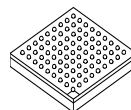


## MC9328MX21



**Package Information**

(MAPBGA-289)

## MC9328MX21

**Ordering Information: See Table 1 on page 4**

### 1 Introduction

Freescale's i.MX family of microprocessors has demonstrated leadership in the portable handheld market. Building on the success of the MX (Media Extensions) series, the i.MX21 (MC9328MX21) provides a leap in performance with an ARM926EJ-S™ microprocessor core that provides native security and accelerated Java support in addition to highly integrated system functions. The i.MX products specifically address the needs of the smartphone and portable product markets with their intelligent integrated peripherals, advanced processor core, and power management capabilities.

The i.MX21 features the advanced and power-efficient ARM926EJ-S core operating at speeds up to 266 MHz and is part of a growing family of *Smart Speed* products that offer high performance processing optimized for lowest power consumption. On-chip modules such as a video accelerator module, LCD controller, USB On-The-Go, CMOS sensor interface, and two synchronous serial interfaces offer designers a rich suite of peripherals that can enhance any product seeking to provide a rich

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## Introduction

multimedia experience. In addition, the i.MX21 provides optional hardware enabled security features including high assurance boot mode, unique processor IDs, secret key support, secure RAM, and a security monitor. These optional features enable secure e-commerce, digital rights management (DRM), information encryption, and secure software downloads.

For cost sensitive applications, the NAND Flash controller allows the use of low-cost NAND Flash devices to be used as primary or secondary non-volatile storage. The on-chip error correction code (ECC) and parity checking circuitry of the NAND Flash controller frees the CPU for other tasks. WLAN, Bluetooth and expansion options are provided through PCMCIA/CF, USB, and MMC/SD host controllers.

The i.MX21 is packaged in a 289-pin MAPBGA.

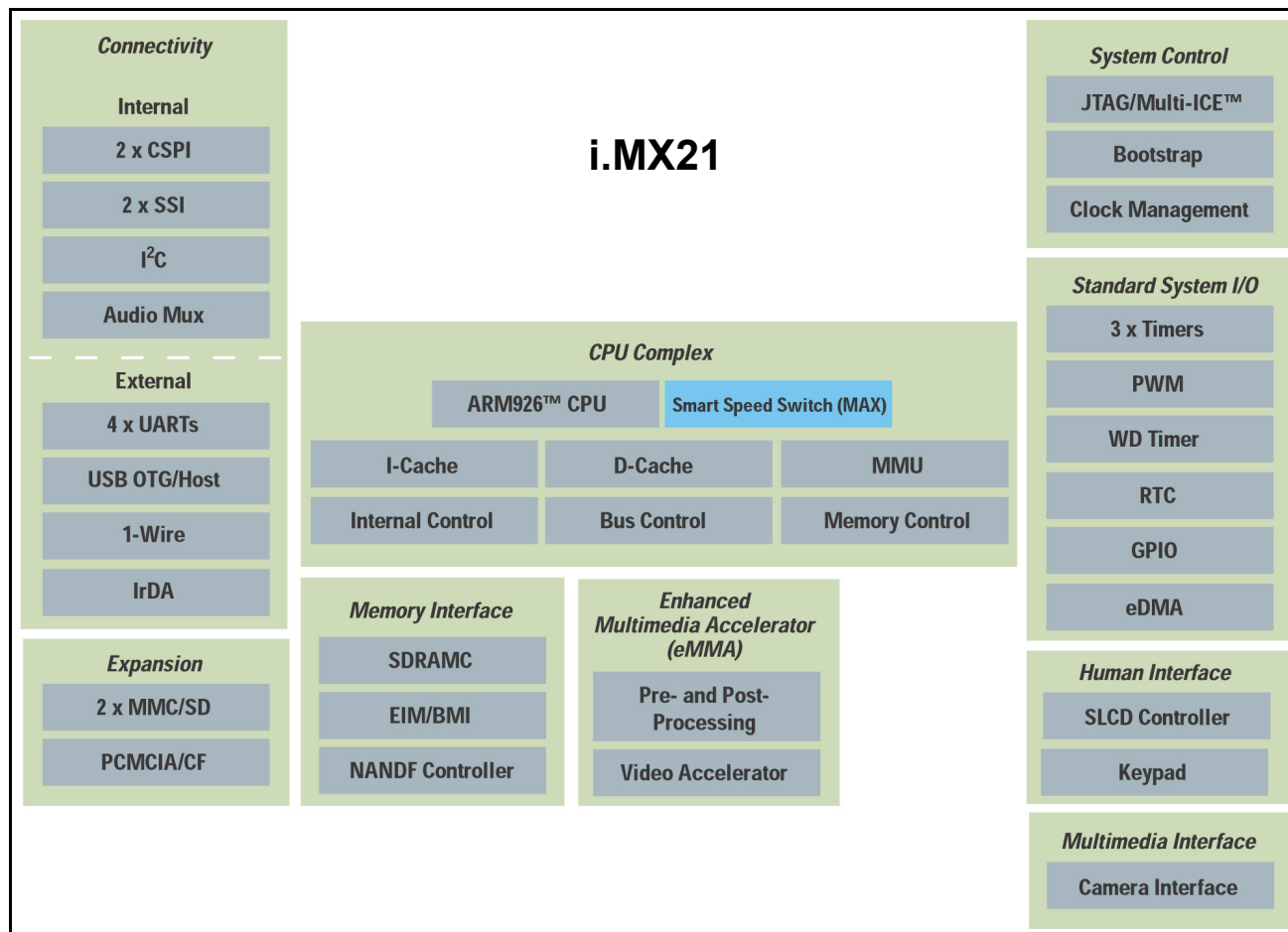


Figure 1. i.MX21 Functional Block Diagram

## 1.1 Conventions

This document uses the following conventions:

- $\overline{\text{OVERBAR}}$  is used to indicate a signal that is active when pulled low: for example,  $\overline{\text{RESET}}$ .
- *Logic level one* is a voltage that corresponds to Boolean true (1) state.
- *Logic level zero* is a voltage that corresponds to Boolean false (0) state.
- To *set* a bit or bits means to establish logic level one.
- To *clear* a bit or bits means to establish logic level zero.
- A *signal* is an electronic construct whose state conveys or changes in state convey information.
- A *pin* is an external physical connection. The same pin can be used to connect a number of signals.
- *Asserted* means that a discrete signal is in active logic state.
  - *Active low* signals change from logic level one to logic level zero.
  - *Active high* signals change from logic level zero to logic level one.
- *Negated* means that an asserted discrete signal changes logic state.
  - *Active low* signals change from logic level zero to logic level one.
  - *Active high* signals change from logic level one to logic level zero.
- LSB means *least significant bit* or *bits*, and MSB means *most significant bit* or *bits*. References to low and high bytes or words are spelled out.
- Numbers preceded by a percent sign (%) are binary. Numbers preceded by a dollar sign (\$) or 0x are hexadecimal.

## 1.2 Target Applications

The i.MX21 is targeted for advanced information appliances, smart phones, Web browsers, digital MP3 audio players, handheld computers based on the popular Palm OS platform, and messaging applications.

## 1.3 Reference Documentation

The following documents are required for a complete description of the i.MX21 and are necessary to design properly with the device. Especially for those not familiar with the ARM926EJ-S processor or previous DragonBall products, the following documents are helpful when used in conjunction with this manual.

*ARM Architecture Reference Manual* (ARM Ltd., order number ARM DDI 0100)

*ARM7TDMI Data Sheet* (ARM Ltd., order number ARM DDI 0029)

*ARM920T Technical Reference Manual* (ARM Ltd., order number ARM DDI 0151C)

*MC9328MX21 Product Brief* (order number MC9328MX21P/D)

*MC9328MX21 Reference Manual* (order number MC9328MX21RM/D)

*MC9328MX1 Product Brief* (order number MC9328MX1P/D)

*MC9328MX1 Data Sheet* (order number MC9328MX1/D)

*MC9328MX1 Reference Manual* (order number MC9328MX1RM/D)

The Freescale manuals are available on the Freescale Semiconductor Web site at <http://www.freescale.com>. These documents may be downloaded directly from the Freescale Web site, or printed versions may be ordered. The ARM Ltd. documentation is available from <http://www.arm.com>.

## 1.4 Ordering Information

Table 1 provides ordering information for the i.MX21.

**Table 1. i.MX21 Ordering Information**

Marking	Package Size	Package Type	Operating range
MC9328MX21VG	289-lead MAPBGA 0.65mm, 14mm x 14mm	Lead	0°C–70°C
MC9328MX21VK	289-lead MAPBGA 0.65mm, 14mm x 14mm	Lead-free	0°C–70°C
MC9328MX21VH	289-lead MAPBGA 0.8mm, 17mm x 17mm	Lead	0°C–70°C
MC9328MX21VM	289-lead MAPBGA 0.8mm, 17mm x 17mm	Lead-free	0°C–70°C
MC9328MX21DVG	289-lead MAPBGA 0.65mm, 14mm x 14mm	Lead	-30°C–70°C
MC9328MX21DVK	289-lead MAPBGA 0.65mm, 14mm x 14mm	Lead-free	-30°C–70°C
MC9328MX21DVH	289-lead MAPBGA 0.8mm, 17mm x 17mm	Lead	-30°C–70°C
MC9328MX21DVM	289-lead MAPBGA 0.8mm, 17mm x 17mm	Lead-free	-30°C–70°C
MC9328MX21CVG	289-lead MAPBGA 0.65mm, 14mm x 14mm	Lead	-40°C–85°C
MC9328MX21CVK	289-lead MAPBGA 0.65mm, 14mm x 14mm	Lead-free	-40°C–85°C
MC9328MX21CVH	289-lead MAPBGA 0.8mm, 17mm x 17mm	Lead	-40°C–85°C
MC9328MX21CVM	289-lead MAPBGA 0.8mm, 17mm x 17mm	Lead-free	-40°C–85°C

## 1.5 Features

The i.MX21 boasts a robust array of features that can support a wide variety of applications. Below is a brief description of i.MX21 features.

- ARM926EJ-S Core Complex
- enhanced Multimedia Accelerator (eMMA)
- Optional Security System
- Display and Video Modules
  - LCD Controller (LCDC)
  - Smart LCD Controller (SLCDC)
  - CMOS Sensor Interface (CSI)
- Bus Master Interface (BMI)

- Wireless Connectivity
  - Fast Infra-Red Interface (Fast IR)
- Wired Connectivity
  - USB On-The-Go (USBOTG) Controller
  - Four Universal Asynchronous Receiver/Transmitters (UART1, UART2, UART3, and UART4)
  - Two Configurable Serial Peripheral Interfaces (CSPI1 and CSPI2) for High Speed Data Transfer
  - Inter-IC (I<sup>2</sup>C) Bus Module
  - Two Synchronous Serial Interfaces (SSI) with Inter-IC Sound (I<sup>2</sup>S)
  - Digital Audio Mux
  - One-Wire Controller
  - Keypad Interface
- Memory Expansion and I/O Card Support
  - Two Multimedia Card and Secure Digital (MMC/SD) Host Controller Modules
- Memory Interface
  - External Interface Module (EIM)
  - SDRAM Controller (SDRAMC)
  - NAND Flash Controller (NFC)
  - PCMCIA/CF Interface
- Standard System Resources
  - Clock Generation Module (CGM) and Power Control Module
  - Three General-Purpose 32-Bit Counters/Timers
  - Watchdog Timer
  - Real-Time Clock/Sampling Timer (RTC)
  - Pulse-Width Modulator (PWM) Module
  - Direct Memory Access Controller (DMAC)
  - General-Purpose I/O (GPIO) Ports
  - Debug Capability

## 2 Signal Descriptions

This section identifies and describes the i.MX21 signals and their pin assignments. The i.MX21 signals are listed in Table 2.

**Table 2. i.MX21 Signal Descriptions**

Signal Name	Function/Notes
<b>External Bus/Chip Select (EIM)</b>	
A [25:0]	Address bus signals
D [31:0]	Data bus signals
$\overline{\text{EB0}}$	MSB Byte Strobe—Active low external enable byte signal that controls D [31:24], shared with SDRAM DQM0.
$\overline{\text{EB1}}$	Byte Strobe—Active low external enable byte signal that controls D [23:16], shared with SDRAM DQM1.

Table 2. i.MX21 Signal Descriptions (Continued)

Signal Name	Function/Notes
$\overline{EB2}$	Byte Strobe—Active low external enable byte signal that controls D [15:8], shared with SDRAM DQM2 and PCMCIA $\overline{PC\_REG}$ .
$\overline{EB3}$	LSB Byte Strobe—Active low external enable byte signal that controls D [7:0], shared with SDRAM DQM3 and PCMCIA $\overline{PC\_IORD}$ .
$\overline{OE}$	Memory Output Enable—Active low output enables external data bus, shared with PCMCIA $\overline{PC\_IOWR}$ .
$\overline{CS}$ [5:0]	Chip Select—The chip select signals $\overline{CS}$ [3:2] are multiplexed with $\overline{CSD}$ [1:0] and are selected by the Function Multiplexing Control Register (FMCR) in the System Control chapter. By default $\overline{CSD}$ [1:0] is selected. $\overline{DTACK}$ is multiplexed with $\overline{CS4}$ .
$\overline{ECB}$	Active low input signal sent by flash device to the EIM whenever the flash device must terminate an on-going burst sequence and initiate a new (long first access) burst sequence.
$\overline{LBA}$	Active low signal sent by flash device causing the external burst device to latch the starting burst address.
BCLK	Clock signal sent to external synchronous memories (such as burst flash) during burst mode.
$\overline{RW}$	$\overline{RW}$ signal—Indicates whether external access is a read (high) or write (low) cycle. This signal is also shared with the PCMCIA $\overline{PC\_WE}$ .
$\overline{DTACK}$	$\overline{DTACK}$ signal—External input data acknowledge signal, multiplexed with $\overline{CS4}$ .
<b>Bootstrap</b>	
BOOT [3:0]	System Boot Mode Select—The operational system boot mode of the i.MX21 upon system reset is determined by the settings of these pins.
<b>SDRAM Controller</b>	
SDBA [4:0]	SDRAM non-interleave mode bank address signals. These signals are multiplexed with address signals A[20:16].
SDIBA [3:0]	SDRAM interleave addressing mode bank address signals. These signals are multiplexed with address signals A[24:21].
MA [11:0]	SDRAM address signals. MA[9:0] are multiplexed with address signals A[10:1].
DQM [3:0]	SDRAM data qualifier mask multiplexed with $\overline{EB}$ [3:0]. DQM3 corresponds to D[31:24], DQM2 corresponds to D[23:16], DQM1 corresponds to D[15:8], and DQM0 corresponds to D[7:0].
$\overline{CSD0}$	SDRAM Chip Select signal. This signal is multiplexed with the $\overline{CS2}$ signal. This signal is selectable by programming the Function Multiplexing Control Register in the System Control chapter.
$\overline{CSD1}$	SDRAM Chip Select signal. This signal is multiplexed with the $\overline{CS3}$ signal. This signal is selectable by programming the Function Multiplexing Control Register in the System Control chapter.
$\overline{RAS}$	SDRAM Row Address Select signal
$\overline{CAS}$	SDRAM Column Address Select signal
$\overline{SDWE}$	SDRAM Write Enable signal
SDCKE0	SDRAM Clock Enable 0
SDCKE1	SDRAM Clock Enable 1
SDCLK	SDRAM Clock

**Table 2. i.MX21 Signal Descriptions (Continued)**

Signal Name	Function/Notes
<b>Clocks and Resets</b>	
EXTAL26M	Crystal input (26MHz), or a 16 MHz to 32 MHz oscillator (or square-wave) input when internal oscillator circuit is shut down.
XTAL26M	Oscillator output to external crystal
EXTAL32K	32 kHz crystal input
XTAL32K	Oscillator output to 32 kHz crystal
CLKO	Clock Out signal selected from internal clock signals. Please refer to clock controller for internal clock selection.
EXT_48M	This is a special factory test signal. To ensure proper operation, connect this signal to ground.
EXT_266M	This is a special factory test signal. To ensure proper operation, connect this signal to ground.
RESET_IN	Master Reset—External active low Schmitt trigger input signal. When this signal goes active, all modules (except the reset module, SDRAMC module, and the clock control module) are reset.
RESET_OUT	Reset Out—Internal active low output signal from the Watchdog Timer module and is asserted from the following sources: Power-on reset, External reset (RESET_IN), and Watchdog time-out.
POR	Power On Reset—Active low Schmitt trigger input signal. The $\overline{\text{POR}}$ signal is normally generated by an external RC circuit designed to detect a power-up event.
CLKMODE[1:0]	These are special factory test signals. To ensure proper operation, leave these signals as no connects.
OSC26M_TEST	This is a special factory test signal. To ensure proper operation, leave this signal as a no connect.
TEST_WB[2:0]	These are special factory test signals. However, these signals are also multiplexed with GPIO PORT E as well as alternate keypad signals. If not utilizing these signals for GPIO functionality or for it's other multiplexed function, then configure as GPIO input with pull up enabled, and leave as a no connect.
TEST_WB[4:3]	These are special factory test signals. To ensure proper operation, leave these signals as no connects.
WKGD	Battery indicator input used to qualify the walk-up process. Also multiplexed with TIN.
<b>JTAG</b>	
TRST	Test Reset Pin—External active low signal used to asynchronously initialize the JTAG controller.
TDO	Serial Output for test instructions and data. Changes on the falling edge of TCK.
TDI	Serial Input for test instructions and data. Sampled on the rising edge of TCK.
TCK	Test Clock to synchronize test logic and control register access through the JTAG port.
TMS	Test Mode Select to sequence the JTAG test controller's state machine. Sampled on the rising edge of TCK.
JTAG_CTRL	JTAG Controller select signal—JTAG_CTRL is sampled during the rising edge of TRST. Must be pulled to logic high for proper JTAG interface to debugger. Pulling JTAG_CTRL low is for internal test purposes only.
RTCK	JTAG Return Clock used to enhance stability of JTAG debug interface devices. This signal is multiplexed with OWIRE, hence utilizing OWIRE will render RTCK unusable and vice versa.

**Table 2. i.MX21 Signal Descriptions (Continued)**

Signal Name	Function/Notes
<b>CMOS Sensor Interface</b>	
CSI_D [7:0]	Sensor port data
CSI_MCLK	Sensor port master clock
CSI_VSYNC	Sensor port vertical sync
CSI_HSYNC	Sensor port horizontal sync
CSI_PIXCLK	Sensor port data latch clock
<b>LCD Controller</b>	
LD [17:0]	LCD Data Bus—All LCD signals are driven low after reset and when LCD is off. LD[15:0] signals are multiplexed with SLCDC1_DAT[15:0] from SLCDC1 and BMI_D[15:0]. LD[17] signal is multiplexed with BMI_WRITE of BMI. LD[16] signal is multiplexed with BMI_READ_REQ of BMI and EXT_DMAGRANT signals.
FLM_VSYNC (or simply referred to as VSYNC)	Frame Sync or Vsync—This signal also serves as the clock signal output for gate driver (dedicated signal SPS for Sharp panel HR-TFT). This signal is multiplexed with BMI_RXF_FULL and BMI_WAIT of the BMI.
LP_HSYNC (or simply referred to as HSYNC)	Line Pulse or HSync
LSCLK	Shift Clock. This signal is multiplexed with the BMI_CLK_CS from BMI.
OE_ACD	Alternate Crystal Direction/Output Enable.
CONTRAST	This signal is used to control the LCD bias voltage as contrast control. This signal is multiplexed with the BMI_READ from BMI.
SPL_SPR	Sampling start signal for left and right scanning. This signal is multiplexed with the SLCDC1_CLK.
PS	Control signal output for source driver (Sharp panel dedicated signal). This signal is multiplexed with the SLCDC1_CS.
CLS	Start signal output for gate driver. This signal is invert version of PS (Sharp panel dedicated signal). This signal is multiplexed with the SLCDC1_RS.
REV	Signal for common electrode driving signal preparation (Sharp panel dedicated signal). This signal is multiplexed with SLCDC1_D0.
<b>Smart LCD Controller</b>	
SLCDC1_CLK	SLCDC Clock output signal. This signal is multiplexed and available at 2 alternate locations. These are SPL_SPR and SD2_CLK signals of LCDC and SD2, respectively.
SLCDC1_CS	SLCDC Chip Select output signal. This signal is multiplexed and available at 2 alternate signal locations. These are PS and SD2_CMD signals of LCDC and SD2, respectively.
SLCDC1_RS	SLCDC Register Select output signal. This signal is multiplexed and available at 2 alternate signal locations. These are CLS and SD2_D3 signals of LCDC and SD2, respectively.
SLCDC1_D0	SLCDC serial data output signal. This signal is multiplexed and available at 2 alternate signal locations. These are REV and SD2_D2 signals of LCDC and SD2, respectively. This signal is inactive when a parallel data interface is used.



Table 2. i.MX21 Signal Descriptions (Continued)

Signal Name	Function/Notes
SLCDC1_DAT[15:0]	SLCDC Data output signals for connection to a parallel SLCD panel interface. These signals are multiplexed with LD[15:0] while an alternate 8-bit SLCD muxing is available on LD[15:8]. Further alternate muxing of these signals are available on some of the USB OTG and USBH1 signals.
SLCDC2_CLK	SLCDC Clock input signal for pass through to SLCD device. This signal is multiplexed with SSI3_CLK signal from SSI3.
SLCDC2_CS	SLCDC Chip Select input signal for pass through to SLCD device. This signal is multiplexed with SSI3_TXD signal from SSI3.
SLCDC2_RS	SLCDC Register Select input signal for pass through to SLCD device. This signal is multiplexed with SSI3_RXD signal from SSI3.
SLCDC2_D0	SLCD Data input signal for pass through to SLCD device. This signal is multiplexed with SSI3_FS signal from SSI3.
<b>Bus Master Interface (BMI)</b>	
BMI_D[15:0]	BMI bidirectional data bus. Bus width is programmable between 8-bit or 16-bit. These signals are multiplexed with LD[15:0] and SLCDC_DAT[15:0].
BMI_CLK_CS	BMI bidirectional clock or chip select signal. This signal is multiplexed with LSCLK of LCDC.
$\overline{\text{BMI\_WRITE}}$	BMI bidirectional signal to indicate read or write access. This is an input signal when the BMI is a slave and an output signal when BMI is the master of the interface. $\overline{\text{BMI\_WRITE}}$ is asserted for write and negated for read. This signal is muxed with LD[17] of LCDC.
$\overline{\text{BMI\_READ}}$	BMI output signal to enable data read from external slave device. This signal is not used and driven high when BMI is slave. This signal is multiplexed with CONTRAST signal of LCDC.
BMI_READ_REQ	BMI Read request output signal to external bus master. This signal is active when the data in the TXFIFO is larger or equal to the data transfer size of a single external BMI access. This signal is muxed with LD[16] of LCDC.
BMI_RXF_FULL	BMI Receive FIFO full active high output signal to reflect if the Rx FIFO reaches water mark value. This signal is muxed with VSYNC of the LCDC.
$\overline{\text{BMI\_WAIT}}$	BMI Wait—Active low signal to wait for data ready (read cycle) or accepted (write cycle). Also multiplexed with VSYNC.
<b>External DMA</b>	
$\overline{\text{EXT\_DMAREQ}}$	External DMA Request input signal. This signal is multiplexed with CSPI1_RDY.
$\overline{\text{EXT\_DMAGRANT}}$	External DMA Grant output signal. This signal is multiplexed with LD[16].
<b>NAND Flash Controller</b>	
NF_CLE	NAND Flash Command Latch Enable output signal. This signal is multiplexed with PC_POE of PCMCIA.
$\overline{\text{NF\_CE}}$	NAND Flash Chip Enable output signal. This signal is multiplexed with PC_CE1 of PCMCIA.
$\overline{\text{NF\_WP}}$	NAND Flash Write Protect output signal. This signal is multiplexed with PC_CE2 of PCMCIA.
NF_ALE	NAND Flash Address Latch Enable output signal. This signal is multiplexed with $\overline{\text{PC\_OE}}$ of PCMCIA.
$\overline{\text{NF\_RE}}$	NAND Flash Read Enable output signal. This signal is multiplexed with $\overline{\text{PC\_RW}}$ of PCMCIA.

Table 2. i.MX21 Signal Descriptions (Continued)

Signal Name	Function/Notes
NF_WE	NAND Flash Write Enable output signal. This signal is multiplexed with PC_BVD2 of PCMCIA.
NF_RB	NAND Flash Ready Busy input signal. This signal is multiplexed with PC_RST of PCMCIA.
NF_IO[15:0]	NAND Flash Data input and output signals. NF_IO[15:7] signals are multiplexed with A[25:21] and A[15:13]. NF_IO[7:0] signals are multiplexed with several PCMCIA signals.
<b>PCMCIA Controller</b>	
PC_A[25:0]	PCMCIA Address signals. These signals are multiplexed with A[25:0].
PC_D[15:0]	PCMCIA Data input and output signals. These signals are multiplexed with D[15:0].
PC_CD1	PCMCIA Card Detect1 input signal. This signal is multiplexed with NFIO[7] signal of NF.
PC_CD2	PCMCIA Card Detect2 input signal. This signal is multiplexed with NFIO[6] signal of NF.
PC_WAIT	PCMCIA Wait input signal to extend current access. This signal is multiplexed with NFIO[5] signal of NF.
PC_READY	PCMCIA Ready input signal to indicate card is ready for access. This signal is multiplexed with NFIO[4] signal of NF.
PC_RST	PCMCIA Reset output signal. This signal is multiplexed with NFRB signal of NF.
PC_OE	PCMCIA Memory Read Enable output signal asserted during common or attribute memory read cycles. This signal is multiplexed with NFALE signal of NF.
PC_WE	PCMCIA Memory Write Enable output signal asserted during common or attribute memory cycles. This signal is shared with RW of the EIM.
PC_VS1	PCMCIA Voltage Sense1 input signal. This signal is multiplexed with NFIO[2] signal of NF.
PC_VS2	PCMCIA Voltage Sense2 input signal. This signal is multiplexed with NFIO[1] signal of NF.
PC_BVD1	PCMCIA Battery Voltage Detect1 input signal. This signal is multiplexed with NFIO[0] signal of NF.
PC_BVD2	PCMCIA Battery Voltage Detect2 input signal. This signal is multiplexed with NF_WE signal of NF.
PC_SPKOUT	PCMCIA Speaker Out output signal. This signal is multiplexed with PWMO signal.
PC_REG	PCMCIA Register Select output signal. This signal is shared with EB2 of EIM.
PC_CE1	PCMCIA Card Enable1 output signal. This signal is multiplexed with NFCE signal of NF.
PC_CE2	PCMCIA Card Enable2 output signal. This signal is multiplexed with NFWP signal of NF.
PC_IORD	PCMCIA IO Read output signal. This signal is shared with EB3 of EIM.
PC_IOWR	PCMCIA IO Write output signal. This signal is shared with OE signal of EIM.
PC_WP	PCMCIA Write Protect input signal. This signal is multiplexed with NFIO[3] signal of NF.
PC_POE	PCMCIA Output Enable signal to enable voltage translation buffers and transceivers. This signal is multiplexed with NFCLE signal of NF.
PC_RW	PCMCIA Read Write output signal to control external transceiver direction. Asserted high for read access and negated low for write access. This signal is multiplexed with NFRE signal of NF.
PC_PWRON	PCMCIA input signal to indicate that the card power has been applied and stabilized.

**Table 2. i.MX21 Signal Descriptions (Continued)**

Signal Name	Function/Notes
<b>CSPI</b>	
CSPI1_MOSI	Master Out/Slave In signal
CSPI1_MISO	Master In/Slave Out signal
CSPI1_SS[2:0]	Slave Select (Selectable polarity) signal. CSPI1_SS2 is also multiplexed with USBG_RXDAT.
CSPI1_SCLK	Serial Clock signal
CSPI1_RDY	Serial Data Ready signal. Also multiplexed with EXT_DMAREQ.
CSPI2_MOSI	Master Out/Slave In signal. This signal is multiplexed with USBH2_TXDP signal of USB OTG.
CSPI2_MISO	Master In/Slave Out signal. This signal is multiplexed with USBH2_TXDM signal of USB OTG.
CSPI2_SS[2:0]	Slave Select (Selectable polarity) signals. These signals are multiplexed with USBH2_FS, USBH2_RXDP and USBH2_RXDM signal of USB OTG
CSPI2_SCLK	Serial Clock signal. This signal is multiplexed with USBH2_OE signal of USB OTG
CSPI3_MOSI	Master Out/Slave In signal. This signal is multiplexed with SD1_CMD.
CSPI3_MISO	Master In/Slave Out signal. This signal is multiplexed with SD1_D0.
CSPI3_SS	Slave Select (Selectable polarity) signal multiplexed with SD1_D3.
CSPI3_SCLK	Serial Clock signal. This signal is multiplexed with SD1_CLK.
<b>General Purpose Timers</b>	
TIN	Timer Input Capture or Timer Input Clock—The signal on this input is applied to all 3 timers simultaneously. This signal is muxed with the Walk-up Guard Mode WKGD signal in the PLL, Clock, and Reset Controller module.
TOUT1 (or simply TOUT)	Timer Output signal from General Purpose Timer1 (GPT1). This signal is multiplexed with SSI1_MCLK and SSI2_MCLK signal of SSI1 and SSI2. The pin name of this signal is simply TOUT.
TOUT2	Timer Output signal from General Purpose Timer1 (GPT2). This signal is multiplexed with PWMO.
TOUT3	Timer Output signal from General Purpose Timer1 (GPT3). This signal is multiplexed with PWMO.
<b>USB On-The-Go</b>	
USB_BYP	USB Bypass input active low signal.
USB_PWR	USB Power output signal
USB_OC	USB Over current input signal
USBG_RXDP	USB OTG Receive Data Plus input signal. This signal is muxed with SLCDC1_DAT15.
USBG_RXDM	USB OTG Receive Data Minus input signal. This signal is muxed with SLCDC1_DAT14.
USBG_TXDP	USB OTG Transmit Data Plus output signal. This signal is muxed with SLCDC1_DAT13.
USBG_TXDM	USB OTG Transmit Data Minus output signal. This signal is muxed with SLCDC1_DAT12.
USBG_RXDAT	USB OTG Transceiver differential data receive signal. Multiplexed with CSPI1_SS2.
USBG_OE	USB OTG Output Enable signal. This signal is muxed with SLCDC1_DAT11.

**Table 2. i.MX21 Signal Descriptions (Continued)**

Signal Name	Function/Notes
USBG_ON	USB OTG Transceiver ON output signal. This signal is muxed with SLCDC1_DAT9.
USBG_FS	USB OTG Full Speed output signal. This signal is multiplexed with external transceiver USBG_TXR_INT signal of USB OTG. This signal is muxed with SLCDC1_DAT10.
USBH1_RXDP	USB Host1 Receive Data Plus input signal. This signal is multiplexed with UART4_RXD and SLCDC1_DAT6. It also provides an alternative multiplex for UART4_RTS, where this signal is selectable by programming the Function Multiplexing Control Register in the System Control chapter.
USBH1_RXDM	USB Host1 Receive Data Minus input signal. This signal is muxed with SLCDC1_DAT5. It also provides an alternative multiplex for UART4_CTS.
USBH1_TXDP	USB Host1 Transmit Data Plus output signal. This signal is multiplexed with UART4_CTS and SLCDC1_DAT4. It also provides an alternative multiplex for UART4_RXD, where this signal is selectable by programming the Function Multiplexing Control Register in the System Control chapter.
USBH1_TXDM	USB Host1 Transmit Data Minus output signal. This signal is multiplexed with UART4_TXD and SLCDC1_DAT3.
USBH1_RXDAT	USB Host1 Transceiver differential data receive signal. Multiplexed with USBH1_FS.
USBH1_OE	USB Host1 Output Enable signal. This signal is muxed with SLCDC1_DAT2.
USBH1_FS	USB Host1 Full Speed output signal. This signal is multiplexed with UART4_RTS and SLCDC1_DAT1 and USBH1_RXDAT.
USBH_ON	USB Host transceiver ON output signal. This signal is muxed with SLCDC1_DAT0.
USBH2_RXDP	USB Host2 Receive Data Plus input signal. This signal is multiplexed with CSPI2_SS[1] of CSPI2.
USBH2_RXDM	USB Host2 Receive Data Minus input signal. This signal is multiplexed with CSPI2_SS[2] of CSPI2.
USBH2_TXDP	USB Host2 Transmit Data Plus output signal. This signal is multiplexed with CSPI2_MOSI of CSPI2.
USBH2_TXDM	USB Host2 Transmit Data Minus output signal. This signal is multiplexed with CSPI2_MISO of CSPI2.
USBH2_OE	USB Host2 Output Enable signal. This signal is multiplexed with CSPI2_SCLK of CSPI2.
USBH2_FS	USB Host2 Full Speed output signal. This signal is multiplexed with CSPI2_SS[0] of CSPI2.
USBG_SCL	USB OTG I <sup>2</sup> C Clock Output signal. This signal is multiplexed with SLCDC1_DAT8.
USBG_SDA	USB OTG I <sup>2</sup> C Data Input/Output signal. This signal is multiplexed with SLCDC1_DAT7.
USBG_TXR_INT	USB OTG transceiver Interrupt input. Multiplexed with USBG_FS.
<b>Secure Digital Interface</b>	
SD1_CMD	SD Command bidirectional signal—If the system designer does not want to make use of the internal pull-up, via the Pull-up enable register, a 4.7K–69K external pull up resistor must be added. This signal is multiplexed with CSPI3_MOSI.
SD1_CLK	SD Output Clock. This signal is multiplexed with CSPI3_SCLK.

