## Micropower 400 mA LDO Linear Regulators with ENABLE, DELAY, and RESET

The NCV8505 is a family of precision micropower voltage regulators. Their output current capability is 400 mA. The family has output voltage options for Adjustable, 2.5 V, 3.3 V and 5.0 V.

The output voltage is accurate within  $\pm$  2.0% with a maximum dropout voltage of 0.6 V at 400 mA. Low quiescent current is a feature drawing less than 1.0  $\mu$ A with ENABLE = 0 V. With ENABLE = 5.0 V, the part only draws 200  $\mu$ A with 100  $\mu$ A load. This part is ideal for any and all battery operated microprocessor equipment.

Microprocessor control logic includes an active  $\overline{RESET}$  (with DELAY).

The active  $\overline{RESET}$  circuit operates correctly at an output voltage as low as 1.0 V. The  $\overline{RESET}$  function is activated during the power up sequence or during normal operation if the output voltage drops below the regulation limits.

The regulator is protected against reverse battery, short circuit, and thermal overload conditions. The device can withstand load dump transients making it suitable for use in automotive environments. The device has also been optimized for EMC conditions.

#### Features

- Output Voltage Options: Adjustable, 2.5 V, 3.3 V, 5.0 V
- ± 2.0% Output
- Low < 1.0 µA Sleep Current
- Low 200 μA Quiescent Current
- Fixed or Adjustable Output Voltage
- Active RESET
- ENABLE
- 400 mA Output Current Capability
- Fault Protection
  - ♦ +60 V Peak Transient Voltage
  - → -15 V Reverse Voltage
  - ♦ Short Circuit
  - Thermal Overload
- NCV Prefix for Automotive and Other Applications Requiring Site and Change Control



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#### MARKING DIAGRAM



D<sup>2</sup>PAK-7 DPS SUFFIX CASE 936AB



x = Voltage Ratings as Indicated Below:

A = Adjustable

2 = 2.5 V

3 = 3.3 V5 = 5.0 V

= Assembly Location

WL = Wafer Lot

YY = Year

WW = Work Week

#### ORDERING INFORMATION

See detailed ordering and shipping information in the package dimensions section on page 11 of this data sheet.

#### **PIN CONNECTIONS**

#### **ADJUSTABLE OUTPUT FIXED OUTPUT** Tab = GNDTab = GNDLead 1. DELAY Lead 1. DELAY 0 0 2. ENABLE 2. ENABLE 3. RESET 3. RESET 4. GND 4. GND 5. SENSE 5. V<sub>ADJ</sub> 6. V<sub>OUT</sub> 6. V<sub>OUT</sub> 7. V<sub>IN</sub> 7. V<sub>IN</sub>

