

## 6A, 600V, UFS Series N-Channel IGBT with Anti-Parallel Hyperfast Diodes

The HGTP3N60C3D, and HGT1S3N60C3DS are MOS gated high voltage switching devices combining the best features of MOSFETs and bipolar transistors. These devices have the high input impedance of a MOSFET and the low on-state conduction loss of a bipolar transistor. The much lower on-state voltage drop varies only moderately between 25°C and 150°C. The IGBT used is the development type TA49113. The diode used in anti-parallel with the IGBT is the development type TA49055.

The IGBT is ideal for many high voltage switching applications operating at moderate frequencies where low conduction losses are essential.

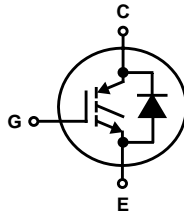
Formerly Developmental Type TA49119.

## Ordering Information

PART NUMBER	PACKAGE	BRAND
HGTP3N60C3D	TO-220AB	G3N60C3D
HGT1S3N60C3DS	TO-263AB	G3N60C3D

NOTE: When ordering, use the entire part number. Add the suffix 9A to obtain the TO-263AB variant in tape and reel, i.e., HGT1S3N60C3DS9A.

## Symbol

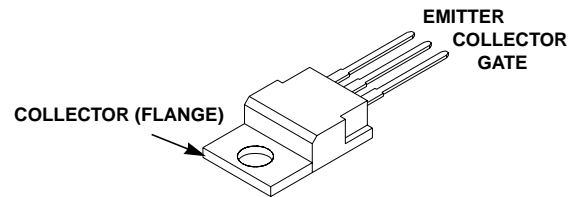


## Features

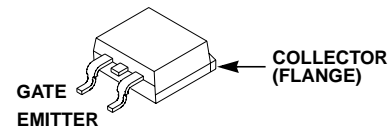
- 6A, 600V at  $T_C = 25^\circ\text{C}$
- 600V Switching SOA Capability
- Typical Fall Time . . . . . 130ns at  $T_J = 150^\circ\text{C}$
- Short Circuit Rating
- Low Conduction Loss
- Hyperfast Anti-Parallel Diode

## Packaging

**JEDEC TO-220AB**



**JEDEC TO-263AB**



### INTERSIL CORPORATION IGBT PRODUCT IS COVERED BY ONE OR MORE OF THE FOLLOWING U.S. PATENTS

4,364,073	4,417,385	4,430,792	4,443,931	4,466,176	4,516,143	4,532,534	4,587,713
4,598,461	4,605,948	4,620,211	4,631,564	4,639,754	4,639,762	4,641,162	4,644,637
4,682,195	4,684,413	4,694,313	4,717,679	4,743,952	4,783,690	4,794,432	4,801,986
4,803,533	4,809,045	4,809,047	4,810,665	4,823,176	4,837,606	4,860,080	4,883,767
4,888,627	4,890,143	4,901,127	4,904,609	4,933,740	4,963,951	4,969,027	

# HGTP3N60C3D, HGT1S3N60C3DS

## Absolute Maximum Ratings $T_C = 25^\circ\text{C}$ , Unless Otherwise Specified

	HGTP3N60C3D, HGT1S3N60C3DS	UNITS
Collector to Emitter Voltage	$BV_{CES}$ 600	V
Collector Current Continuous		
At $T_C = 25^\circ\text{C}$	$I_{C25}$ 6	A
At $T_C = 110^\circ\text{C}$	$I_{C110}$ 3	A
Collector Current Pulsed (Note 1)	$I_{CM}$ 24	A
Gate to Emitter Voltage Continuous	$V_{GES}$ $\pm 20$	V
Gate to Emitter Voltage Pulsed	$V_{GEM}$ $\pm 30$	V
Switching Safe Operating Area at $T_J = 150^\circ\text{C}$ (Figure 14)	SSOA 18A at 480V	
Power Dissipation Total at $T_C = 25^\circ\text{C}$	$P_D$ 33	W
Power Dissipation Derating $T_C > 25^\circ\text{C}$	0.27	W/ $^\circ\text{C}$
Operating and Storage Junction Temperature Range	$T_J, T_{STG}$ -40 to 150	$^\circ\text{C}$
Maximum Lead Temperature for Soldering	$T_L$ 260	$^\circ\text{C}$
Short Circuit Withstand Time (Note 2) at $V_{GE} = 10\text{V}$ (Figure 6)	$t_{SC}$ 8	$\mu\text{s}$

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

### NOTES:

1. Repetitive Rating: Pulse width limited by maximum junction temperature.
2.  $V_{CE(PK)} = 360\text{V}$ ,  $T_J = 125^\circ\text{C}$ ,  $R_G = 82\Omega$ .