**Table 2. i.MX21 Signal Descriptions (Continued)**

Signal Name	Function/Notes
SD1_D[3:0]	SD Data bidirectional signals—If the system designer does not want to make use of the internal pull-up, via the Pull-up enable register, a 50 K–69K external pull up resistor must be added. SD1_D[3] is muxed with CSPI3_SS while SD1_D[0] is muxed with CSPI3_MISO.
SD2_CMD	SD Command bidirectional signal. This signal is multiplexed with SLCDC1_CS signal from SLCDC1.
SD2_CLK	SD Output Clock signal. This signal is multiplexed with SLCDC1_CLK signal from SLCDC1.
SD2_D[3:0]	SD Data bidirectional signals. SD2_D[3:2] are which are multiplexed with SLCDC1_RS and SLCDC_D0 signals from SLCDC1.
<b>UARTs – IrDA/Auto-Bauding</b>	
UART1_RXD	Receive Data input signal
UART1_TXD	Transmit Data output signal
UART1_RTS	Request to Send input signal
UART1_CTS	Clear to Send output signal
UART2_RXD	Receive Data input signal. This signal is multiplexed with KP_ROW6 signal from KPP.
UART2_TXD	Transmit Data output signal. This signal is multiplexed with KP_COL6 signal from KPP.
UART2_RTS	Request to Send input signal. This signal is multiplexed with KP_ROW7 signal from KPP.
UART2_CTS	Clear to Send output signal. This signal is multiplexed with KP_COL7 signal from KPP.
UART3_RXD	Receive Data input signal. This signal is multiplexed with IR_RXD from FIRI.
UART3_TXD	Transmit Data output signal. This signal is multiplexed with IR_TXD from FIRI.
UART3_RTS	Request to Send input signal
UART3_CTS	Clear to Send output signal
UART4_RXD	Receive Data input signal which is multiplexed with USBH1_RXDP and USBH1_TXDP.
UART4_TXD	Transmit Data output signal which is multiplexed with USBH1_TXDM.
UART4_RTS	Request to Send input signal which is multiplexed with USBH1_FS and USBH1_RXDP.
UART4_CTS	Clear to Send output signal which is multiplexed with USBH1_TXDP and USBH1_RXDM.
<b>Serial Audio Port – SSI (configurable to I<sup>2</sup>S protocol and AC97)</b>	
SSI1_CLK	Serial clock signal which is output in master or input in slave
SSI1_TXD	Transmit serial data
SSI1_RXD	Receive serial data
SSI1_FS	Frame Sync signal which is output in master and input in slave
SSI1_MCLK	SSI1 master clock. Multiplexed with TOUT.
SSI2_CLK	Serial clock signal which is output in master or input in slave.
SSI2_TXD	Transmit serial data signal
SSI2_RXD	Receive serial data

**Table 2. i.MX21 Signal Descriptions (Continued)**

Signal Name	Function/Notes
SSI2_FS	Frame Sync signal which is output in master and input in slave.
SSI2_MCLK	SSI2 master clock. Multiplexed with TOUT.
SSI3_CLK	Serial clock signal which is output in master or input in slave. This signal is multiplexed with SLCDC2_CLK
SSI3_TXD	Transmit serial data signal which is multiplexed with SLCDC2_CS
SSI3_RXD	Receive serial data which is multiplexed with SLCDC2_RS
SSI3_FS	Frame Sync signal which is output in master and input in slave. This signal is multiplexed with SLCDC2_D0.
SAP_CLK	Serial clock signal which is output in master or input in slave.
SAP_TXD	Transmit serial data
SAP_RXD	Receive serial data
SAP_FS	Frame Sync signal which is output in master and input in slave.
<b>I<sup>2</sup>C</b>	
I2C_CLK	I <sup>2</sup> C Clock
I2C_DATA	I <sup>2</sup> C Data
<b>1-Wire</b>	
OWIRE	One wire input and output signal. This signal is multiplexed with JTAG RTCK.
<b>PWM</b>	
PWMO	PWM Output. This signal is multiplexed with PC_SPKOUT of PCMCIA, as well as TOUT2 and TOUT3 of the General Purpose Timer module.
<b>Keypad</b>	
KP_COL[7:0]	Keypad Column selection signals. KP_COL[7:6] are multiplexed with $\overline{\text{UART2\_CTS}}$ and UART2_TXD respectively. Alternatively, KP_COL6 is also available on the internal factory test signal TEST_WB2. The Function Multiplexing Control Register in the System Control chapter must be used in conjunction with programming the GPIO multiplexing (to select the alternate signal multiplexing) to choose which signal KP_COL6 is available.
KP_ROW[7:0]	Keypad Row selection signals. KP_ROW[7:6] are multiplexed with $\overline{\text{UART2\_RTS}}$ and UART2_RXD signals respectively. Alternatively, KP_ROW7 and KP_ROW6 are available on the internal factory test signals TEST_WB0 and TEST_WB1 respectively. The Function Multiplexing Control Register in the System Control chapter must be used in conjunction with programming the GPIO multiplexing (to select the alternate signal multiplexing) to choose which signals KP_ROW6 and KP_ROW7 are available.
<b>Noisy Supply Pins</b>	
NVDD	Noisy Supply for the I/O pins. There are six (6) I/O voltage rings, NVDD1 through NVDD6.
NVSS	Noisy Ground for the I/O pins

**Table 2. i.MX21 Signal Descriptions (Continued)**

Signal Name	Function/Notes
<b>Supply Pins – Analog Modules</b>	
VDDA (formally AVDD)	Supply for analog blocks
QVSS (internally connected to AVSS)	Quiet GND for analog blocks (QVSS and AVSS are synonymous)
<b>Internal Power Supply</b>	
QVDD	Power supply pins for silicon internal circuitry
QVSS	Quiet GND pins for silicon internal circuitry
QVDDX	Power supply pin for the ARM core, connect directly to QVDD

## 3 Specifications

This section contains the electrical specifications and timing diagrams for the i.MX21 processor.

### 3.1 Maximum Ratings

Table 3 provides information on maximum ratings.

**Table 3. Maximum Ratings**

Rating	Symbol	Minimum	Maximum	Unit
Supply voltage	$V_{dd}$	-0.3	3.3	V
Maximum operating temperature range of i.MX21	$T_A$	- 40 / -30 / 0	70 / 85	°C
Storage temperature	Test	-55	150	°C

### 3.2 Recommended Operating Range

Table 4 provides the recommended operating ranges for the supply voltages. The i.MX21 processor has multiple pairs of VDD and VSS power supply and return pins. QVDD, QVDDx, and QVSS pins are used for internal logic. All other VDD and VSS pins are for the I/O pads voltage supply, and each pair of VDD and VSS provides power to the enclosed I/O pads. This design allows different peripheral supply voltage levels in a system.

Because AVDD pins are supply voltages to the analog pads, it is recommended to isolate and noise-filter the AVDD pins from other VDD pins.

For more information about I/O pads grouping per VDD, please refer to Table 4 on page 15.

**Table 4. Recommended Operating Range**

Rating	Symbol	Minimum	Maximum	Unit
I/O supply voltage	NVDD 2, 3, 4, 5, 6	2.70	3.30	V
I/O supply voltage	NVDD 1	1.70	3.30	V

Table 4. Recommended Operating Range (Continued)

Rating	Symbol	Minimum	Maximum	Unit
Internal supply voltage (Core = 266 MHz)	QVDD, QVDDx	1.45	1.65	V
Analog supply voltage	AVDD	1.70	3.30	V

### 3.3 DC Electrical Characteristics

Table 5 contains both maximum and minimum DC characteristics of the i.MX21.

Table 5. Maximum and Minimum DC Characteristics

Number or Symbol	Parameter	Minimum	Typical	Maximum	Unit
I <sub>op</sub>	Full running operating current QVDD & QVDDx=1.65V, NVDD1=1.8V, NVDD2-6 & AVDD=3.1V, Full run: Core=266MHz, System=133MHz, Doze: Core=266MHz, System=53MHz, MPEG4 Playback (QVGA) from MMC/SD card, 30fps, 44.1kHz audio)	–	120mA (QVDD+QVDDx),  8mA (NVDD1)  6.6mA (NVDD2-6+AVDD)	–	mA
I <sub>std</sub>	Standby current (QVDD, QVDDx= 1.55V)	–	360	–	μA
V <sub>IH</sub>	Input high voltage	0.7NVDD	–	NVDD	V
V <sub>IL</sub>	Input low voltage	0	–	0.3NVDD	V
V <sub>OH</sub>	Output high voltage	0.8NVDD	–	–	V
V <sub>OL</sub>	Output low voltage	–	–	0.2NVDD	V
V <sub>it+</sub>	Positive input threshold voltage, V <sub>i</sub> = V <sub>ih</sub>	–	–	2.15	V
V <sub>it-</sub>	Negative input threshold voltage, V <sub>i</sub> = V <sub>il</sub>	0.75	–	–	V
V <sub>hys</sub>	Hysteresis (V <sub>it+</sub> – V <sub>it-</sub> ) = V <sub>ih</sub>	–	0.3	–	–
I <sub>IL</sub>	Input low leakage current (V <sub>IN</sub> = GND, no pull-up or pull-down)	–	–	±1	μA
I <sub>IH</sub>	Input high leakage current (V <sub>IN</sub> = V <sub>DD</sub> , no pull-up or pull-down)	–	–	±1	μA
I <sub>OH</sub>	Output high current VO = VOH	–	–	Slow Pad: -6 Fast Pad: -5	mA
I <sub>OL</sub>	Output low current VO = VOL	Slow Pad: 6 Fast Pad: 5	–	–	mA
I <sub>OZ</sub>	Output leakage current (V <sub>out</sub> = V <sub>DD</sub> , output is tri-stated)	–	–	±5	μA
C <sub>i</sub>	Input capacitance	–	–	5	pF
C <sub>o</sub>	Output capacitance	–	–	5	pF



### 3.4 AC Electrical Characteristics

The AC characteristics consist of output delays, input setup and hold times, and signal skew times. All signals are specified relative to an appropriate edge of other signals. All timing specifications are specified at a system operating frequency from 0 MHz to 133 MHz (core operating frequency 266 MHz) with an operating supply voltage from  $V_{DD\ min}$  to  $V_{DD\ max}$  under an operating temperature from  $T_L$  to  $T_H$ . All timing is measured at 30 pF loading.

**Table 6. Tri-State Signal Timing**

Pin	Parameter	Minimum	Maximum	Unit
TRISTATE	Time from TRISTATE activate until I/O becomes Hi-Z	–	20.8	ns

**Table 7. 32k/26M Oscillator Signal Timing**

Parameter	Minimum	RMS	Maximum	Unit
EXTAL32k input jitter (peak to peak) for both System PLL and MCUPLL	–	5	20	ns
EXTAL32k input jitter (peak to peak) for MCUPLL only	–	5	100	ns
EXTAL32k startup time	800	–	–	ms

**Table 8. CLKO Rise/Fall Time (at 30pF Loaded)**

	Best Case	Typical	Worst Case	Units
Rise Time	0.80	1.00	1.40	ns
Fall Time	0.74	1.08	1.67	ns

### 3.5 DPLL Timing Specifications

Parameters of the DPLL are given in Table 9. In this table,  $T_{ref}$  is a reference clock period after the predivider and  $T_{dck}$  is the output double clock period.

**Table 9. DPLL Specifications**

Parameter	Test Conditions	Minimum	Typical	Maximum	Unit
Reference clock frequency range	$V_{cc} = 1.5V$	16	–	320	MHz
Pre-divider output clock frequency range	$V_{cc} = 1.5V$	16	–	32	MHz
Double clock frequency range	$V_{cc} = 1.5V$	160	–	560	MHz
Pre-divider factor (PD)	–	1	–	16	–
Total multiplication factor (MF)	Includes both integer and fractional parts	5	–	15	–
MF integer part	–	5	–	15	–
MF numerator	Should be less than the denominator	0	–	1022	–
MF denominator	–	1	–	1023	–
Frequency lock-in time after full reset	FOL mode for non-integer MF (does not include pre-multi lock-in time)	350	400	450	$T_{ref}$
Frequency lock-in time after partial reset	FOL mode for non-integer MF (does not include pre-multi lock-in time)	220	280	330	$T_{ref}$
Phase lock-in time after full reset	FPL mode and integer MF (does not include pre-multi lock-in time)	480	530	580	$T_{ref}$
Phase lock-in time after partial reset	FPL mode and integer MF (does not include pre-multi lock-in time)	360	410	460	$T_{ref}$
Frequency jitter (p-p)	–	–	0.02	0.03	$2 \cdot T_{dck}$
Phase jitter (p-p)	Integer MF, FPL mode, $V_{cc}=1.5V$	–	1.0	1.5	ns
Power dissipation	FOL mode, integer MF, $f_{dck} = 560 \text{ MHz}$ , $V_{cc} = 1.5V$	–	1.5	–	mW (Avg)

### 3.6 Reset Module

The timing relationships of the Reset module with the  $\overline{\text{POR}}$  and  $\overline{\text{RESET\_IN}}$  are shown in Figure 2 and Figure 3 on page 20. Be aware that NVDD must ramp up to at least 1.7V for NVDD1 and 2.7V for NVDD2-6 before QVDD is powered up to prevent forward biasing.

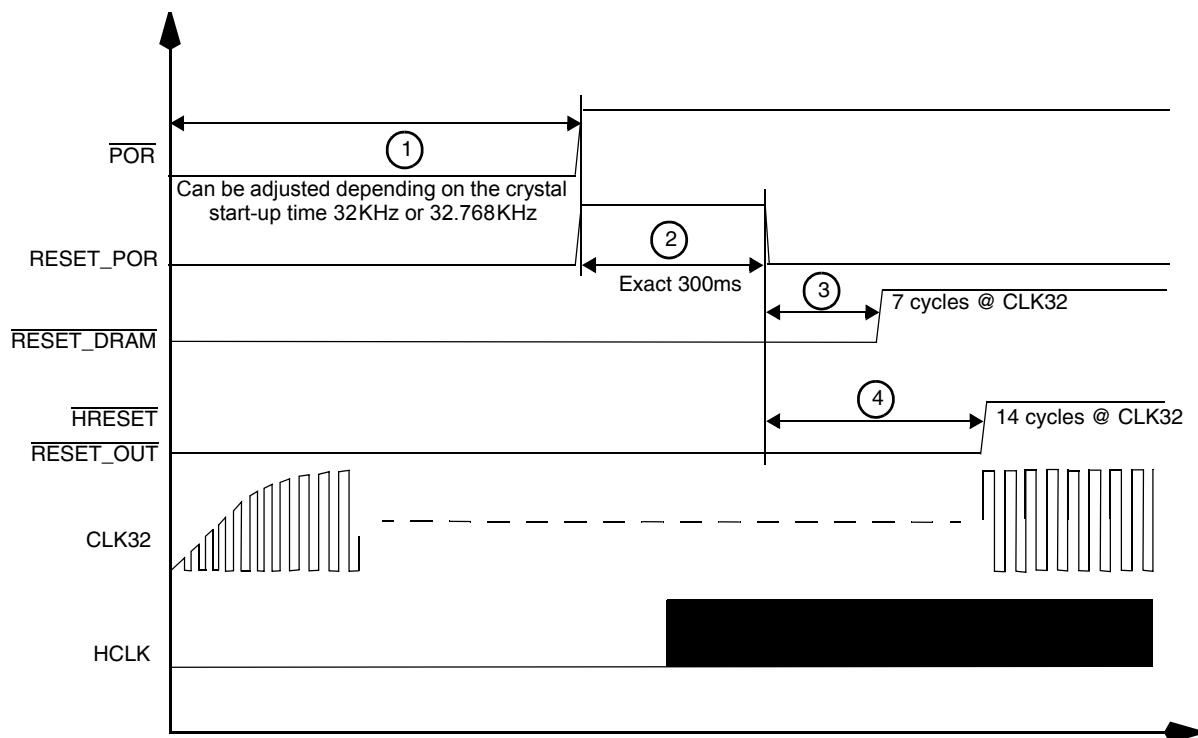


Figure 2. Timing Relationship with  $\overline{\text{POR}}$

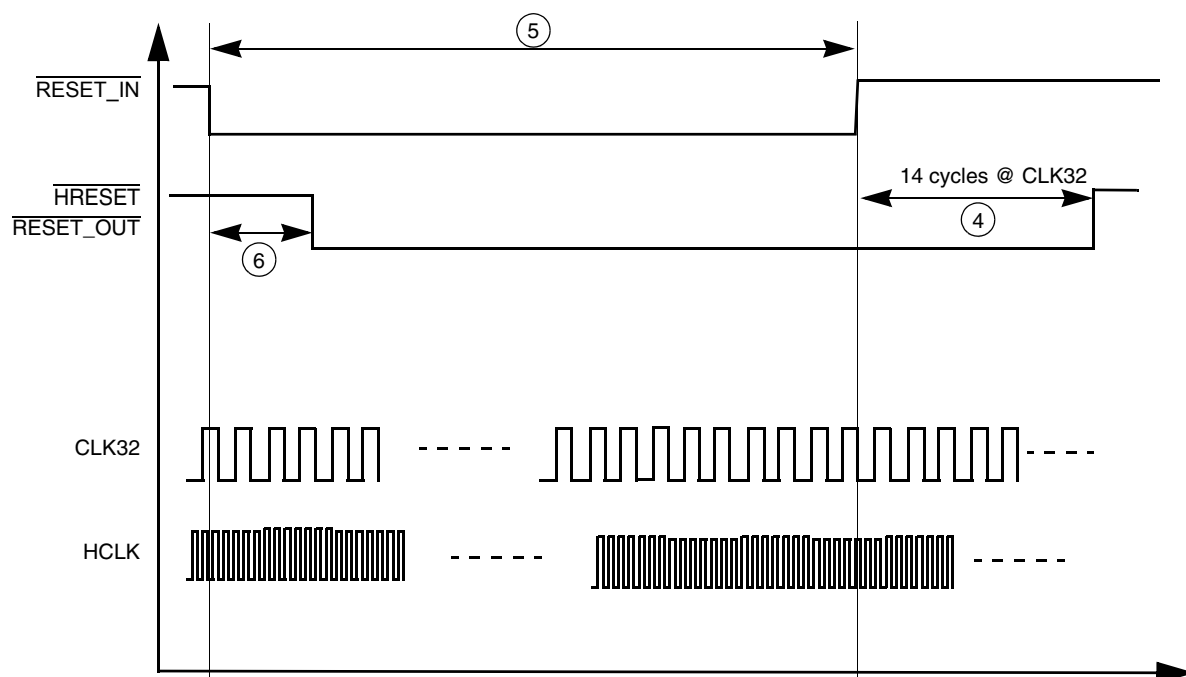
Figure 3. Timing Relationship with  $\overline{\text{RESET\_IN}}$ 

Table 10. Reset Module Timing Parameter Table

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Min	Max	Min	Max	
1	Width of input $\overline{\text{POWER\_ON\_RESET}}$	800	–	800	–	ms
2	Width of internal $\overline{\text{POWER\_ON\_RESET}}$ (CLK32 at 32 KHz)	300	300	300	300	ms
3	7K to 32K-cycle stretcher for SDRAM reset	7	7	7	7	Cycles of CLK32
4	14K to 32K-cycle stretcher for internal system reset $\overline{\text{HRESET}}$ and output reset at pin $\overline{\text{RESET\_OUT}}$	14	14	14	14	Cycles of CLK32
5	Width of external hard-reset $\overline{\text{RESET\_IN}}$	4	–	4	–	Cycles of CLK32
6	4K to 32K-cycle qualifier	4	4	4	4	Cycles of CLK32

### 3.7 External DMA Request and Grant

The External DMA request is an active low signal to be used by devices external to i.MX21 processor to request the DMAC for data transfer.

After assertion of External DMA request the DMA burst will start when the channel on which the External request is the source (as per the RSSR settings) becomes the current highest priority channel. The external device using the External DMA request should keep its request asserted until it is serviced by the DMAC. One External DMA request will initiate one DMA burst.

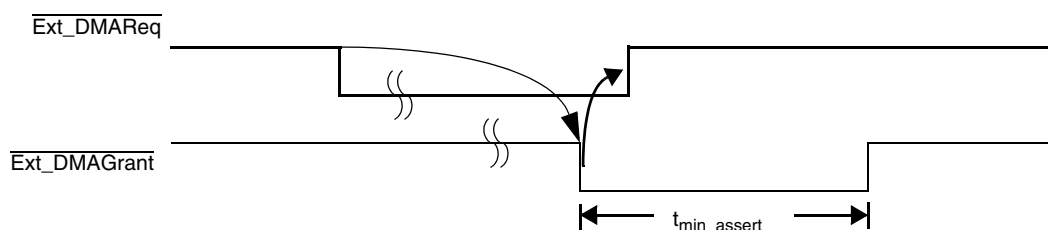
The output External Grant signal from the DMAC is an active-low signal. When the following conditions are true, the External DMA Grant signal is asserted with the initiation of the DMA burst.

- The DMA channel for which the DMA burst is ongoing has request source as external DMA Request (as per source select register setting).
- REN and CEN bit of this channel are set.
- External DMA Request is asserted.

After the grant is asserted, the External DMA request will not be sampled until completion of the DMA burst. As the external request is synchronized, the request synchronization will not be done during this period. The priority of the external request becomes low for the next consecutive burst, if another DMA request signal is asserted.

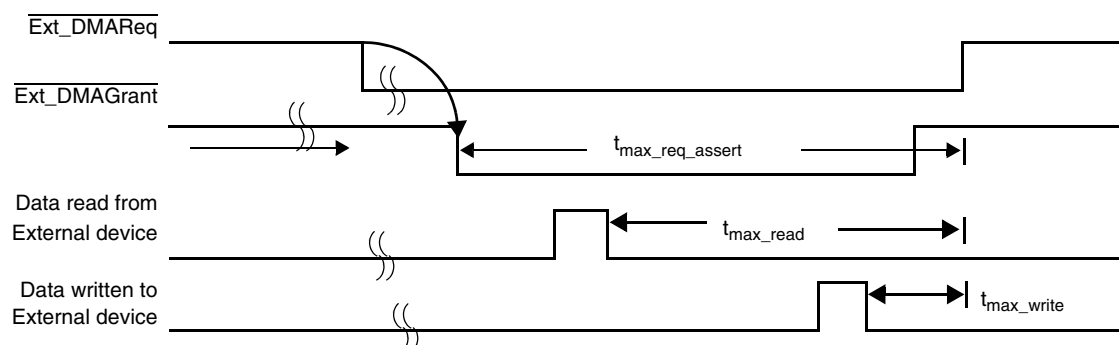
Worst case—that is, the smallest burst (1 byte read/write) timing diagrams are shown in Figure 4 and Figure 5 on page 21. Minimum and maximum timings for the External request and External grant signals are present in Table 11 on page 22.

Figure 4 shows the minimum time for which the External Grant signal remains asserted when an External DMA request is de-asserted immediately after sensing grant signal active.



**Figure 4. Assertion of DMA External Grant Signal**

Figure 5 shows the safe maximum time for which External DMA request can be kept asserted, after sensing grant signal active such that a new burst is not initiated.



NOTE: Assuming in worst case the data is read/written from/to External device as per the above waveform.

**Figure 5. Safe Maximum Timings for External Request De-Assertion**

**Table 11. DMA External Request and Grant Timing Parameter Table**

Parameter	Description	3.0 V		1.8 V		Unit
		WCS	BCS	WCS	BCS	
$t_{\min\_assert}$	Minimum assertion time of External Grant signal	8 hclk + 8.6	8 hclk + 2.74	8 hclk + 7.17	8 hclk + 3.25	ns
$t_{\max\_req\_assert}$	Maximum External request assertion time after assertion of Grant signal	9 hclk - 20.66	9 hclk - 6.7	9 hclk - 17.96	9 hclk - 8.16	ns
$t_{\max\_read}$	Maximum External request assertion time after first read completion	8 hclk - 6.21	8 hclk - 0.77	8 hclk - 5.84	8 hclk - 0.66	ns
$t_{\max\_write}$	Maximum External request assertion time after completion of first write	3 hclk - 15.87	3 hclk - 8.83	3 hclk - 15.9	3 hclk - 9.12	ns

## 3.8 BMI Interface Timing Diagram

### 3.8.1 Connecting BMI to ATI MMD Devices

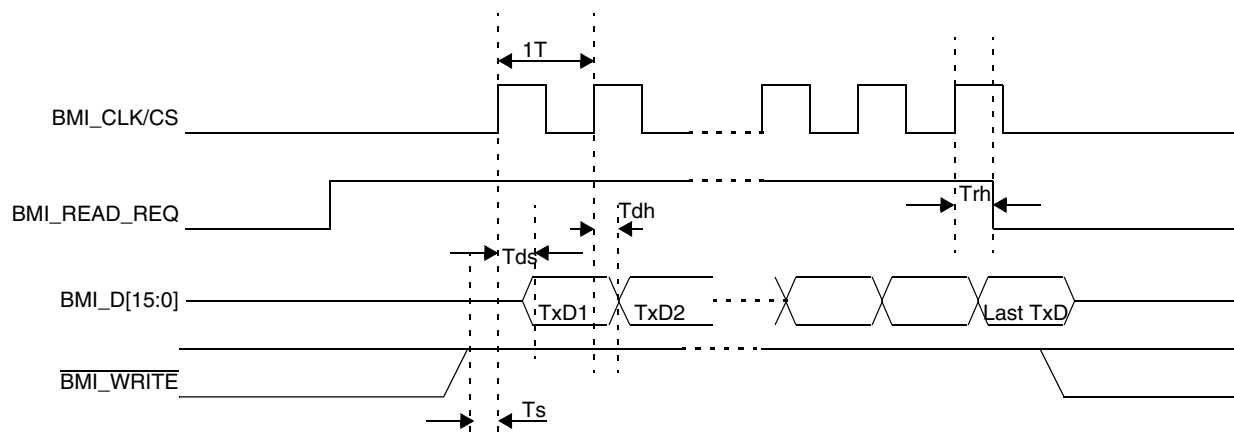
#### 3.8.1.1 ATI MMD Devices Drive the BMI\_CLK/CS

In this mode MMD\_MODE\_SEL bit is set and MMD\_CLKOUT bit is cleared.  $\overline{\text{BMI\_WRITE}}$  and BMI\_CLK/CS are input signals to BMI driving by ATI MMD chip set. Output signal BMI\_READ\_REQ can be used as interrupt signal to inform MMD that data is ready in BMI TxFIFO for read access. MMD can write data to BMI RxFIFO anytime as CPU or DMA can move data out from RxFIFO much faster than the BMI interface. Overflow interrupt is generated if RxFIFO overflow is detected. Once this happens, the new coming data is ignored.

##### 3.8.1.1.1 MMD Read BMI Timing

Figure 6 shows the MMD read BMI timing when the MMD drives clock.

On each rising edge of BMI\_CLK/CS BMI checks the  $\overline{\text{BMI\_WRITE}}$  logic level to determine if the current cycle is a read cycle. It puts data into the data bus and enables the data out on the rising edge of BMI\_CLK/CS if BMI\_WRITE is logic high. The BMI\_READ\_REQ is negated one hclk cycle after the BMI\_CLK/CS rising edge of last data read. The MMD cannot issues read command when BMI\_READ\_REQ is low (no data in TxFIFO).



**Figure 6. MMD (ATI) Drives Clock, MMD Read BMI Timing  
(MMD\_MODE\_SEL=1, MASTER\_MODE\_SEL=0, MMD\_CLKOUT=0)**

**Table 12. MMD Read BMI Timing Table when MMD Drives Clock**

Item	Symbol	Minimum	Typical	Maximum	Unit
Clock period	1T	33.3	–	–	ns
$\overline{\text{write}}$ setup time	Ts	11	–	–	ns
read_req hold time	Trh	6	–	24	ns
transfer data setup time	Tds	6	–	14	ns
transfer data hold time	Tdh	6	–	14	ns

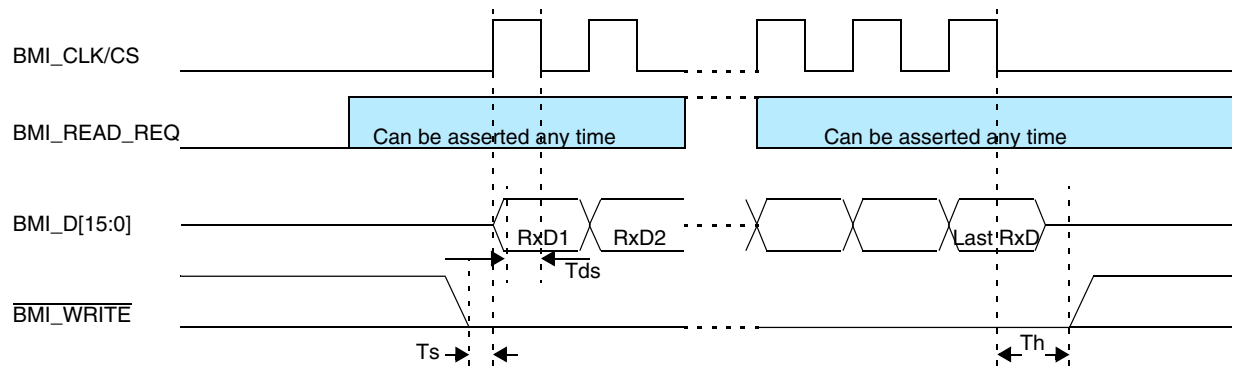
**Note:** All the timings assume that the hclk is running at 133 MHz.

**Note:** The MIN period of the 1T is assumed that MMD latch data at falling edge.

**Note:** If the MMD latch data at next rising edge, the ideally max clock can be as much as double, but because the BMI data pads are slow pads and its max frequency can only up to 18MHz, the max clock frequency can only up to 36 MHz.

### 3.8.1.1.2 MMD Write BMI Timing

Figure 7 on page 24 shows the MMD write BMI timing when MMD drives clock. On each falling edge of BMI\_CLK/CS BMI checks the  $\overline{\text{BMI\_WRITE}}$  logic level to determine if the current cycle is a write cycle. If the  $\overline{\text{BMI\_WRITE}}$  is logic low, it latches data into the RxFIFO on each falling edge of BMI\_CLK/CS signal.



**Figure 7. MMD (ATI) Drives Clock, MMD Write BMI Timing**  
(MMD\_MODE\_SEL=1, MASTER\_MODE\_SEL=0, MMD\_CLKOUT=0)

**Table 13. MMD Write BMI Timing**

Item	Symbol	Minimum	Typical	Maximum	Unit
write setup time	Ts	11	–	–	ns
write hold time	Th	0	–	–	ns
receive data setup time	Tds	5	–	–	ns

**Note:** All timings assume that the hclk is running at 133 MHz.

**Note:** At this mode, the maximum frequency of the BMI\_CLK/CS can be up to 36 MHz (doubles as maximum data pad speed).

### 3.8.1.2 BMI Drives the BMI\_CLK/CS

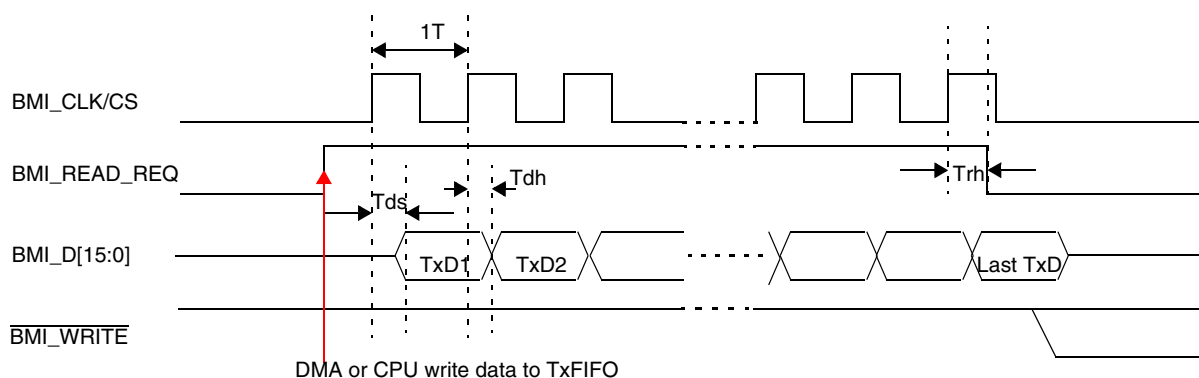
In this mode MMD\_MODE\_SEL and MMD\_CLKOUT are both set. The software must know which mode it is now (READ or WRITE). When the BMI\_WRITE is high, BMI drives BMI\_CLK/CS out if the TxFIFO is not emptied. When BMI\_WRITE is low, user can write a 1 to READ bit of control register1 to issue a write cycle (MMD write BMI).

### 3.8.1.3 MMD Read BMI Timing

Figure 13 on page 29 shows the MMD read BMI timing when BMI drives the BMI\_CLK/CS. When the BMI\_WRITE is high, the BMI drives BMI\_CLK/CS out if data is written to TxFIFO (BMI\_READ\_REQ become high), BMI puts data into data bus and enable data out on the rising edge of BMI\_CLK/CS. The MMD devices can latch the data on each falling edge of BMI\_CLK/CS.

It is recommended that the MMD do not change the BMI\_WRITE signal from high to low when the BMI\_READ\_REQ is asserted. If user writes data to the TxFIFO when the BMI\_WRITE is low, the BMI will drive BMI\_CLK/CS out once the BMI\_WRITE is changed from low to high.





**Figure 8. BMI Drives Clock, MMD Read BMI Timing**  
(MASTER\_MODE\_SEL=0, MMD\_MODE\_SEL=1, MMD\_CLKOUT=1)

**Table 14. MMD Read BMI Timing Table when BMI Drives Clock**

Item	Symbol	MIN	TYP	MAX	Unit
transfer data setup time	Tds	2	–	8	ns
transfer data hold time	Tdh	2	–	8	ns
read_req hold time	Trh	2	–	18	ns

**Note:** In this mode, the max frequency of the BMI\_CLK/CS can be up to 36Mhz(double as max data pad speed).

**Note:** The BMI\_CLK/CS can only be divided by 2,4,8,16 from HCLK.

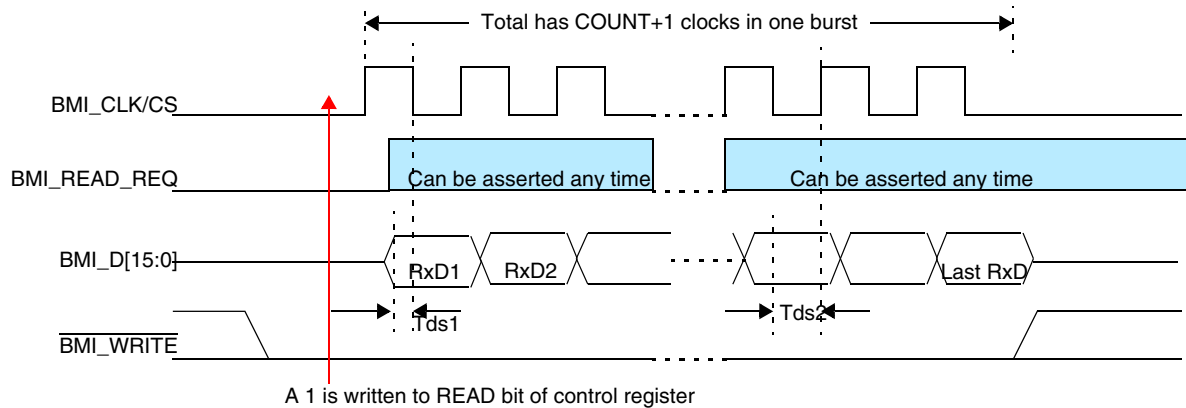
### 3.8.1.4 MMD Write BMI Timing

Figure on page 26 shows the MMD write BMI timing when BMI drives BMI\_CLK/CS.