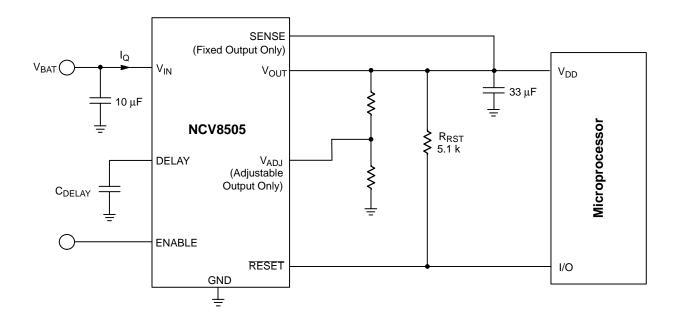


Figure 1. Application Diagram

#### **MAXIMUM RATINGS\***†

Rat	Value	Unit	
V <sub>IN</sub> (DC)		-15 to 45	V
Peak Transient Voltage (46 V Load Dump @ V <sub>IN</sub> =	= 14 V)	60	V
Operating Voltage		45	V
V <sub>OUT</sub> (DC)		16	V
Voltage Range (RESET, DELAY)		-0.3 to 10	V
Input Voltage Range: V <sub>ADJ</sub>		-0.3 to 16	V
Input Voltage Range (ENABLE)		-0.3 to 10**	V
ESD Susceptibility (Human Body Model) (Machine Model)		4.0 200	kV V
Junction Temperature, T <sub>J</sub>		-40 to +150	°C
Storage Temperature, T <sub>S</sub>		-55 to 150	°C
Package Thermal Resistance, 7 Lead D <sup>2</sup> PAK	$\begin{array}{c} \text{Junction-to-Case, R}_{\theta \text{JC}} \\ \text{Junction-to-Ambient, R}_{\theta \text{JA}} \end{array}$	2.0 10–50***	°C/W °C/W
Lead Temperature Soldering:	Reflow: (SMD styles only) (Note 1)	240 peak (Note 2)	°C

 $\textbf{ELECTRICAL CHARACTERISTICS} \text{ ($I_{OUT} = 1.0$ mA, ENABLE = $5.0$ V, $-40^{\circ}$C} \leq T_{J} \leq 150^{\circ}$C; $V_{IN} = \text{dependent on voltage option } T_{J} \leq 150^{\circ}$C; $V_{IN} = T_{J} \leq 150^$ (Note 3); unless otherwise specified.)

Characteristic	Test Conditions	Min	Тур	Max	Unit
Output Stage					
Output Voltage for 2.5 V Option (V <sub>O</sub> )	$6.5 \text{ V} < \text{V}_{\text{IN}} < 16 \text{ V}, \ 1.0 \text{ mA} \le \text{I}_{\text{OUT}} \le 400 \text{ mA}$ $\textbf{4.5 V} < \textbf{V}_{\text{IN}} < 26 \text{ V}, \ 1.0 \text{ mA} \le \text{I}_{\text{OUT}} \le 400 \text{ mA}$	2.450 2.425	2.5 2.5	2.550 2.575	V V
Output Voltage for 3.3 V Option (V <sub>O</sub> )	7.3 V < $V_{IN}$ < 16 V, 1.0 mA $\leq$ $I_{OUT}$ $\leq$ 400 mA 4.5 V < $V_{IN}$ < 26 V, 1.0 mA $\leq$ $I_{OUT}$ $\leq$ 400 mA	3.234 3.201	3.3 3.3	3.366 3.399	V V
Output Voltage for 5.0 V Option (V <sub>O</sub> )	9.0 V < $V_{IN}$ < 16 V, 1.0 mA $\leq$ $I_{OUT}$ $\leq$ 400 mA <b>6.0 V &lt; <math>V_{IN}</math> &lt; 26 V,</b> 1.0 mA $\leq$ $I_{OUT}$ $\leq$ 400 mA	4.90 4.85	5.0 5.0	5.10 5.15	V V
Output Voltage for Adjustable Option (V <sub>O</sub> )	$V_{OUT} = V_{ADJ}$ (Unity Gain) 6.5 V < $V_{IN}$ < 16 V, 1.0 mA < $I_{OUT}$ < 400 mA 4.5 V < $V_{IN}$ < 26 V, 1.0 mA < $I_{OUT}$ < 400 mA	1.274 1.261	1.300 1.300	1.326 1.339	V V
Dropout Voltage (V <sub>IN</sub> – V <sub>OUT</sub> ) (5.0 V and Adj. > 5.0 V Options Only)	I <sub>OUT</sub> = 400 mA I <sub>OUT</sub> = 1.0 mA	_ _	400 30	600 150	mV mV
Load Regulation	$V_{IN} = 14 \text{ V}, 5.0 \text{ mA} \le I_{OUT} \le 400 \text{ mA}$	-30	5.0	30	mV
Line Regulation (2.5 V, 3.3 V, and Adjustable Options)	4.5 V < V <sub>IN</sub> < 26 V, I <sub>OUT</sub> = 1.0 mA	_	5.0	25	mV
Line Regulation (5.0 V Option)	6.0 V < V <sub>IN</sub> < 26 V, I <sub>OUT</sub> = 1.0 mA	-	5.0	25	mV
Quiescent Current, (I <sub>Q</sub> ) Active Mode	$I_{OUT} = 100 \mu A, V_{IN} = 12 V$ $I_{OUT} = 75 \text{ mA}, V_{IN} = 14 V$ $I_{OUT} \le 400 \text{ mA}, V_{IN} = 14 V$	- - -	200 2.5 25	350 5.0 45	μA mA mA
Quiescent Current, (I <sub>Q</sub> ) Sleep Mode	ENABLE = 0 V, $V_{IN}$ = 12 V, $-40^{\circ}$ C $\leq T_{J} \leq 125^{\circ}$ C	-	_	1.0	μΑ
Current Limit	-	425	800	-	mA
Short Circuit Output Current	V <sub>OUT</sub> = 0 V	100	500	-	mA
Thermal Shutdown	(Guaranteed by Design)	150	180	_	°C