## Electrical Specifications $T_C = 25^\circ\text{C}$ , Unless Otherwise Specified

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNITS
Collector to Emitter Breakdown Voltage	$BV_{CES}$	$I_C = 250\mu\text{A}$ , $V_{GE} = 0\text{V}$	600	-	-	V
Collector to Emitter Leakage Current	$I_{CES}$	$V_{CE} = BV_{CES}$ , $T_C = 25^\circ\text{C}$	-	-	250	$\mu\text{A}$
		$V_{CE} = BV_{CES}$ , $T_C = 150^\circ\text{C}$	-	-	2.0	mA
Collector to Emitter Saturation Voltage	$V_{CE(SAT)}$	$I_C = I_{C110}$ , $V_{GE} = 15\text{V}$ , $T_C = 25^\circ\text{C}$	-	1.65	2.0	V
		$T_C = 150^\circ\text{C}$	-	1.85	2.2	V
Gate to Emitter Threshold Voltage	$V_{GE(TH)}$	$I_C = 250\mu\text{A}$ , $V_{CE} = V_{GE}$ , $T_C = 25^\circ\text{C}$	3.0	5.5	6.0	V
Gate to Emitter Leakage Current	$I_{GES}$	$V_{GE} = \pm 25\text{V}$	-	-	$\pm 250$	nA
Switching SOA	SSOA	$T_J = 150^\circ\text{C}$ , $R_G = 82\Omega$ , $V_{GE} = 15\text{V}$ , $L = 1\text{mH}$ , $V_{CE(PK)} = 480\text{V}$	18	-	-	A
		$V_{CE(PK)} = 600\text{V}$	2	-	-	A
Gate to Emitter Plateau Voltage	$V_{GEP}$	$I_C = I_{C110}$ , $V_{CE} = 0.5 BV_{CES}$	-	8.3	-	V
On-State Gate Charge	$Q_{G(ON)}$	$I_C = I_{C110}$ , $V_{CE} = 0.5 BV_{CES}$ , $V_{GE} = 15\text{V}$	-	10.8	13.5	nC
		$V_{GE} = 20\text{V}$	-	13.8	17.3	nC
Current Turn-On Delay Time	$t_{d(ON)I}$	$T_J = 150^\circ\text{C}$	-	5	-	ns
Current Rise Time	$t_{rI}$	$I_{CE} = I_{C110}$	-	10	-	ns
Current Turn-Off Delay Time	$t_{d(OFF)I}$	$V_{CE(PK)} = 0.8 BV_{CES}$	-	325	400	ns
Current Fall Time	$t_{fI}$	$V_{GE} = 15\text{V}$	-	130	275	ns
Turn-On Energy	$E_{ON}$	$R_G = 82\Omega$ , $L = 1\text{mH}$	-	85	-	$\mu\text{J}$
Turn-Off Energy (Note 3)	$E_{OFF}$		-	245	-	$\mu\text{J}$
Diode Forward Voltage	$V_{EC}$	$I_{EC} = 3\text{A}$	-	2.0	2.5	V
Diode Reverse Recovery Time	$t_{RR}$	$I_{EC} = 3\text{A}$ , $dI_{EC}/dt = 200\text{A}/\mu\text{s}$	-	22	28	ns
		$I_{EC} = 1\text{A}$ , $dI_{EC}/dt = 200\text{A}/\mu\text{s}$	-	17	22	ns
Thermal Resistance	$R_{\theta JC}$	IGBT	-	-	3.75	$^\circ\text{C}/\text{W}$
		Diode	-	-	3.0	$^\circ\text{C}/\text{W}$

### NOTE:

3. Turn-Off Energy Loss ( $E_{OFF}$ ) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero ( $I_{CE} = 0\text{A}$ ). The HGTP3N60C3D and HGT1S3N60C3DS were tested per JEDEC standard No. 24-1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss. Turn-On losses include diode losses.