When the  $\overline{\text{BMI\_WRITE}}$  signal is asserted, the BMI can write a 1 to READ bit of control register to issue a WRITE cycle. This bit is cleared automatically when the WRITE operation is completed. In a WRITE burst the MMD will write COUNT+1 data to the BMI. The user can issue another WRITE operation if the MMD still has data to write after the first operation completed.

The BMI can latch the data either at falling edge or the next rising edge of the BMI\_CLK/CS according to the DATA\_LATCH bit. When the DATA\_LATCH bit is set, the BMI latch data at the next rising edge and latch the last data using the internal clock.

$\overline{\text{BMI\_WRITE}}$  signal can not be negated when the WRITE operation is proceeding.



**Figure 9. BMI Drives Clock, MMD Write BMI Timing**  
(MASTER\_MODE\_SEL=0, MMD\_MODE\_SEL=1, MMD\_CLKOUT=1)

**Table 15. MMD Write BMI Timing Table when BMI Drives Clock**

Item	Symbol	Minimum	Typical	Maximum	Unit
receive data setup time1	Tds1	14	—	—	ns
receive data setup time2	Tds2	14	—	—	ns

**Note:** The BMI\_CLK/CS can only be up to 30Mhz if BMI latch data at the falling edge and can be up to 36Mhz (double as max data pad speed) if BMI latch data at the next rising edge.

**Note:** Tds1 is the receive data setup time when BMI latch data at the falling edge.

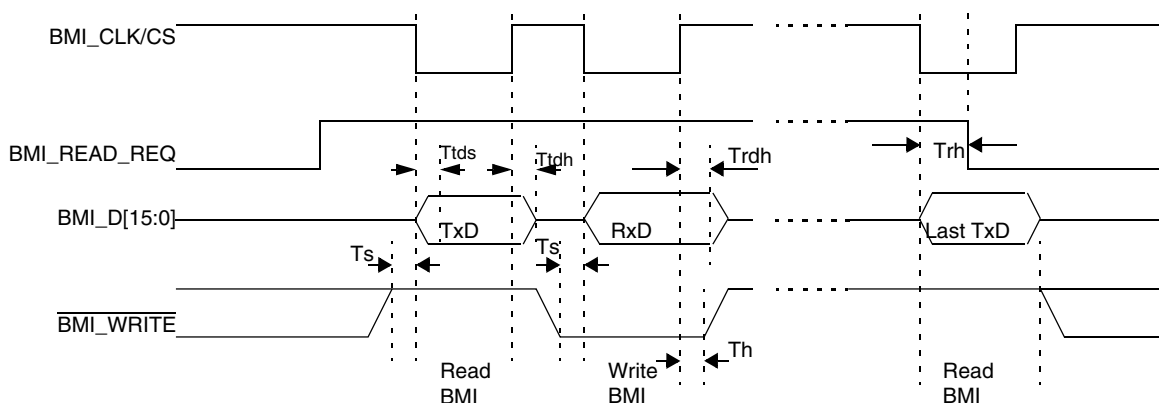
**Note:** Tds2 is the receive data setup time when BMI latch data at the next rising edge.

### 3.8.2 Connecting BMI to External Bus Master Devices

In this mode both MASTER\_SEL bit and MMD\_MODE\_SEL bit are cleared and the MMD\_CLKOUT bit is no useful. BMI\_WRITE and BMI\_CLK/CS are input signals driving by the external bus master. The Output signal BMI\_READ\_REQ can be used as an interrupt signal to inform external bus master that data is ready in the BMI Tx FIFO for a read access. The external bus master can write data to the BMI Rx FIFO anytime since the CPU or DMA can move data out from Rx FIFO much faster than the BMI interface. An overflow interrupt is generated if Rx FIFO overflow is detected. Once this happens, the new coming data is ignored.

Each falling edge of BMI\_CLK/CS will determine if the current cycle is read or write cycle. It drives data and enables data out if BMI\_WRITE is logic high. The D\_EN signal remains active only while BMI\_CLK/CS is logic low and BMI\_WRITE is logic high.

Each rising edge of BMI\_CLK/CS will determine if data should be latched to Rx FIFO from the data bus.



**Figure 10. Memory Interface Slave Mode, External Bus Master Read/Write to BMI Timing (MMD\_MODE\_SEL=0, MASTER\_MODE\_SEL=0)**

**Table 16. External Bus Master Read/Write to BMI Timing Table**

Item	Symbol	Minimum	Typical	Maximum	Unit
write setup time	Ts	11	—	—	ns
write hold time	Th	0	—	—	ns
receive data hold time	Trdh	3	—	—	ns
transfer data setup time	Ttds	6	—	14	ns
transfer data hold time	Ttdh	6	—	14	ns
read_req hold time	Trh	6	—	24	ns

**Note:** All the timings are assumed that the hclk is running at 133 MHz.

### 3.8.3 Connecting BMI to External Bus Slave Devices

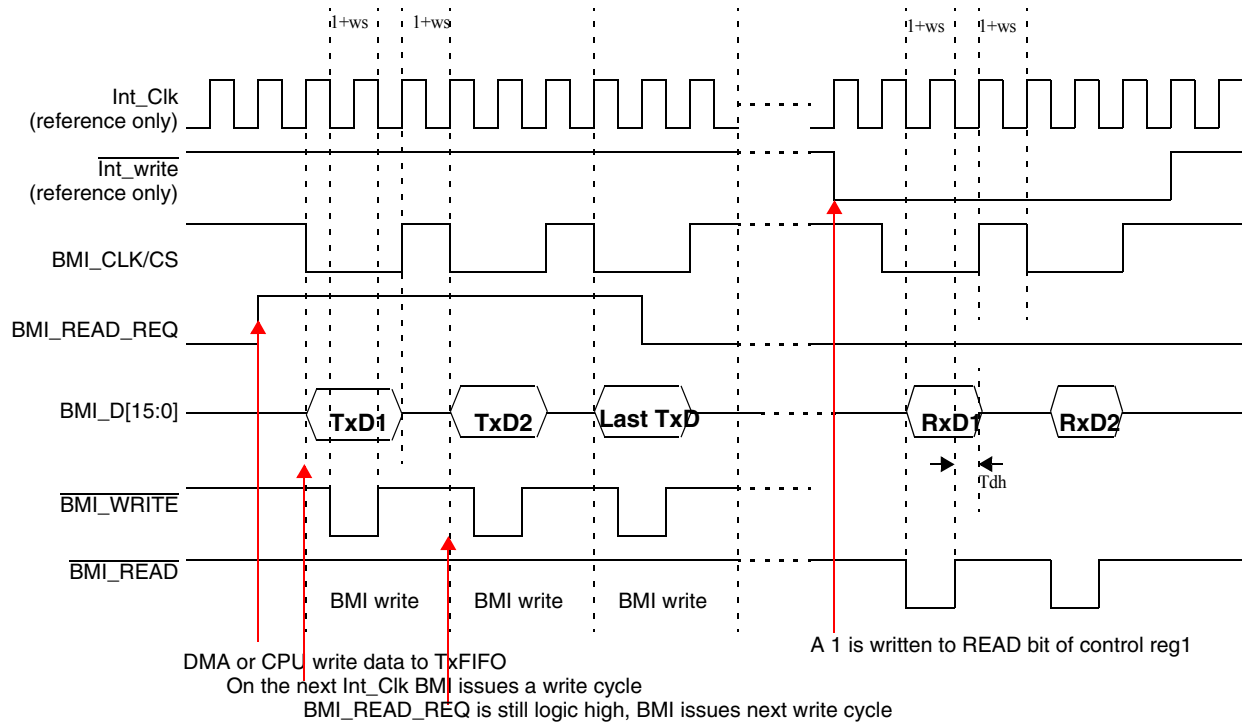
In this mode the  $\overline{\text{BMI\_WRITE}}$ ,  $\overline{\text{BMI\_READ}}$  and BMI\_CLK/CS are output signals driving by the BMI module. The output signal BMI\_READ\_REQ is still driving active-in on a write cycle, but it can be ignored in this case. Instead, it is used to trigger internal logic to generate the read or write signals. Data write cycles are continuously generated when TxFIFO is not emptied.

To issue a read cycle, the user can write a value of 1 to the READ bit of control register. This bit is cleared automatically when the read operation is completed. A read cycle reads COUNT+1 data from the external bus slave. The user can write a 1 to the READ bit while there is still data in the TxFIFO, but the read cycle will not start until all data in the TxFIFO is emptied. If the read cycle begins, the write operation also cannot begin until this read cycle complete.

In this master mode operation, Int\_Clk is derived from HCLK through an integer divider DIV of BMI control register and it is used to control the read/write cycle timing by generate  $\overline{\text{WRITE}}$  and CLK/CS signals.

### 3.8.3.1 Memory Interface Master Mode Without $\overline{\text{WAIT}}$ Signal

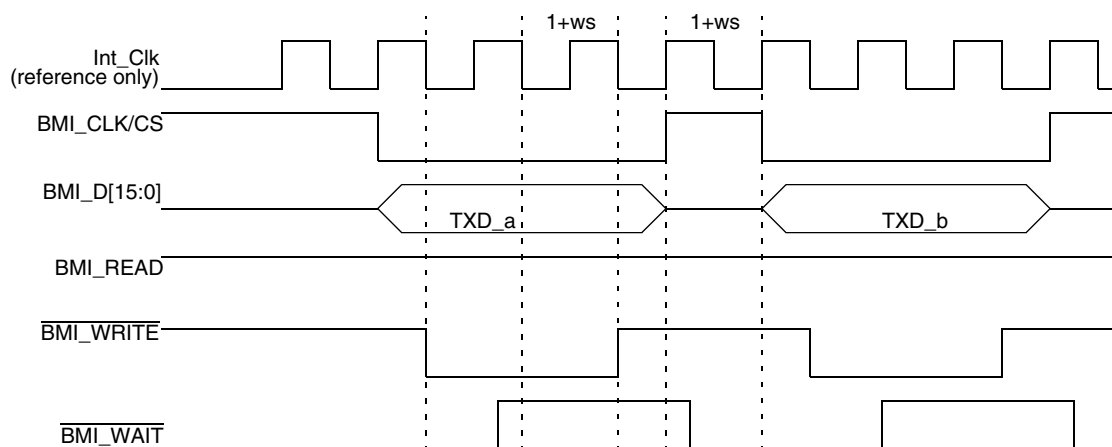
The  $\overline{\text{WAIT}}$  control bit (BMICTLR1[29]) is used in this mode. When this bit is cleared (default), the  $\overline{\text{BMI\_WAIT}}$  signal is ignored and the CS cycle is terminated by Wait State (WS) control bits. Figure 11 shows the BMI timing when the  $\overline{\text{WAIT}}$  bit is cleared.



**Figure 11. Memory Interface Master Mode, BMI Read/Write to External Slave Device Timing without  $\overline{\text{Wait}}$  Signal (MMD\_MODE\_SEL=0, MASTER\_MODE\_SEL=1)**

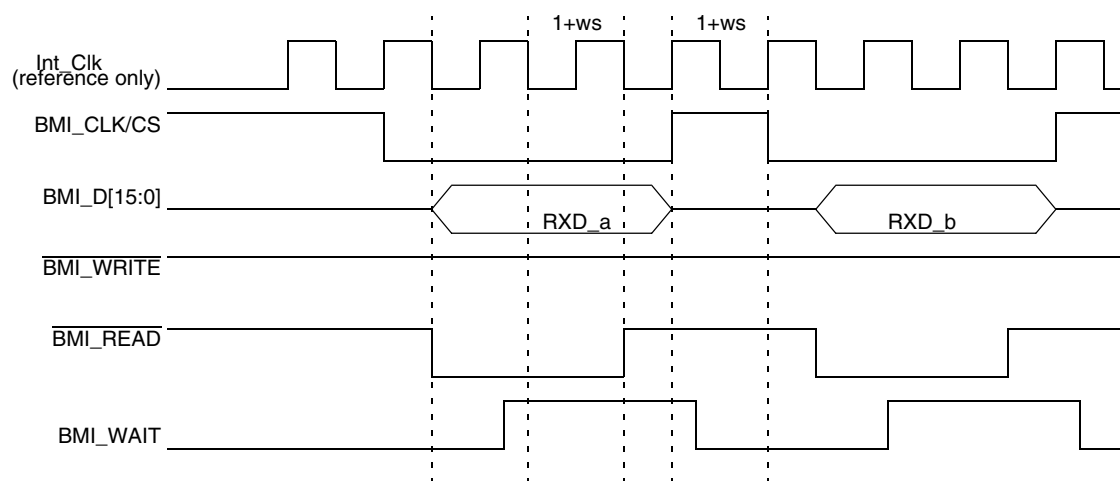
### 3.8.3.2 Memory Interface Master Mode with $\overline{\text{WAIT}}$ Signal

When the  $\overline{\text{WAIT}}$  control bit is set, the  $\overline{\text{BMI\_WAIT}}$  signal is used and the CS cycle is terminated upon sampling a logic high  $\overline{\text{BMI\_WAIT}}$  signal. Figure 12 shows the BMI write timing when the  $\overline{\text{WAIT}}$  bit is set. When the  $\overline{\text{BMI\_WRITE}}$  is asserted, the BMI will detect the  $\overline{\text{BMI\_WAIT}}$  signal on every falling edge of the Int\_Clk. When it detected the high level of the  $\overline{\text{BMI\_WAIT}}$ , the  $\overline{\text{BMI\_WRITE}}$  will be negated after 1+WS Int\_Clk period. If the  $\overline{\text{BMI\_WAIT}}$  is always high or already high before  $\overline{\text{BMI\_WRITE}}$  is asserted, this timing will same as without  $\overline{\text{WAIT}}$  signal. So the  $\overline{\text{BMI\_WRITE}}$  will be asserted at least for 1+WS Int\_Clk period.



**Figure 12. Memory Interface Master Mode, BMI Write to External Slave Device Timing with Wait Signal (MMD\_MODE\_SEL=0, MASTER\_MODE\_SEL=1, WAIT=1)**

Figure 13 shows the BMI read timing when the WAIT bit is set. As write timing, when the BMI\_READ is asserted, the BMI will detect the BMI\_WAIT signal on every falling edge of the Int\_Clk. When it detected the high level of the BMI\_WAIT, the BMI\_READ will be negated after  $1+WS$  Int\_Clk period. If the BMI\_WAIT is always high or already high before BMI\_READ is asserted, this timing will be the same as without WAIT signal. So the BMI\_READ will be asserted at least for  $1+WS$  Int\_Clk period.



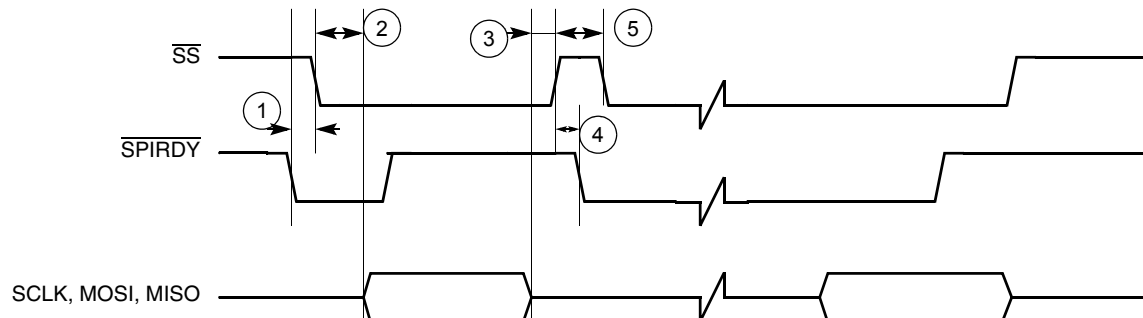
**Figure 13. Memory Interface Master Mode, BMI Read to External Slave Device Timing with Wait Signal (MMD\_MODE\_SEL=0, MASTER\_MODE\_SEL=1, WAIT=1)**

### 3.9 SPI Timing Diagrams

To use the internal transmit (TX) and receive (RX) data FIFOs when the SPI 1 module is configured as a master, two control signals are used for data transfer rate control: the  $\overline{SS}$  signal (output) and the SPI\_RDY signal (input). The SPI 1 Sample Period Control Register (PERIODREG1) and the SPI 2 Sample Period Control Register (PERIODREG2) can also be programmed to a fixed data transfer rate for either SPI 1 or SPI 2. When the SPI 1 module is configured as a slave, the user can configure the SPI 1 Control Register (CONTROLREG1) to match the external SPI master's timing. In this configuration,  $\overline{SS}$  becomes an input

## Specifications

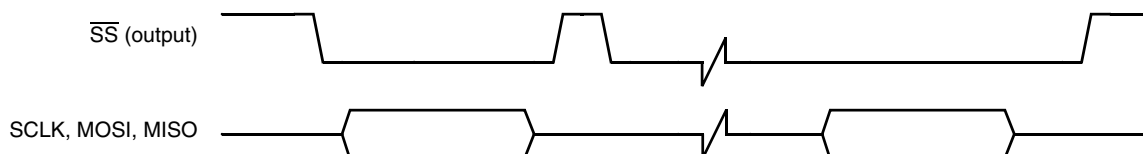
signal, and is used to latch data into or load data out to the internal data shift registers, as well as to increment the data FIFO.



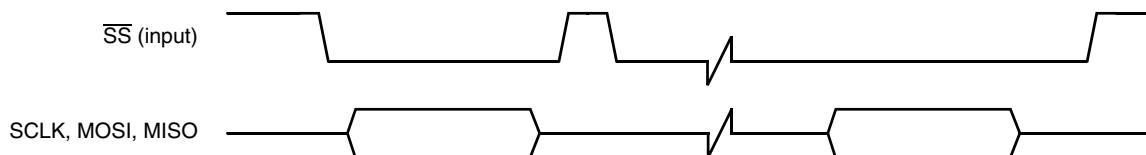
**Figure 14. Master SPI Timing Diagram Using  $\overline{\text{SPI\_RDY}}$  Edge Trigger**



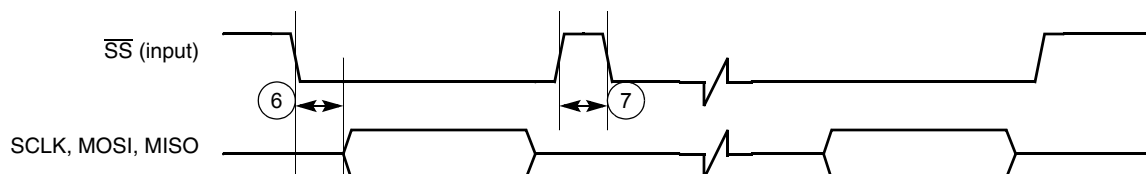
**Figure 15. Master SPI Timing Diagram Using  $\overline{\text{SPI\_RDY}}$  Level Trigger**



**Figure 16. Master SPI Timing Diagram Ignore  $\overline{\text{SPI\_RDY}}$  Level Trigger**



**Figure 17. Slave SPI Timing Diagram FIFO Advanced by BIT COUNT**



**Figure 18. Slave SPI Timing Diagram FIFO Advanced by  $\overline{\text{SS}}$  Rising Edge**

**Table 17. Timing Parameter Table for Figure 14 through Figure 18**

Ref No.	Parameter	Minimum	Maximum	Unit
1	$\overline{\text{SPI\_RDY}}$ to $\overline{\text{SS}}$ output low	$2T^1$	—	ns
2	$\overline{\text{SS}}$ output low to first SCLK edge	$3 \cdot T_{\text{sclk}}^2$	—	ns
3	Last SCLK edge to $\overline{\text{SS}}$ output high	$2 \cdot T_{\text{sclk}}$	—	ns
4	$\overline{\text{SS}}$ output high to $\overline{\text{SPI\_RDY}}$ low	0	—	ns
5	$\overline{\text{SS}}$ output pulse width	$T_{\text{sclk}} + \text{WAIT}^3$	—	ns
6	$\overline{\text{SS}}$ input low to first SCLK edge	T	—	ns
7	$\overline{\text{SS}}$ input pulse width	T	—	ns

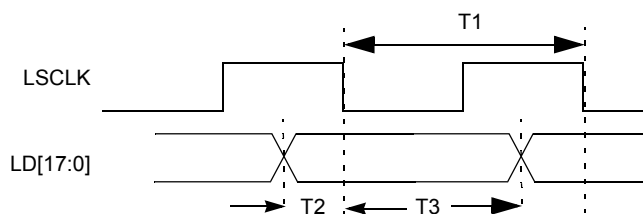
1. T = CSPI system clock period (PERCLK2).

2.  $T_{\text{sclk}}$  = Period of SCLK.

3. WAIT = Number of bit clocks (SCLK) or 32.768 KHz clocks per Sample Period Control Register.

### 3.10 LCD Controller

This section includes timing diagrams for the LCD controller. For detailed timing diagrams of the LCD controller with various display configurations, refer to the LCD controller chapter of the *i.MX21 Reference Manual*.

**Figure 19. SCLK to LD Timing Diagram****Table 18. LCDC SCLK Timing Parameter Table**

Symbol	Parameter	3.0 +/- 0.3V		Unit
		Minimum	Maximum	
T1	SCLK period	23	2000	ns
T2	Pixel data setup time	11	—	ns
T3	Pixel data up time	11	—	ns

The pixel clock is equal to  $\text{LCDC\_CLK} / (\text{PCD} + 1)$ .

When it is in CSTN, TFT or monochrome mode with bus width = 1, SCLK is equal to the pixel clock.

When it is in monochrome with other bus width settings, SCLK is equal to the pixel clock divided by bus width.

The polarity of SCLK and LD can also be programmed.

Maximum frequency of SCLK is  $\text{HCLK} / 3$  for TFT and CSTN, otherwise LD output will be incorrect.

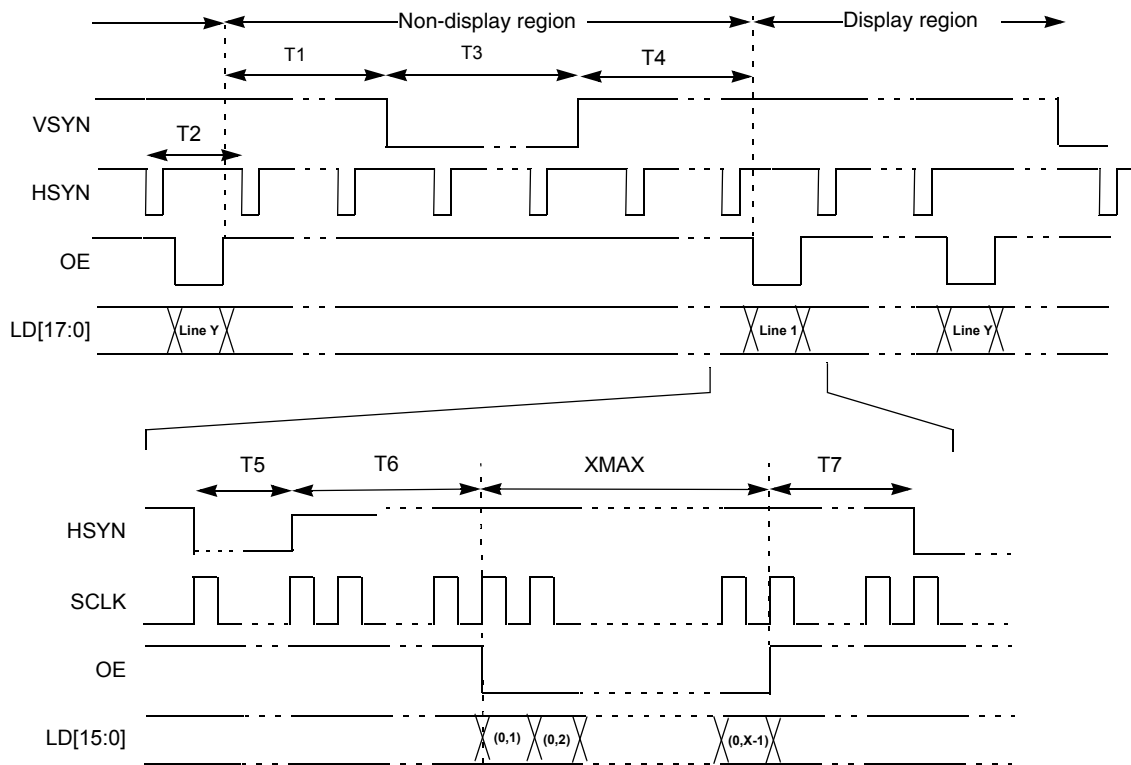


Figure 20. 4/8/12/16/18 Bit/Pixel TFT Color Mode Panel Timing

Table 19. 4/8/12/16/18 Bit/Pixel TFT Color Mode Panel Timing

Symbol	Description	Minimum	Value	Unit
T1	End of OE to beginning of VSYN	$T5+T6+T7-1$	$(VWAIT1 \cdot T2)+T5+T6+T7-1$	$T_s$
T2	HSYN period	—	$XMAX+T5+T6+T7$	$T_s$
T3	VSYN pulse width	$T2$	$VWIDTH \cdot T2$	$T_s$
T4	End of VSYN to beginning of OE	1	$(VWAIT2 \cdot T2)+1$	$T_s$
T5	HSYN pulse width	1	$HWIDTH+1$	$T_s$
T6	End of HSYN to beginning to OE	3	$HWAIT2+3$	$T_s$
T7	End of OE to beginning of HSYN	1	$HWAIT1+1$	$T_s$

**Note:**

- $T_s$  is the SCLK period.
- VSYN, HSYN and OE can be programmed as active high or active low. In Figure 20, all 3 signals are active low.
- SCLK can be programmed to be deactivated during the VSYN pulse or the OE deasserted period. In Figure 20, SCLK is always active.
- XMAX is defined in number of pixels in one line.



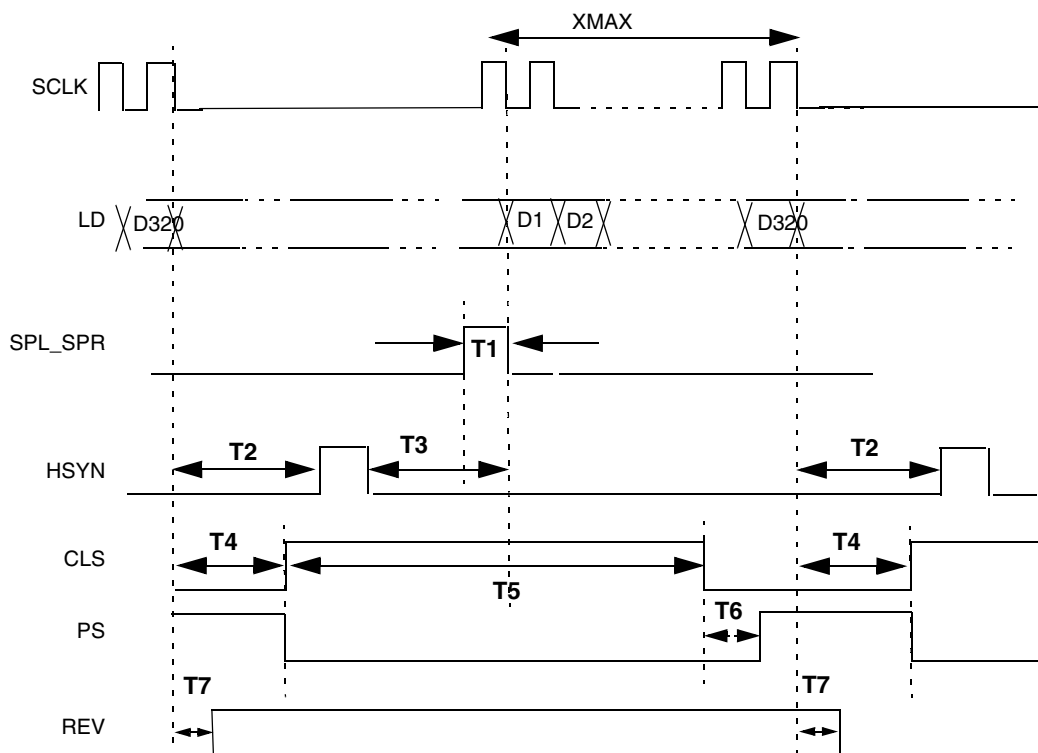


Figure 21. Sharp TFT Panel Timing

Table 20. Sharp TFT Panel Timing

Symbol	Description	Minimum	Value	Unit
T1	SPL/SPR pulse width	—	1	Ts
T2	End of LD of line to beginning of HSYN	1	HWAIT1+1	Ts
T3	End of HSYN to beginning of LD of line	4	HWAIT2 + 4	Ts
T4	CLS rise delay from end of LD of line	3	CLS_RISE_DELAY+1	Ts
T5	CLS pulse width	1	CLS_HI_WIDTH+1	Ts
T6	PS rise delay from CLS negation	0	PS_RISE_DELAY	Ts
T7	REV toggle delay from last LD of line	1	REV_TOGGLE_DELAY+1	Ts

**Note:**

- Falling of SPL/SPR aligns with first LD of line.
- Falling of PS aligns with rising edge of CLS.
- REV toggles in every HSYN period.

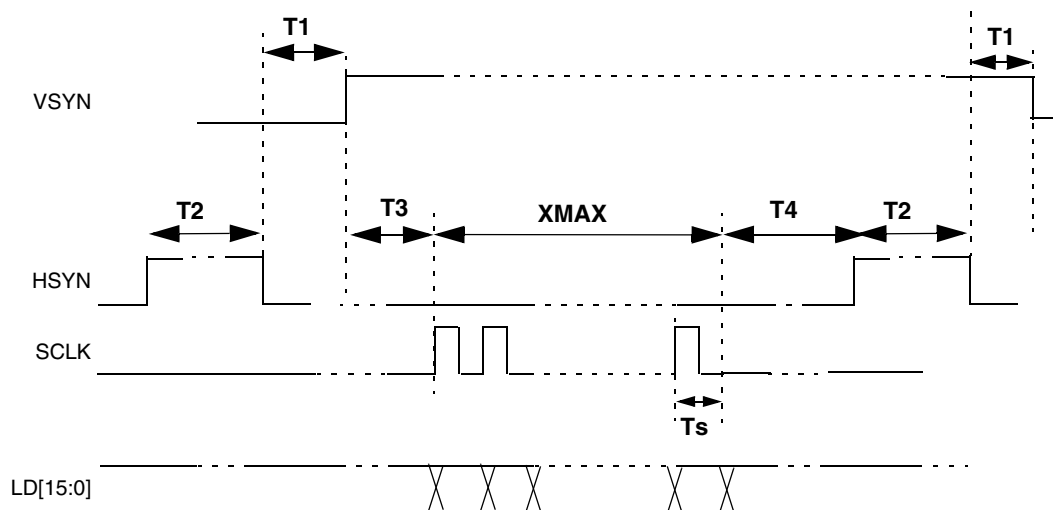


Figure 22. Non-TFT Mode Panel Timing

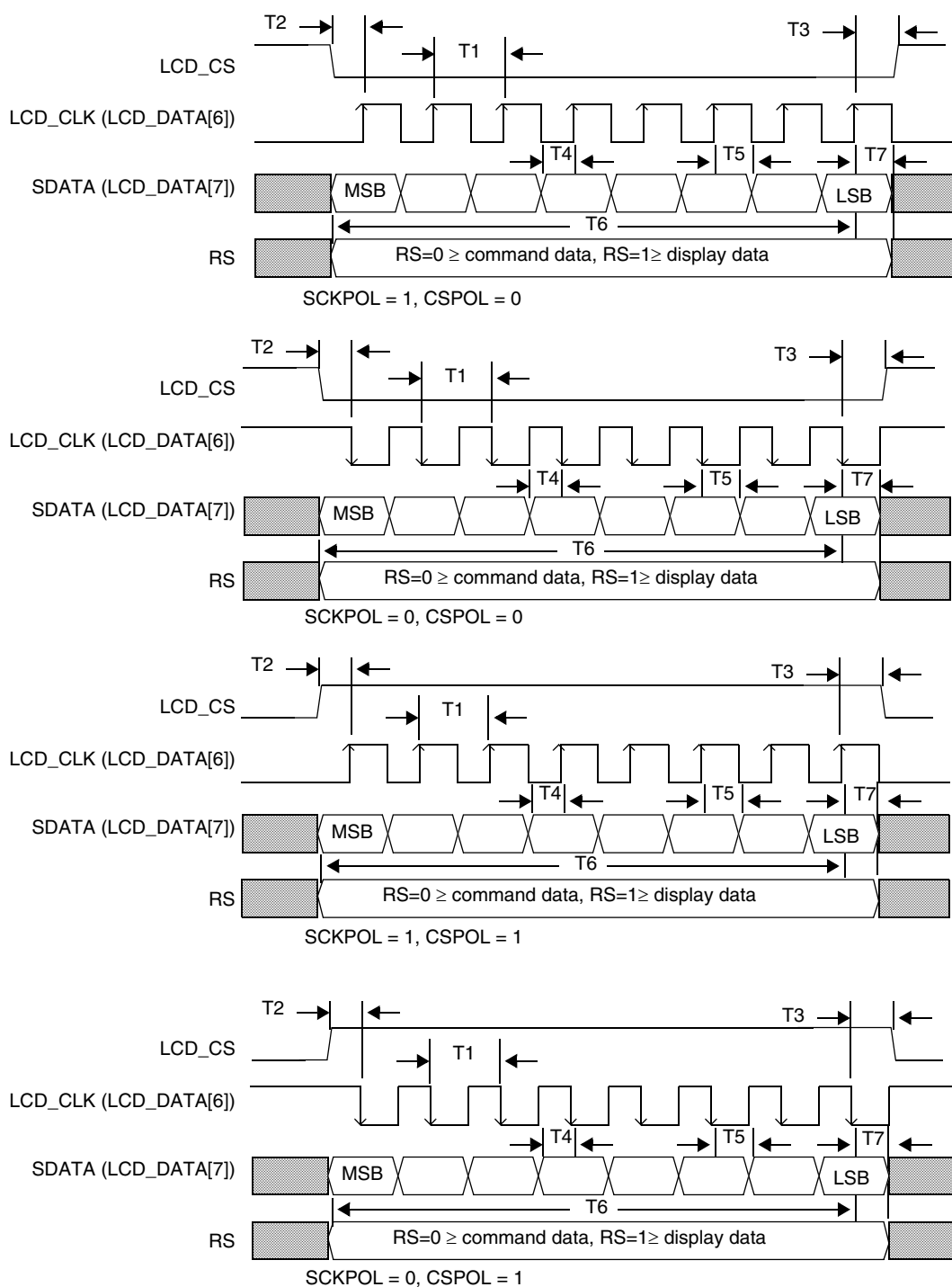
Table 21. Non-TFT Mode Panel Timing

Symbol	Description	Minimum	Value	Unit
T1	HSYN to VSYN delay	2	$H_{WAIT2}+2$	Tpix
T2	HSYN pulse width	1	$H_{WIDTH}+1$	Tpix
T3	VSYN to SCLK	–	$0 \leq T_3 \leq T_s$	–
T4	SCLK to HSYN	1	$H_{WAIT1}+1$	Tpix

**Note:**

- $T_s$  is the SCLK period while Tpix is the pixel clock period.
- VSYN, HSYN and SCLK can be programmed as active high or active low. In Figure 67 on page 83, all these 3 signals are active high.
- When it is in CSTN mode or monochrome mode with bus width = 1,  $T_3 = T_{pix} = T_s$ .
- When it is in monochrome mode with bus width = 2, 4, and 8,  $T_3 = 1, 2$  and  $4$  Tpix respectively.

