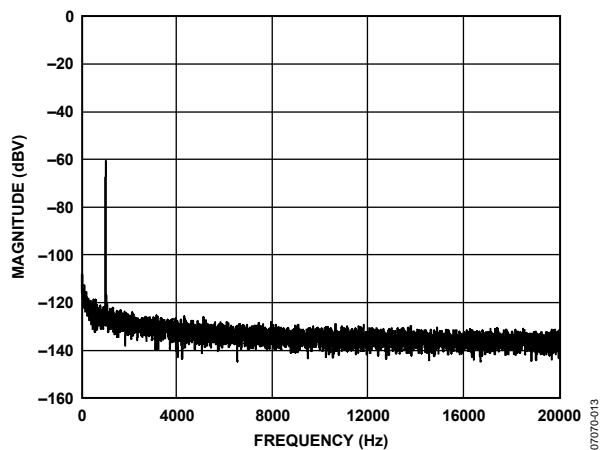


Figure 16. DAC Dynamic Range

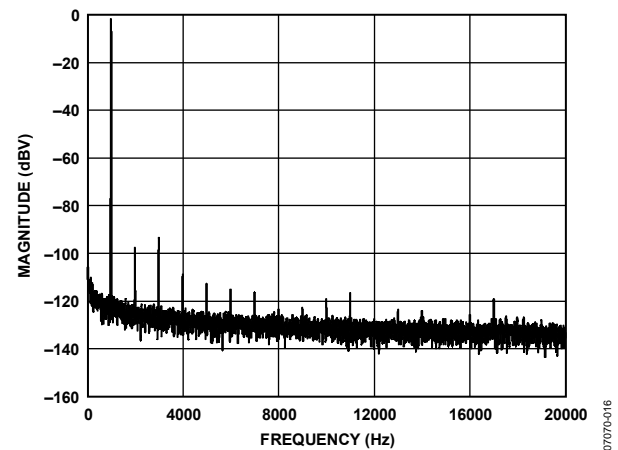


Figure 19. ADC Total Harmonic Distortion + Noise

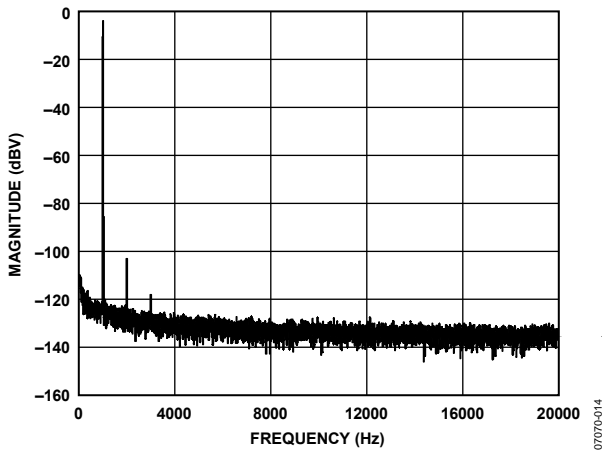


Figure 17. DAC Total Harmonic Distortion + Noise

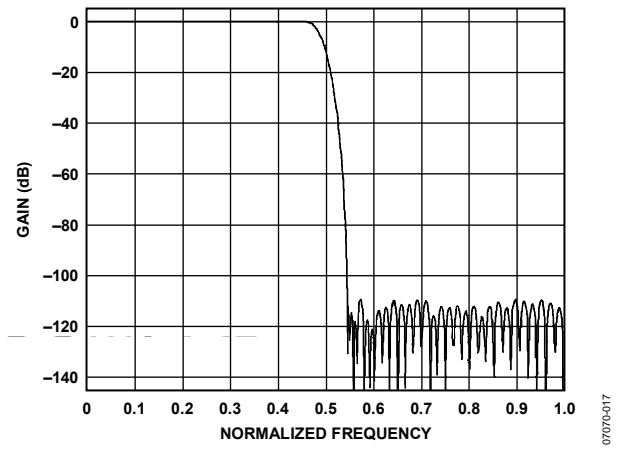


Figure 20. Sample Rate Converter Transfer Function

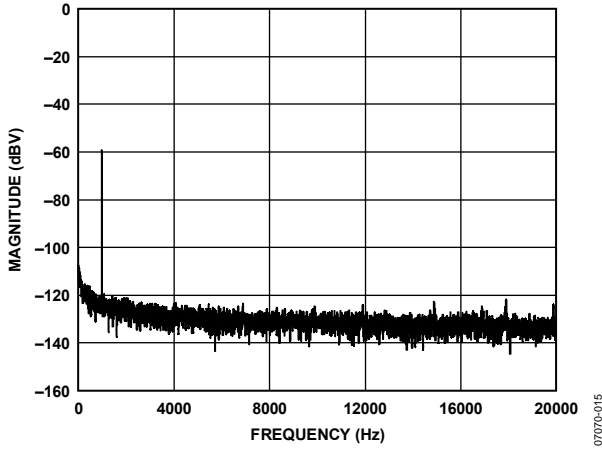


Figure 18. ADC Dynamic Range

## TERMINOLOGY

### Dynamic Range

The ratio of a full-scale input signal to the integrated input noise in the pass band (20 Hz to 20 kHz), expressed in decibels (dB). Dynamic range is measured with a –60 dB input signal and is equal to  $(S/[THD+N]) + 60$  dB. Note that spurious harmonics are below the noise with a –60 dB input, so the noise level establishes the dynamic range. The dynamic range is specified with and without an A-weight filter applied.

