

LMC6009MTX

Tape and Reel

www.national.com

Absolute Maximum Ratings (Note 1)

•

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

1.0 kV
$\begin{array}{l} \text{GND-0.3V} \leq \text{V}^{+} \leq \\ \text{V}_{\text{DD}} + 0.3 \text{V}_{\text{DC}} \end{array}$
-0.3 to +6.5 V _{DC}
-20°C to +75°C
–55°C to +150°C

 $\begin{array}{ll} \mbox{Maximum Junction Temperature} (T_J) & +150 \mbox{`C} \\ \mbox{Maximum Power Dissipation} (P_D) & 1.09 \mbox{W} \end{array}$

Operating Ratings (Note 1)

Supply Voltage Frequency Thermal Resistance (θ_{JA}) Derating

 $\begin{array}{l} \text{2.7V} \leq \text{V}_{\text{DD}} \leq \text{5.5V} \\ \text{DC-50 kHz} \end{array}$

8.70 mW/°C

3V DC Electrical Characteristics

Unless otherwise specified, all limits are guaranteed for $T_J = 25^{\circ}C$, and $V_{DD} = 3.0 V_{DC}$

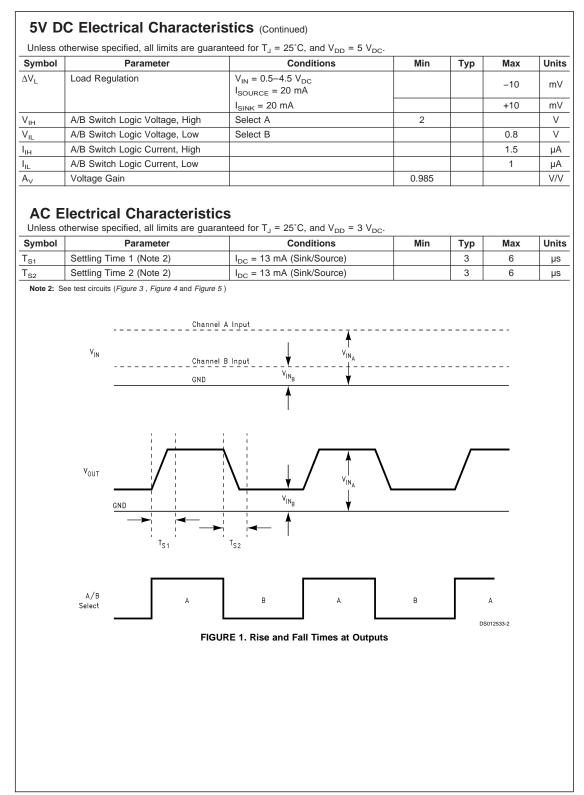
Symbol	Parameter	Conditions	Min	Тур	Max	Units
V _{DD}	Supply Voltage		2.7	3.0	3.3	V
Vos	Offset Voltage	R _S = 10k			20	mV
I _B	Input Bias Current				1500	nA
V _{OL}	Output Voltage, Low	Amp A8 and A9 I _{SINK} = 13 mA			GND + 0.2	V
		Amp A1–A7 I _{SINK} = 13 mA			GND + 0.6	V
V _{OH}	Output Voltage, High	Amp A1 and A2 I _{SOURCE} = 13 mA	V _{DD} -0.2			V
		Amp A3–A9 I _{SOURCE} = 13 mA	V _{DD} -0.6			V
I _{sc}	Output Short Circuit Current	V _{OUT} - 1.65V (Note 1)	80	150		mA
I _{DD}	Supply Current	No Load		3.5	5	mA
ΔV_L	Load Regulation	$V_{IN} = 0.3-3 V_{DC}$ $I_{SOURCE} = 13 \text{ mA}$			-10	mV
		I _{SINK} = 13 mA			+10	mV
V _{IH}	A/B Switch Logic Voltage, High	Select A	2			V
V _{IL}	A/B Switch Logic Voltage, Low	Select B			0.8	V
н	A/B Switch Logic Current, High				1.5	μΑ
IL	A/B Switch Logic Current, Low				1	μA
A _V	Voltage Gain		0.985			V/V

Note 1: See Test Circuit (Figure 2)