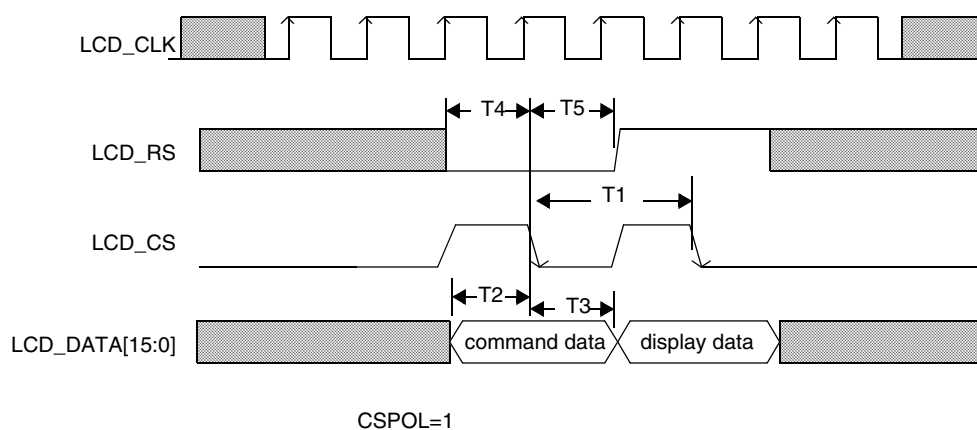
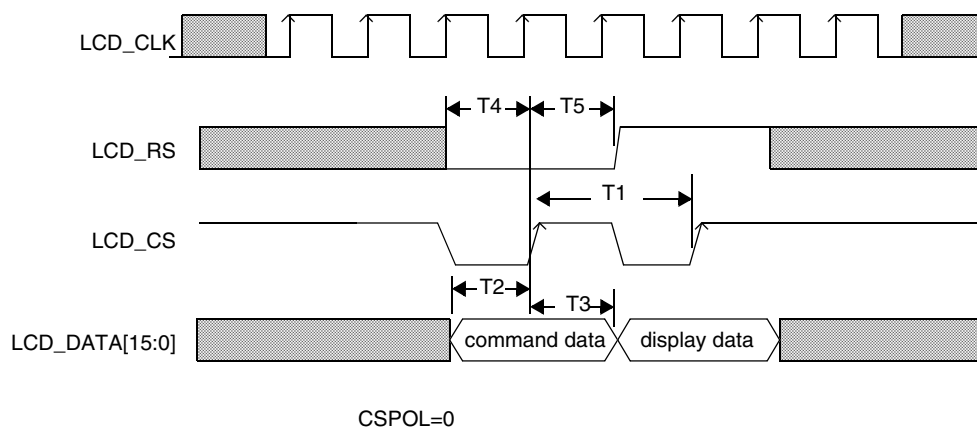
### 3.11 Smart LCD Controller



**Figure 23. SLCDC Serial Transfer Timing**

**Table 22. SLCDC Serial Transfer Timing**

Symbol	Description	Minimum	Maximum	Unit
T1	Pixel clock period	42	962	ns
T2	Chip select setup time	5	–	ns
T3	Chip select hold time	5	–	ns
T4	Data setup time	5	–	ns
T4	Data hold time	5	–	ns
T6	Register select setup time	5	–	ns
T7	Register select hold time	5	–	ns

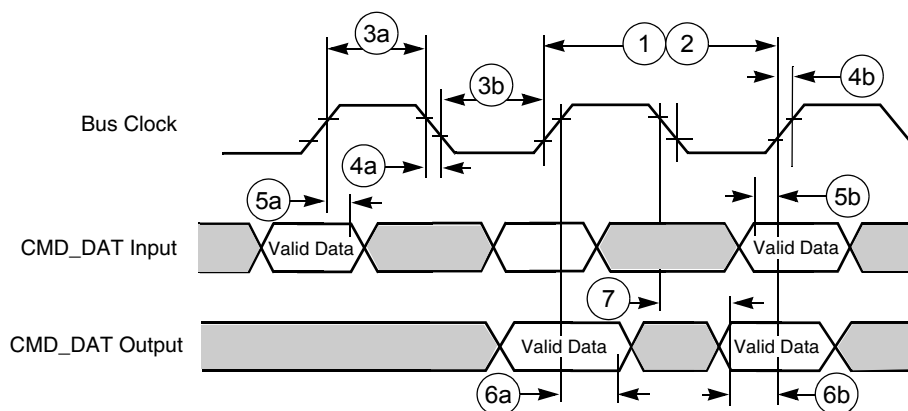
**Figure 24. SLCDC Parallel Transfers Timing**

**Table 23. SLCDC Parallel Transfers Timing**

Symbol	Description	Minimum	Maximum	Unit
T1	Pixel clock period	23	962	ns
T2	Data setup time	5	—	ns
T3	Data hold time	5	—	ns
T4	Register select setup time	5	—	ns
T5	Register select hold time	5	—	ns

### 3.12 Multimedia Card/Secure Digital Host Controller

The DMA interface block controls all data routing between the external data bus (DMA access), internal MMC/SD module data bus, and internal system FIFO access through a dedicated state machine that monitors the status of FIFO content (empty or full), FIFO address, and byte/block counters for the MMC/SD module (inner system) and the application (user programming).

**Figure 25. Chip-Select Read Cycle Timing Diagram****Table 24. SDHC Bus Timing Parameter Table**

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Min	Max	Min	Max	
1	CLK frequency at Data transfer Mode (PP) <sup>1</sup> —10/30 cards	0	25/5	0	25/5	MHz
2	CLK frequency at Identification Mode <sup>2</sup>	0	400	0	400	KHz
3a	Clock high time <sup>1</sup> —10/30 cards	6/33	—	10/50	—	ns
3b	Clock low time <sup>1</sup> —10/30 cards	15/75	—	10/50	—	ns
4a	Clock fall time <sup>1</sup> —10/30 cards	—	10/50 (5.00) <sup>3</sup>	—	10/50	ns
4b	Clock rise time <sup>1</sup> —10/30 cards	—	14/67 (6.67) <sup>3</sup>	—	10/50	ns

Table 24. SDHC Bus Timing Parameter Table (Continued)

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Min	Max	Min	Max	
5a	Input hold time <sup>3</sup> —10/30 cards	5.7/5.7	—	5/5	—	ns
5b	Input setup time <sup>3</sup> —10/30 cards	5.7/5.7	—	5/5	—	ns
6a	Output hold time <sup>3</sup> —10/30 cards	5.7/5.7	—	5/5	—	ns
6b	Output setup time <sup>3</sup> —10/30 cards	5.7/5.7	—	5/5	—	ns
7	Output delay time <sup>3</sup>	0	16	0	14	ns

1.  $C_L \leq 100 \text{ pF}$  /  $250 \text{ pF}$  (10/30 cards)

2.  $C_L \leq 250 \text{ pF}$  (21 cards)

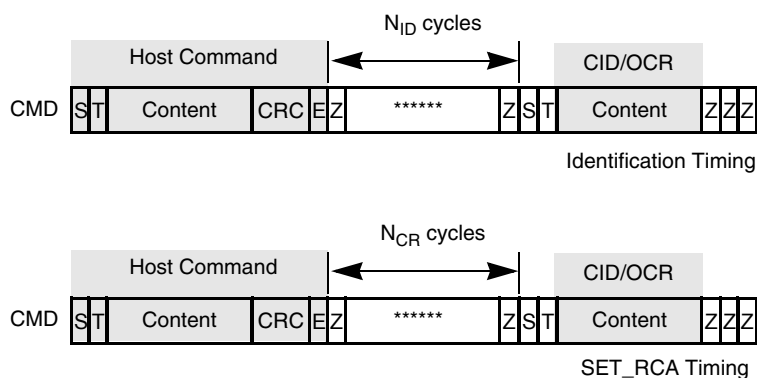
3.  $C_L \leq 25 \text{ pF}$  (1 card)

### 3.12.1 Command Response Timing on MMC/SD Bus

The card identification and card operation conditions timing are processed in open-drain mode. The card response to the host command starts after exactly  $N_{ID}$  clock cycles. For the card address assignment, SET\_RCA is also processed in the open-drain mode. The minimum delay between the host command and card response is NCR clock cycles as illustrated in Figure 26. The symbols for Figure 26 through Figure 30 are defined in Table 25.

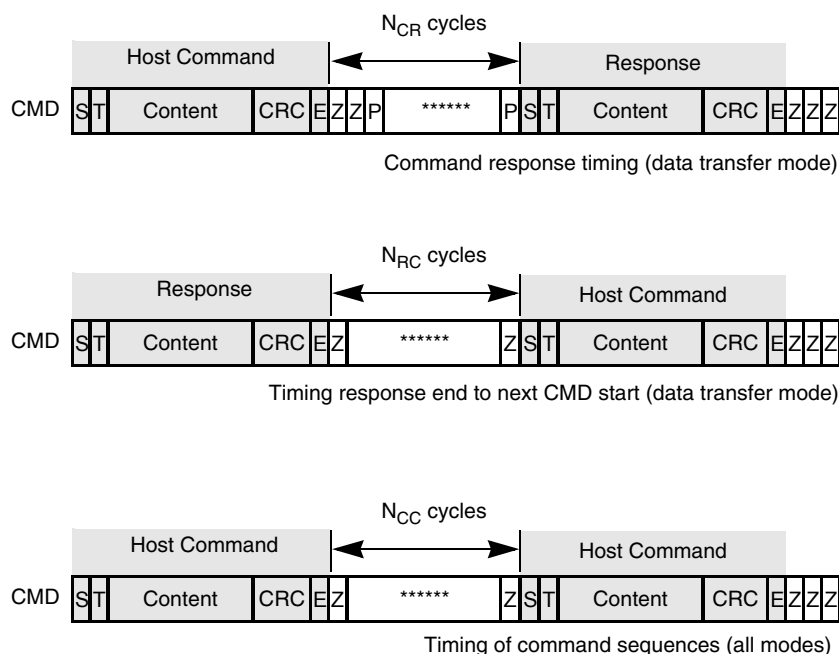
Table 25. State Signal Parameters for Figure 26 through Figure 30

Card Active		Host Active	
Symbol	Definition	Symbol	Definition
Z	High impedance state	S	Start bit (0)
D	Data bits	T	Transmitter bit (Host = 1, Card = 0)
*	Repetition	P	One-cycle pull-up (1)
CRC	Cyclic redundancy check bits (7 bits)	E	End bit (1)



**Figure 26. Timing Diagrams at Identification Mode**

After a card receives its RCA, it switches to data transfer mode. As shown on the first diagram in Figure 27 on page 39, SD\_CMD lines in this mode are driven with push-pull drivers. The command is followed by a period of two Z bits (allowing time for direction switching on the bus) and then by P bits pushed up by the responding card. The other two diagrams show the separating periods  $N_{RC}$  and  $N_{CC}$ .



**Figure 27. Timing Diagrams at Data Transfer Mode**

Figure 28 on page 40 shows basic read operation timing. In a read operation, the sequence starts with a single block read command (which specifies the start address in the argument field). The response is sent on the SD\_CMD lines as usual. Data transmission from the card starts after the access time delay  $N_{AC}$ , beginning from the last bit of the read command. If the system is in multiple block read mode, the card sends a continuous flow of data blocks with distance  $N_{AC}$  until the card sees a stop transmission command. The data stops two clock cycles after the end bit of the stop command.

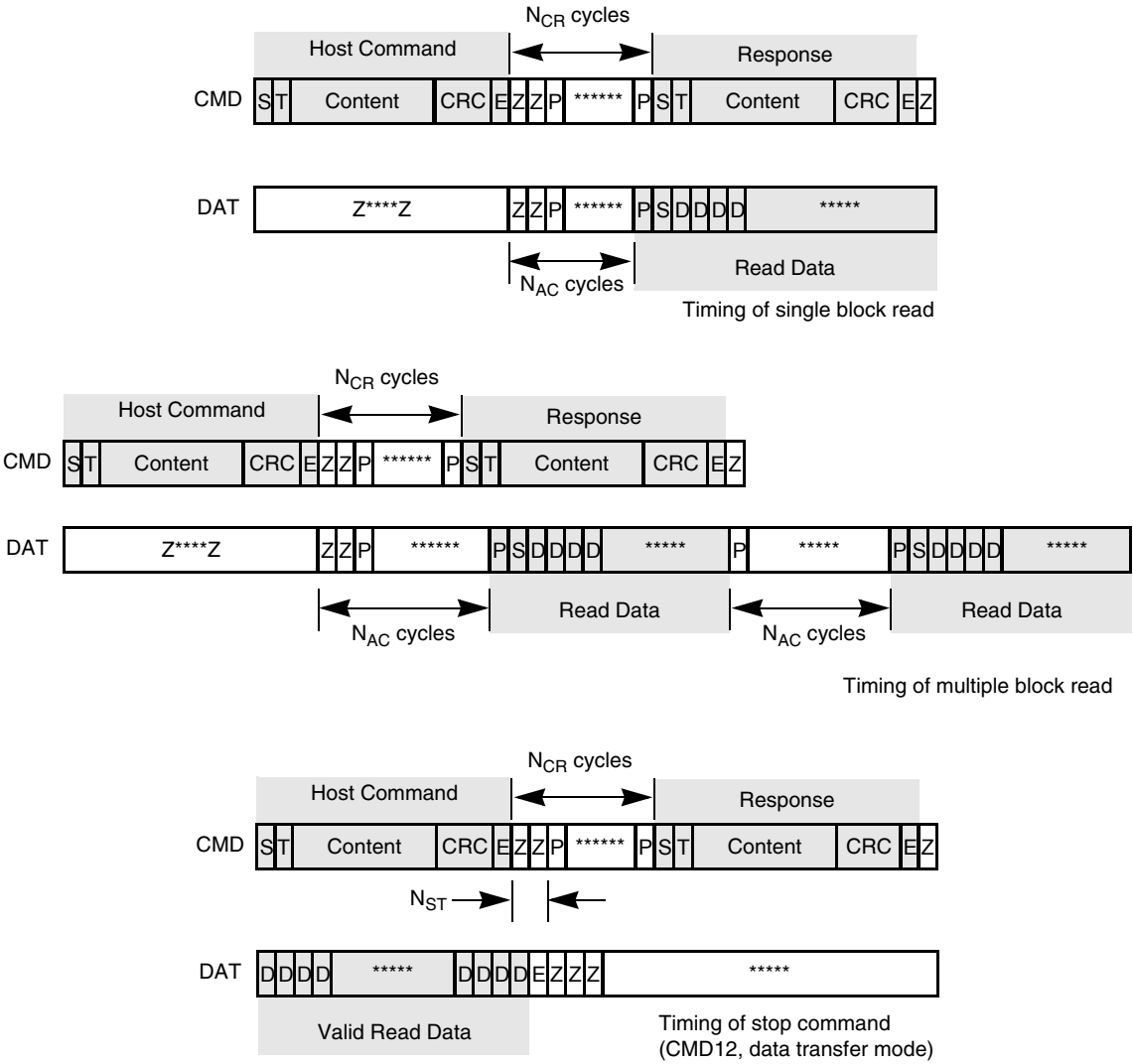


Figure 28. Timing Diagrams at Data Read

Figure 29 on page 41 shows the basic write operation timing. As with the read operation, after the card response, the data transfer starts after  $N_{WR}$  cycles. The data is suffixed with CRC check bits to allow the card to check for transmission errors. The card sends back the CRC check result as a CC status token on the data line. If there was a transmission error, the card sends a negative CRC status (101); otherwise, a positive CRC status (010) is returned. The card expects a continuous flow of data blocks if it is configured to multiple block mode, with the flow terminated by a stop transmission command.





Freescale Semiconductor

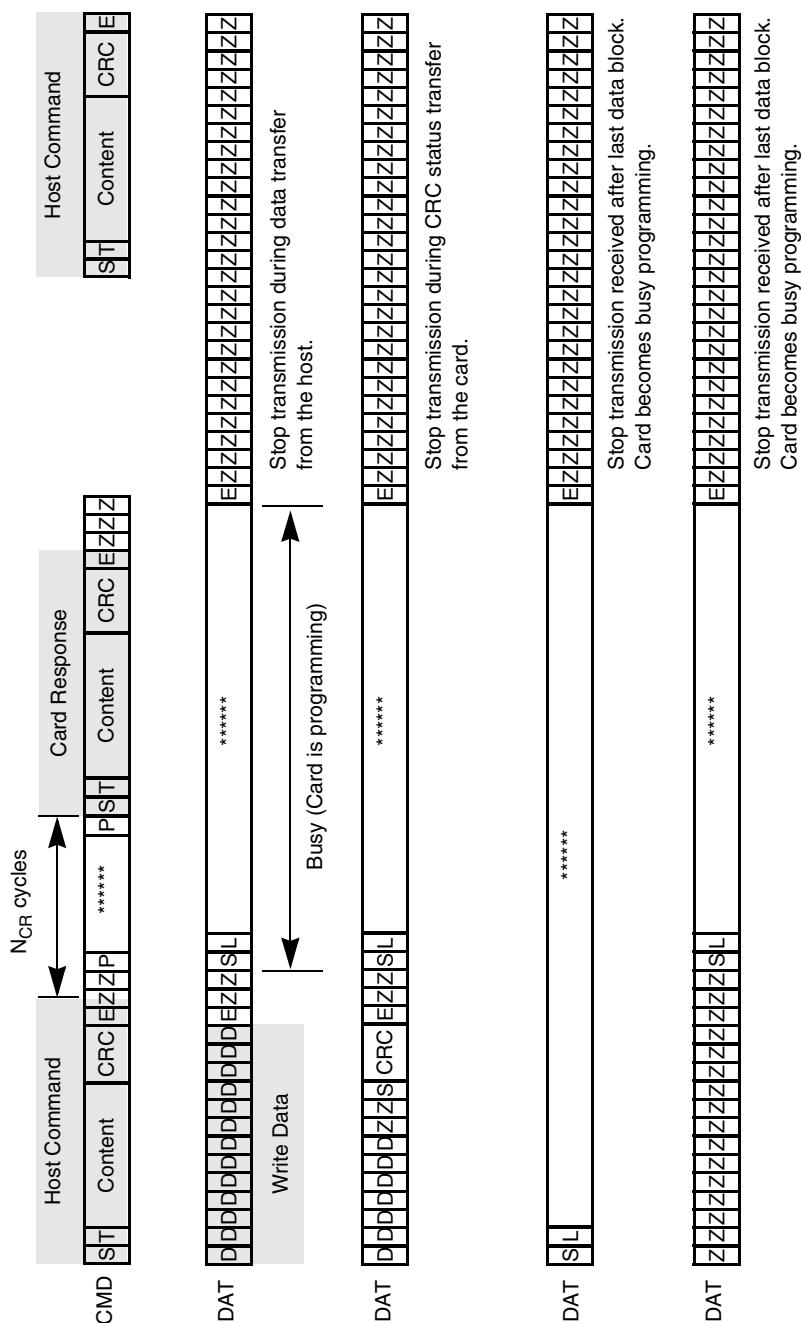


Figure 30. Stop Transmission During Different Scenarios

Table 26. Timing Values for Figure 26 through Figure 30

Parameter	Symbol	Minimum	Maximum	Unit
MMC/SD bus clock, CLK (All values are referred to minimum (VIH) and maximum (VIL))				
Command response cycle	NCR	2	64	Clock cycles
Identification response cycle	NID	5	5	Clock cycles

Table 26. Timing Values for Figure 26 through Figure 30 (Continued)

Parameter	Symbol	Minimum	Maximum	Unit
Access time delay cycle	NAC	2	TAAC + NSAC	Clock cycles
Command read cycle	NRC	8	–	Clock cycles
Command-command cycle	NCC	8	–	Clock cycles
Command write cycle	NWR	2	–	Clock cycles
Stop transmission cycle	NST	2	2	Clock cycles
TAAC: Data read access time -1 defined in CSD register bit[119:112] NSAC: Data read access time -2 in CLK cycles (NSAC-100) defined in CSD register bit[111:104]				

### 3.12.2 SDIO-IRQ and ReadWait Service Handling

In SDIO, there is a 1-bit or 4-bit interrupt response from the SDIO peripheral card. In 1-bit mode, the interrupt response is simply that the SD\_DAT[1] line is held low. The SD\_DAT[1] line is not used as data in this mode. The memory controller generates an interrupt according to this low and the system interrupt continues until the source is removed (SD\_DAT[1] returns to its high level).

In 4-bit mode, the interrupt is less simple. The interrupt triggers at a particular period called the *Interrupt Period* during the data access, and the controller must sample SD\_DAT[1] during this short period to determine the IRQ status of the attached card. The interrupt period only happens at the boundary of each block (512 bytes).

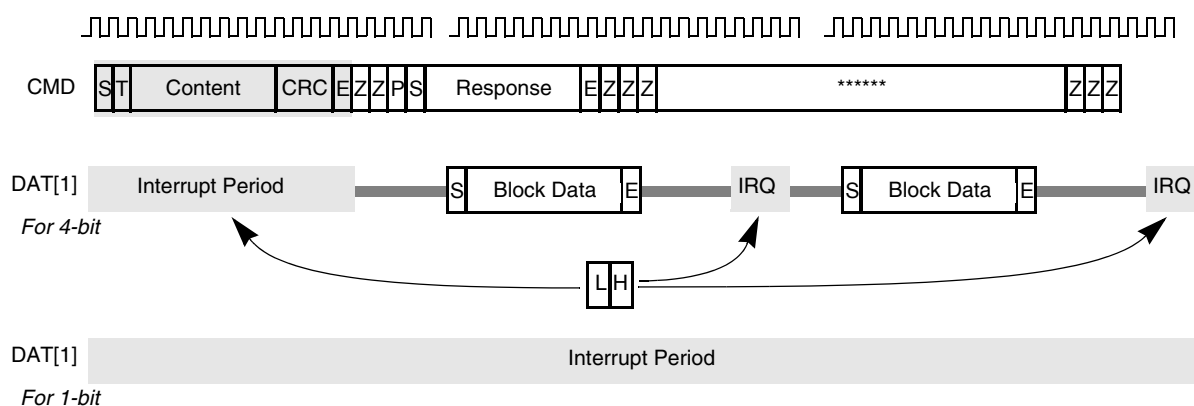
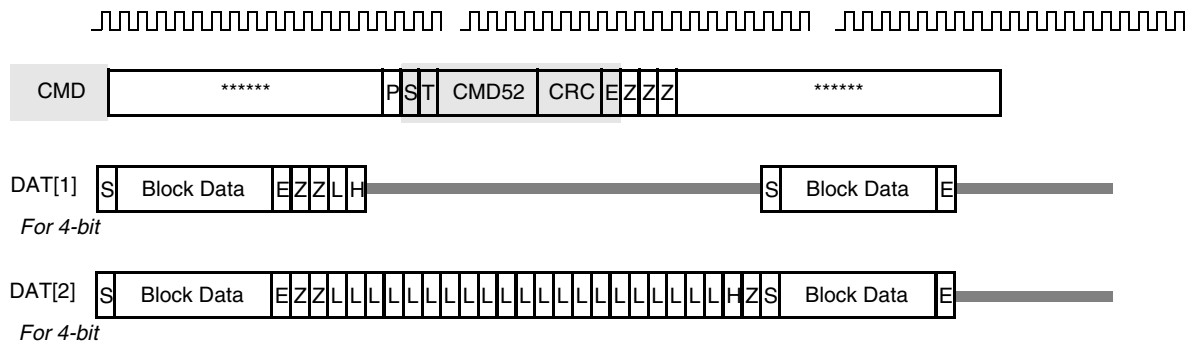


Figure 31. SDIO IRQ Timing Diagram

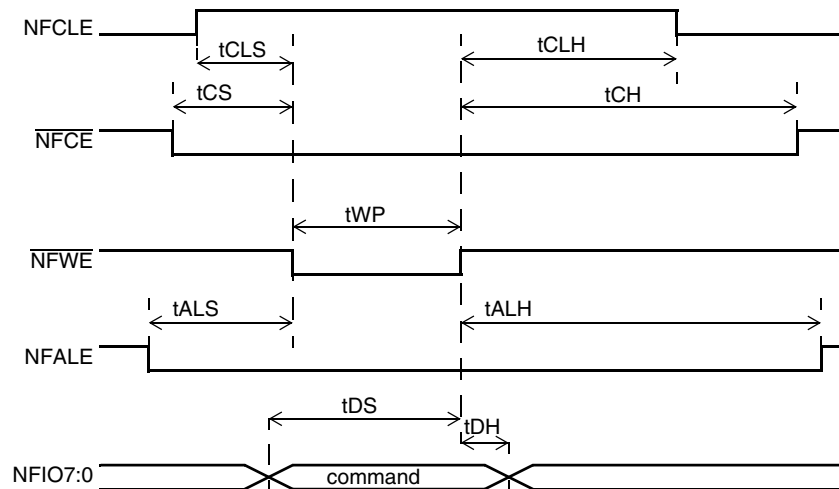
ReadWait is another feature in SDIO that allows the user to submit commands during the data transfer. In this mode, the block temporarily pauses the data transfer operation counter and related status, yet keeps the clock running, and allows the user to submit commands as normal. After all commands are submitted, the user can switch back to the data transfer operation and all counter and status values are resumed as access continues.



### Figure 32. SDIO ReadWait Timing Diagram

### 3.13 NAND-Flash Controller Interface

The timing diagrams Figure 33 through Figure 36 shows the timing of the NAND Flash controller. Table 27 on page 46 provides the relative timing requirement for the different signals of NFC at the i.MX21 module level.



### Figure 33. Command Latch Cycle Timing

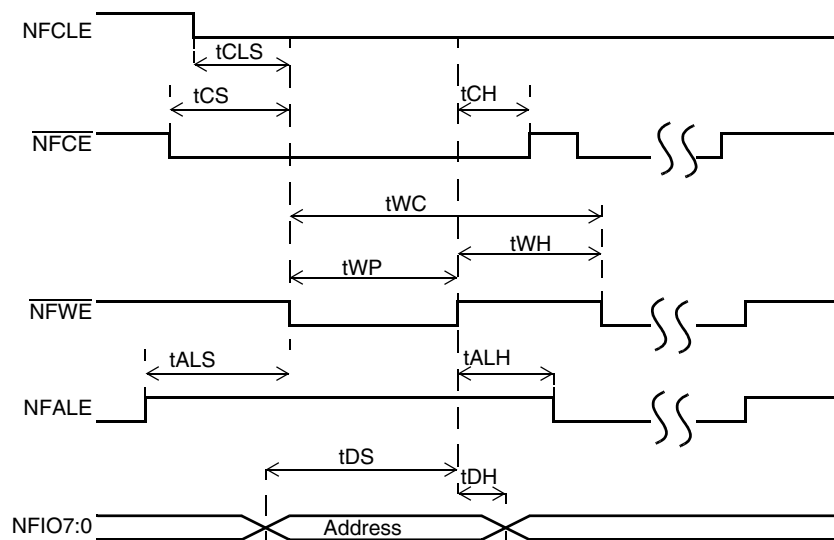


Figure 34. Address Latch Cycle Timing

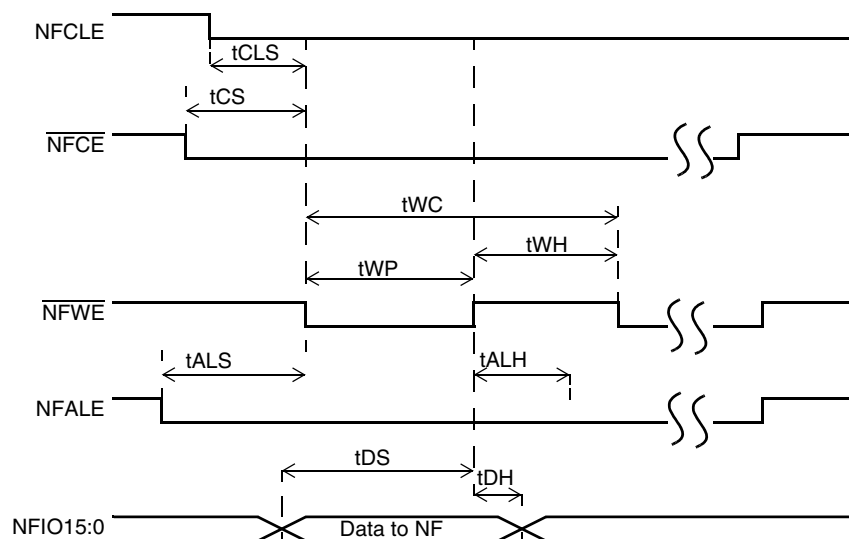
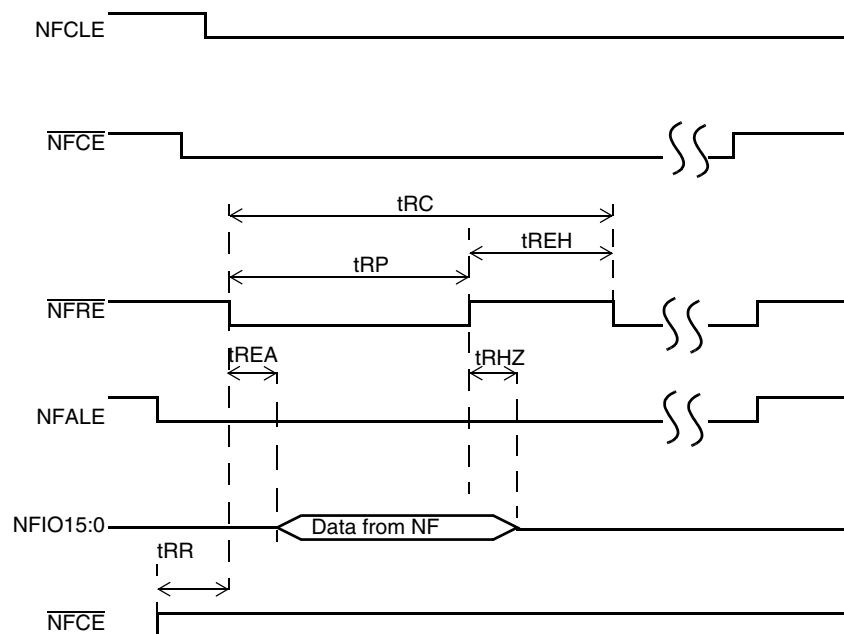


Figure 35. Input Data Latch Cycle Timing



**Figure 36. Output Data Latch Cycle Timing**

**Note:** The data shown in Figure 36 is generated using the NAND Flash device and sampled with `IPP_FLASH_CLK`.

**Table 27. Timing Characteristics**

Number	Timing Parameter	Minimum	Maximum
1	$t_{CLS}$	0	—
2	$t_{CLH}$	10	—
3	$t_{CS}$	0	—
4	$t_{CH}$	10	—
5	$t_{WP}$	25	—
6	$t_{ALS}$	0	—
7	$t_{ALH}$	10	—
8	$t_{DS}$	20	—
9	$t_{DH}$	10	—
10	$t_{WC}$	45	—
11	$t_{WH}$	15	—
12	$t_{AR}$	10	—
13	$t_{CLR}$	10	—
14	$t_{RR}$	20	—
15	$t_{RP}$	25	—
16	$t_{WB}$	—	100

Table 27. Timing Characteristics (Continued)

Number	Timing Parameter	Minimum	Maximum
17	t <sub>RC</sub>	50	–
18	t <sub>CEA</sub>	–	45
19	t <sub>REA</sub>	–	30
20	t <sub>RHZ</sub>	–	30
21	t <sub>CHZ</sub>	–	20
22	t <sub>OH</sub>	15	–
23	t <sub>REH</sub>	15	–
24	t <sub>IR</sub>	0	–
25	t <sub>WHR</sub>	60	–

### 3.14 Pulse-Width Modulator

The PWM can be programmed to select one of two clock signals as its source frequency. The selected clock signal is passed through a divider and a prescaler before being input to the counter. The output is available at the pulse-width modulator output (PWMO) external pin.