### Pass Band

The region of the frequency spectrum unaffected by the attenuation of the digital decimator's filter.

### Pass-Band Ripple

The peak-to-peak variation in amplitude response from equal amplitude input signal frequencies within the pass band, expressed in decibels.

### Stop Band

The region of the frequency spectrum attenuated by the digital decimator's filter to the degree specified by stop-band attenuation.

### Gain Error

With a near full-scale input, the ratio of the actual output to the expected output, expressed in dB.

### Interchannel Gain Mismatch

With identical near full-scale inputs, the ratio of the outputs of the two stereo channels, expressed in decibels.

### Crosstalk

Ratio of response on one channel with a grounded input to a full-scale 1 kHz sine wave input on the other channel, expressed in decibels.

### Power Supply Rejection

With no analog input, the signal present at the output when a 300 mV p-p signal is applied to power supply pins, expressed in decibels of full scale.

### Group Delay

Intuitively, the time interval required for an input pulse to appear at the converter's output, expressed in milliseconds (ms). More precisely, the derivative of radian phase with respect to radian frequency at a given frequency.

## PIN FUNCTIONS

Table 5 shows the pin numbers, mnemonics, and descriptions for the ADAV4601. The input pins have a logic threshold compatible with 3.3 V input levels.

### SDIN0, SDIN1, SDIN2, AND SDIN3/SPDIF\_IN0

Serial data inputs. These input pins provide the digital audio data to the signal processing core. Any of the inputs can be routed to either of the SRCs for conversion; this input is then not available as a synchronous input to the audio processor but only as an input through the selected SRC. The serial format for the synchronous data is selected by Bits [3:2] of the serial port control register. If the SRCs are required, the serial format is selected by Bits [12:9] of the same register. The synchronous inputs are capable of using any pair of serial clocks, LRCLK0/BCLK0, LRCLK1/BCLK1, or LRCLK2/BCLK2. By default, they use LRCLK1 and BCLK1. See Figure 23 for more details regarding the configuration of the synchronous inputs.

SDIN3 is a shared pin with SPDIF\_IN0. If SDIN3 is not in use, this pin can be used to connect an S/PDIF signal from an external source, such as an MPEG decoder, to the ADAV4601 on-chip S/PDIF output multiplexer. If SPDIF\_OUT is selected from one of the SPDIF\_IN (external) signals, the signal is simply passed through from input to output.

### LRCLK0, BCLK0, LRCLK1, BCLK1, LRCLK2, AND BCLK2

By default, LRCLK1 and BCLK1 are associated with the synchronous inputs, LRCLK0 and BCLK0 are associated with SRC1, and LRCLK2 and BCLK2 are associated with SRC2. However, the SRCs and synchronous inputs can use any of the serial clocks (see Figure 23 for more details). LRCLK0, BCLK0, LRCLK1, BCLK1, LRCLK2, and BCLK2 are shared pins with SPDIF\_IN1, SPDIF\_IN2, SPDIF\_IN3, SPDIF\_IN4, SPDIF\_IN5, and SPDIF\_IN6, respectively. If LRCLK0/LRCLK1/LRCLK2 or BCLK0/BCLK1/BCLK2 are not in use, these pins can be used to connect an S/PDIF signal from an external source, such as an MPEG decoder, to the ADAV4601 on-chip S/PDIF output multiplexer. If SPDIF\_OUT is selected from one of the SPDIF\_IN (external) signals, the signal is simply passed through from input to output.

### SDO0/AD0

Serial data output. This pin can output two channels of digital audio using a variety of standard 2-channel formats. The clocks for SDO0 are always the same as those used by the synchronous inputs; this means that LRCLK1 and BCLK1 are used by default, although SDO0 is capable of using any pair of serial clocks, LRCLK0/BCLK0, LRCLK1/BCLK1, or LRCLK2/BCLK2. The serial port control register selects the serial format for the synchronous output. On reset, the SDO0 pin duplicates as the I<sup>2</sup>C® address select pin. In this mode, the logical state of the pin is polled for four MCLKI cycles following reset. The address select bit is set as the majority poll of the pin's logic level after the four MCLKI cycles.

### SPDIF\_OUT (SDO1)

The ADAV4601 contains an S/PDIF multiplexer functionality that allows the SPDIF\_OUT signal to be chosen from an internally generated S/PDIF signal or from the S/PDIF signal of an external source, which is connected via one of the SPDIF\_IN pins. This pin can also be configured as an additional serial data output (SDO1) as an alternate function.

### MCLKI/XIN

Master clock input. The ADAV4601 uses a PLL to generate the appropriate internal clock for the audio processing core. A clock signal of a suitable frequency can be connected directly to this pin, or a crystal can be connected between MCLKI/XIN and XOUT together with the appropriate capacitors to DGND to generate a suitable clock signal.

### XOUT

This pin is used in conjunction with MCLKI/XIN to generate a clock signal for the ADAV4601.

### MCLK\_OUT

This pin can be used to output MCLKI or one of the internal system clocks. It should be noted that the output level of this pin is referenced to DVDD (1.8 V) and not ODVDD (3.3 V) like all other digital inputs and outputs.

### SDA

Serial data input for the I<sup>2</sup>C control port. SDA features a glitch elimination filter that removes spurious pulses that are less than 50 ns wide.