<sup>3.</sup> Voltage range specified in the Output Stage of the Electrical Characteristics in boldface type.

 <sup>60</sup> second maximum above 183°C.
 -5°C/+0°C allowable conditions.
 \*The maximum package power dissipation must be observed.
 \*\*Reference Figure 14 for switched-battery ENABLE application.

<sup>\*\*\*</sup>Depending on thermal properties of substrate,  $R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$ .
†During the voltage range which exceeds the maximum tested voltage of  $V_{IN}$ , operation is assured, but not specified. Wider limits may apply. Thermal dissipation must be observed closely.

**ELECTRICAL CHARACTERISTICS (continued)** ( $I_{OUT} = 1.0 \text{ mA}$ , ENABLE = 5.0 V,  $-40 ^{\circ}\text{C} \le T_{J} \le 150 ^{\circ}\text{C}$ ;  $V_{IN} = \text{dependent on } 1.0 ^{\circ}\text{C}$ voltage option (Note 4); unless otherwise specified.)

Characteristic Test Conditions		Min	Тур	Max	Unit
Reset Function (RESET)					
RESET Threshold for 2.5 V Option HIGH (V <sub>RH</sub> ) LOW (V <sub>RL</sub> ) Hysteresis	V <sub>IN</sub> = 4.5 V (Note 5) (Note 6) V <sub>OUT</sub> Increasing V <sub>OUT</sub> Decreasing		- - -	1.0 × V <sub>O</sub>	V V mV
RESET Threshold for 3.3 V Option HIGH (V <sub>RH</sub> ) LOW (V <sub>RL</sub> ) Hysteresis	V <sub>IN</sub> = 4.5 V (Note 5) (Note 6) V <sub>OUT</sub> Increasing V <sub>OUT</sub> Decreasing		- - -	1.0 × V <sub>O</sub>	V V mV
RESET Threshold for 5.0 V Option HIGH (V <sub>RH</sub> ) LOW (V <sub>RL</sub> ) Hysteresis	V <sub>IN</sub> = 6.0 V (Note 6) V <sub>OUT</sub> Increasing V <sub>OUT</sub> Decreasing	4.70 4.60 50		1.0 × V <sub>O</sub>	V V mV
$\label{eq:RESET} \hline \textbf{RESET Threshold for Adjustable Option} \\ \hline \textbf{HIGH (V}_{RH}) \\ \hline \textbf{LOW (V}_{RL}) \\ \hline \textbf{Hysteresis} \\ \hline \\ \hline $	V <sub>IN</sub> = 4.5 V (Note 5) (Note 6) V <sub>OUT</sub> Increasing V <sub>OUT</sub> Decreasing	1.22 1.19 10	111	1.0 × V <sub>O</sub>	V V mV
Output Voltage Low (V <sub>RLO</sub> )	$V_{IN}$ = Minimum (Note 6) (Note 7) 1.0 V $\leq$ V <sub>OUT</sub> $\leq$ V <sub>RL</sub> , R <sub>RESET</sub> = 5.1 k	1	0.1	0.4	V
DELAY Switching Threshold (V <sub>DT</sub> ) (2.5 V, 3.3 V, and 5.0 V Options)	V <sub>IN</sub> = Minimum (Note 6) (Note 7)	1.4	1.8	2.2	V
DELAY Switching Threshold (V <sub>DT</sub> ) (Adjustable Option)	V <sub>IN</sub> = Minimum (Note 6) (Note 7)	1.0	1.3	1.6	V
DELAY Low Voltage	V <sub>IN</sub> = Minimum (Note 6) (Note 7) V <sub>OUT</sub> < RESET Threshold Low(min)	-	-	0.2	V
DELAY Charge Current	V <sub>IN</sub> = Minimum (Note 6) (Note 7) DELAY = 1.0 V, V <sub>OUT</sub> > V <sub>RH</sub>	2.5	4.0	5.5	μΑ
DELAY Discharge Current	V <sub>IN</sub> = Minimum (Note 6) (Note 7) DELAY = 1.0 V, V <sub>OUT</sub> < V <sub>RL</sub>	5.0	-	-	mA
Voltage Adjust (Adjustable Output only)					
Input Current	V <sub>ADJ</sub> = 1.25 V, V <sub>IN</sub> = Minimum (Note 6) (Note 7)	-0.5	-	0.5	μΑ
ENABLE				•	
Input Threshold Low, V <sub>IN</sub> = 14 V (Note 6) High, V <sub>IN</sub> = 14 V (Note 6)		- 2.0	- -	1.0	V V
Input Current	ENABLE = 5.0 V, V <sub>IN</sub> = 14 V (Note 6)	_	30	75	μΑ