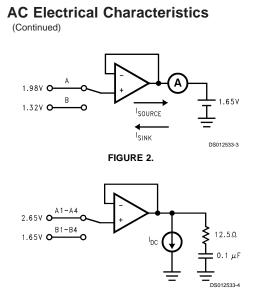
5V DC Electrical Characteristics

Unless otherwise specified, all limits are guaranteed for T_J = 25 $^\circ\text{C},$ and V_{DD} = 5 $V_{DC}.$

Symbol	Parameter	Conditions	Min	Тур	Max	Units
V _{DD}	Supply Voltage		4.5	5	5.5	V
Vos	Offset Voltage	R _s = 10k			20	mV
I _B	Input Bias Current				1500	nA
V _{OL}		Amp A8 and A9			GND +	V
		I _{SINK} = 20 mA			0.2	v
		Amp A1–A7			GND +	V
		I _{SINK} = 20 mA			1.0	v
V _{он}	H Output Voltage, High Amp A1 and A2 I _{SOURCE} = 20 mA	Amp A1 and A2	V _{DD} -0.2			V
		V _{DD} =0.2			v	
		Amp A3–A9	V _{DD} -1.0			V
		I _{SOURCE} = 20 mA				
I _{sc}	Output Short Circuit Current	V _{OUT} - 1.65V (Note 1)	120	200		mA
I _{DD}	Supply Current	No Load		4.5	6	mA



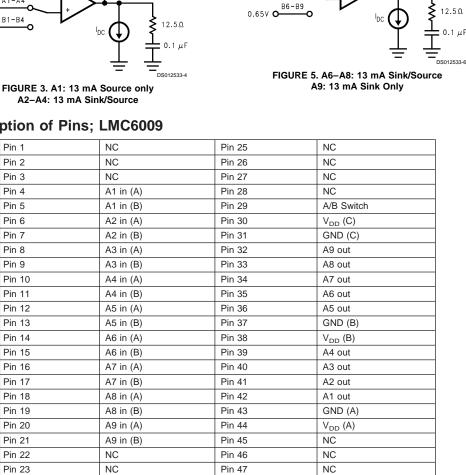
www.national.com



•

FIGURE 3. A1: 13 mA Source only

Description of Pins; LMC6009



Α5

Β5

A6-A9

0

FIGURE 4. 13 mA Sink/Source

7.5Ω

0.1 μl

DS012533-5

2.65V O

1.65V **O**-

1.65V O

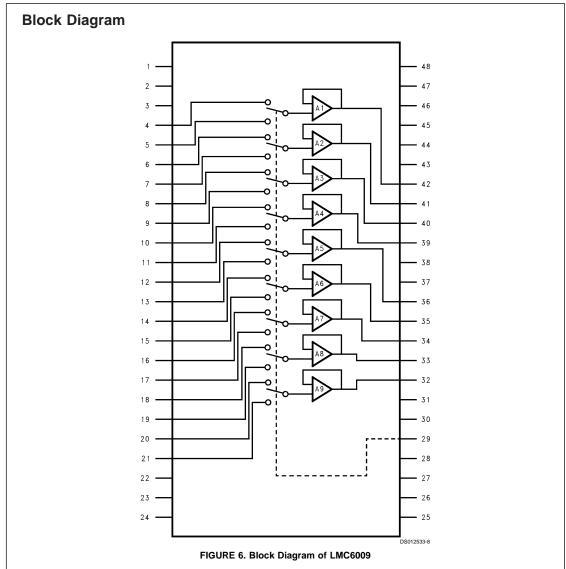
www.national.com

Pin 24

NC

Pin 48

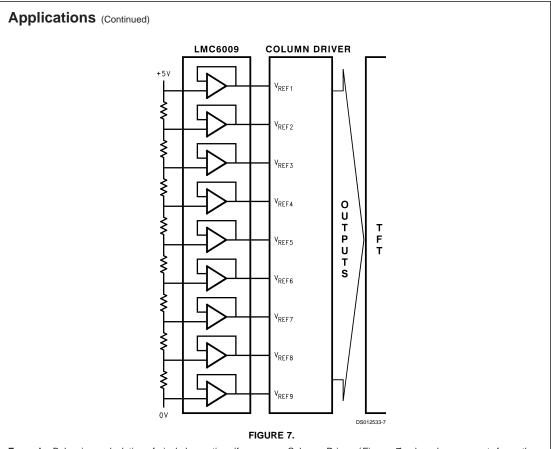
NC



Applications

The LMC6009 is useful for buffering the nine reference voltages for gamma correction in a TFT-LCD as shown in *Figure 7*. The A/B channel inputs allow the user to alternate two sets of gamma references to compensate for asymmetrical Gamma characteristic during Row Inversion. The LMC6009 eliminates the need for nine external switches or an 18-to-9 multiplexer.