## Typical Performance Curves

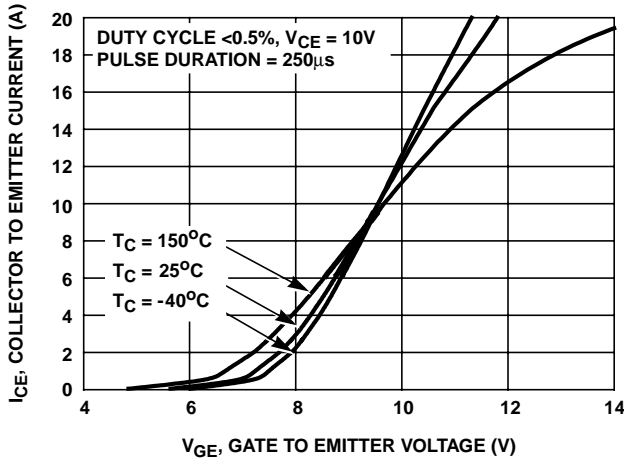


FIGURE 1. TRANSFER CHARACTERISTICS

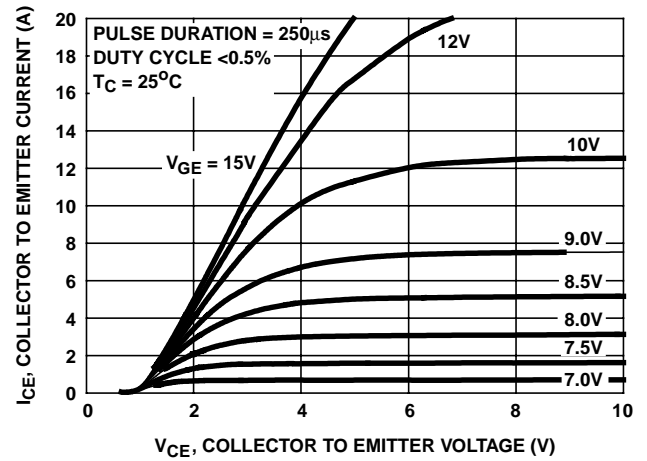


FIGURE 2. SATURATION CHARACTERISTICS

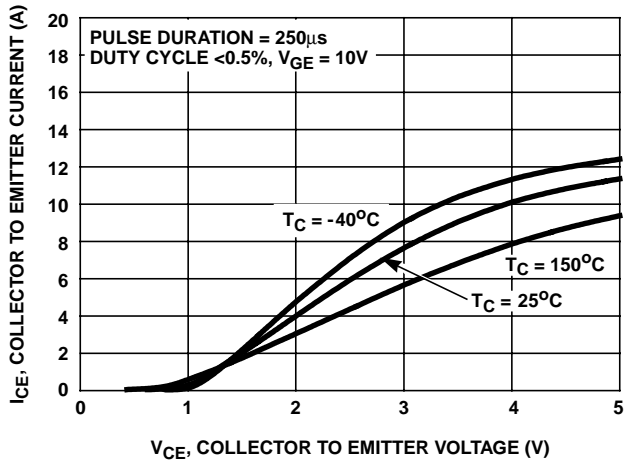


FIGURE 3. COLLECTOR TO EMITTER ON-STATE VOLTAGE

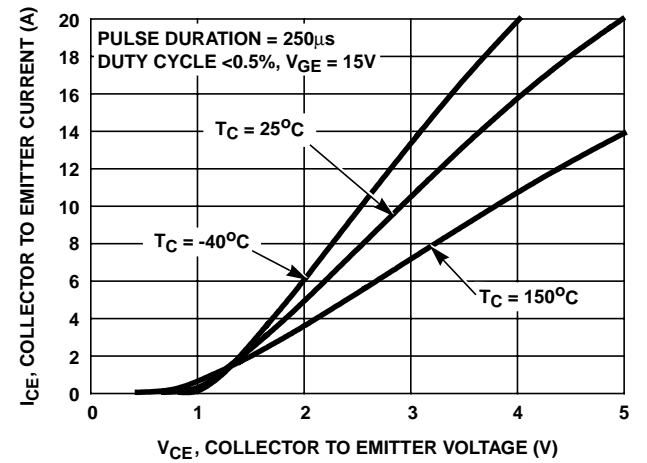


FIGURE 4. COLLECTOR TO EMITTER ON-STATE VOLTAGE

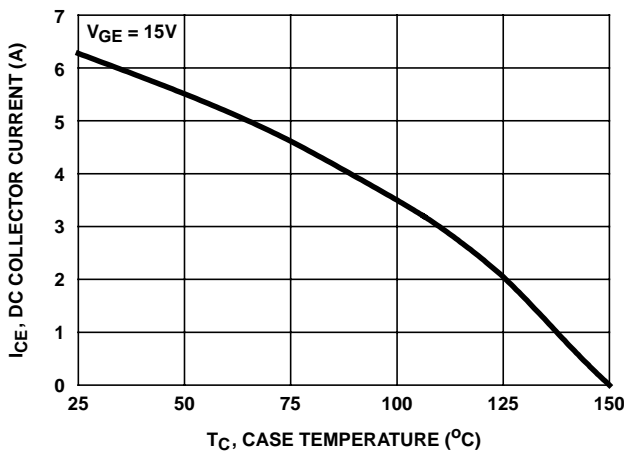


FIGURE 5. MAXIMUM DC COLLECTOR CURRENT vs CASE TEMPERATURE

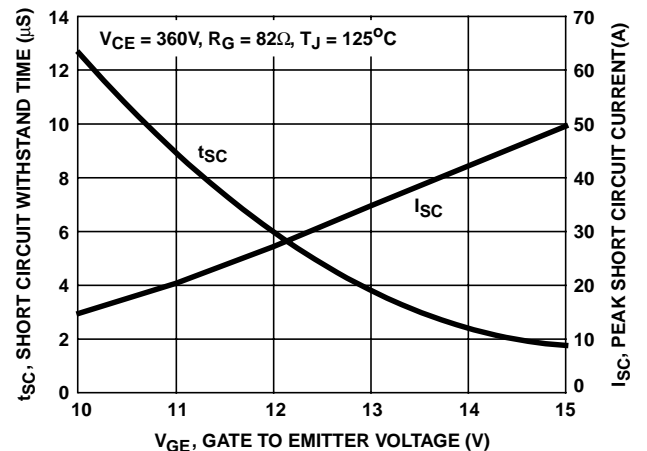


FIGURE 6. SHORT CIRCUIT WITHSTAND TIME

Typical Performance Curves (Continued)