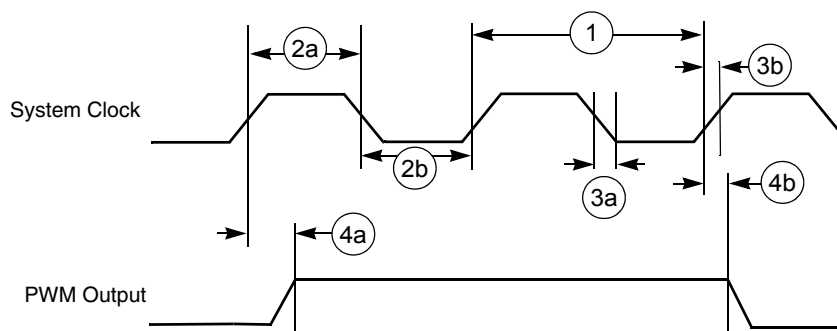


Figure 37. PWM Output Timing Diagram

Table 28. PWM Output Timing Parameter Table

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Minimum	Maximum	Minimum	Maximum	
1	System CLK frequency <sup>1</sup>	0	45	0	45	MHz
2a	Clock high time <sup>1</sup>	12.29	–	12.29	–	ns
2b	Clock low time <sup>1</sup>	9.91	–	9.91	–	ns
3a	Clock fall time <sup>1</sup>	–	0.5	–	0.5	ns
3b	Clock rise time <sup>1</sup>	–	0.5	–	0.5	ns

Table 28. PWM Output Timing Parameter Table (Continued)

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Minimum	Maximum	Minimum	Maximum	
4a	Output delay time <sup>1</sup>	9.37	—	3.61	—	ns
4b	Output setup time <sup>1</sup>	8.71	—	3.03	—	ns

1.  $C_L$  of PWMO = 30 pF

### 3.15 SDRAM Memory Controller

The following figures (Figure 38 through Figure 41 on page 52) and their associated tables specify the timings related to the SDRAMC module in the i.MX21.

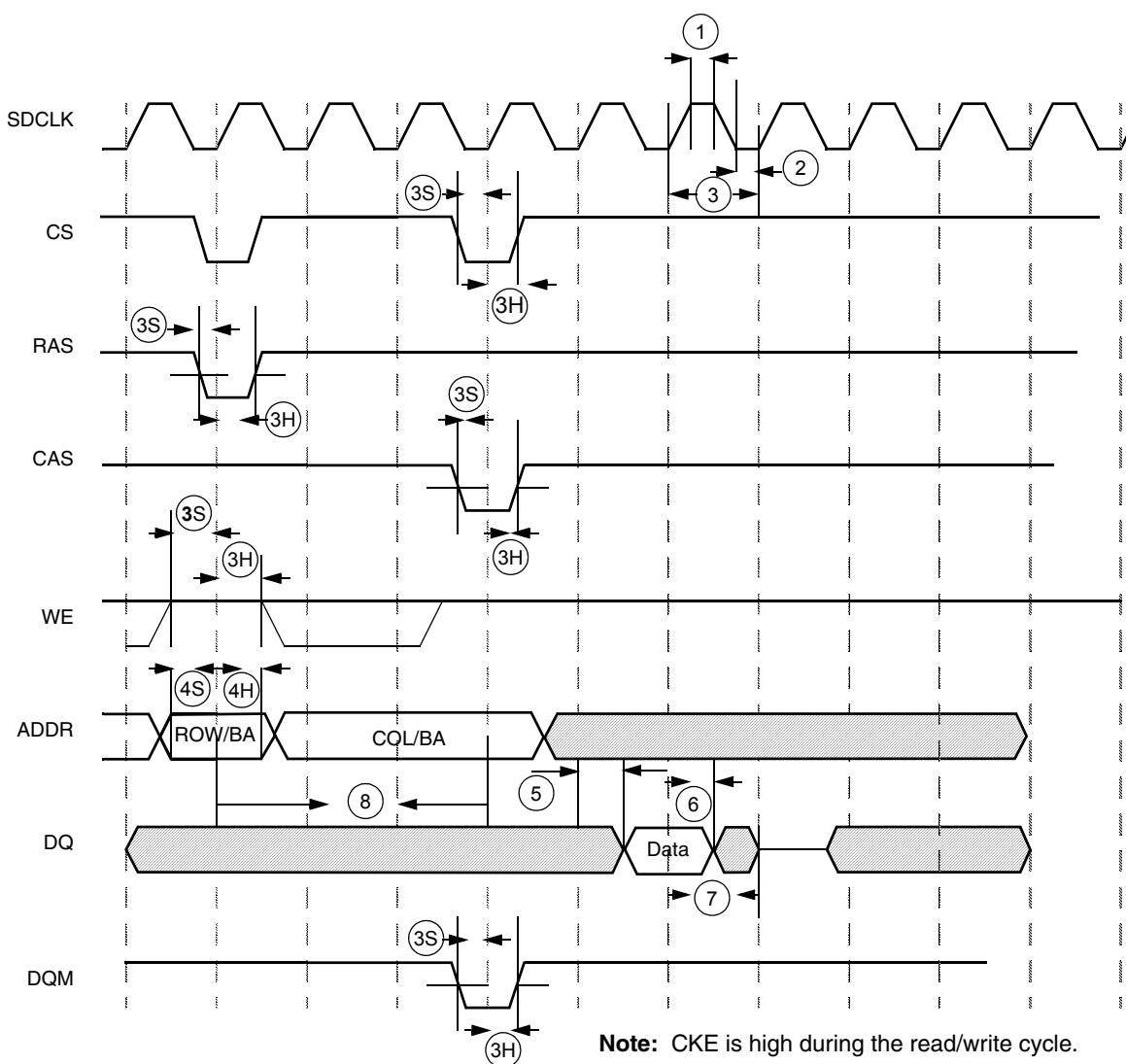


Figure 38. SDRAM Read Cycle Timing Diagram



Table 29. SDRAM Timing Parameter Table

Ref No.	Parameter	1.8V		3.0V +/-10%		Unit
		Minimum	Maximum	Minimum	Maximum	
1	SDRAM clock high-level width	3.00	—	4	—	ns
2	SDRAM clock low-level width	3.00	—	4	—	ns
3	SDRAM clock cycle time	11.1	—	7.5	—	ns
3S	CS, RAS, CAS, WE, DQM setup time	4.78	—	3	—	ns
3H	CS, RAS, CAS, WE, DQM hold time	3.03	—	2	—	ns
4S	Address setup time	3.67	—	3	—	ns
4H	Address hold time	2.95	—	2	—	ns
5	SDRAM access time (CL = 3)	—	5.4	—	5.4	ns
5	SDRAM access time (CL = 2)	—	6.0	—	6.0	ns
5	SDRAM access time (CL = 1)	—	—	—	—	ns
6	Data out hold time	3.0	—	3.0	—	ns
7	Data out high-impedance time (CL = 3)	—	$t_{HZ}^1$	—	$t_{HZ}^1$	ns
7	Data out high-impedance time (CL = 2)	—	$t_{HZ}^1$	—	$t_{HZ}^1$	ns
7	Data out high-impedance time (CL = 1)	—	—	—	—	ns
8	Active to read/write command period (RC = 1)	$t_{RCD}^2$	—	$t_{RCD}^2$	—	ns

1.  $t_{HZ}$  = SDRAM data out high-impedance time, external SDRAM memory device dependent parameter.

2.  $t_{RCD}$  = SDRAM clock cycle time. The  $t_{RCD}$  setting can be found in the i.MX21 reference manual.

## Specifications

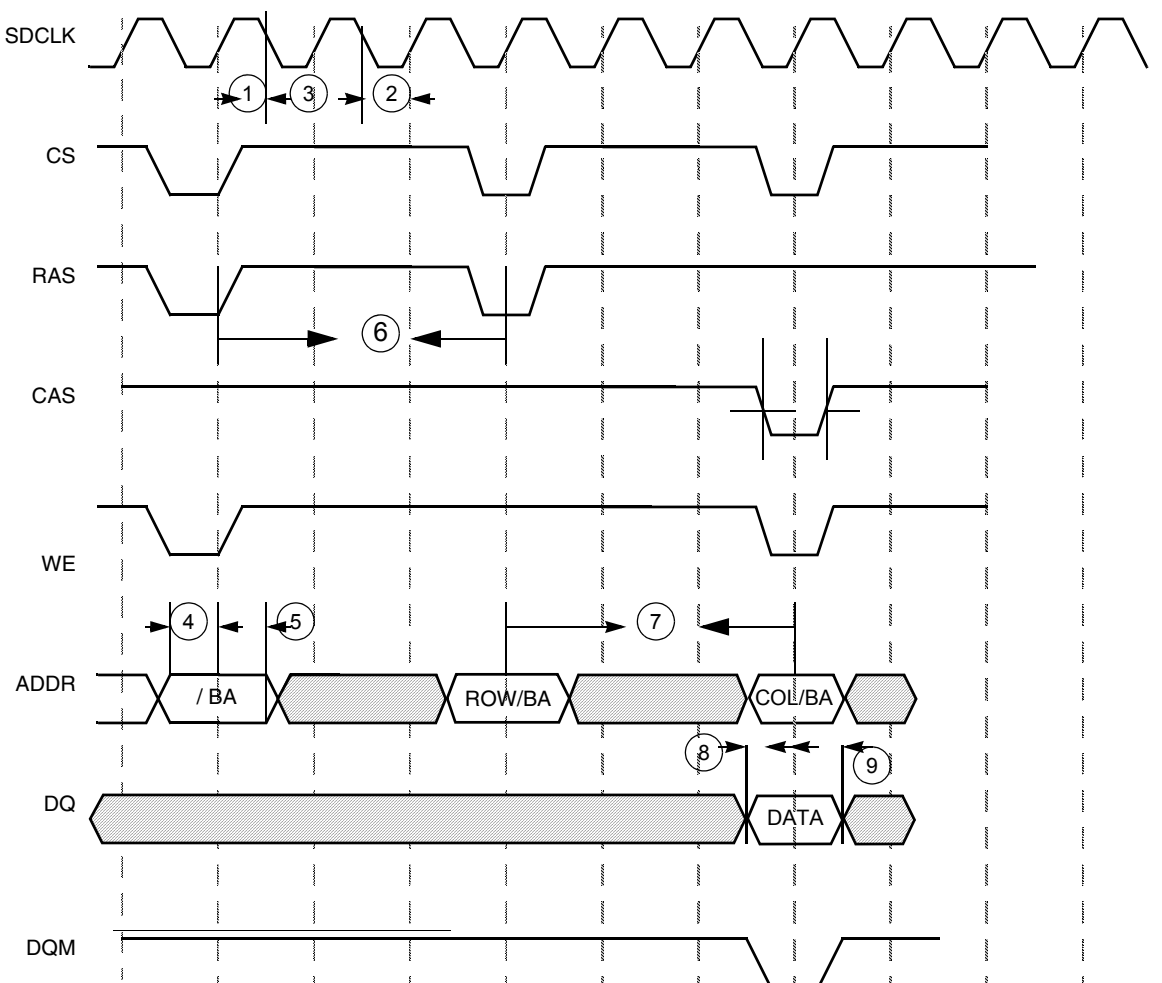


Figure 39. SDRAM Write Cycle Timing Diagram

Table 30. SDRAM Write Timing Parameter Table

Ref No.	Parameter	1.8V		3.0V +/-10%		Unit
		Minimum	Maximum	Minimum	Maximum	
1	SDRAM clock high-level width	3.00	—	4	—	ns
2	SDRAM clock low-level width	3.00	—	4	—	ns
3	SDRAM clock cycle time	11.1	—	7.5	—	ns
4	Address setup time	3.67	—	3	—	ns
5	Address hold time	2.95	—	2	—	ns
6	Precharge cycle period <sup>1</sup>	$t_{RP}^2$	—	$t_{RP}^2$	—	ns
7	Active to read/write command delay	$t_{RCD}^2$	—	$t_{RCD}^2$	—	ns
8	Data setup time	3.41	—	2	—	ns

Table 30. SDRAM Write Timing Parameter Table (Continued)

Ref No.	Parameter	1.8V		3.0V +/-10%		Unit
		Minimum	Maximum	Minimum	Maximum	
9	Data hold time	2.45	—	2	—	ns

1. Precharge cycle timing is included in the write timing diagram.

2.  $t_{RP}$  and  $t_{RCD}$  = SDRAM clock cycle time. These settings can be found in the i.MX21 reference manual.

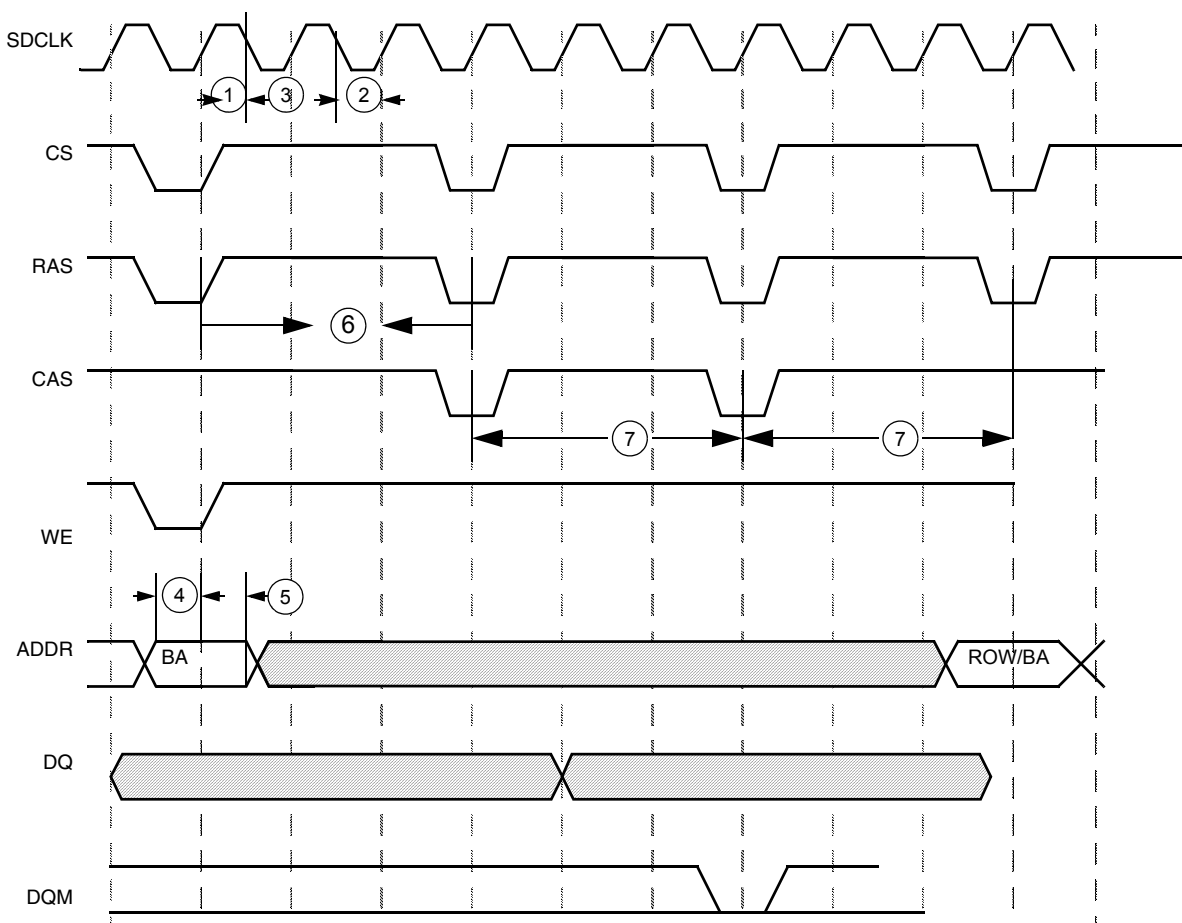


Figure 40. SDRAM Refresh Timing Diagram

Table 31. SDRAM Refresh Timing Parameter Table

Ref No.	Parameter	1.8V		3.0V +/-10%		Unit
		Minimum	Maximum	Minimum	Maximum	
1	SDRAM clock high-level width	3.00	—	4	—	ns
2	SDRAM clock low-level width	3.00	—	4	—	ns
3	SDRAM clock cycle time	11.1	—	7.5	—	ns

Table 31. SDRAM Refresh Timing Parameter Table (Continued)

Ref No.	Parameter	1.8V		3.0V +/-10%		Unit
		Minimum	Maximum	Minimum	Maximum	
4	Address setup time	3.67	—	3	—	ns
5	Address hold time	2.95	—	2	—	ns
6	Precharge cycle period	$t_{RP}^1$	—	$t_{RP}^1$	—	ns
7	Auto precharge command period	$t_{RC}^1$	—	$t_{RC}^1$	—	ns

1.  $t_{RP}$  and  $t_{RC}$  = SDRAM clock cycle time. These settings can be found in the i.MX21 reference manual.

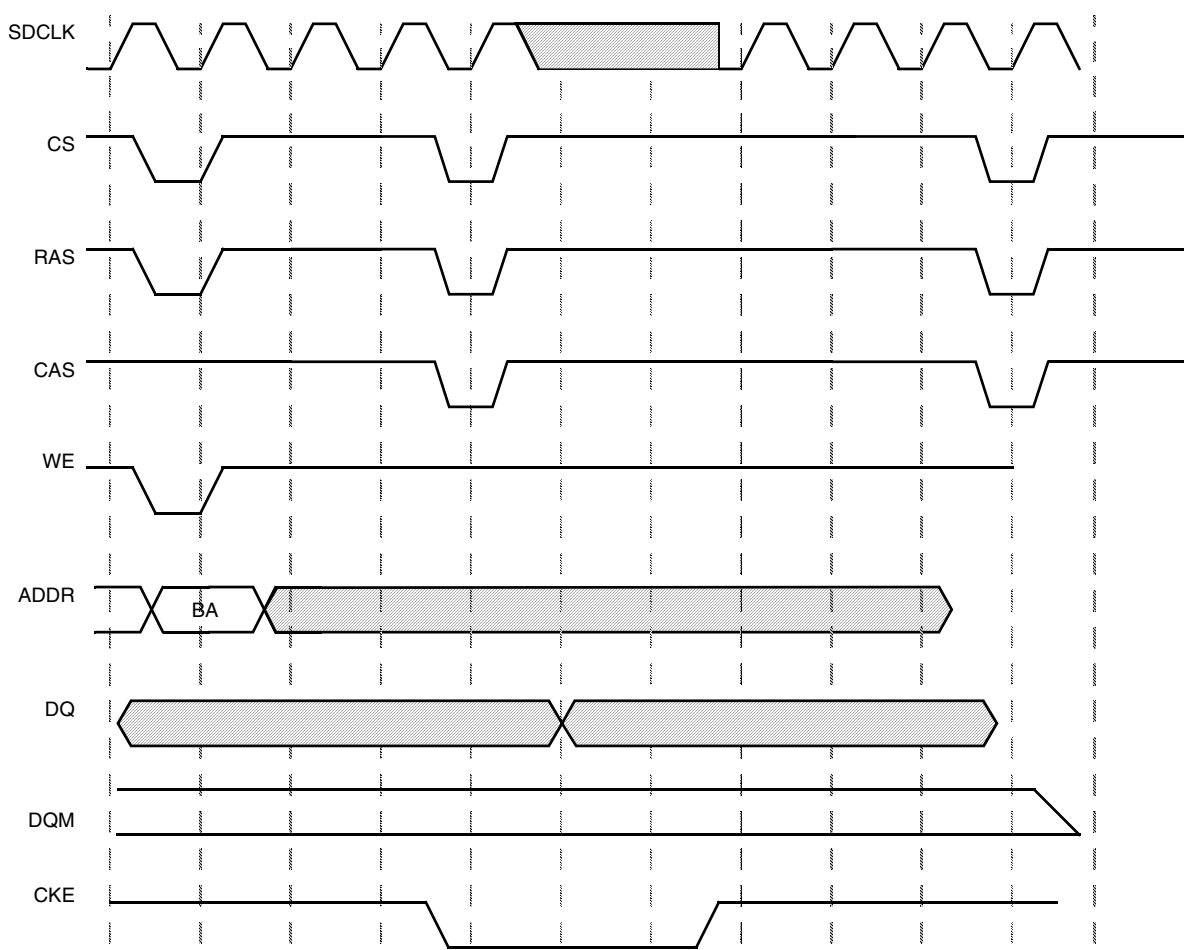


Figure 41. SDRAM Self-Refresh Cycle Timing Diagram

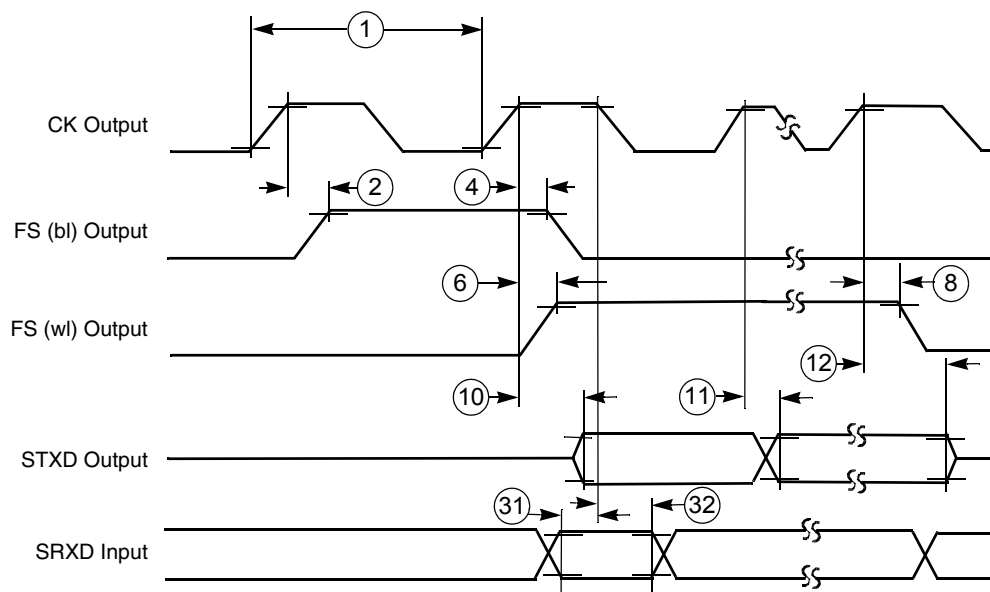
### 3.16 Synchronous Serial Interface

The transmit and receive sections of the SSI can be synchronous or asynchronous. In synchronous mode, the transmitter and the receiver use a common clock and frame synchronization signal. In asynchronous

mode, the transmitter and receiver each have their own clock and frame synchronization signals. Continuous or gated clock mode can be selected. In continuous mode, the clock runs continuously. In gated clock mode, the clock functions only during transmission. The internal and external clock timing diagrams are shown in Figure 42 through Figure 45 on page 54.

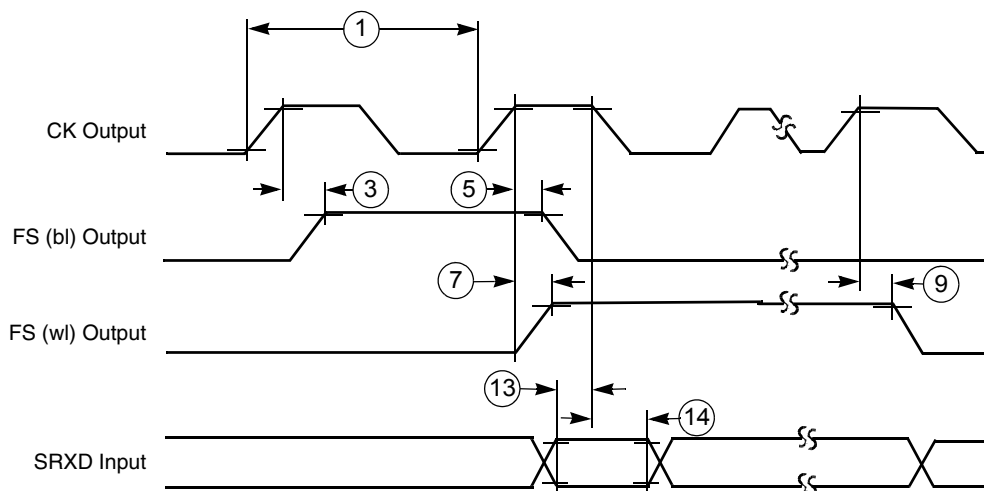
Normal or network mode can also be selected. In normal mode, the SSI functions with one data word of I/O per frame. In network mode, a frame can contain between 2 and 32 data words. Network mode is typically used in star or ring-time division multiplex networks with other processors or codecs, allowing interface to time division multiplexed networks without additional logic. Use of the gated clock is not allowed in network mode. These distinctions result in the basic operating modes that allow the SSI to communicate with a wide variety of devices.

The SSI can be connected to 4 set of ports, SAP, SSI1, SSI2 and SSI3.

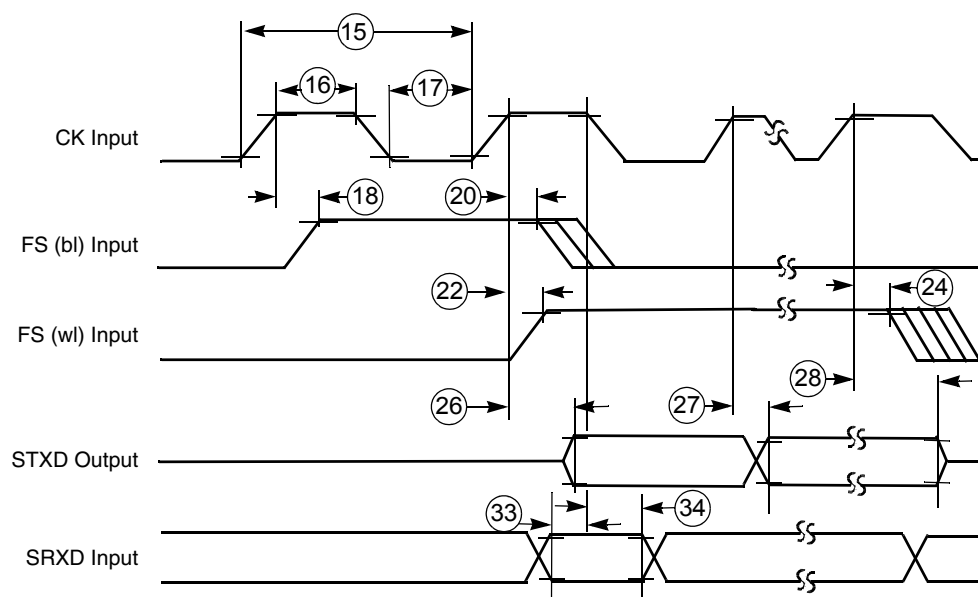


**Note:** SRXD input in synchronous mode only.

**Figure 42. SSI Transmitter Internal Clock Timing Diagram**



**Figure 43. SSI Receiver Internal Clock Timing Diagram**



Note: SRXD Input in Synchronous mode only

Figure 44. SSI Transmitter External Clock Timing Diagram

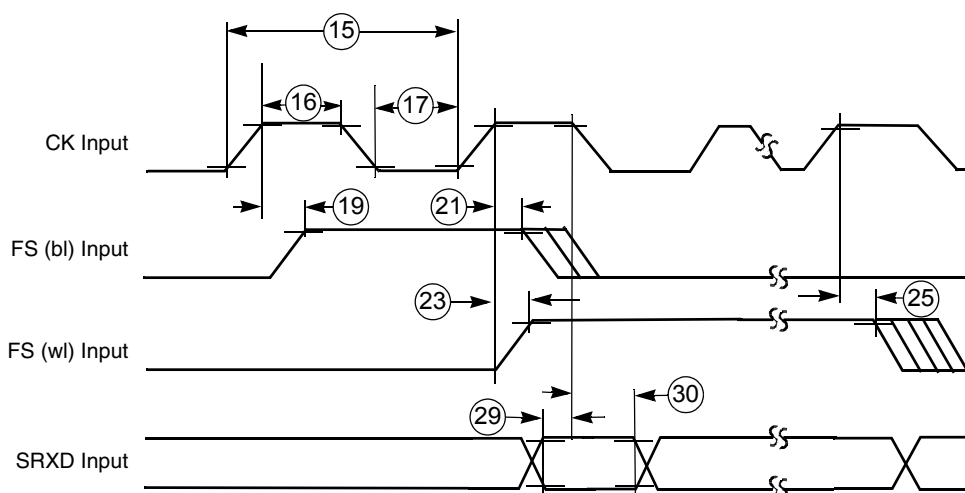


Figure 45. SSI Receiver External Clock Timing Diagram

Table 32. SSI to SAP Ports Timing Parameter Table

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Minimum	Maximum	Minimum	Maximum	
Internal Clock Operation <sup>1</sup> (SAP Ports)						
1	(Tx/Rx) CK clock period <sup>1</sup>	90.91	–	90.91	–	ns
2	(Tx) CK high to FS (bl) high	-3.30	-1.16	-2.98	-1.10	ns
3	(Rx) CK high to FS (bl) high	-3.93	-1.34	-4.18	-1.43	ns

Table 32. SSI to SAP Ports Timing Parameter Table (Continued)

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Minimum	Maximum	Minimum	Maximum	
4	(Tx) CK high to FS (bl) low	-3.30	-1.16	-2.98	-1.10	ns
5	(Rx) CK high to FS (bl) low	-3.93	-1.34	-4.18	-1.43	ns
6	(Tx) CK high to FS (wl) high	-3.30	-1.16	-2.98	-1.10	ns
7	(Rx) CK high to FS (wl) high	-3.93	-1.34	-4.18	-1.43	ns
8	(Tx) CK high to FS (wl) low	-3.30	-1.16	-2.98	-1.10	ns
9	(Rx) CK high to FS (wl) low	-3.93	-1.34	-4.18	-1.43	ns
10	(Tx) CK high to STXD valid from high impedance	-2.44	-0.60	-2.65	-0.98	ns
11a	(Tx) CK high to STXD high	-2.44	-0.60	-2.65	-0.98	ns
11b	(Tx) CK high to STXD low	-2.44	-0.60	-2.65	-0.98	ns
12	(Tx) CK high to STXD high impedance	-2.67	-0.99	-2.65	-0.98	ns
13	SRXD setup time before (Rx) CK low	23.68	–	22.09	–	ns
14	SRXD hold time after (Rx) CK low	0	–	0	–	ns
<b>External Clock Operation (SAP Ports)</b>						
15	(Tx/Rx) CK clock period <sup>1</sup>	90.91	–	90.91	–	ns
16	(Tx/Rx) CK clock high period	36.36	–	36.36	–	ns
17	(Tx/Rx) CK clock low period	36.36	–	36.36	–	ns
18	(Tx) CK high to FS (bl) high	10.24	19.50	7.16	8.65	ns
19	(Rx) CK high to FS (bl) high	10.89	21.27	7.63	9.12	ns
20	(Tx) CK high to FS (bl) low	10.24	19.50	7.16	8.65	ns
21	(Rx) CK high to FS (bl) low	10.89	21.27	7.63	9.12	ns
22	(Tx) CK high to FS (wl) high	10.24	19.50	7.16	8.65	ns
23	(Rx) CK high to FS (wl) high	10.89	21.27	7.63	9.12	ns
24	(Tx) CK high to FS (wl) low	10.24	19.50	7.16	8.65	ns
25	(Rx) CK high to FS (wl) low	10.89	21.27	7.63	9.12	ns
26	(Tx) CK high to STXD valid from high impedance	12.08	19.36	7.71	9.20	ns
27a	(Tx) CK high to STXD high	10.80	19.36	7.71	9.20	ns

Table 32. SSI to SAP Ports Timing Parameter Table (Continued)

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Minimum	Maximum	Minimum	Maximum	
27b	(Tx) CK high to STXD low	10.80	19.36	7.71	9.20	ns
28	(Tx) CK high to STXD high impedance	12.08	19.36	7.71	9.20	ns
29	SRXD setup time before (Rx) CK low	0.37	–	0.42	–	ns
30	SRXD hold time after (Rx) CK low	0	–	0	–	ns
<b>Synchronous Internal Clock Operation (SAP Ports)</b>						
31	SRXD setup before (Tx) CK falling	23.00	–	21.41	–	ns
32	SRXD hold after (Tx) CK falling	0	–	0	–	ns
<b>Synchronous External Clock Operation (SAP Ports)</b>						
33	SRXD setup before (Tx) CK falling	1.20	–	0.88	–	ns
34	SRXD hold after (Tx) CK falling	0	–	0	–	ns

1. All the timings for the SSI are given for a non-inverted serial clock polarity (TSCKP/RSCKP = 0) and a non-inverted frame sync (TFSS/RFSI = 0). If the polarity of the clock and/or the frame sync have been inverted, all the timing remains valid by inverting the clock signal STCK/SRCK and/or the frame sync STFS/SRFS shown in the tables and in the figures.

Table 33. SSI to SSI1 Ports Timing Parameter Table

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Minimum	Maximum	Minimum	Maximum	
Internal Clock Operation <sup>1</sup> (SSI1 Ports)						
1	(Tx/Rx) CK clock period <sup>1</sup>	90.91	–	90.91	–	ns
2	(Tx) CK high to FS (bl) high	-0.68	-0.15	-0.68	-0.15	ns
3	(Rx) CK high to FS (bl) high	-0.96	-0.27	-0.96	-0.27	ns
4	(Tx) CK high to FS (bl) low	-0.68	-0.15	-0.68	-0.15	ns
5	(Rx) CK high to FS (bl) low	-0.96	-0.27	-0.96	-0.27	ns
6	(Tx) CK high to FS (wl) high	-0.68	-0.15	-0.68	-0.15	ns
7	(Rx) CK high to FS (wl) high	-0.96	-0.27	-0.96	-0.27	ns
8	(Tx) CK high to FS (wl) low	-0.68	-0.15	-0.68	-0.15	ns
9	(Rx) CK high to FS (wl) low	-0.96	-0.27	-0.96	-0.27	ns



Table 33. SSI to SSI1 Ports Timing Parameter Table (Continued)

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Minimum	Maximum	Minimum	Maximum	
10	(Tx) CK high to STXD valid from high impedance	-1.68	-0.36	-1.68	-0.36	ns
11a	(Tx) CK high to STXD high	-1.68	-0.36	-1.68	-0.36	ns
11b	(Tx) CK high to STXD low	-1.68	-0.36	-1.68	-0.36	ns
12	(Tx) CK high to STXD high impedance	-1.58	-0.31	-1.58	-0.31	ns
13	SRXD setup time before (Rx) CK low	20.41	–	20.41	–	ns
14	SRXD hold time after (Rx) CK low	0	–	0	–	ns
<b>External Clock Operation (SSI1 Ports)</b>						
15	(Tx/Rx) CK clock period <sup>1</sup>	90.91	–	90.91	–	ns
16	(Tx/Rx) CK clock high period	36.36	–	36.36	–	ns
17	(Tx/Rx) CK clock low period	36.36	–	36.36	–	ns
18	(Tx) CK high to FS (bl) high	10.22	17.63	8.82	16.24	ns
19	(Rx) CK high to FS (bl) high	10.79	19.67	9.39	18.28	ns
20	(Tx) CK high to FS (bl) low	10.22	17.63	8.82	16.24	ns
21	(Rx) CK high to FS (bl) low	10.79	19.67	9.39	18.28	ns
22	(Tx) CK high to FS (wl) high	10.22	17.63	8.82	16.24	ns
23	(Rx) CK high to FS (wl) high	10.79	19.67	9.39	18.28	ns
24	(Tx) CK high to FS (wl) low	10.22	17.63	8.82	16.24	ns
25	(Rx) CK high to FS (wl) low	10.79	19.67	9.39	18.28	ns
26	(Tx) CK high to STXD valid from high impedance	10.05	15.75	8.66	14.36	ns
27a	(Tx) CK high to STXD high	10.00	15.63	8.61	14.24	ns
27b	(Tx) CK high to STXD low	10.00	15.63	8.61	14.24	ns
28	(Tx) CK high to STXD high impedance	10.05	15.75	8.66	14.36	ns
29	SRXD setup time before (Rx) CK low	0.78	–	0.47	–	ns
30	SRXD hold time after (Rx) CK low	0	–	0	–	ns
<b>Synchronous Internal Clock Operation (SSI1 Ports)</b>						
31	SRXD setup before (Tx) CK falling	19.90	–	19.90	–	ns

Table 33. SSI to SSI1 Ports Timing Parameter Table (Continued)

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Minimum	Maximum	Minimum	Maximum	
32	SRXD hold after (Tx) CK falling	0	–	0	–	ns
<b>Synchronous External Clock Operation (SSI1 Ports)</b>						
33	SRXD setup before (Tx) CK falling	2.59	–	2.28	–	ns
34	SRXD hold after (Tx) CK falling	0	–	0	–	ns

1. All the timings for the SSI are given for a non-inverted serial clock polarity (TSCKP/RSCCKP = 0) and a non-inverted frame sync (TFSI/RFSI = 0). If the polarity of the clock and/or the frame sync have been inverted, all the timing remains valid by inverting the clock signal STCK/SRCK and/or the frame sync STFS/SRFS shown in the tables and in the figures.