Voltage range specified in the Output Stage of the Electrical Characteristics in boldface type.
 For V<sub>IN</sub> ≤ 4.5 V, a RESET = Low may occur with the output in regulation.
 Part is guaranteed by design to meet specification over the entire V<sub>IN</sub> voltage range, but is production tested only at the specified V<sub>IN</sub> voltage.
 Minimum V<sub>IN</sub> = 4.5 V for 2.5 V, 3.3 V, and Adjustable options. Minimum V<sub>IN</sub> = 6.0 V for 5.0 V option.

#### PACKAGE PIN DESCRIPTION, ADJUSTABLE OUTPUT

Pin Number	Pin Symbol	Function	
1	DELAY	Timing capacitor for RESET function.	
2	ENABLE	ENABLE control for the IC. A high powers the device up.	
3	RESET	Active reset (accurate to V <sub>OUT</sub> ≥ 1.0 V)	
4	GND	Ground. All GND leads must be connected to Ground	
5	V <sub>ADJ</sub>	Voltage Adjust. A resistor divider from V <sub>OUT</sub> to this lead sets the output voltage.	
6	V <sub>OUT</sub>	±2.0%, 400 mA output.	
7	V <sub>IN</sub>	Input Voltage.	

### PACKAGE PIN DESCRIPTION, FIXED OUTPUT

Pin Number	Pin Symbol	Function	
1	DELAY	Timing capacitor for RESET function.	
2	ENABLE	ENABLE control for the IC. A high powers the device up.	
3	RESET	Active reset (accurate to V <sub>OUT</sub> ≥ 1.0 V)	
4	GND	Ground. All GND leads must be connected to Ground	
5	SENSE	Kelvin connection which allows remote sensing of output voltage for improved regulation. If remote sensing is not desired, connect to V <sub>OUT</sub> .	
6	V <sub>OUT</sub>	±2.0%, 400 mA output.	
7	V <sub>IN</sub>	Input Voltage.	