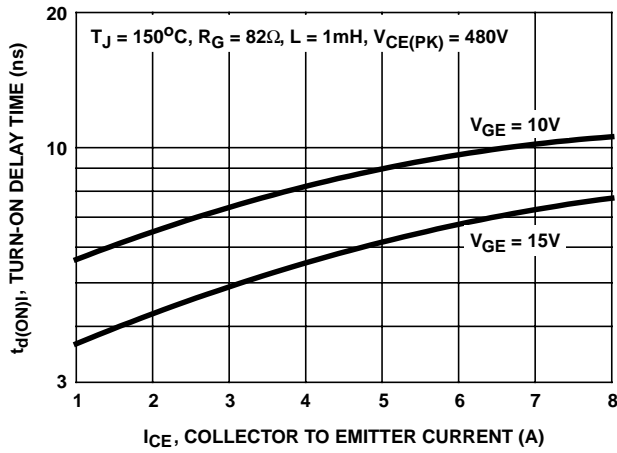


FIGURE 7. TURN-ON DELAY TIME vs COLLECTOR TO EMITTER CURRENT

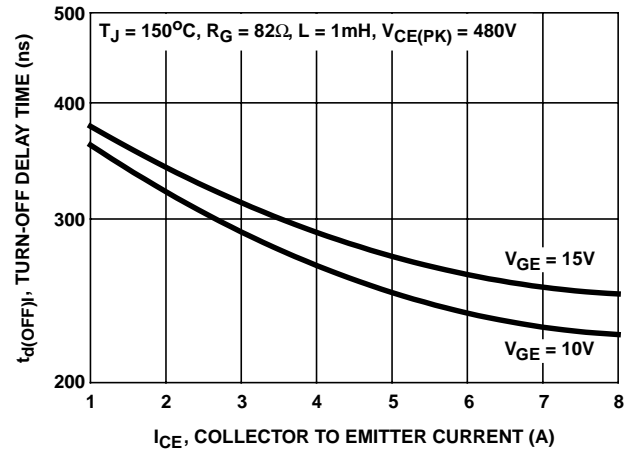


FIGURE 8. TURN-OFF DELAY TIME vs COLLECTOR TO EMITTER CURRENT

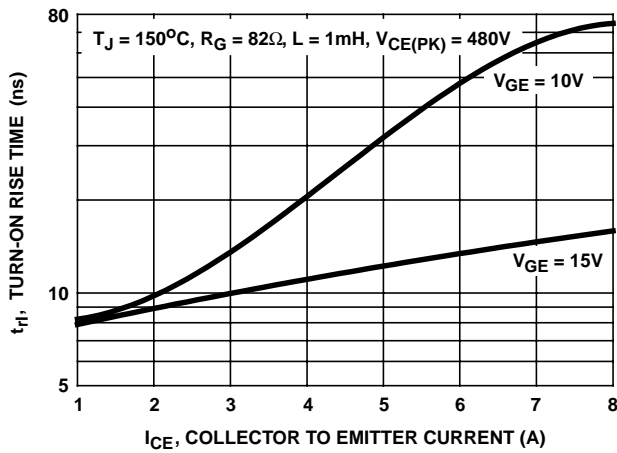


FIGURE 9. TURN-ON RISE TIME vs COLLECTOR TO EMITTER CURRENT

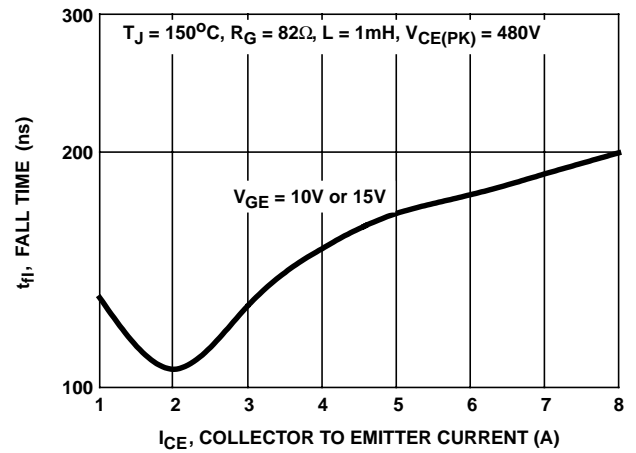