Table 34. SSI to SSI2 Ports Timing Parameter Table

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Minimum	Maximum	Minimum	Maximum	
Internal Clock Operation <sup>1</sup> (SSI2 Ports)						
1	(Tx/Rx) CK clock period <sup>1</sup>	90.91	–	90.91	–	ns
2	(Tx) CK high to FS (bl) high	0.01	0.15	0.01	0.15	ns
3	(Rx) CK high to FS (bl) high	-0.21	0.05	-0.21	0.05	ns
4	(Tx) CK high to FS (bl) low	0.01	0.15	0.01	0.15	ns
5	(Rx) CK high to FS (bl) low	-0.21	0.05	-0.21	0.05	ns
6	(Tx) CK high to FS (wl) high	0.01	0.15	0.01	0.15	ns
7	(Rx) CK high to FS (wl) high	-0.21	0.05	-0.21	0.05	ns
8	(Tx) CK high to FS (wl) low	0.01	0.15	0.01	0.15	ns
9	(Rx) CK high to FS (wl) low	-0.21	0.05	-0.21	0.05	ns
10	(Tx) CK high to STXD valid from high impedance	0.34	0.72	0.34	0.72	ns
11a	(Tx) CK high to STXD high	0.34	0.72	0.34	0.72	ns
11b	(Tx) CK high to STXD low	0.34	0.72	0.34	0.72	ns
12	(Tx) CK high to STXD high impedance	0.34	0.48	0.34	0.48	ns
13	SRXD setup time before (Rx) CK low	21.50	–	21.50	–	ns
14	SRXD hold time after (Rx) CK low	0	–	0	–	ns

Table 34. SSI to SSI2 Ports Timing Parameter Table (Continued)

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Minimum	Maximum	Minimum	Maximum	
External Clock Operation (SSI2 Ports)						
15	(Tx/Rx) CK clock period <sup>1</sup>	90.91	–	90.91	–	ns
16	(Tx/Rx) CK clock high period	36.36	–	36.36	–	ns
17	(Tx/Rx) CK clock low period	36.36	–	36.36	–	ns
18	(Tx) CK high to FS (bl) high	10.40	17.37	8.67	15.88	ns
19	(Rx) CK high to FS (bl) high	11.00	19.70	9.28	18.21	ns
20	(Tx) CK high to FS (bl) low	10.40	17.37	8.67	15.88	ns
21	(Rx) CK high to FS (bl) low	11.00	19.70	9.28	18.21	ns
22	(Tx) CK high to FS (wl) high	10.40	17.37	8.67	15.88	ns
23	(Rx) CK high to FS (wl) high	11.00	19.70	9.28	18.21	ns
24	(Tx) CK high to FS (wl) low	10.40	17.37	8.67	15.88	ns
25	(Rx) CK high to FS (wl) low	11.00	19.70	9.28	18.21	ns
26	(Tx) CK high to STXD valid from high impedance	9.59	17.08	7.86	15.59	ns
27a	(Tx) CK high to STXD high	9.59	17.08	7.86	15.59	ns
27b	(Tx) CK high to STXD low	9.59	17.08	7.86	15.59	ns
28	(Tx) CK high to STXD high impedance	9.59	16.84	7.86	15.35	ns
29	SRXD setup time before (Rx) CK low	2.52	–	2.52	–	ns
30	SRXD hole time after (Rx) CK low	0	–	0	–	ns
Synchronous Internal Clock Operation (SSI2 Ports)						
31	SRXD setup before (Tx) CK falling	20.78	–	20.78	–	ns
32	SRXD hold after (Tx) CK falling	0	–	0	–	ns
Synchronous External Clock Operation (SSI2 Ports)						
33	SRXD setup before (Tx) CK falling	4.42	–	4.42	–	ns
34	SRXD hold after (Tx) CK falling	0	–	0	–	ns

1. All the timings for the SSI are given for a non-inverted serial clock polarity (TSCKP/RSCKP = 0) and a non-inverted frame sync (TFSI/RFSI = 0). If the polarity of the clock and/or the frame sync have been inverted, all the timing remains valid by inverting the clock signal STCK/SRCK and/or the frame sync STFS/SRFS shown in the tables and in the figures.

Table 35. SSI to SSI3 Ports Timing Parameter Table

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Minimum	Maximum	Minimum	Maximum	
Internal Clock Operation <sup>1</sup> (SSI3 Ports)						
1	(Tx/Rx) CK clock period <sup>1</sup>	90.91	–	90.91	–	ns
2	(Tx) CK high to FS (bl) high	-2.09	-0.66	-2.09	-0.66	ns
3	(Rx) CK high to FS (bl) high	-2.74	-0.84	-2.74	-0.84	ns
4	(Tx) CK high to FS (bl) low	-2.09	-0.66	-2.09	-0.66	ns
5	(Rx) CK high to FS (bl) low	-2.74	-0.84	-2.74	-0.84	ns
6	(Tx) CK high to FS (wl) high	-2.09	-0.66	-2.09	-0.66	ns
7	(Rx) CK high to FS (wl) high	-2.74	-0.84	-2.74	-0.84	ns
8	(Tx) CK high to FS (wl) low	-2.09	-0.66	-2.09	-0.66	ns
9	(Rx) CK high to FS (wl) low	-2.74	-0.84	-2.74	-0.84	ns
10	(Tx) CK high to STXD valid from high impedance	-1.73	-0.26	-1.73	-0.26	ns
11a	(Tx) CK high to STXD high	-2.87	-0.80	-2.87	-0.80	ns
11b	(Tx) CK high to STXD low	-2.87	-0.80	-2.87	-0.80	ns
12	(Tx) CK high to STXD high impedance	-1.73	-0.26	-1.73	-0.26	ns
13	SRXD setup time before (Rx) CK low	22.77	–	22.77	–	ns
14	SRXD hold time after (Rx) CK low	0	–	0	–	ns
External Clock Operation (SSI3 Ports)						
15	(Tx/Rx) CK clock period <sup>1</sup>	90.91	–	90.91	–	ns
16	(Tx/Rx) CK clock high period	36.36	–	36.36	–	ns
17	(Tx/Rx) CK clock low period	36.36	–	36.36	–	ns
18	(Tx) CK high to FS (bl) high	9.62	17.10	7.90	15.61	ns
19	(Rx) CK high to FS (bl) high	10.30	19.54	8.58	18.05	ns
20	(Tx) CK high to FS (bl) low	9.62	17.10	7.90	15.61	ns
21	(Rx) CK high to FS (bl) low	10.30	19.54	8.58	18.05	ns
22	(Tx) CK high to FS (wl) high	9.62	17.10	7.90	15.61	ns
23	(Rx) CK high to FS (wl) high	10.30	19.54	8.58	18.05	ns

Table 35. SSI to SSI3 Ports Timing Parameter Table (Continued)

Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Minimum	Maximum	Minimum	Maximum	
24	(Tx) CK high to FS (wl) low	9.62	17.10	7.90	15.61	ns
25	(Rx) CK high to FS (wl) low	10.30	19.54	8.58	18.05	ns
26	(Tx) CK high to STXD valid from high impedance	9.02	16.46	7.29	14.97	ns
27a	(Tx) CK high to STXD high	8.48	15.32	6.75	13.83	ns
27b	(Tx) CK high to STXD low	8.48	15.32	6.75	13.83	ns
28	(Tx) CK high to STXD high impedance	9.02	16.46	7.29	14.97	ns
29	SRXD setup time before (Rx) CK low	1.49	–	1.49	–	ns
30	SRXD hole time after (Rx) CK low	0	–	0	–	ns
<b>Synchronous Internal Clock Operation (SSI3 Ports)</b>						
31	SRXD setup before (Tx) CK falling	21.99	–	21.99	–	ns
32	SRXD hold after (Tx) CK falling	0	–	0	–	ns
<b>Synchronous External Clock Operation (SSI3 Ports)</b>						
33	SRXD setup before (Tx) CK falling	3.80	–	3.80	–	ns
34	SRXD hold after (Tx) CK falling	0	–	0	–	ns

1. All the timings for the SSI are given for a non-inverted serial clock polarity (TSCKP/RSCCKP = 0) and a non-inverted frame sync (TFSI/RFSI = 0). If the polarity of the clock and/or the frame sync have been inverted, all the timing remains valid by inverting the clock signal STCK/SRCK and/or the frame sync STFS/SRFS shown in the tables and in the figures.

## 3.17 1-Wire Interface Timing

### 3.17.1 Reset Sequence with Reset Pulse Presence Pulse

To begin any communications with the DS2502, it is required that an initialization procedure be issued. A reset pulse must be generated and then a presence pulse must be detected. The minimum reset pulse length is 480 us. The bus master (one-wire) will generate this pulse, then after the DS2502 detects a rising edge on the one-wire bus, it will wait 15-60 us before it will transmit back a presence pulse. The presence pulse will exist for 60-240 us.

The timing diagram for this sequence is shown in Figure 46.

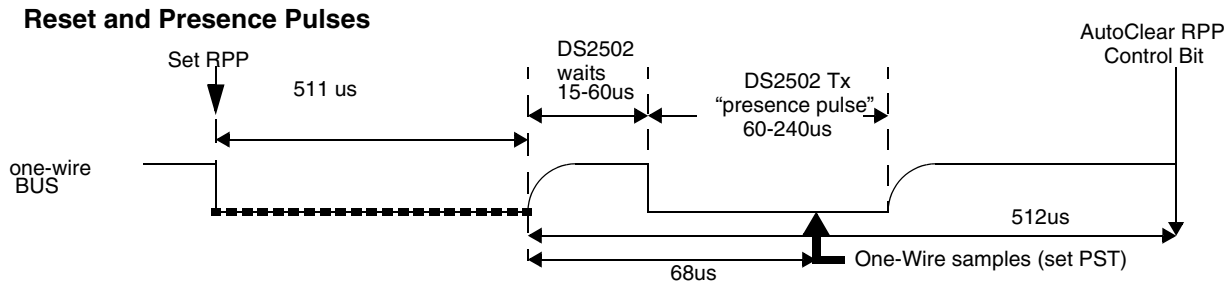


Figure 46. 1-Wire Initialization

The reset pulse begins the initialization sequence and it is initiated when the RPP control register bit is set. When the presence pulse is detected, this bit will be cleared. The presence pulse is used by the bus master to determine if at least one DS2502 is connected. Software will determine if more than one DS2502 exists. The one-wire will sample for the DS2502 presence pulse. The presence pulse is latched in the one-wire control register PST. When the PST bit is set to a one, it means that a DS2502 is present; if the bit is set to a zero, then no device was found.

### 3.17.2 Write 0

The Write 0 function simply writes a zero bit to the DS2502. The sequence takes 117 us. The one-wire bus is held low for 100us.

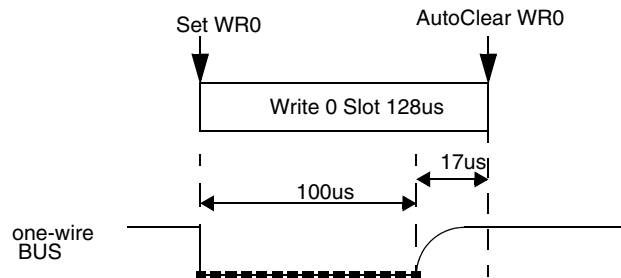


Figure 47. Write 0 Timing

The Write 0 pulse sequence is initiated when the WR0 control bit register is set. When the write is complete, the WR0 register will be auto cleared.

### 3.17.3 Write 1/Read Data

The Write 1 and Read timing is identical. The time slot is first driven low. According to the DS2502 documentation, the DS2502 has a delay circuit which is used to synchronize the DS2502 with the bus master (one-wire). This delay circuit is triggered by the falling edge of the data line and is used to decide when the DS2502 should sample the line. In the case of a write 1 or read 1, after a delay, a 1 will be transmitted / received. When a read 0 slot is issued, the delay circuit will hold the data line low to override the 1 generated by the bus master (one-wire).

For the Write 1 or Read, the control register WR1/RD is set and auto-cleared when the sequence has been completed. After a Read, the control register RDST bit is set to the value of the read.

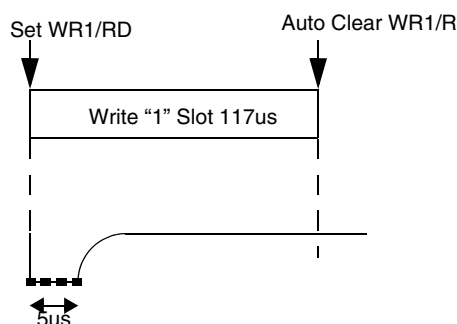


Figure 48. Write 1 Timing

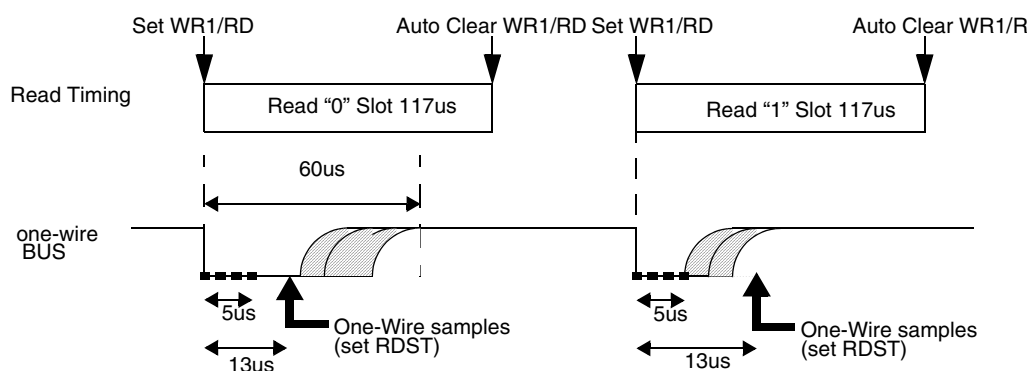


Figure 49. Read Timing

The precision of the generated clock is very important to get a proper behavior of the one-wire module. This module is based on a state machine which undertakes actions at defined times.

Table 36. System Timing Requirements

Times	Values (Microsec)	Minimum (Microsec)	Maximum (microsec)	Absolute Precision	Relative Precision
RSTL	511	480	–	31	0.0645
PST	68	60	75	7	0.1
RSTH	512	480	–	32	0.0645
LOW0	100	60	120	20	0.2
LOWR	5	1	15	4	0.8
READ_sample	13	–	15	2	0.15

The most stringent constraint is 0.0645 as a relative time imprecision.

The time relative precision is directly derived from the frequency of the derivative clock (f):

$$\text{Time relative precision} = 1/f - 1 = \text{divider/clock (MHz)} - 1$$

The Table 37 gathers relative time precision for different main clock frequencies.

**Table 37. System Clock Requirements**

Main Clock Frequency (MHz)	13	16.8	19.44
Clock divide ratio	13	17	19
Generated frequency (MHz)	1	0.9882	1.023
Relative time imprecision	0	0.0117	0.023

This shows that the user should take care of the main clock frequency when using the one-wire module. If the main clock is an exact integer multiple of 1 MHz, then the generated frequency will be exactly 1 MHz.

**NOTE:**

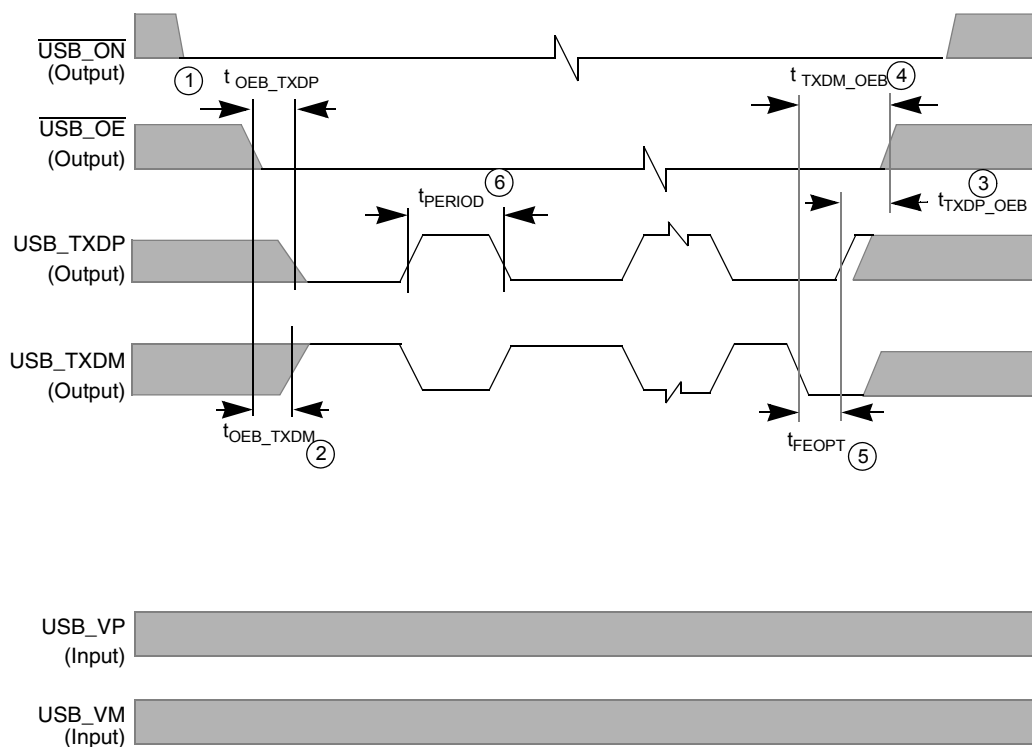
A main clock frequency below 10 MHz might cause a misbehavior of the module.

### 3.18 USB On-The-Go

Four types of data transfer modes exist for the USB module: control transfers, bulk transfers, isochronous transfers and interrupt transfers. From the perspective of the USB module, the interrupt transfer type is identical to the bulk data transfer mode, and no additional hardware is supplied to support it. This section covers the transfer modes and how they work from the ground up.

Data moves across the USB in packets. Groups of packets are combined to form data transfers. The same packet transfer mechanism applies to bulk, interrupt, and control transfers. Isochronous data is also moved in the form of packets, but because isochronous pipes are given a fixed portion of the USB bandwidth at all times, there is no end-of-transfer.





**Figure 50. USB Timing Diagram for Data Transfer to USB Transceiver (TX)**

**Table 38. USB Timing Parameter Table for Data Transfer to USB Transceiver (TX)**

Ref No.	Parameter	3.0 +/- 0.3V		Unit
		Minimum	Maximum	
1	$t_{OEB\_TXDP}$ ; $\overline{USBD\_OE}$ active to USBD_TXDP low	83.14	83.47	ns
2	$t_{OEB\_TXDM}$ ; $\overline{USBD\_OE}$ active to USBD_TXDM high	81.55	81.98	ns
3	$t_{TXDP\_OEB}$ ; USBD_TXDP high to $\overline{USBD\_OE}$ deactivated	83.54	83.8	ns
4	$t_{TXDM\_OEB}$ ; USBD_TXDM low to $\overline{USBD\_OE}$ deactivated (includes SE0)	248.9	249.13	ns
5	$t_{FEOPT}$ ; SE0 interval of EOP	160	175	ns
6	$t_{PERIOD}$ ; Data transfer rate	11.97	12.03	Mb/s

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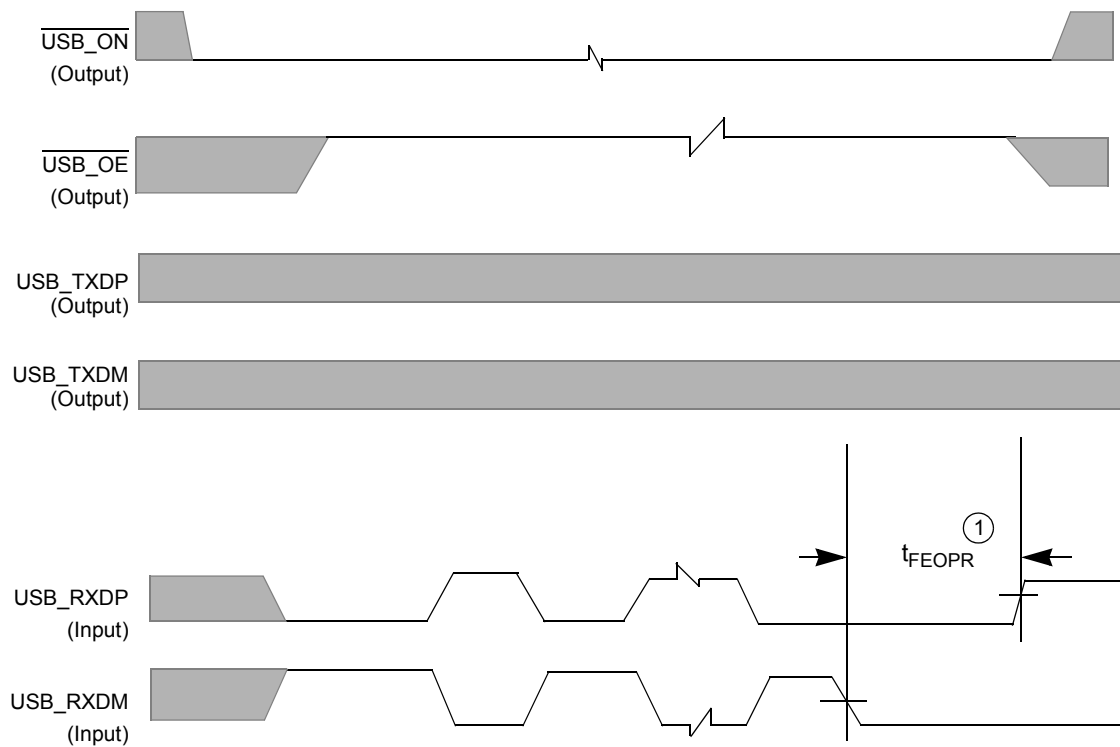


Figure 51. USB Timing Diagram for Data Transfer from USB Transceiver (RX)

Table 39. USB Timing Parameter Table for Data Transfer from USB Transceiver (RX)

Ref No.	Parameter	3.0 +/- 0.3V		Unit
		Minimum	Maximum	
1	t <sub>FEOPR</sub> ; Receiver SE0 interval of EOP	82	—	ns

The USBOTG I<sup>2</sup>C communication protocol consists of six components: START, Data Source/Recipient, Data Direction, Slave Acknowledge, Data, Data Acknowledge, and STOP.

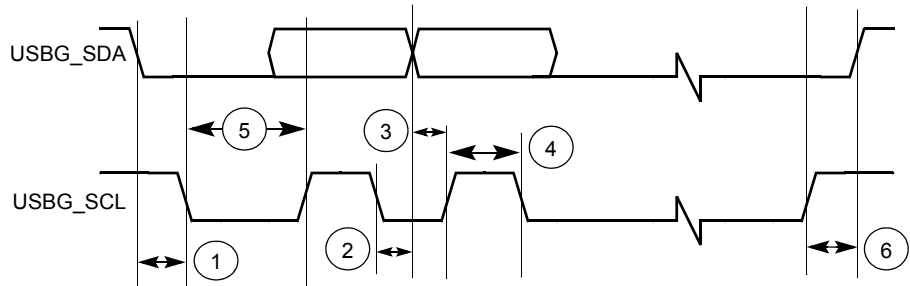


Figure 52. USB Timing Diagram for Data Transfer from USB Transceiver (I<sup>2</sup>C)

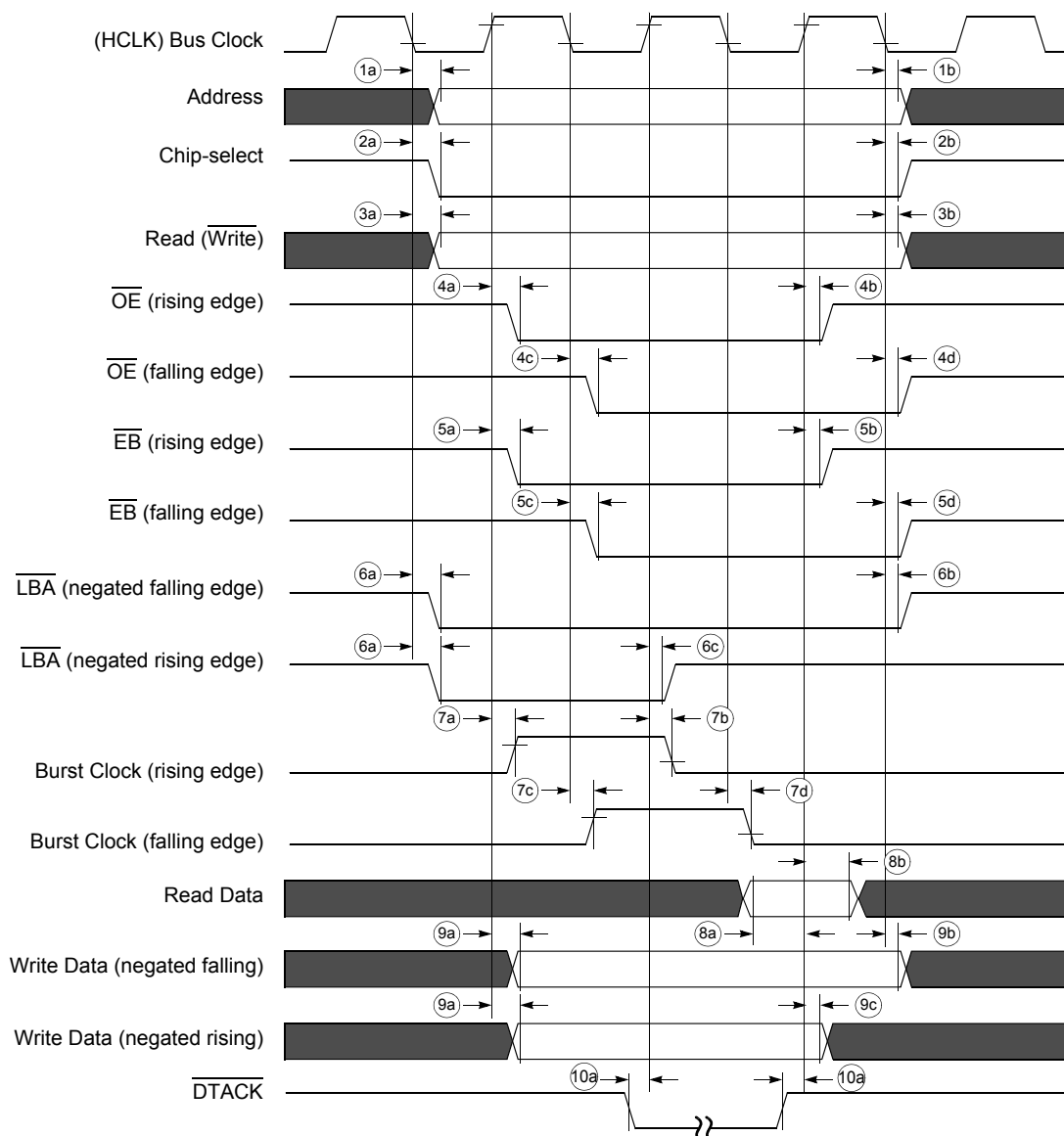
**Table 40. USB Timing Parameter Table for Data Transfer from USB Transceiver (I<sup>2</sup>C)**

Ref No.	Parameter	1.8 +/- 0.10V		Unit
		Minimum	Maximum	
1	Hold time (repeated) START condition	188	–	ns
2	<u>Data hold time</u>	0	188	ns
3	Data setup time	88	–	ns
4	HIGH period of the SCL clock	500	–	ns
5	LOW period of the SCL clock	500	–	ns
6	Setup time for STOP condition	185	–	ns

### 3.19 External Interface Module (EIM)

The External Interface Module (EIM) handles the interface to devices external to the i.MX21, including generation of chip-selects for external peripherals and memory. The timing diagram for the EIM is shown in Figure 53, and Table 41 on page 68 defines the parameters of signals.

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**Figure 53. EIM Bus Timing Diagram**

**Table 41. EIM Bus Timing Parameters**

Ref No.	Parameter	1.8V +/- 0.1V			3.0V +/- 0.3V			Unit
		Min	Typical	Max	Min	Typical	Max	
1a	Clock fall to address valid	3.97	6.02	9.89	3.83	5.89	9.79	ns
1b	Clock fall to address invalid	3.93	6.00	9.86	3.81	5.86	9.76	ns
2a	Clock fall to chip-select valid	3.47	5.59	8.62	3.30	5.09	8.45	ns
2b	Clock fall to chip-select invalid	3.39	5.09	8.27	3.15	4.85	8.03	ns
3a	Clock fall to Read (Write) Valid	3.51	5.56	8.79	3.39	5.39	8.51	ns

Table 41. EIM Bus Timing Parameters (Continued)

Ref No.	Parameter	1.8V +/- 0.1V			3.0V +/- 0.3V			Unit
		Min	Typical	Max	Min	Typical	Max	
3b	Clock fall to Read ( $\overline{\text{Write}}$ ) Invalid	3.59	5.37	9.14	3.36	5.20	8.50	ns
4a	Clock <sup>1</sup> rise to Output Enable Valid	3.62	5.49	8.98	3.46	5.33	9.02	ns
4b	Clock <sup>1</sup> rise to Output Enable Invalid	3.70	5.61	9.26	3.46	5.37	8.81	ns
4c	Clock <sup>1</sup> fall to Output Enable Valid	3.60	5.48	8.77	3.44	5.30	8.88	ns
4d	Clock <sup>1</sup> fall to Output Enable Invalid	3.69	5.62	9.12	3.42	5.36	8.60	ns
5a	Clock <sup>1</sup> rise to Enable Bytes Valid	3.69	5.46	8.71	3.46	5.25	8.54	ns
5b	Clock <sup>1</sup> rise to Enable Bytes Invalid	4.64	5.47	8.70	3.46	5.25	8.54	ns
5c	Clock <sup>1</sup> fall to Enable Bytes Valid	3.52	5.06	8.39	3.41	5.18	8.36	ns
5d	Clock <sup>1</sup> fall to Enable Bytes Invalid	3.50	5.05	8.27	3.41	5.18	8.36	ns
6a	Clock <sup>1</sup> fall to Load Burst Address Valid	3.65	5.28	8.69	3.30	5.23	8.81	ns
6b	Clock <sup>1</sup> fall to Load Burst Address Invalid	3.65	5.67	9.36	3.41	5.43	9.13	ns
6c	Clock <sup>1</sup> rise to Load Burst Address Invalid	3.66	5.69	9.48	3.33	5.47	9.25	ns
7a	Clock <sup>1</sup> rise to Burst Clock rise	3.50	5.22	8.42	3.26	4.99	8.19	ns
7b	Clock <sup>1</sup> rise to Burst Clock fall	3.49	5.19	8.30	3.31	5.03	8.17	ns
7c	Clock <sup>1</sup> fall to Burst Clock rise	3.50	5.22	8.39	3.26	4.98	8.15	ns
7d	Clock <sup>1</sup> fall to Burst Clock fall	3.49	5.19	8.29	3.31	5.02	8.12	ns
8a	Read Data setup time	4.54	–	–	4.54	–	–	ns
8b	Read Data hold time	0.5	–	–	0.5	–	–	ns
9a	Clock <sup>1</sup> rise to Write Data Valid	4.13	5.86	9.16	3.95	6.36	10.31	ns
9b	Clock <sup>1</sup> fall to Write Data Invalid	4.10	5.79	9.15	4.04	6.27	9.16	ns
9c	Clock <sup>1</sup> rise to Write Data Invalid	4.02	5.81	9.37	4.22	5.29	9.24	ns
10a	$\overline{\text{DTACK}}$ setup time	2.65	4.63	8.40	2.64	4.61	8.41	ns

1. Clock refers to the system clock signal, HCLK, generated from the System DPLL

### 3.19.1 EIM External Bus Timing Diagrams

The following timing diagrams show the timing of accesses to memory or a peripheral.