# ADAV4601

## SCL

Serial clock for the I<sup>2</sup>C control port. SCL features a glitch elimination filter that removes spurious pulses that are less than 50 ns wide.

## MUTE

Mute input request. This active-low input pin controls the muting of the output ports (both analog and digital) from the ADAV4601. When low, it asserts mute on the outputs that are enabled in the audio flow.

## RESET

Active-low reset signal. After  $\overline{\text{RESET}}$  goes high, all the circuit blocks are powered down. The blocks can be individually powered up with software. When the part is powered up, it takes approximately 3072 internal clocks to initialize the internal circuitry. The internal system clock is equal to MCLKI until the PLL is powered and enabled, after which the internal system clock becomes  $2560 \times f_s$  (122.88 MHz). Once the PLL is powered up and enabled after reset, it takes approximately 3 ms to lock. When the audio processor is enabled, it takes approximately 32,768 internal system clocks to initialize and load the default flow to the audio processor memory. The audio processor is not available during this time.

## AUXIN1L AND AUXIN1R

Analog inputs to the on-chip ADCs.

## AUXOUT1L, AUXOUT1R, AUXOUT3L, AUXOUT3R, AUXOUT4L, AND AUXOUT4R

Auxiliary DAC analog outputs. These pins can be programmed to supply the outputs of the internal audio processing for line out or record use.

## HPOUT1L AND HPOUT1R

Analog outputs from the headphone amplifiers.

## PLL\_LF

PLL loop filter connection. A 100 nF capacitor and a 2 k $\Omega$  resistor in parallel with a 1 nF capacitor tied to AVDD are required for the PLL loop filter to operate correctly.

## VREF

Voltage reference for DACs and ADCs. This pin is driven by an internal 1.5 V reference voltage.

## FILTA AND FILTD

Decoupling nodes for the ADC and DAC. Decoupling capacitors should be connected between these nodes and AGND, typically 47  $\mu\text{F}$  and 10  $\mu\text{F}$ , respectively.

## PWM1A, PWM1B, PWM2A, PWM2B, PWM3A, PWM3B, PWM4A, AND PWM4B

Differential pulse-width modulation outputs are suitable for driving Class-D amplifiers.

## PWM\_READY

This pin is set high when PWM is enabled and stable.

## AVDD

Analog power supply pins. These pins should be connected to 3.3 V. Each pin should be decoupled with 10  $\mu\text{F}$  and 0.1  $\mu\text{F}$  capacitors to AGND, as close to the pin as possible.

## DVDD

Digital power supply. This pin is connected to a 1.8 V digital supply. Connecting 10  $\mu\text{F}$  and 0.1  $\mu\text{F}$  decoupling capacitors to DGND, as close to the pin as possible is strongly recommended for optimal performance.

## ODVDD

Digital interface power supply pin. This pin should be connected to a 3.3 V digital supply. The pin should be decoupled with 10  $\mu\text{F}$  and 0.1  $\mu\text{F}$  capacitors to DGND, as close to the pin as possible.

## DGND

Digital ground.

## AGND

Analog ground.

## ODGND

Ground for the digital interface power supply.

## ISET

ADC current setting resistor.



## FUNCTIONAL DESCRIPTIONS

### MASTER CLOCK OSCILLATOR

Internally, the ADAV4601 operates synchronously to the master MCLKI input. All internal system clocks are generated from this single clock input using an internal PLL. This MCLKI input can also be generated by an external crystal oscillator connected to the MCLKI/XIN pin or by using a simple crystal resonator connected across MCLKI/XIN and XOUT. By default, the master clock frequency is 24.576 MHz; however, by using the internal dividers, an MCLKI of 12.288 MHz, 6.144 MHz, and 3.072 MHz are also supported.

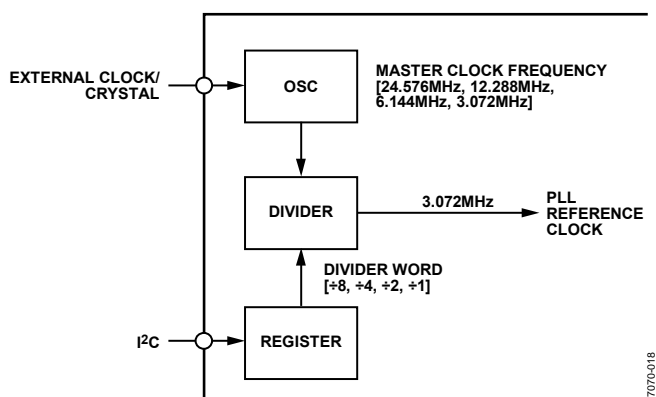


Figure 21. Master Clock

### I<sup>2</sup>C INTERFACE

The ADAV4601 supports a 2-wire serial (I<sup>2</sup>C compatible) microprocessor bus driving multiple peripherals. The ADAV4601 is controlled by an external I<sup>2</sup>C master device, such as a microcontroller. The ADAV4601 is in slave mode on the I<sup>2</sup>C bus, except during self-boot. While the ADAV4601 is self-booting, it becomes the master, and the EEPROM, which contains the ROMs to be booted, is the slave. When the self-boot process is complete, the ADAV4601 reverts to slave mode on the I<sup>2</sup>C bus. No other devices should access the I<sup>2</sup>C bus while the ADAV4601 is self-booting (refer to the Application Layer section and the Loading a Custom Audio Processing Flow section).