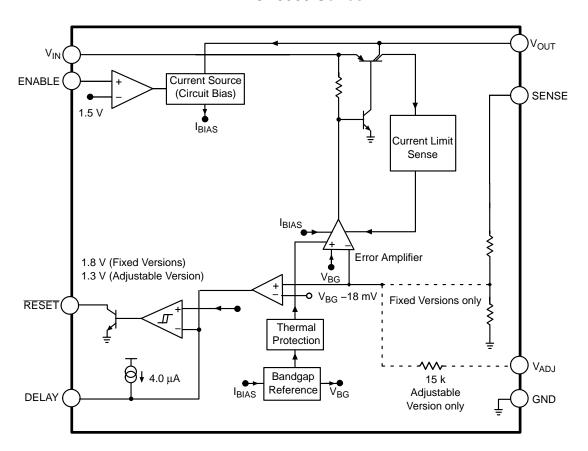


Figure 2. Block Diagram

#### TYPICAL PERFORMANCE CHARACTERISTICS

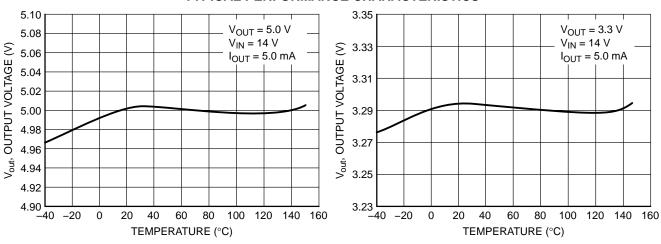


Figure 3. 5 V Output Voltage vs Temperature

Figure 4. 3.3 V Output Voltage vs Temperature

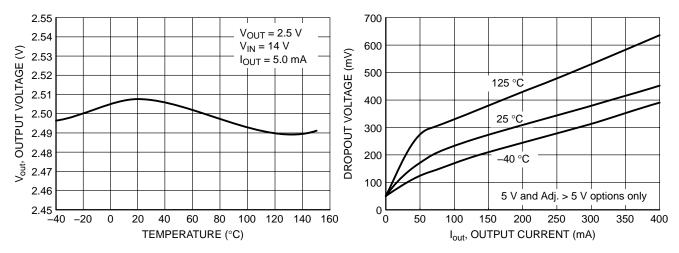


Figure 5. 2.5 V Output Voltage vs Temperature

Figure 6. Dropout Voltage vs Output Current

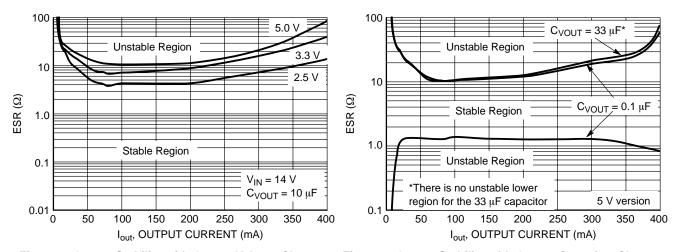
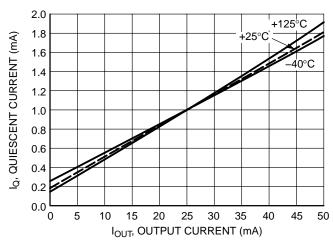


Figure 7. Output Stability with Output Voltage Change

Figure 8. Output Stability with Output Capacitor Change

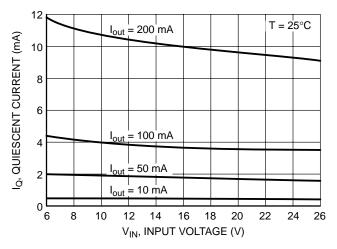
#### TYPICAL PERFORMANCE CHARACTERISTICS



60 (YE) 50 30 30 10 0 50 100 150 200 250 300 350 400 450 500 HOUT, OUTPUT CURRENT (mA)

Figure 9. Quiescent Current vs Output Current

Figure 10. Quiescent Current vs Output Current



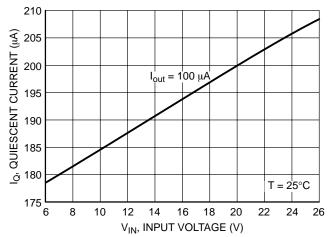


Figure 11. Quiescent Current vs Input Voltage

Figure 12. Quiescent Current vs Input Voltage

#### CIRCUIT DESCRIPTION

#### **REGULATOR CONTROL FUNCTIONS**

The NCV8505 contains the microprocessor compatible control function  $\overline{RESET}$  (Figure 13).