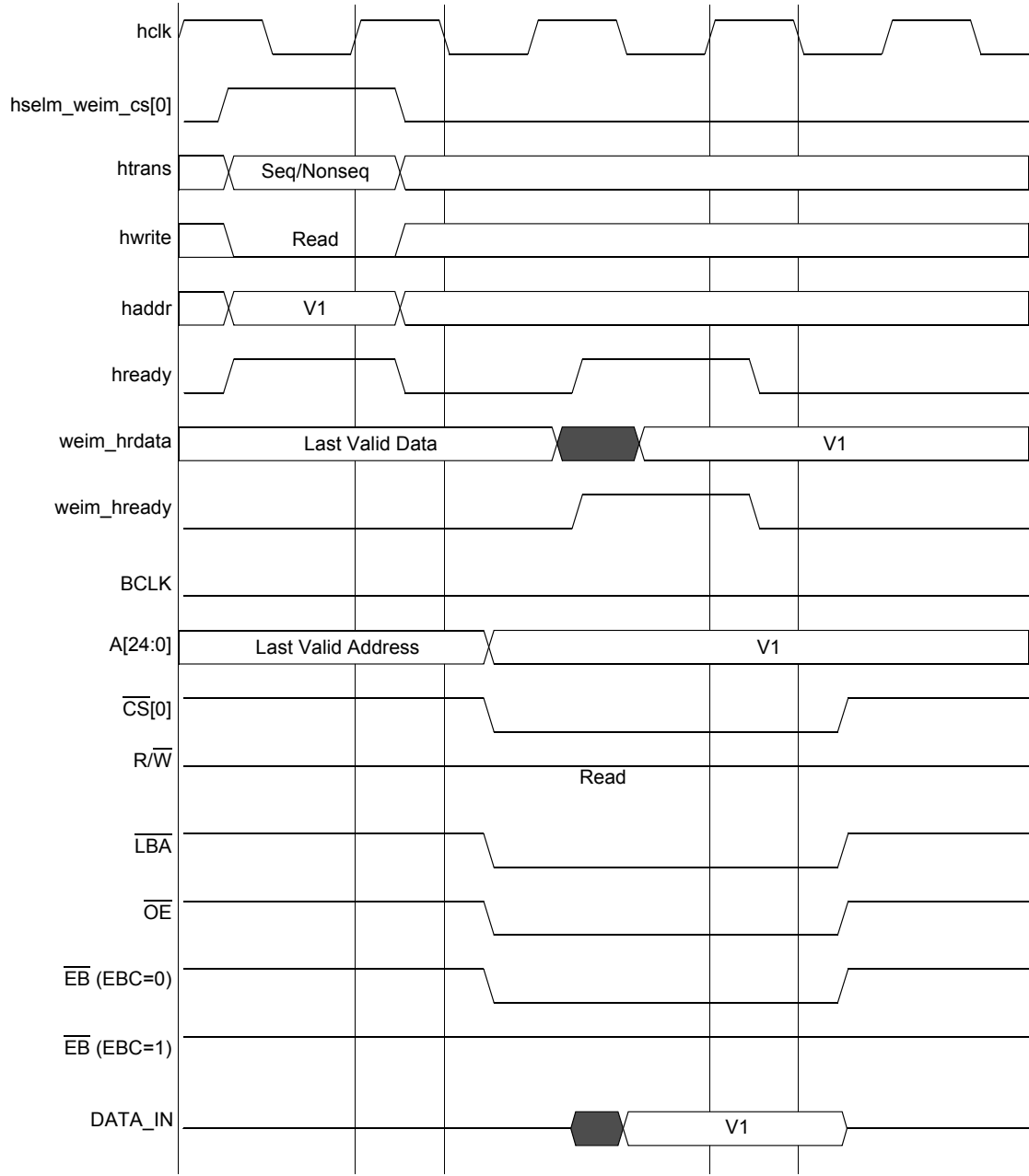
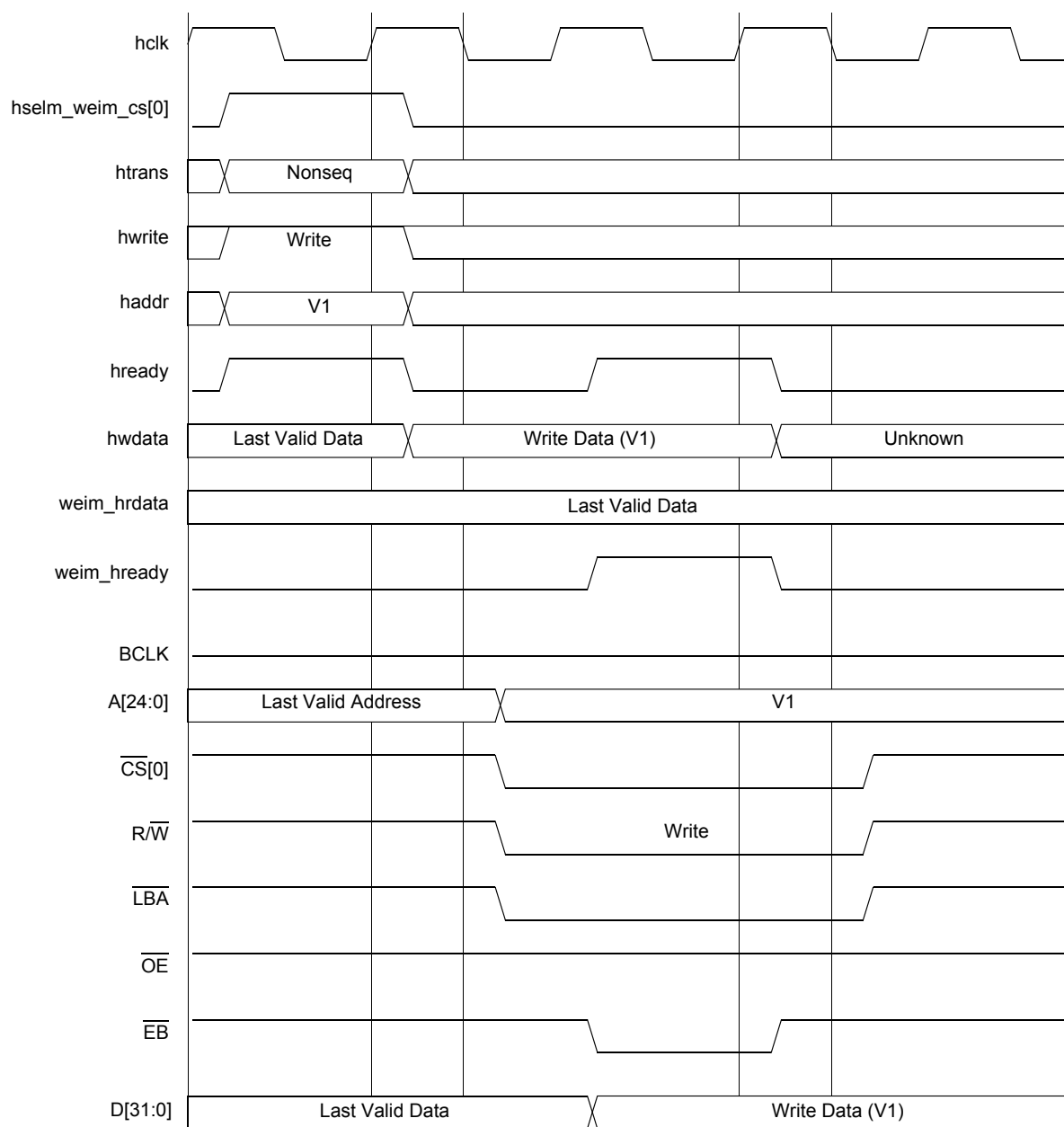


Figure 54. WSC = 1, A.HALF/E.HALF



**Figure 55. WSC = 1, WEA = 1, WEN = 1, A.HALF/E.HALF**

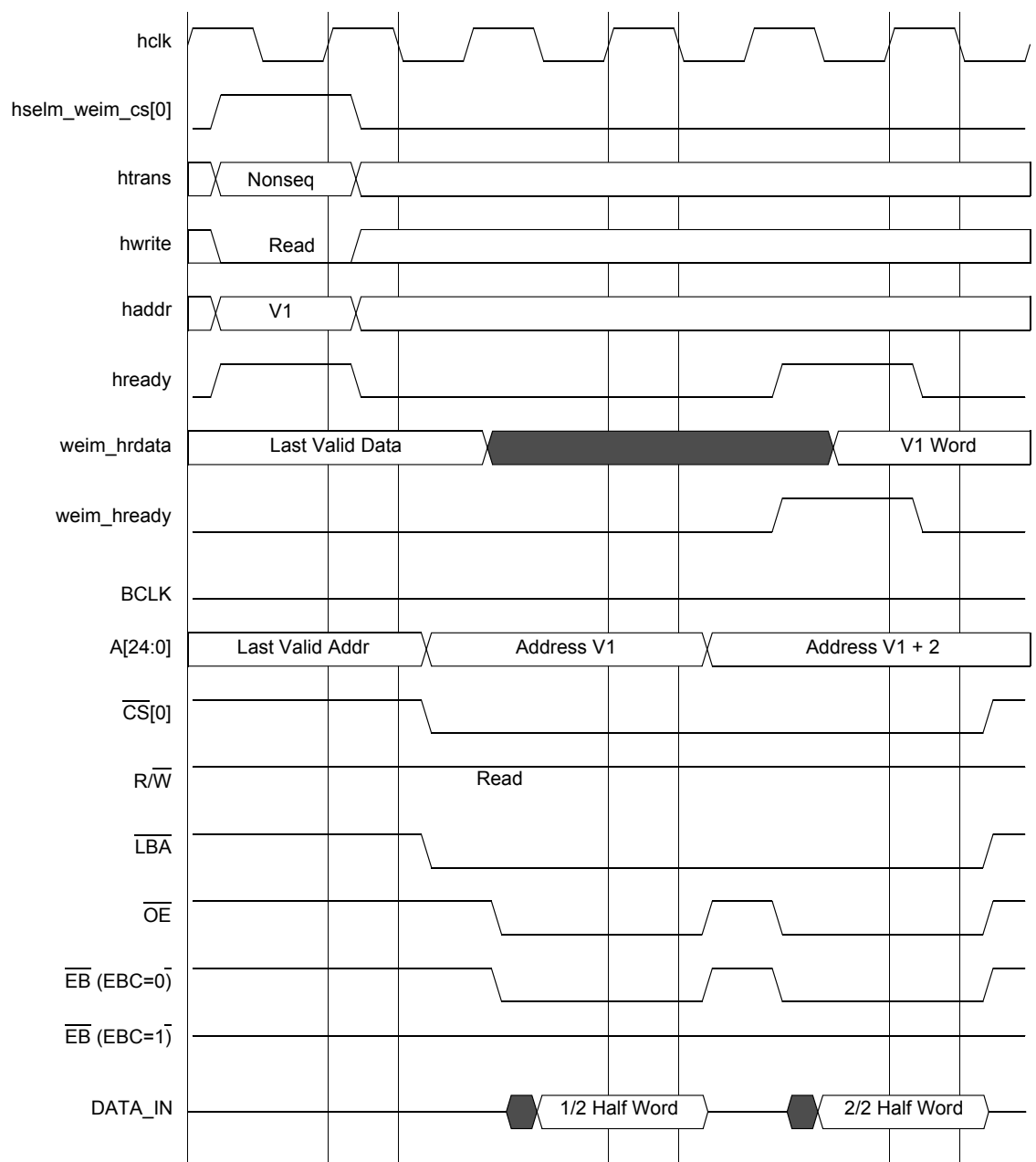


Figure 56.  $WSC = 1$ ,  $OEA = 1$ ,  $A.WORD/E.HALF$



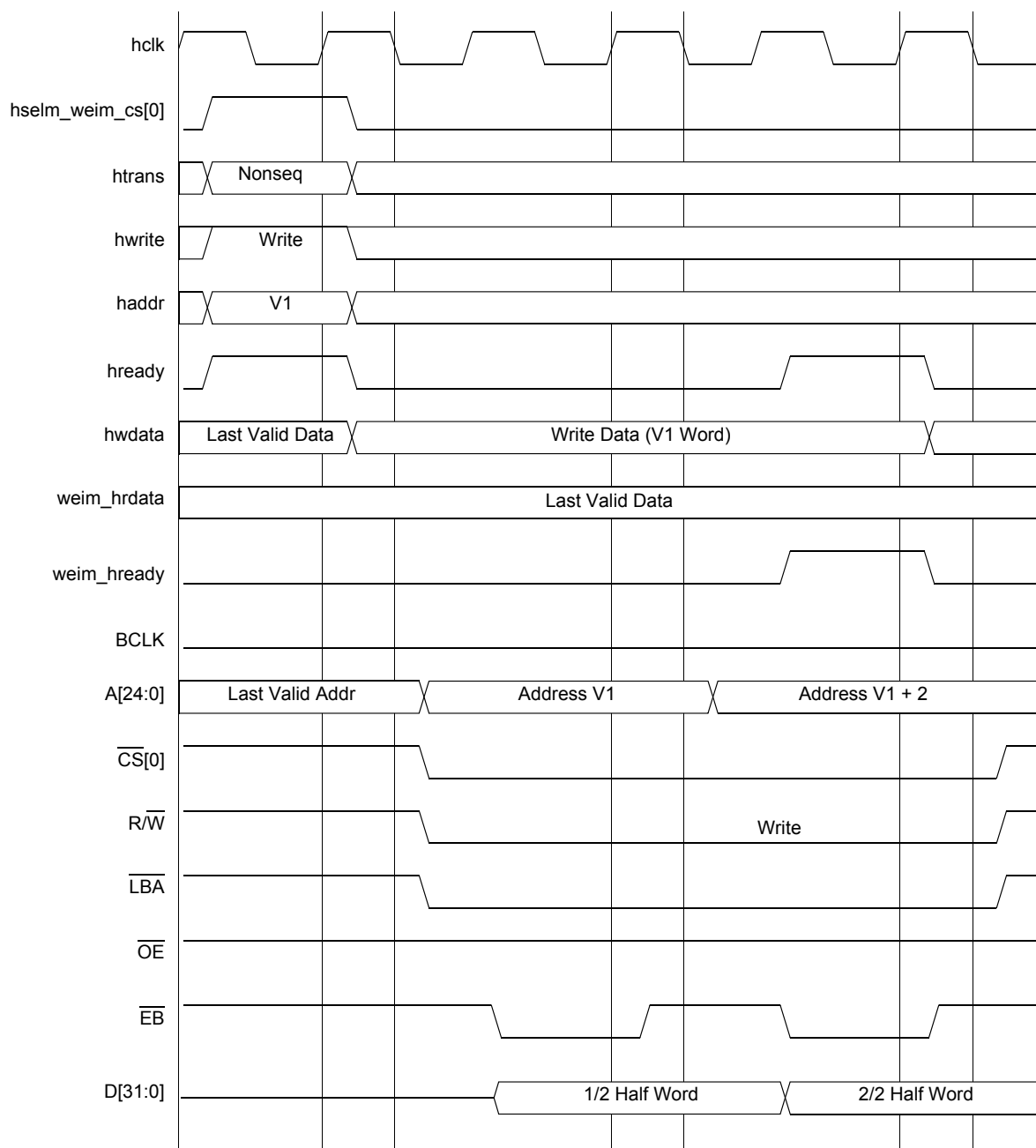


Figure 57. WSC = 1, WEA = 1, WEN = 1, A.WORD/E.HALF

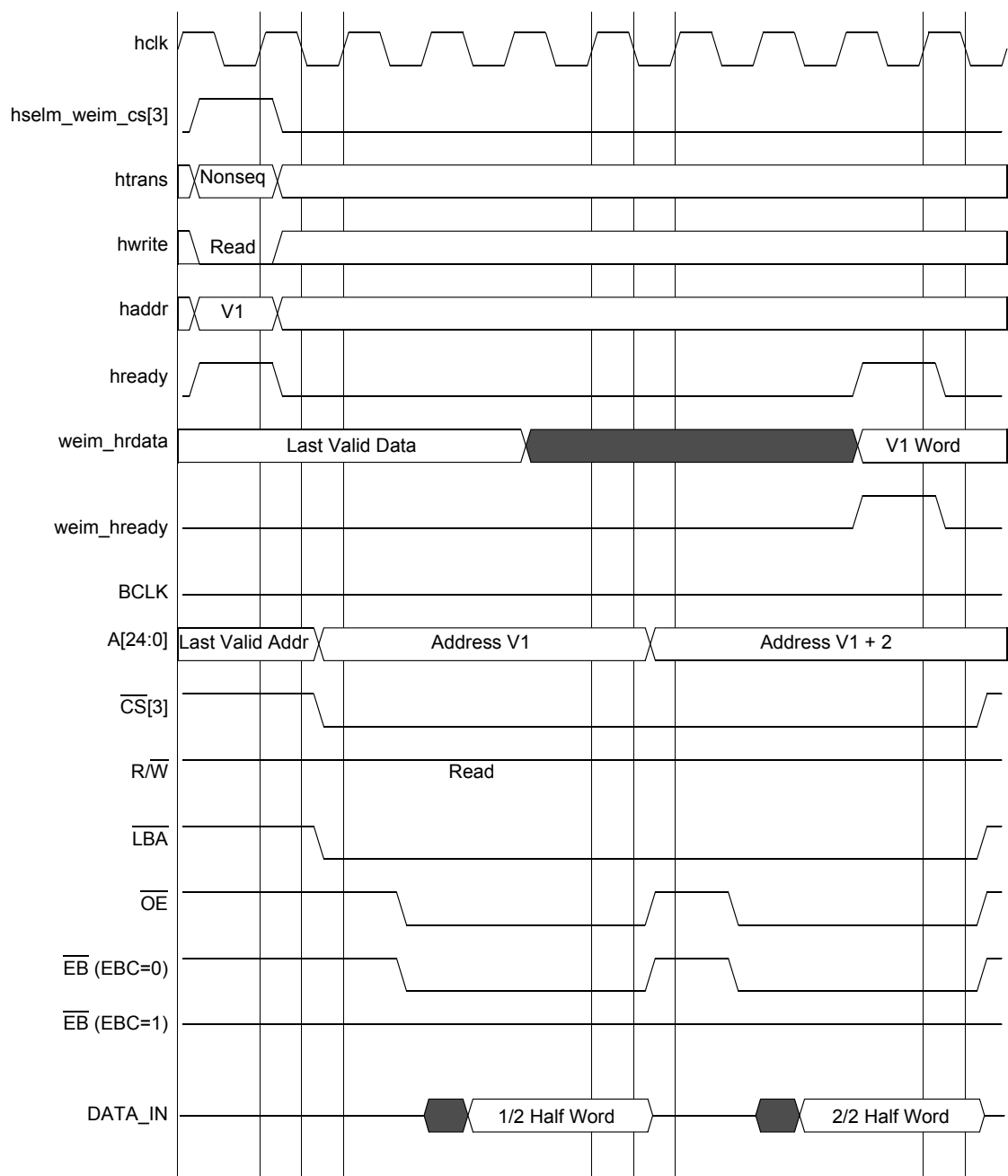


Figure 58. WSC = 3, OEA = 2, A.WORD/E.HALF

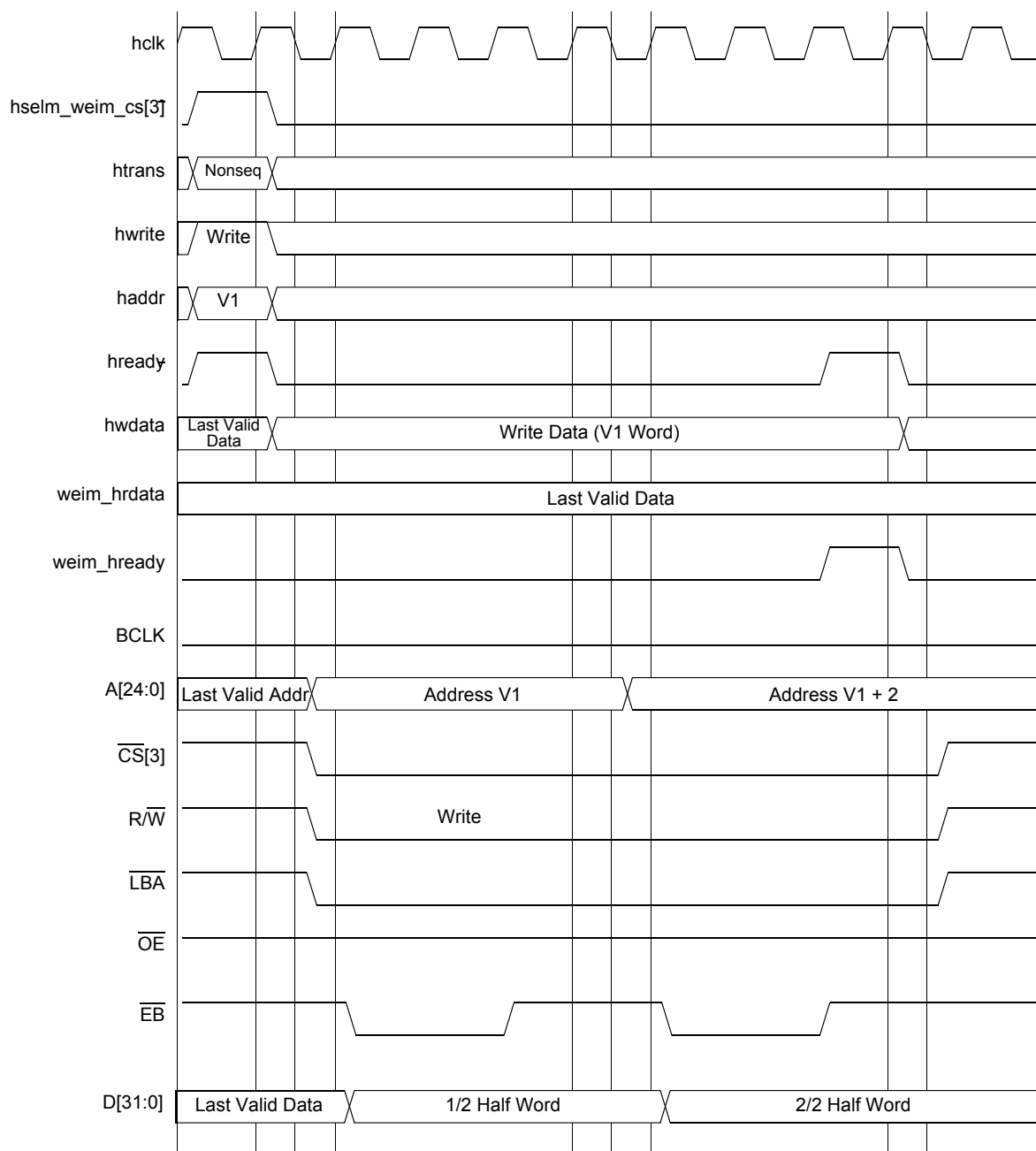


Figure 59. WSC = 3, WEA = 1, WEN = 3, A.WORD/E.HALF

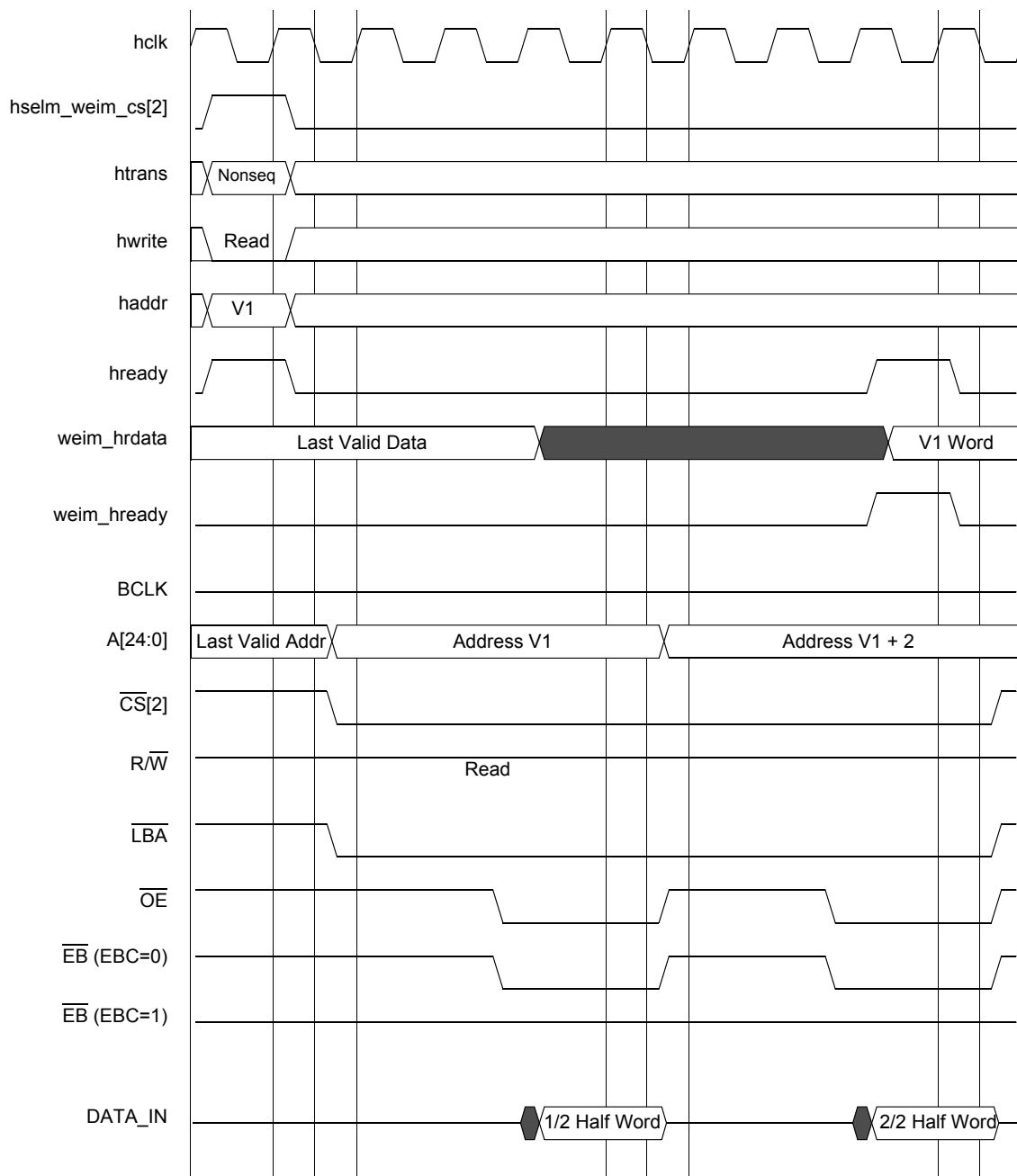
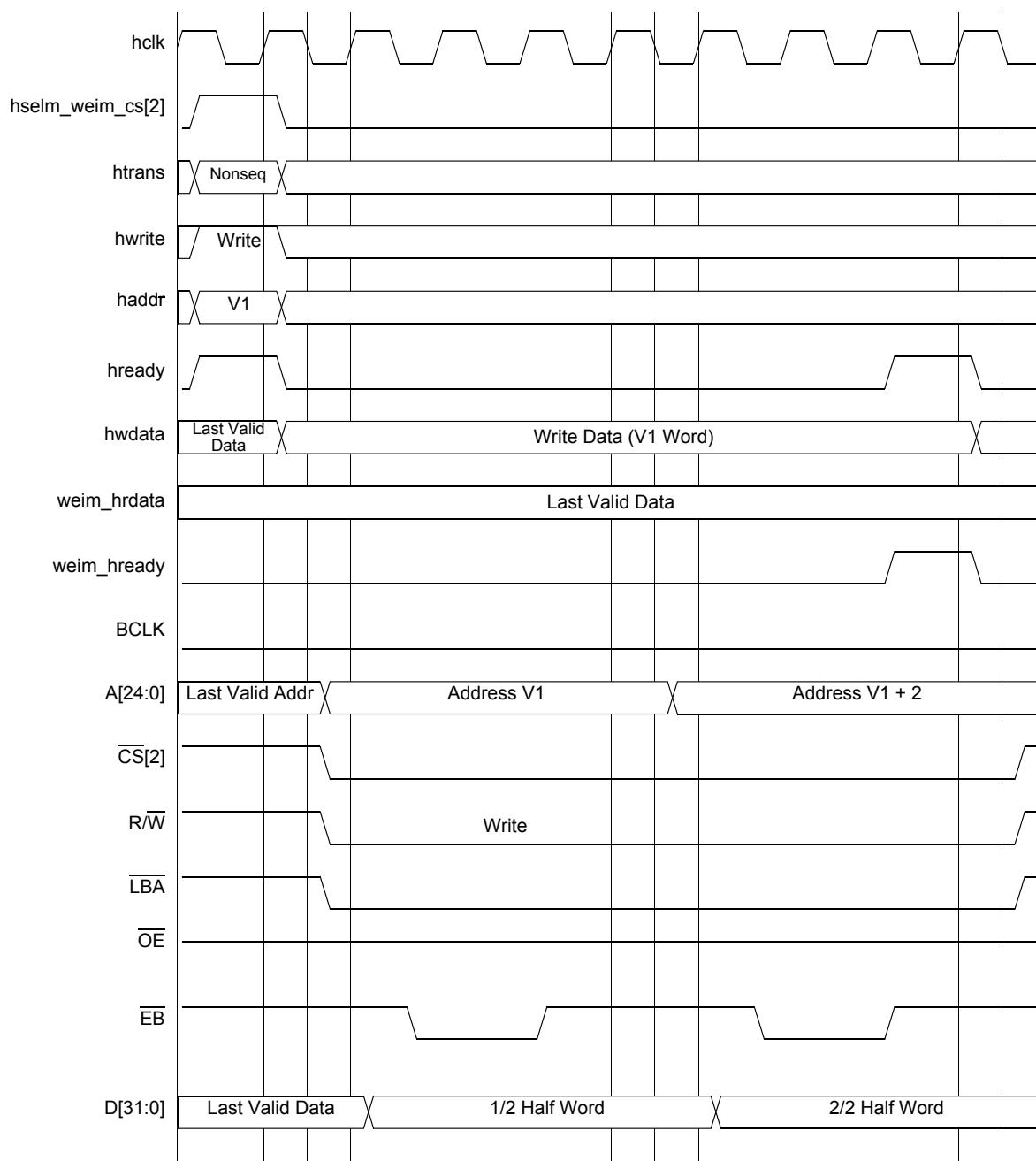