Initially, all devices on the I<sup>2</sup>C bus are in an idle state, wherein the devices monitor the SDA and SCL lines for a start condition and the proper address. The I<sup>2</sup>C master initiates a data transfer by establishing a start condition, defined by a high-to-low transition on SDA while SCL remains high. This indicates that an address/data stream follows. All devices on the bus respond to the start condition and read the next byte (7-bit address plus the R/W bit) MSB first. The device that recognizes the transmitted address responds by pulling the data line low during the ninth clock pulse. This ninth bit is known as an acknowledge bit. All other devices on the bus revert to an idle condition. The R/W bit determines the direction of the data. A Logic Level 0 on the LSB of the first byte means the master writes information to the peripheral. A Logic Level 1 on the LSB of the first byte

means the master reads information from the peripheral. A data transfer takes place until a stop condition is encountered. A stop condition occurs when SDA transitions from low to high while SCL is held high.

The ADAV4601 determines its I<sup>2</sup>C device address by sampling the SDO0 pin after reset. Internally, the SDO0 pin is sampled by four MCLKI edges to determine the state of the pin (high or low). Because the pin has an internal pull-down resistor default, the address of the ADAV4601 is 0x34 (write) and 0x35 (read). An alternate address, 0x36 (write) and 0x37 (read), is available by tying the SDO0 pin to ODVDD via a 10 kΩ resistor. The I<sup>2</sup>C interface supports a clock frequency up to 400 kHz.

### ADC INPUTS

The ADAV4601 has two ADC inputs. By default, these are configured as a single stereo input; however, because the audio processor is programmable, these inputs can be reconfigured.

The ADC inputs are shown in Figure 22. The analog inputs are current inputs (100 μA rms FS) with a 1.5 V dc bias voltage. Any input voltage can be accommodated by choosing a suitable combination of input resistor (R<sub>IN</sub>) and ISET resistor (R<sub>ISET</sub>) using the formulas

$$R_{IN} = V_{FS\ rms}/100\ \mu A\ rms$$

$$R_{ISET} = 2R_{IN}/V_{IN}$$

Resistor matching (typically 1%) between R<sub>IN</sub> and R<sub>ISET</sub> is important to ensure a full-scale signal on the ADC without clipping.

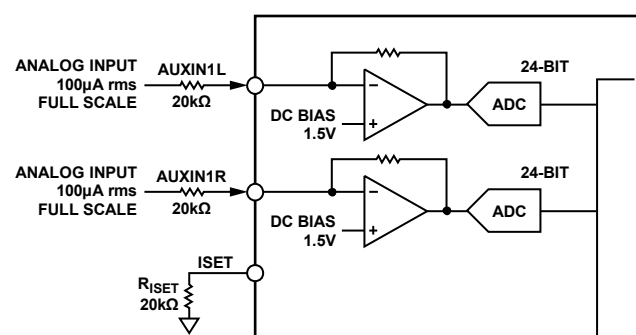


Figure 22. Analog Input Section

### I<sup>2</sup>S DIGITAL AUDIO INPUTS

The ADAV4601 has four I<sup>2</sup>S digital audio inputs that are, by default, synchronous to the master clock. Also available are two SRCs capable of supporting any nonsynchronous input with a sample rate between 5 kHz and 50 kHz. Any of the serial digital inputs can be redirected through the SRC. Figure 23 shows a block diagram of the input serial port.

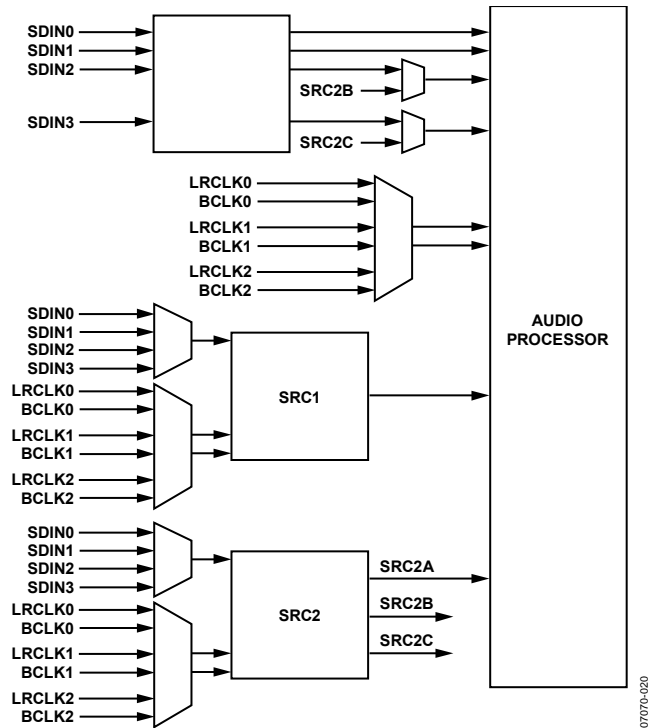


Figure 23. Digital Input Section

## Synchronous Inputs and Outputs

The synchronous digital inputs and outputs can use any of the BCLK or LRCLK inputs as a clock and framing signal. By default, BCLK1 and LRCLK1 are the serial clocks used for the synchronous inputs. The synchronous port for the ADAV4601 is in slave mode by default, which means the user must supply the appropriate serial clocks, BCLK and LRCLK. The synchronous port can also be set to master mode, which means that the appropriate serial clocks, BCLK and LRCLK, can be generated internally from the MCLK; therefore, the user does not need to provide them. The serial data inputs are capable of accepting all the popular audio transmission standards (see the Serial Data Interface section for more details).