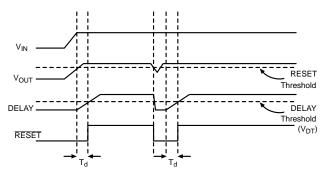


Figure 13. Reset and Delay Circuit Wave Forms

#### **RESET** Function

A  $\overline{RESET}$  signal (low voltage) is generated as the IC powers up until  $V_{OUT}$  is within 1.5% of the regulated output voltage, or when  $V_{OUT}$  drops out of regulation,and is lower than 4.0% below the regulated output voltage. Hysteresis is included in the function to minimize oscillations.

The  $\overline{RESET}$  output is an open collector NPN transistor, controlled by a low voltage detection circuit. The circuit is functionally independent of the rest of the IC thereby guaranteeing that the  $\overline{RESET}$  signal is valid for  $V_{OUT}$  as low as 1.0 V.

#### **ENABLE Function**

The part stays in a low  $I_Q$  sleep mode when the ENABLE pin is held low. The part has an internal pull down if the pin is left floating.

The integrity of the ENABLE pin allows it to be tied to the battery line through an external resistor. It will withstand load dump potentials in this configuration.

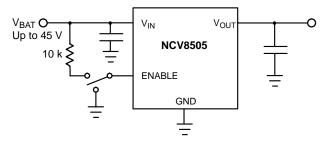


Figure 14. ENABLE Function

#### **DELAY Function**

The reset delay circuit provides a programmable (by external capacitor) delay on the  $\overline{RESET}$  output lead.

The DELAY lead provides source current (typically  $4.0~\mu A$ ) to the external DELAY capacitor during the following proceedings:

- 1. During Power Up (once the regulation threshold has been verified).
- 2. After a reset event has occurred and the device is back in regulation. The DELAY capacitor is discharged when the regulation (RESET threshold) has been violated. This is a latched incident. The capacitor will fully discharge and wait for the device to regulate before going through the delay time event again.

#### **Voltage Adjust**

Figure 15 shows the device setup for a user configurable output voltage. The feedback to the  $V_{\rm ADJ}$  pin is taken from a voltage divider referenced to the output voltage. The loop is balanced around the Unity Gain threshold (1.30 V typical).

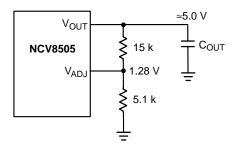


Figure 15. Adjustable Output Voltage

#### **APPLICATION NOTES**

#### **SETTING THE DELAY TIME**

The delay time is controlled by the Reset Delay Low Voltage, Delay Switching Threshold, and the Delay Charge Current. The delay follows the equation:

$$t_{DELAY} = \frac{[C_{DELAY}(V_{dt} - Reset \ Delay \ Low \ Voltage)]}{Delay \ Charge \ Current}$$

Example:

Using  $C_{DELAY} = 33 \text{ nF}$ .

Assume reset Delay Low Voltage = 0.

Use the typical value for  $V_{dt} = 1.8 \text{ V}$  (2.5 V, 3.3 V, and 5.0 V options).

Use the typical value for Delay Charge Current =  $4.2 \,\mu\text{A}$ .

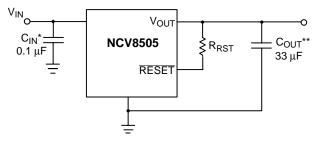
$$t_{DELAY} = \frac{[33 \text{ nF}(1.8 - 0)]}{4.2 \,\mu\text{A}} = 14 \text{ ms}$$

#### STABILITY CONSIDERATIONS

The output or compensation capacitor helps determine three main characteristics of a linear regulator: start-up delay, load transient response and loop stability.