FIGURE 10. TURN-OFF FALL TIME vs COLLECTOR TO EMITTER CURRENT

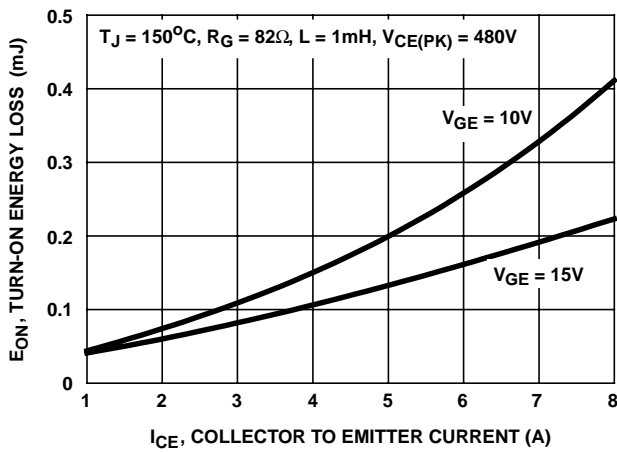


FIGURE 11. TURN-ON ENERGY LOSS vs COLLECTOR TO EMITTER CURRENT

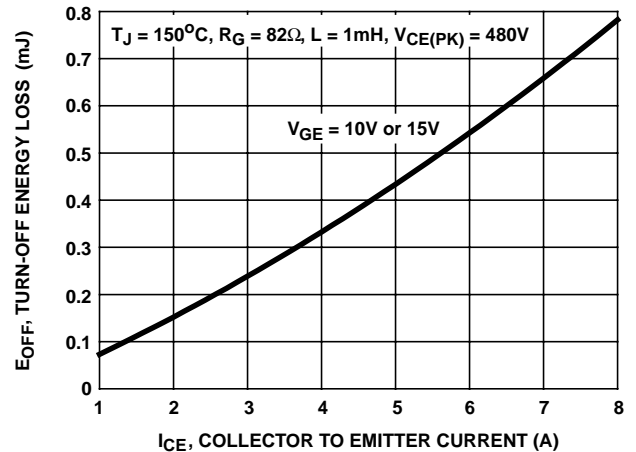


FIGURE 12. TURN-OFF ENERGY LOSS vs COLLECTOR TO EMITTER CURRENT

Typical Performance Curves (Continued)