## Asynchronous Inputs

The ADAV4601 has two SRCs, SRC1 and SRC2, that can be used for converting digital data, which is not synchronous to the master clock. Each SRC can accept input sample rates in the range of 5 kHz to 50 kHz. Data that has been converted by the SRC is inputted to the part and is then synchronous to the internal audio processor.

The SRC1 is a 2-channel (single-stereo) sample rate converter that is capable of using any of the three serial clocks available. The SRC1 can accept data from any of the serial data inputs (SDIN0, SDIN1, SDIN2, and SDIN3). Once selected as an input to the SRC, this SDIN line is assumed to contain asynchronous data and is then masked as an input to the audio processor to ensure that asynchronous data is not processed as synchronous data. By default, SRC1 uses the LRCLK0 and BCLK0 as the clock and framing signals.

The SRC2 is a 6-channel (3-stereo) sample rate converter that is capable of using any of the three serial clocks available. The SRC2 can accept data from any of the serial data inputs (SDIN0, SDIN1, SDIN2, and SDIN3). Once selected as an input to the SRC, this SDIN line is assumed to contain asynchronous data and is then masked internally as an input to the audio processor to ensure that asynchronous data is not processed as synchronous data. By default, SRC2 uses the LRCLK2 and BCLK2 as the clock and framing signals.

The first output (SRC2A) from SRC2 is always available to the audio processor. The other two outputs are muxed with two of the serial inputs before being available to the audio processor. SRC2B is muxed with SDIN2 and SRC2C is muxed with SDIN3. By default, these muxes are configured so that the synchronous inputs are available to the audio processor. The SRC2B and SRC2C channels can be made available to the audio processor simply by enabling them by register write.

When using the ADAV4601 in an asynchronous digital-in-to-digital-out configuration, the input digital data are input to the audio processor core from one of the SRCs, using the assigned BCLK/LRCLK as a framing signal. The digital output is synchronous to the BCLK/LRCLK, which is assigned to the synchronous port; the default clocks in this case are BCLK1 and LRCLK1.

## Serial Data Interface

LRCLK is the framing signal for the left- and right-channel inputs, with a frequency equal to the sampling frequency ( $f_s$ ).

BCLK is the bit clock for the digital interface with a frequency of  $64 \times f_s$  (32 BCLK periods for each of the left and right channels).

The serial data interface supports all the popular audio interface standards, such as I<sup>2</sup>S, left-justified (LJ), and right-justified (RJ). The interface mode is software selectable, and its default is I<sup>2</sup>S. The data sample width is also software selectable from 16 bits, 20 bits, or 24 bits. The default is 24 bits.

## I<sup>2</sup>S Mode

In I<sup>2</sup>S mode, the data are left-justified, MSB first, with the MSB placed in the second BCLK period following the transition of the LRCLK. A high-to-low transition of the LRCLK signifies the beginning of the left channel data transfer, and a low-to-high transition on the LRCLK signifies the beginning of the right channel data transfer (see Figure 24).

### LJ Mode

In LJ mode, the data are left-justified, MSB first, with the MSB placed in the first BCLK period following the transition of the LRCLK. A high-to-low transition of the LRCLK signifies the beginning of the right-channel data transfer, and a low-to-high transition on the LRCLK signifies the beginning of the left-channel data transfer (see Figure 25).

### RJ Mode

In RJ mode, the data are right-justified, LSB last, with the LSB placed in the last BCLK period preceding the transition of LRCLK. A high-to-low transition of the LRCLK signifies the beginning of the right-channel data transfer, and a low-to-high transition on the LRCLK signifies the beginning of the left-channel data transfer (see Figure 26).

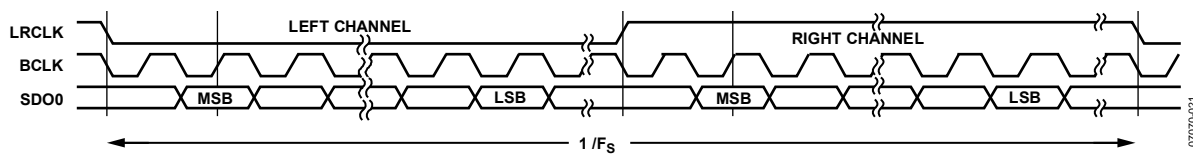


Figure 24. I²S Mode

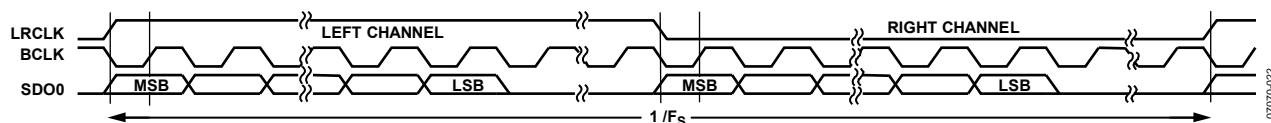


Figure 25. Left-Justified Mode

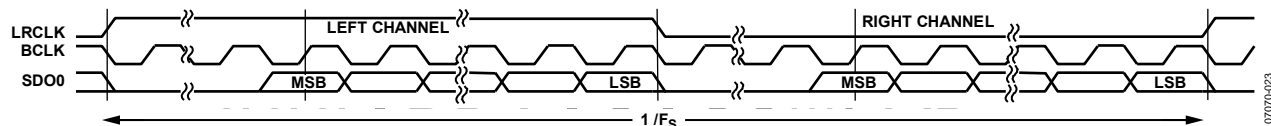


Figure 26. Right-Justified Mode

# ADAV4601

## DAC VOLTAGE OUTPUTS

The ADAV4601 has six DAC outputs, configured as three stereo auxiliary DAC outputs. However, because the flow is customizable, this is programmable. The output level is 1 V rms full scale.