The capacitor value and type should be based on cost, availability, size and temperature constraints. A tantalum or aluminum electrolytic capacitor is best, since a film or ceramic capacitor with almost zero ESR can cause instability. The aluminum electrolytic capacitor is the least expensive solution, but, if the circuit operates at low temperatures (-25°C to -40°C), both the value and ESR of the capacitor will vary considerably. The capacitor manufacturers data sheet usually provides this information.

The value for the output capacitor  $C_{OUT}$  shown in Figure 16 should work for most applications, however it is not necessarily the optimized solution.



\*C<sub>IN</sub> required if regulator is located far from the power supply filter

Figure 16. Test and Application Circuit Showing
Output Compensation

## CALCULATING POWER DISSIPATION IN A SINGLE OUTPUT LINEAR REGULATOR

The maximum power dissipation for a single output regulator (Figure 17) is:

$$P_{D(max)} = [V_{IN(max)} - V_{OUT(min)}]I_{OUT(max)}$$

$$+ V_{IN(max)}I_{Q}$$
(1)

where:

V<sub>IN(max)</sub> is the maximum input voltage,

V<sub>OUT(min)</sub> is the minimum output voltage,

 $I_{OUT(max)}$  is the maximum output current for the application, and

 $I_Q$  is the quiescent current the regulator consumes at  $I_{OUT(max)}$ .

Once the value of  $P_{D(max)}$  is known, the maximum permissible value of  $R_{\theta JA}$  can be calculated:

$$R_{\theta}JA = \frac{150^{\circ}C - T_{A}}{P_{D}}$$
 (2)

The value of  $R_{\theta JA}$  can then be compared with those in the package section of the data sheet. Those packages with  $R_{\theta JA}$ 's less than the calculated value in equation 2 will keep the die temperature below 150°C.

In some cases, none of the packages will be sufficient to dissipate the heat generated by the IC, and an external heatsink will be required.

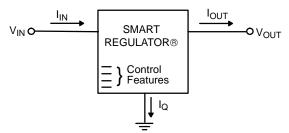


Figure 17. Single Output Regulator with Key Performance Parameters Labeled

#### **HEAT SINKS**

A heat sink effectively increases the surface area of the package to improve the flow of heat away from the IC and into the surrounding air.

Each material in the heat flow path between the IC and the outside environment will have a thermal resistance. Like series electrical resistances, these resistances are summed to determine the value of  $R_{\rm BIA}$ :

$$R_{\theta}JA = R_{\theta}JC + R_{\theta}CS + R_{\theta}SA \tag{3}$$

where:

 $R_{\theta JC}$  = the junction-to-case thermal resistance,

 $R_{\theta CS}$  = the case–to–heatsink thermal resistance, and

 $R_{\theta SA}$  = the heatsink-to-ambient thermal resistance.

 $R_{\theta JC}$  appears in the package section of the data sheet. Like  $R_{\theta JA}$ , it too is a function of package type.  $R_{\theta CS}$  and  $R_{\theta SA}$  are functions of the package type, heatsink and the interface between them. These values appear in heat sink data sheets of heat sink manufacturers.

<sup>\*\*</sup>C<sub>OUT</sub> required for stability. Capacitor must operate at minimum temperature expected.

#### **ORDERING INFORMATION**

Device	Output Voltage	Package	Shipping†
NCV8505D2TADJ	A Prostable		50 Units/Rail
NCV8505D2TADJR4	Adjustable		750 Tape & Reel
NCV8505D2T25	2.5 V		50 Units/Rail
NCV8505D2T25R4	2.5 V	D <sup>2</sup> PAK-7	750 Tape & Reel
NCV8505D2T33	227	D-PAK-1	50 Units/Rail
NCV8505D2T33R4	3.3 V		750 Tape & Reel
NCV8505D2T50	501/		50 Units/Rail
NCV8505D2T50R4	5.0 V		750 Tape & Reel

<sup>†</sup>For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

#### PACKAGE DIMENSIONS

D<sup>2</sup>PAK-7 DPS SUFFIX CASE 936AB-01 ISSUE O

# For D<sup>2</sup>PAK Outline and Dimensions – Contact Factory

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