Since the buffers in the LMC6009 draw extremely low bias current (1.5 μA max), large resistance values can be used in the reference voltage string. This allows the power dissipation in the gamma reference circuit to be minimized. The nine buffers are guaranteed to deliver 80 mA to the load, allowing the pixel voltages of the TFT-LCD to settle very quickly.



Example: Below is a calculation of pixel charge time (for a black to black transition) in a VGA display operating at a vertical refresh rate of 60 Hz, with a panel capacitance of 50 pF per sub-pixel:

A full black to black transition represents the maximum charging time for the panel, since it requires that the panel capacitance be driven by a 4V swing from node V_{REF1} (*Figure 7*).

Total capacitive load presented to the LMC6009 is

C_L = 50 pF x 3 x 640 = 96 nF

Output current of the LMC6009 is:

 I_{SC} = 80 mA

Hence, slew time $t_{SLEW} = (96 \text{ nF x 4V})/80 \text{ mA} = 3.07 \mu \text{s}$ The total line time for a VGA system is approximately 34 µs. Therefore, the LMC6009 easily meets the drive requirements for the application. The input resistance seen between the V_{REFn} and $V_{REF(n+1)}$ inputs, (where n = 0 thru 8) of the Column Driver (*Figure 7*) also draw current from the LMC6009. Thus, the actual current available for charging the panel capacitance is:

lpx = 80 mA - $(V_{VREF1} - V_{VREF2})/R_{CD}$

where

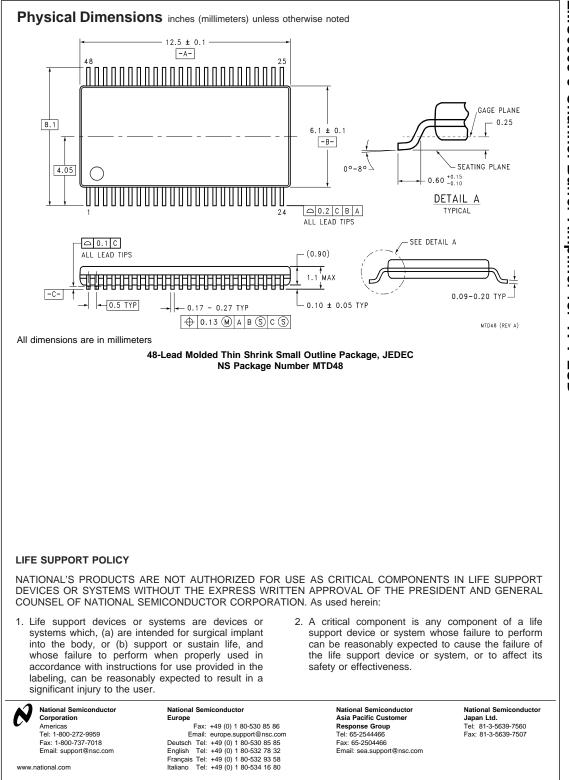
V_{V REFn} = Voltage at node V_{REFn},

 $V_{VREF(n+1)}$ = Voltage at node $V_{REF(n+1)}$, and R_{CD} = Column driver input resistance between

VREFn and VREF(n+1)

Since the LMC6009 is capable of sourcing 80 mA, the pixel charging time is primarily limited only by the length of the $R_{\rm CD}.$ C_L time constant. To implement a high quality display, column drivers that allow the shortest possible time constant (lower values of $R_{\rm CD}$) are desirable. However, lower values of $R_{\rm CD}$ result in increased system quiescent power dissipation. It is therefore important to optimize system performance by carefully considering speed vs power tradeoffs.

www.national.com



National does not assume any responsibility for use of any circuitry described, no circuit patent licenses are implied and National reserves the right at any time without notice to change said circuitry and specifications.