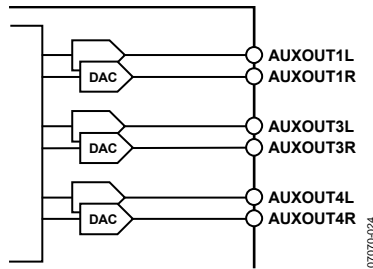


Figure 27. DAC Output Section

## PWM OUTPUTS

In the ADAV4601, the main outputs are available as four PWM output channels, which are suitable for driving Class-D amplifiers. PWM\_Ready is a status pin used to signify that the ADAV4601 PWM outputs are in a valid state. During PWM power-up and power-down, this pin remains low to signify that the outputs are not in a valid state. The output power stage should remain muted until this pin goes high. This functionality helps to eliminate pop/click and other unwanted noise on the outputs.

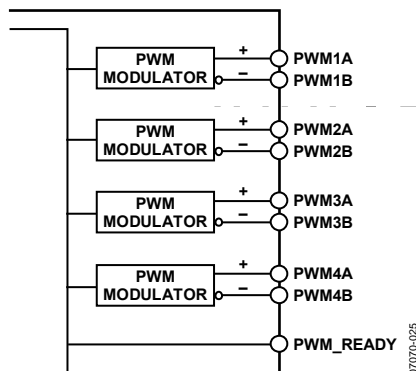


Figure 28. PWM Output Section

Each set of PWM outputs is a complementary output. The modulation frequency is 384 kHz, and the full-scale duty cycle has a ratio of 97:3.

Full details on the use of the PWM outputs are available upon request from [AV.Products@analog.com](mailto:AV.Products@analog.com).

## HEADPHONE OUTPUTS

There is a dedicated stereo headphone amplifier output that is capable of driving 32  $\Omega$  loads at 1 V rms.

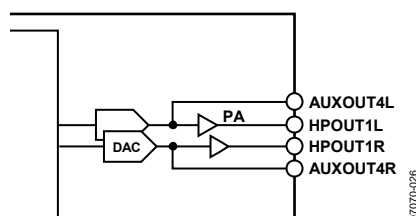


Figure 29. Headphone Output Section

## I<sup>2</sup>S DIGITAL AUDIO OUTPUTS

One I<sup>2</sup>S output, SDO0, uses the same serial clocks as the serial inputs, which is BCLK1 and LRCLK1 by default. If an additional digital output is required, an additional pin can be reconfigured as a serial digital output, as shown in Figure 30.

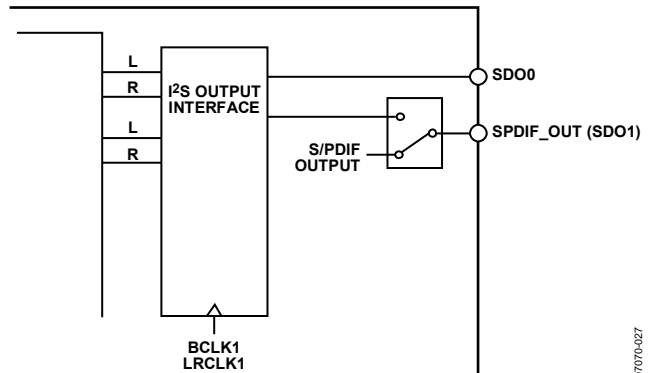


Figure 30. I<sup>2</sup>S Digital Outputs

## S/PDIF INPUT/OUTPUT

The S/PDIF output (SPDIF\_OUT/SDO1) uses a multiplexer to select an output from the audio processor or to pass through the unprocessed SPDIF\_IN signals, as shown in Figure 31. On the ADAV4601, the S/PDIF inputs, SPDIF\_IN0/SPDIF\_IN1/SPDIF\_IN2/SPDIF\_IN3/SPDIF\_IN4/SPDIF\_IN5/SPDIF\_IN6, are available on the SDIN3, LRCLK0, BCLK0, LRCLK1, BCLK1, LRCLK2, and BCLK2 pins, respectively. It is possible to have all seven S/PDIF inputs connected to different S/PDIF signals at one time. A consequence of this setup is that none of the LRCLKs and BCLKs are available for use with the digital inputs SDIN0, SDIN1, SDIN2, and SDIN3. If there is only one S/PDIF input in use, using the SDIN3 pin as the dedicated S/PDIF input is recommended; this enables BCLK0/LRCLK0, BCLK1/LRCLK1, and BCLK2/LRCLK2 to be used as the clock and framing signal for the synchronous and asynchronous port. If SDIN3 is used as an S/PDIF input, it should not be used internally as an input to the audio processor because it contains invalid data. Similarly, if BCLK or LRCLK are used as S/PDIF inputs, they can no longer be used as the clock and framing signals for SDIN0, SDIN1, SDIN2, and SDIN3. The S/PDIF encoder supports only consumer formats that conform to IEC-600958.