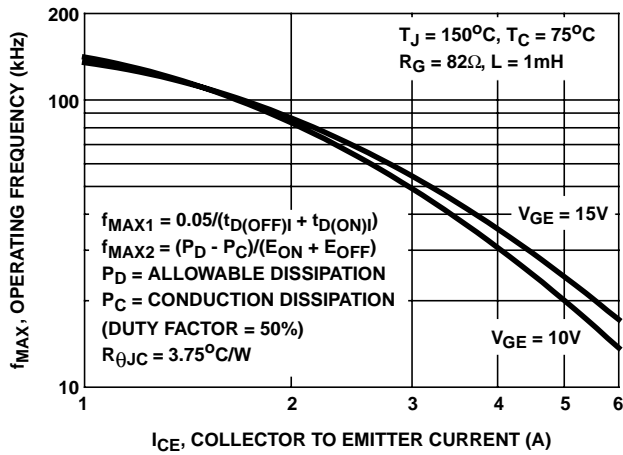


FIGURE 13. OPERATING FREQUENCY vs COLLECTOR TO EMITTER CURRENT

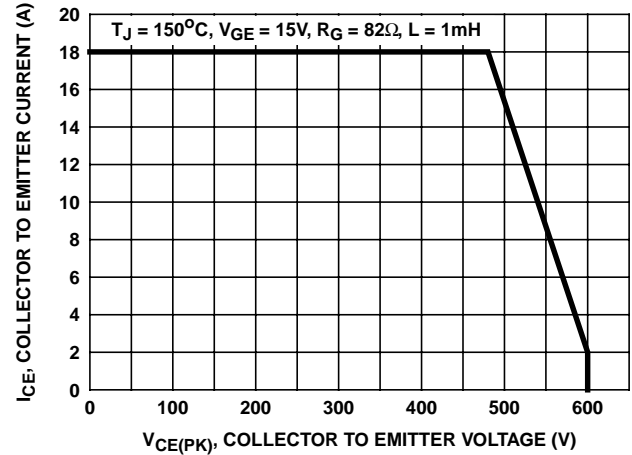


FIGURE 14. MINIMUM SWITCHING SAFE OPERATING AREA

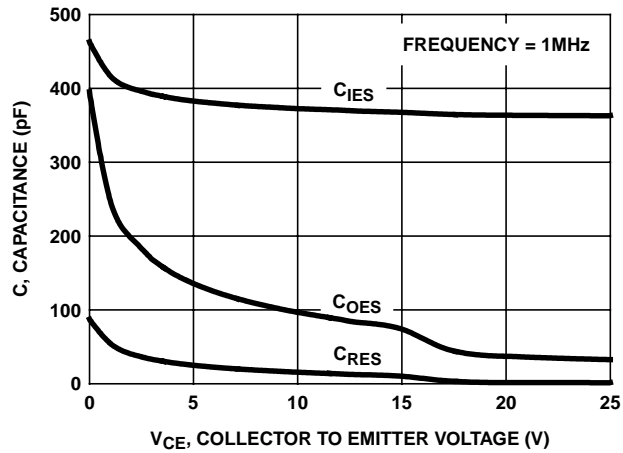


FIGURE 15. CAPACITANCE vs COLLECTOR TO EMITTER VOLTAGE

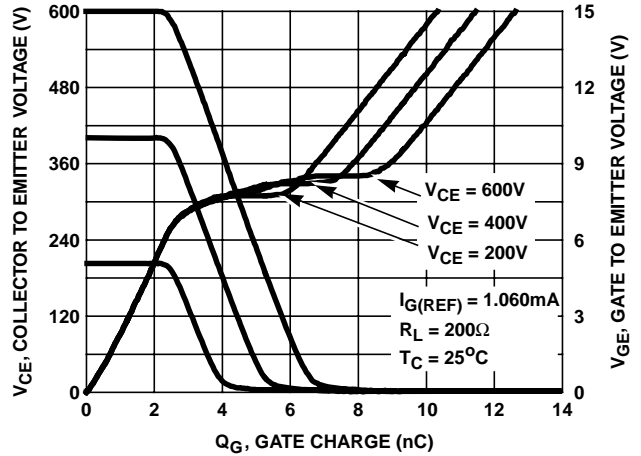


FIGURE 16. GATE CHARGE WAVEFORMS

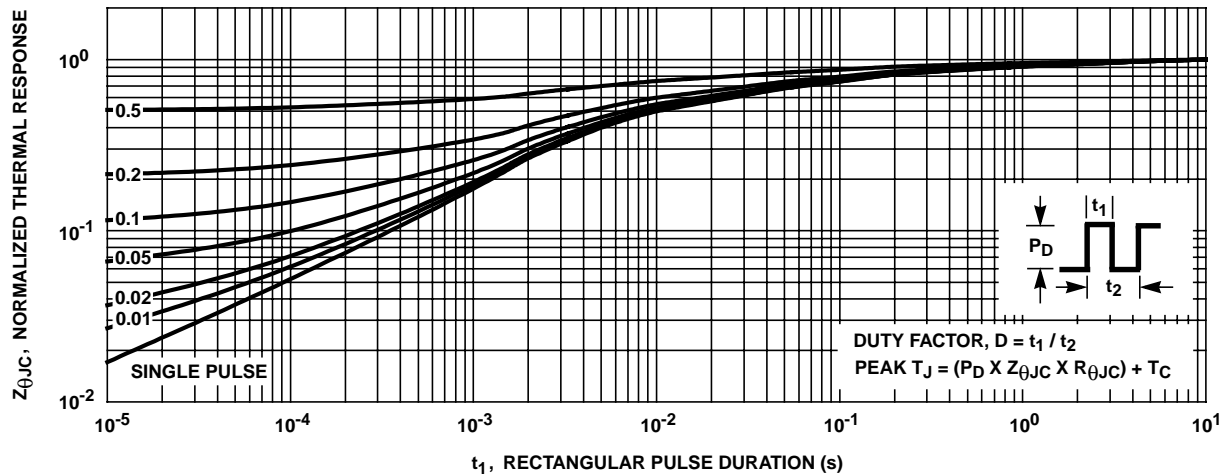


FIGURE 17. IGBT NORMALIZED TRANSIENT THERMAL IMPEDANCE, JUNCTION TO CASE

## Typical Performance Curves (Continued)

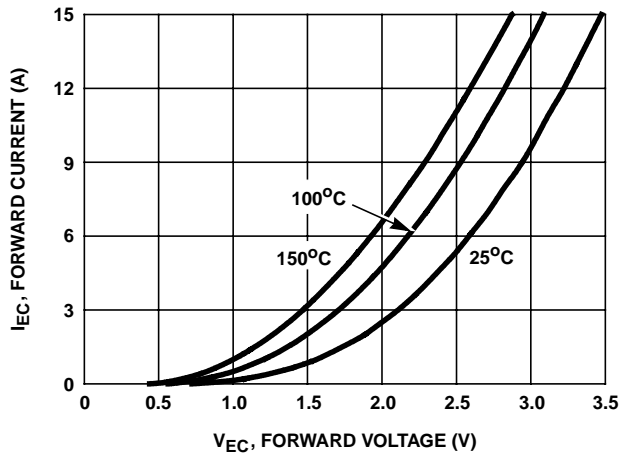


FIGURE 18. DIODE FORWARD CURRENT vs FORWARD VOLTAGE DROP

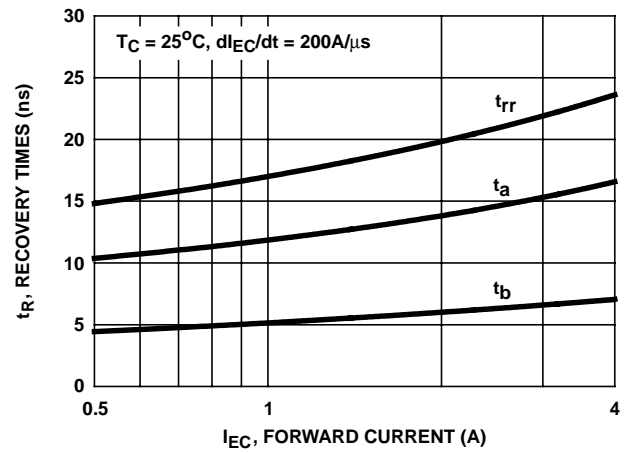


FIGURE 19. RECOVERY TIMES vs FORWARD CURRENT

## Test Circuit and Waveforms

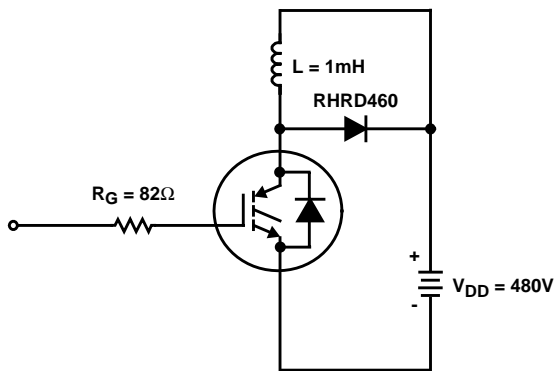


FIGURE 20. INDUCTIVE SWITCHING TEST CIRCUIT

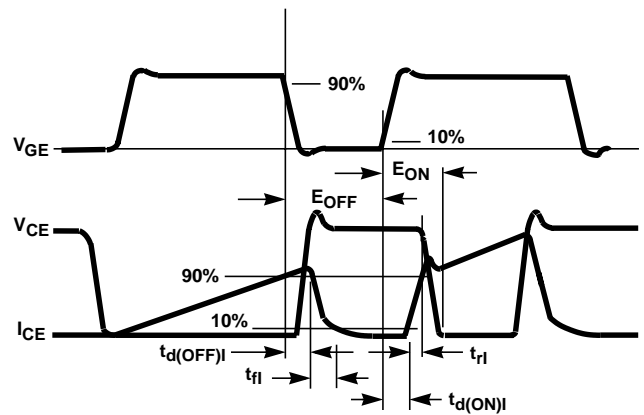


FIGURE 21. SWITCHING TEST WAVEFORMS

## Handling Precautions for IGBTs

Insulated Gate Bipolar Transistors are susceptible to gate-insulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

1. Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as ECCOSORBD™ LD26 or equivalent.
2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means - for example, with a metallic wristband.