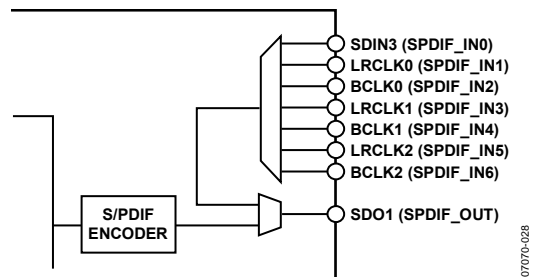


Figure 31. S/PDIF Output

## HARDWARE MUTE CONTROL

The ADAV4601 mute input can be used to mute any of the analog or digital outputs. When the **MUTE** pin goes low, the selected outputs ramp to a muted condition. Unmuting is handled in one of two ways and depends on the register setting. By default, the **MUTE** pin going high causes the outputs to immediately ramp to an unmuted state. However, it is also possible to have the unmute operation controlled by a control register bit. In this scenario, even if the **MUTE** pin goes high, the device does not unmute until a bit in the control register is set. This can be used when the user wants to keep the outputs muted, even after the pin has gone high again, for example, in the case of a fault condition. This allows the system controller total control over the unmute operation.

Full details on register settings and operation of the mute function are available upon request from [AV.Products@analog.com](mailto:AV.Products@analog.com).

## AUDIO PROCESSOR

The internal audio processor runs at  $2560 \times f_s$ ; at 48 kHz, this is 122.88 MHz. Internally, the word size is 28 bits, which allows 24 dB of headroom for internal processing. Designed specifically with audio processing in mind, it can implement complex audio algorithms efficiently.

By default, the ADAV4601 loads a default audio flow, as shown in Figure 33. However, because the audio processor is fully programmable, a custom audio flow can be quickly developed and loaded to the audio processor.

The audio flow is contained in program RAM and parameter RAM. Program RAM contains the instructions to be processed by the audio processor, and parameter RAM contains the coefficients that control the flow, such as volume control, filter coefficients, and enable bits.

## GRAPHICAL PROGRAMMING ENVIRONMENT

Custom flows for the ADAV4601 are created in a powerful drag-and-drop graphical programming application. No knowledge of assembly code is required to program the ADAV4601. Featuring a comprehensive library of audio processing blocks (such as filters, delays, dynamics processors, and third-party algorithms), it allows the quick and simple creation of custom flows. For debugging purposes, run-time control of the audio flow allows the user to fully configure and test the created flow.

Training materials and support are available upon request from [AV.Products@analog.com](mailto:AV.Products@analog.com).

## APPLICATION LAYER

Unique to this family is the embedded application layer, which allows the user to define a custom set of registers to control the audio flow, greatly simplifying the interface between the audio processor and the system controller.

Once a custom flow is created, a user-customized register map can be defined for controlling the flow. Each register is 16 bits, but controls can use only one bit or all 16 bits. Users have full control over which parameters they control and the degree of control they have over those parameters during run time. The combination of the graphical programming environment and the powerful application layer allows the user to quickly develop a custom audio flow and still maintain the usability of a simple register-based device.

Comprehensive documentation on developing a custom audio flow and the definition and creation of the custom application layer for the ADAV4601 is available upon request from [AV.Products@analog.com](mailto:AV.Products@analog.com).

## LOADING A CUSTOM AUDIO PROCESSING FLOW

The ADAV4601 can load a custom audio flow from an external I<sup>2</sup>C ROM. The boot process is initiated by a simple control register write. The EEPROM device address and the EEPROM start address for the audio flow ROMs can all be programmed.

For the duration of the boot sequence, the ADAV4601 becomes the master on the I<sup>2</sup>C bus. Transfer of the ROMs from the EEPROM to the ADAV4601 takes a maximum of 1.06 sec, assuming that the full audio processor memory is required, during which time no other devices should access the I<sup>2</sup>C bus. Once the transfer is complete, the ADAV4601 automatically reverts to slave mode, and the I<sup>2</sup>C bus master can resume sending commands.

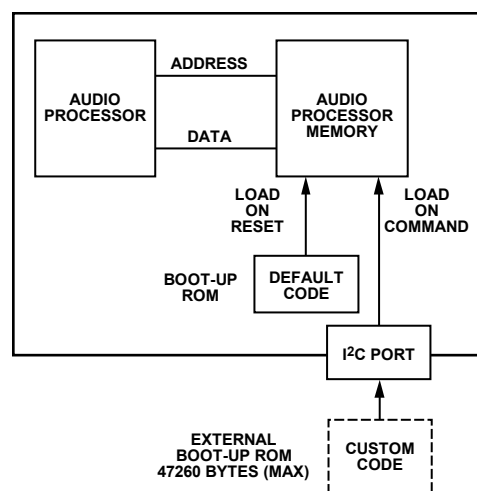


Figure 32. External EEPROM Booting

07070-028

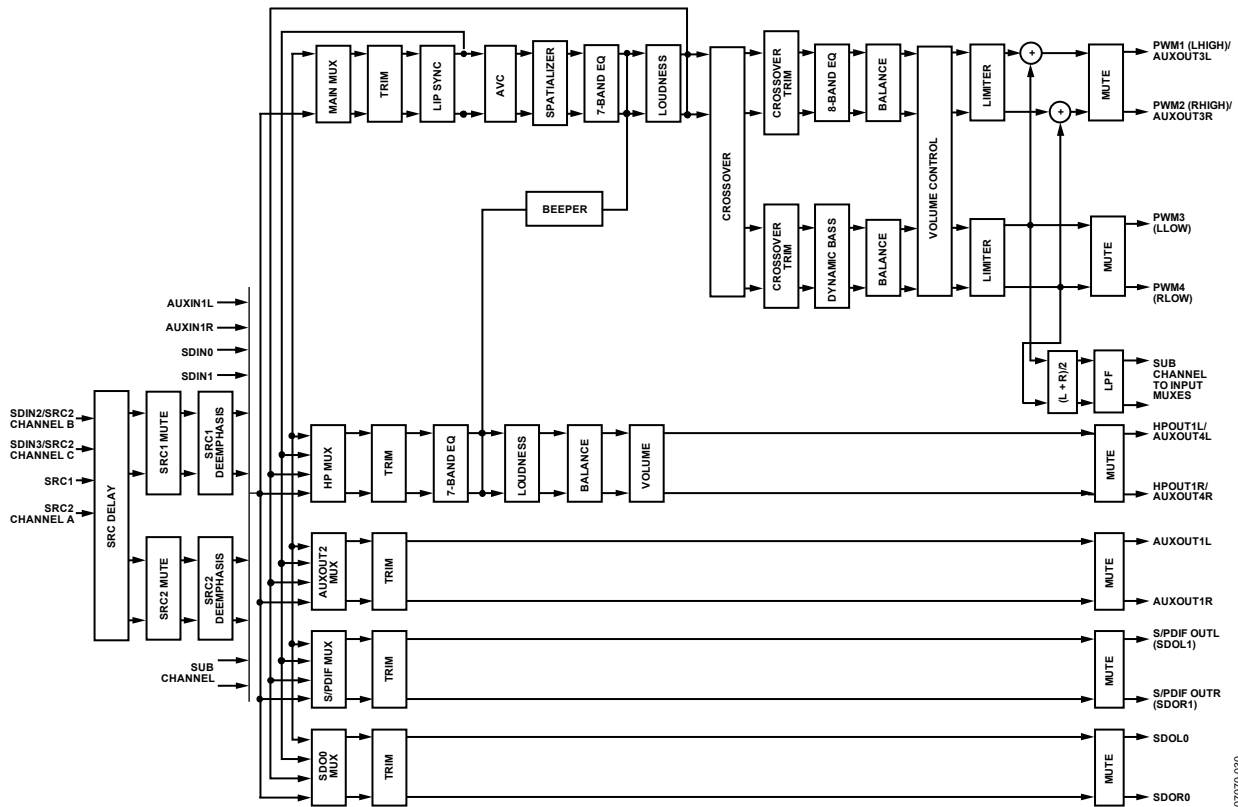
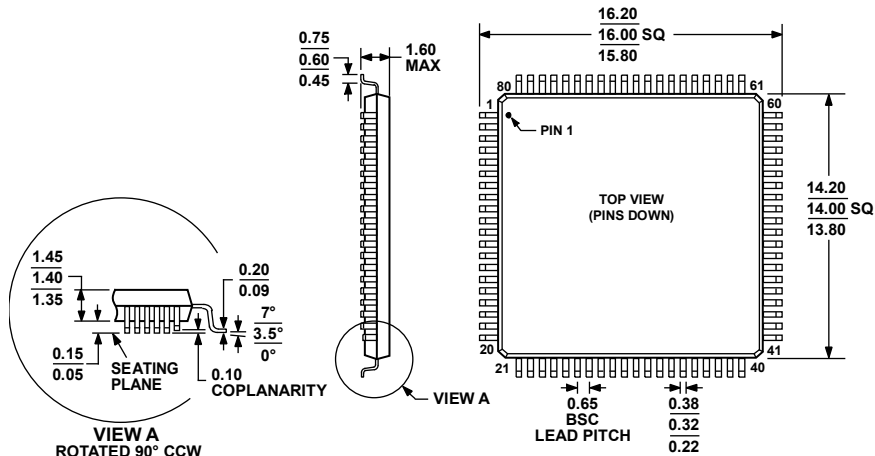


Figure 33. Default Audio Processing Flow

07070-030

# OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MS-026-BEC

Figure 34. 80-Lead Low Profile Quad Flat Package [LQFP]  
(ST-80-2)

Dimensions shown in millimeters

## ORDERING GUIDE

Model	Temperature Range	Package Description	Package Option
ADAV4601BSTZ <sup>1</sup>	−40°C to +85°C	80-Lead Low Profile Quad Flat Package [LQFP]	ST-80-2

<sup>1</sup> Z = RoHS Compliant Part.

In addition, it is backward compatible with conventional SnPb soldering processes. This means the electroplated Sn coating can be soldered with Sn/Pb solder pastes at conventional reflow temperatures of 220°C to 235°C.

## NOTES

Purchase of licensed I<sup>2</sup>C components of Analog Devices or one of its sublicensed Associated Companies conveys a license for the purchaser under the Philips I<sup>2</sup>C Patent Rights to use these components in an I<sup>2</sup>C system, provided that the system conforms to the I<sup>2</sup>C Standard Specification as defined by Philips.