3. Tips of soldering irons should be grounded.
4. Devices should never be inserted into or removed from circuits with power on.
5. **Gate Voltage Rating** - Never exceed the gate-voltage rating of  $V_{GEM}$ . Exceeding the rated  $V_{GE}$  can result in permanent damage to the oxide layer in the gate region.
6. **Gate Termination** - The gates of these devices are essentially capacitors. Circuits that leave the gate open-circuited or floating should be avoided. These conditions can result in turn-on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.
7. **Gate Protection** - These devices do not have an internal monolithic Zener diode from gate to emitter. If gate protection is required an external Zener is recommended.

## Operating Frequency Information

Operating frequency information for a typical device (Figure 13) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current ( $I_{CE}$ ) plots are possible using the information shown for a typical unit in Figures 4, 7, 8, 11 and 12. The operating frequency plot (Figure 13) of a typical device shows  $f_{MAX1}$  or  $f_{MAX2}$ ; whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

$f_{MAX1}$  is defined by  $f_{MAX1} = 0.05 / (t_{d(OFF)I} + t_{d(ON)I})$ . Deadtime (the denominator) has been arbitrarily held to 10% of the on-state time for a 50% duty factor. Other definitions are possible.  $t_{d(OFF)I}$  and  $t_{d(ON)I}$  are defined in Figure 20. Device turn-off delay can establish an additional frequency limiting condition for an application other than  $T_{JM}$ .  $t_{d(OFF)I}$  is important when controlling output ripple under a lightly loaded condition.

$f_{MAX2}$  is defined by  $f_{MAX2} = (P_D - P_C) / (E_{OFF} + E_{ON})$ . The allowable dissipation ( $P_D$ ) is defined by  $P_D = (T_{JM} - T_C) / R_{\theta JC}$ . The sum of device switching and conduction losses must not exceed  $P_D$ . A 50% duty factor was used (Figure 13) and the conduction losses ( $P_C$ ) are approximated by  $P_C = (V_{CE} \times I_{CE}) / 2$ .

$E_{ON}$  and  $E_{OFF}$  are defined in the switching waveforms shown in Figure 20.  $E_{ON}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-on and  $E_{OFF}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-off. All tail losses are included in the calculation for  $E_{OFF}$ ; i.e., the collector current equals zero ( $I_{CE} = 0$ ).

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