Figure 60. WSC = 3, OEA = 4, A.WORD/E.HALF



**Figure 61. WSC = 3, WEA = 2, WEN = 3, A.WORD/E.HALF**

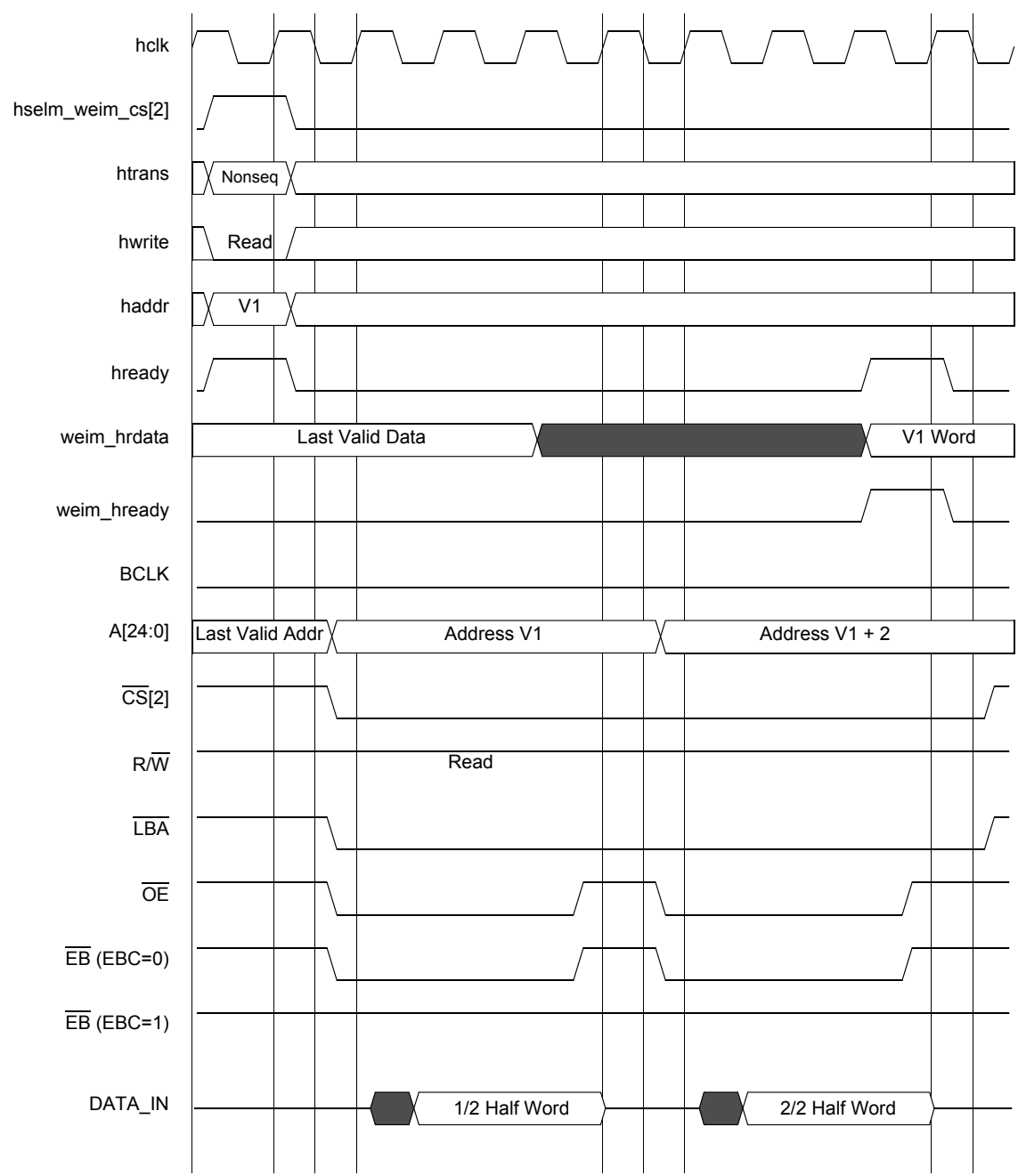


Figure 62. WSC = 3, OEN = 2, A.WORD/E.HALF

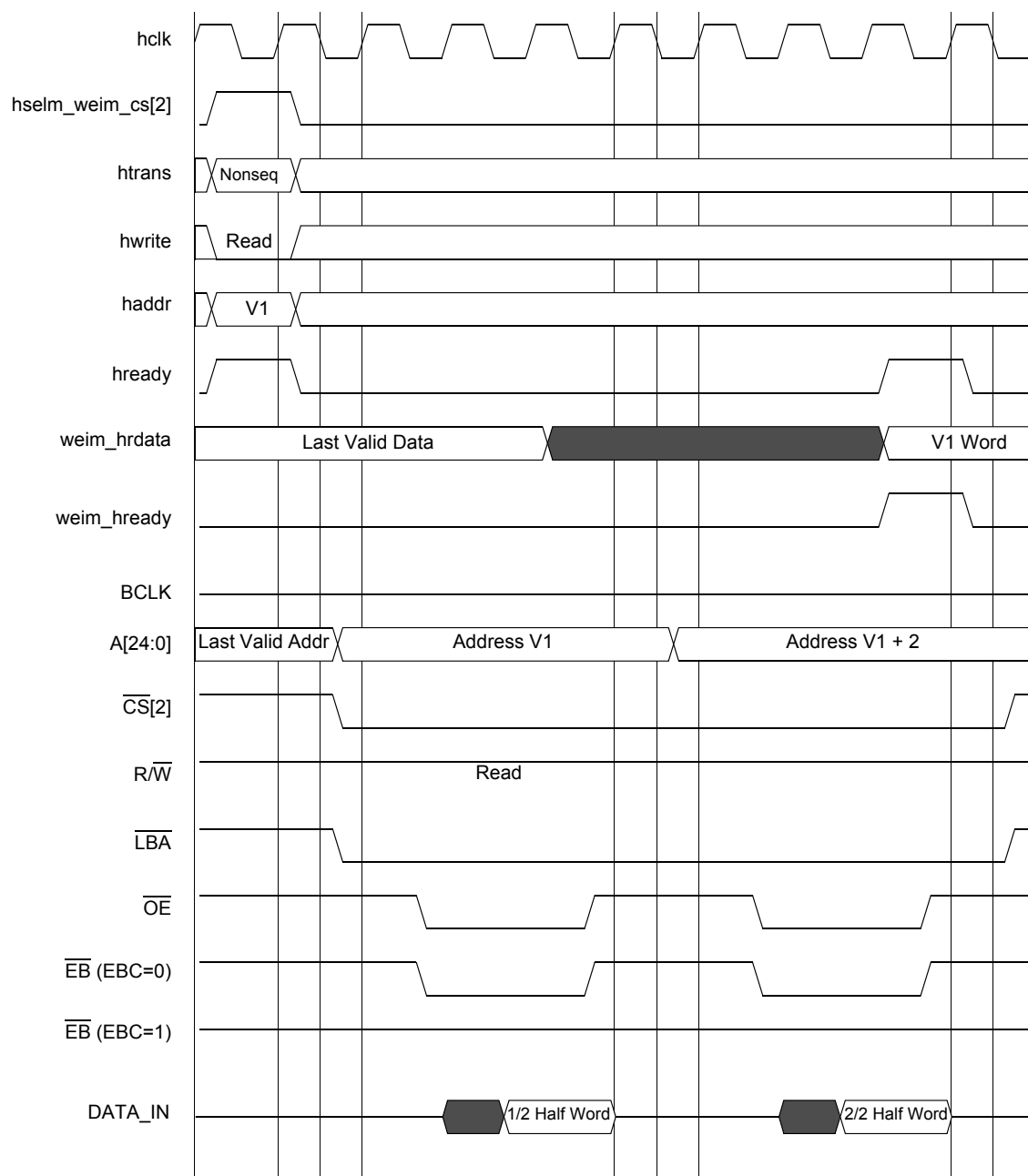


Figure 63. WSC = 3, OEA = 2, OEN = 2, A.WORD/E.HALF

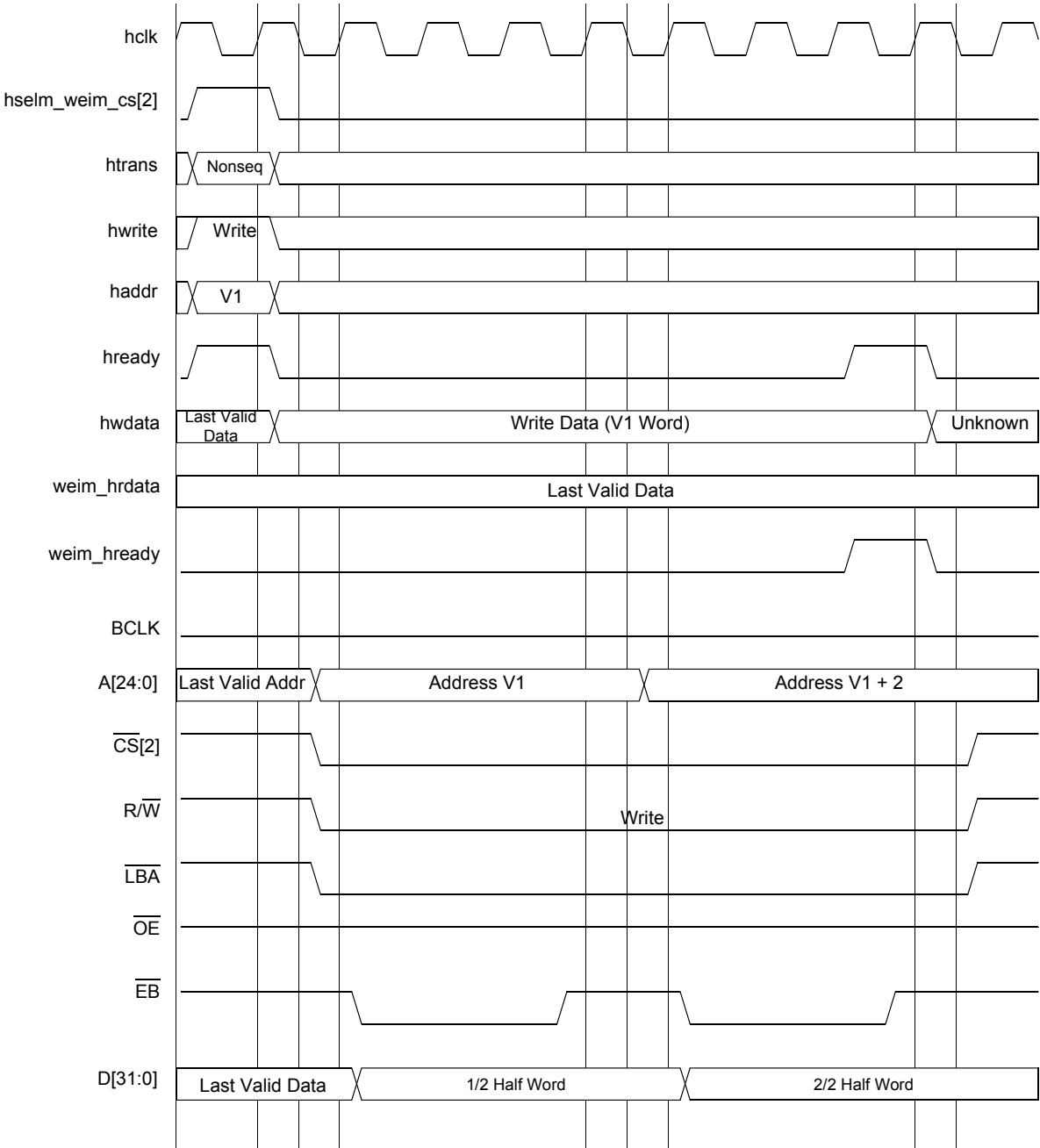


Figure 64. WSC = 2, WWS = 1, WEA = 1, WEN = 2, A.WORD/E.HALF



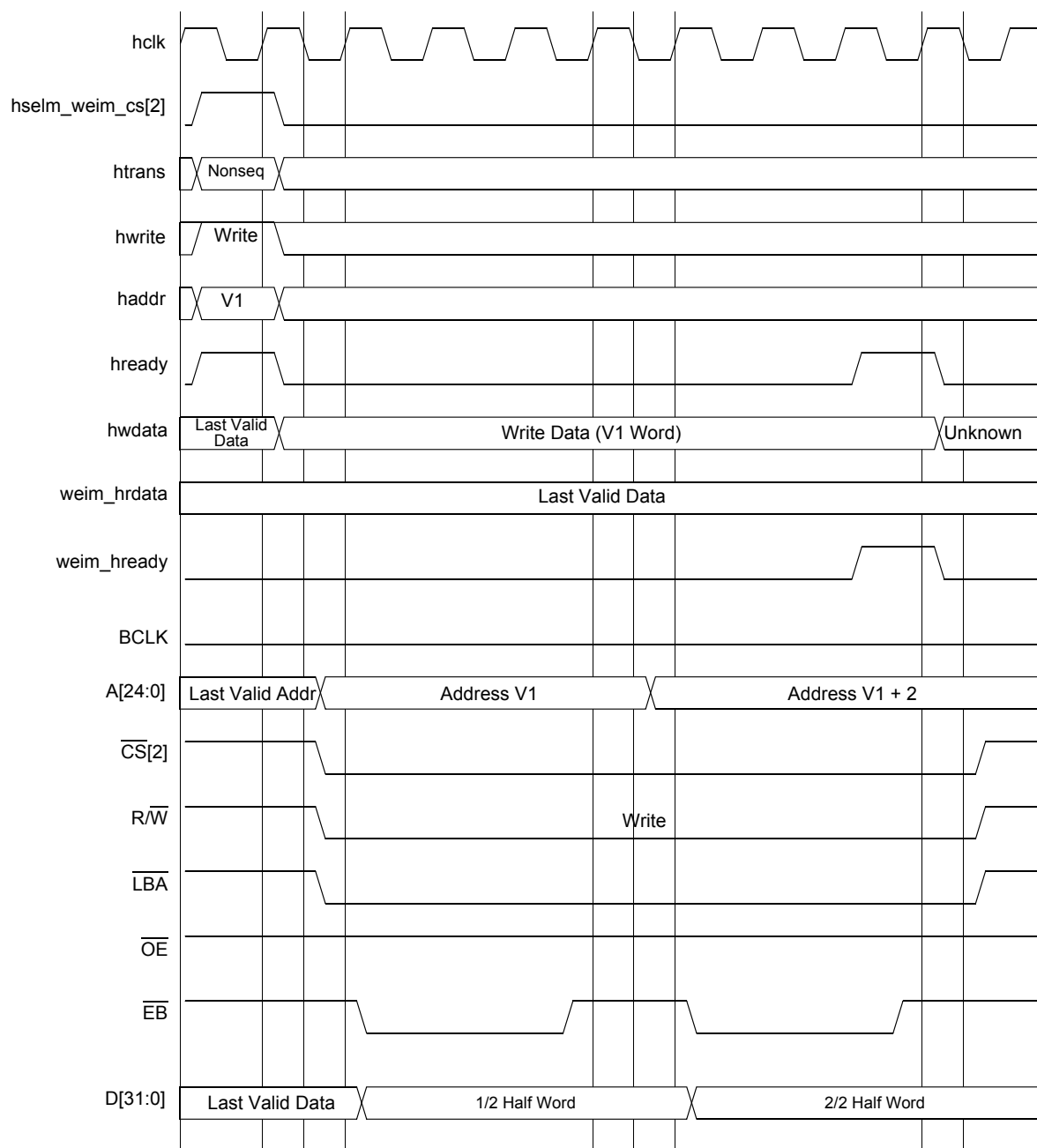


Figure 65. WSC = 1, WWS = 2, WEA = 1, WEN = 2, A.WORD/E.HALF

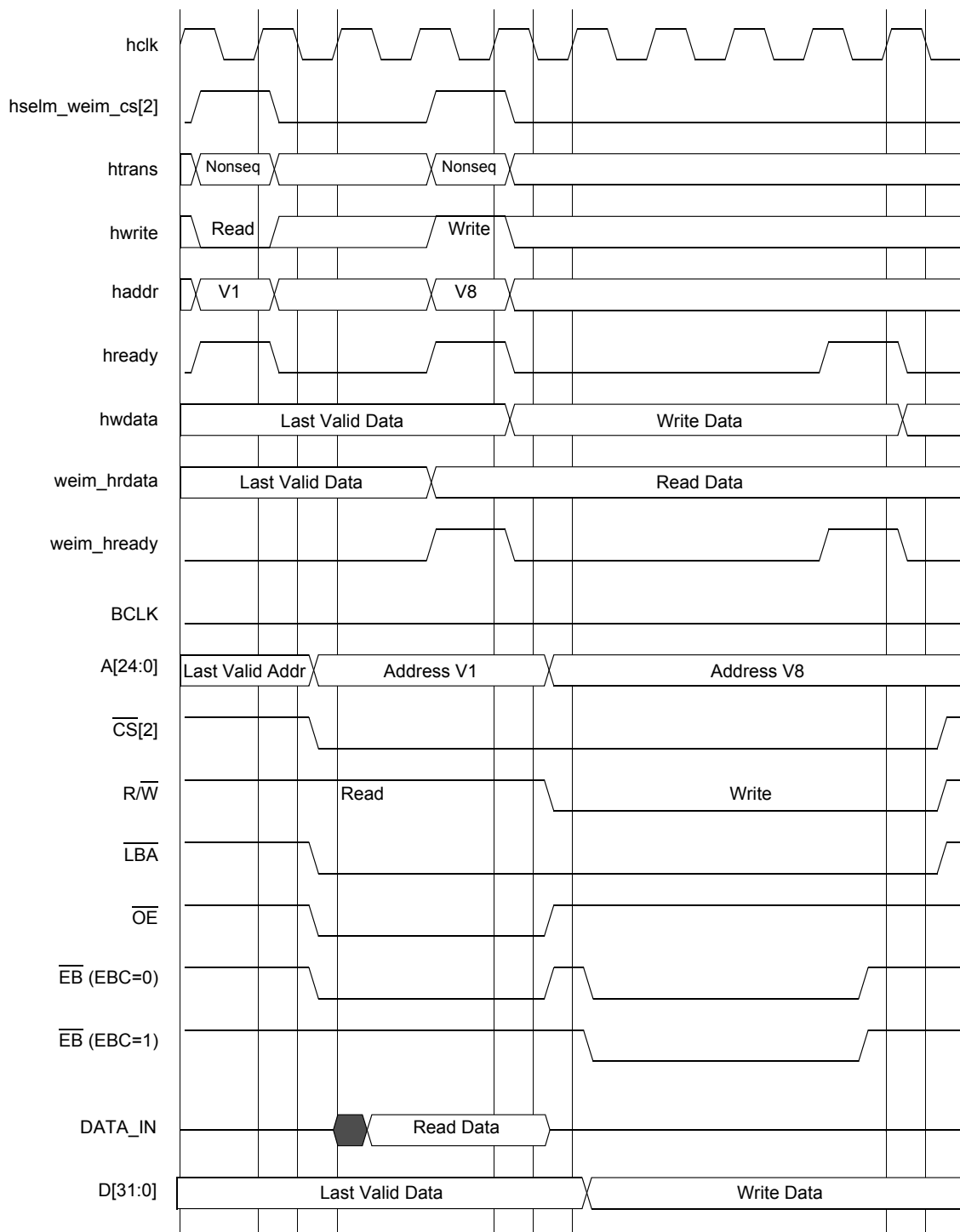
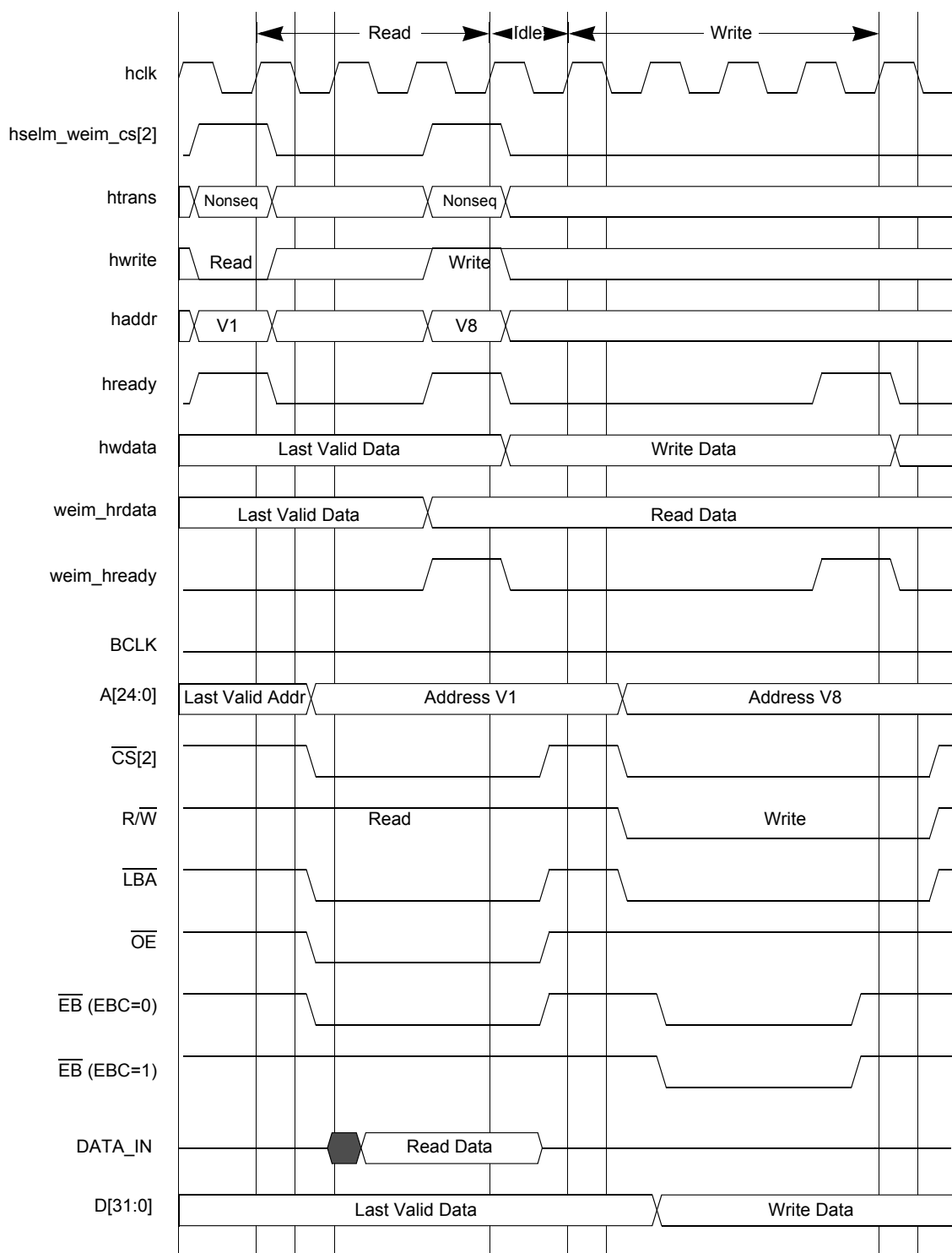


Figure 66. WSC = 2, WWS = 2, WEA = 1, WEN = 2, A.HALF/E.HALF



**Figure 67. WSC = 2, WWS = 1, WEA = 1, WEN = 2, EDC = 1, A.HALF/E.HALF**

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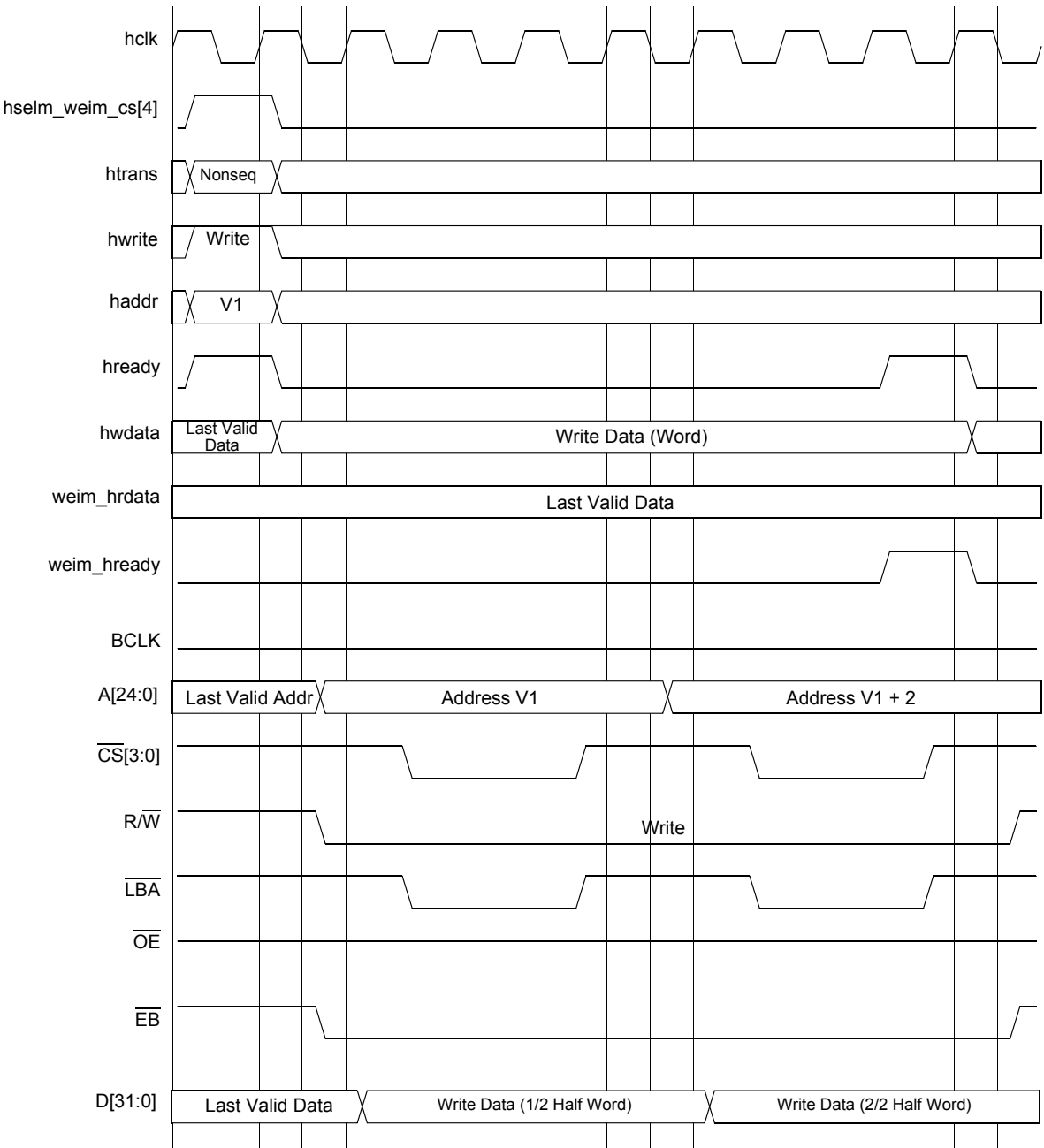


Figure 68. WSC = 2, CSA = 1, WWS = 1, A.WORD/E.HALF

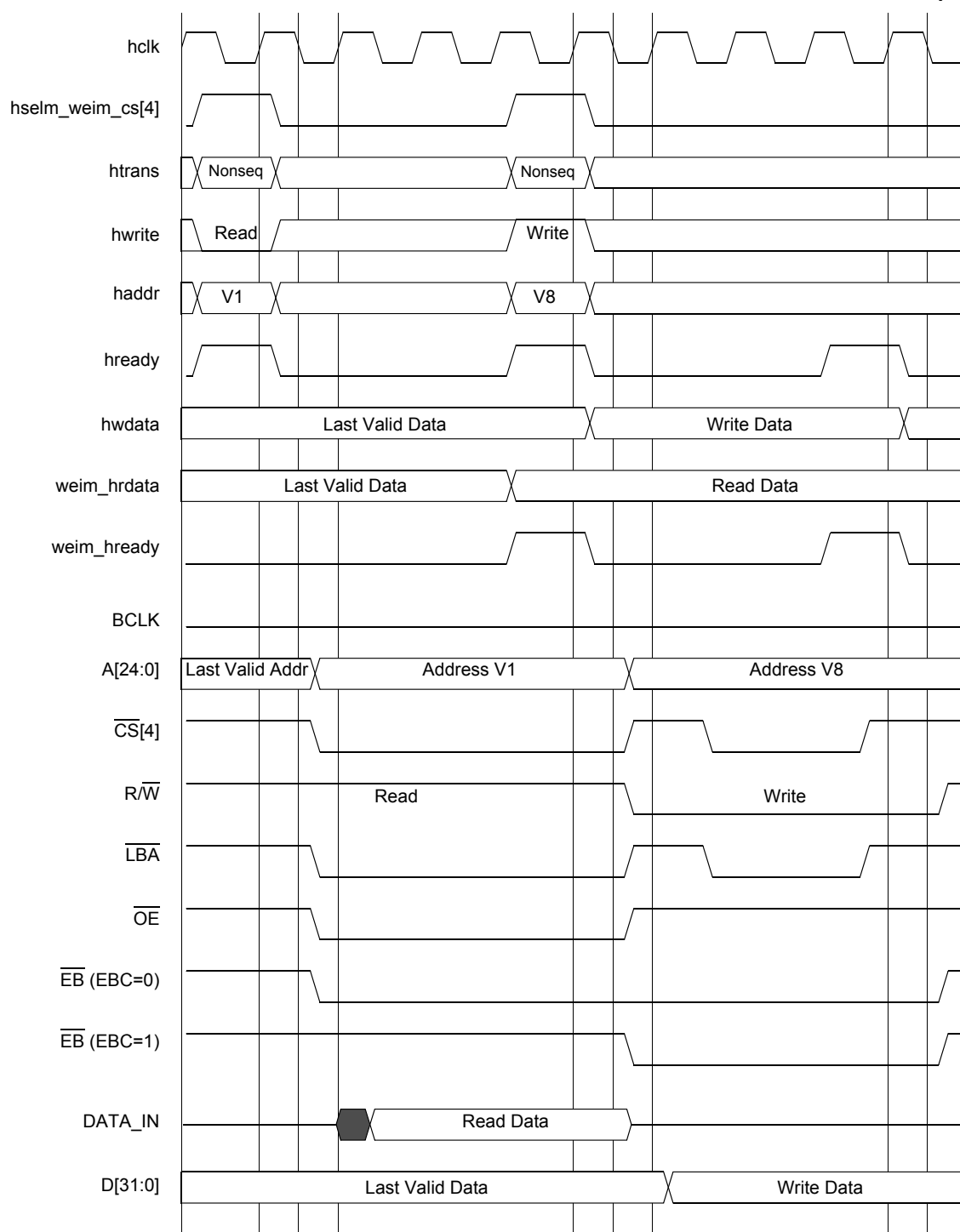


Figure 69. WSC = 3, CSA = 1, A.HALF/E.HALF

Specifications

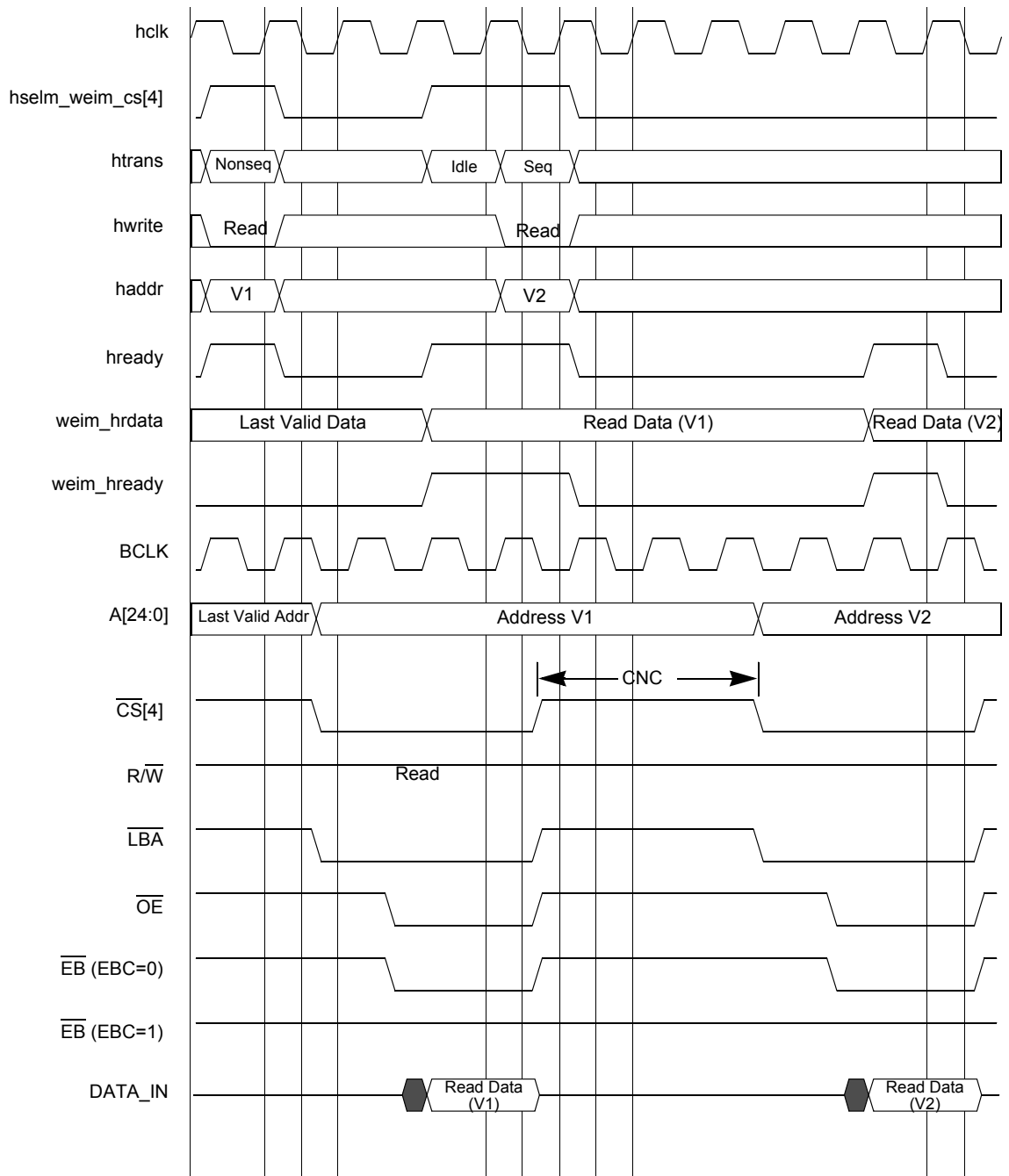


Figure 70. WSC = 2, OEA = 2, CNC = 3, BCM = 1, A.HALF/E.HALF

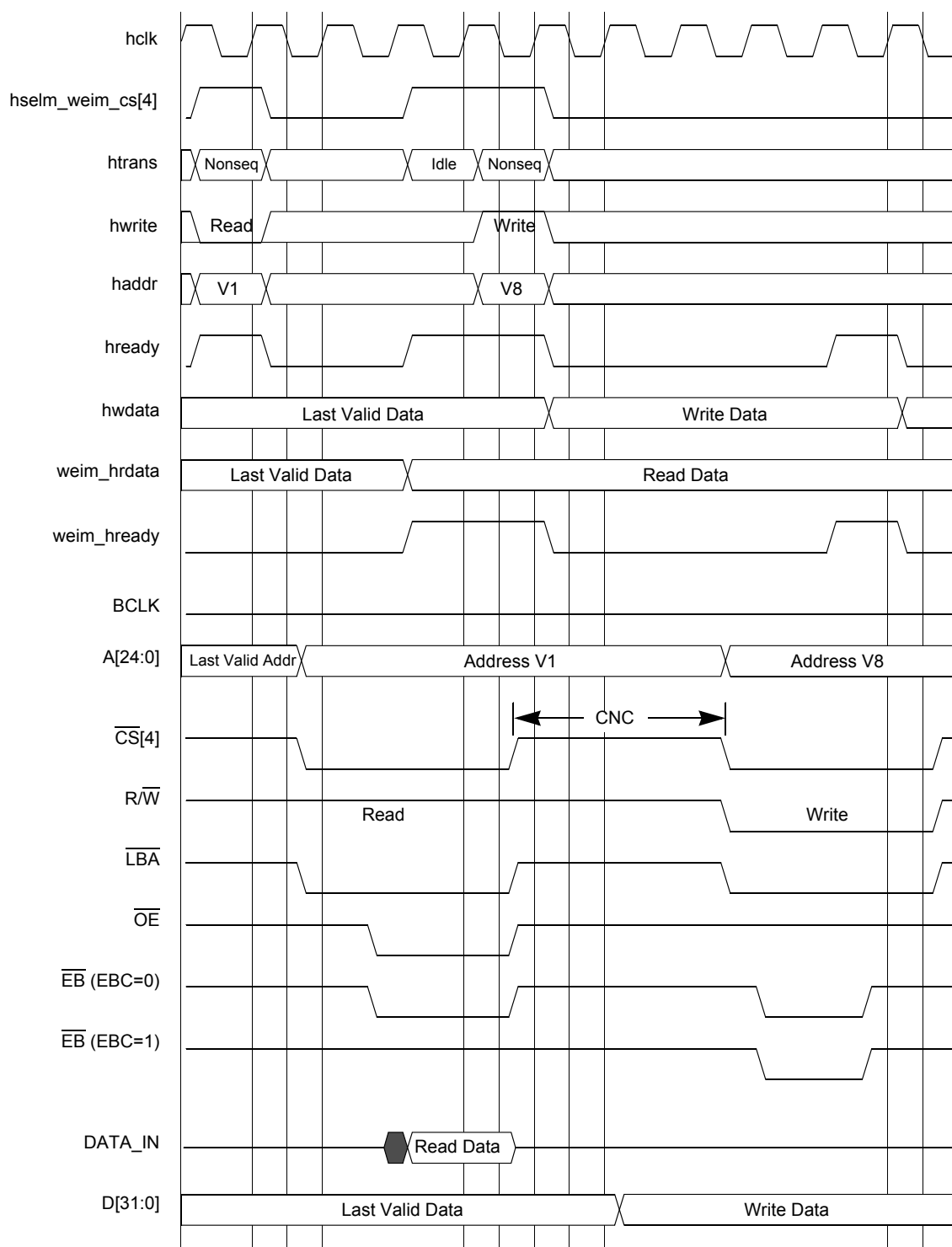


Figure 71. WSC = 2, OEA = 2, WEA = 1, WEN = 2, CNC = 3, A.HALF/E.HALF

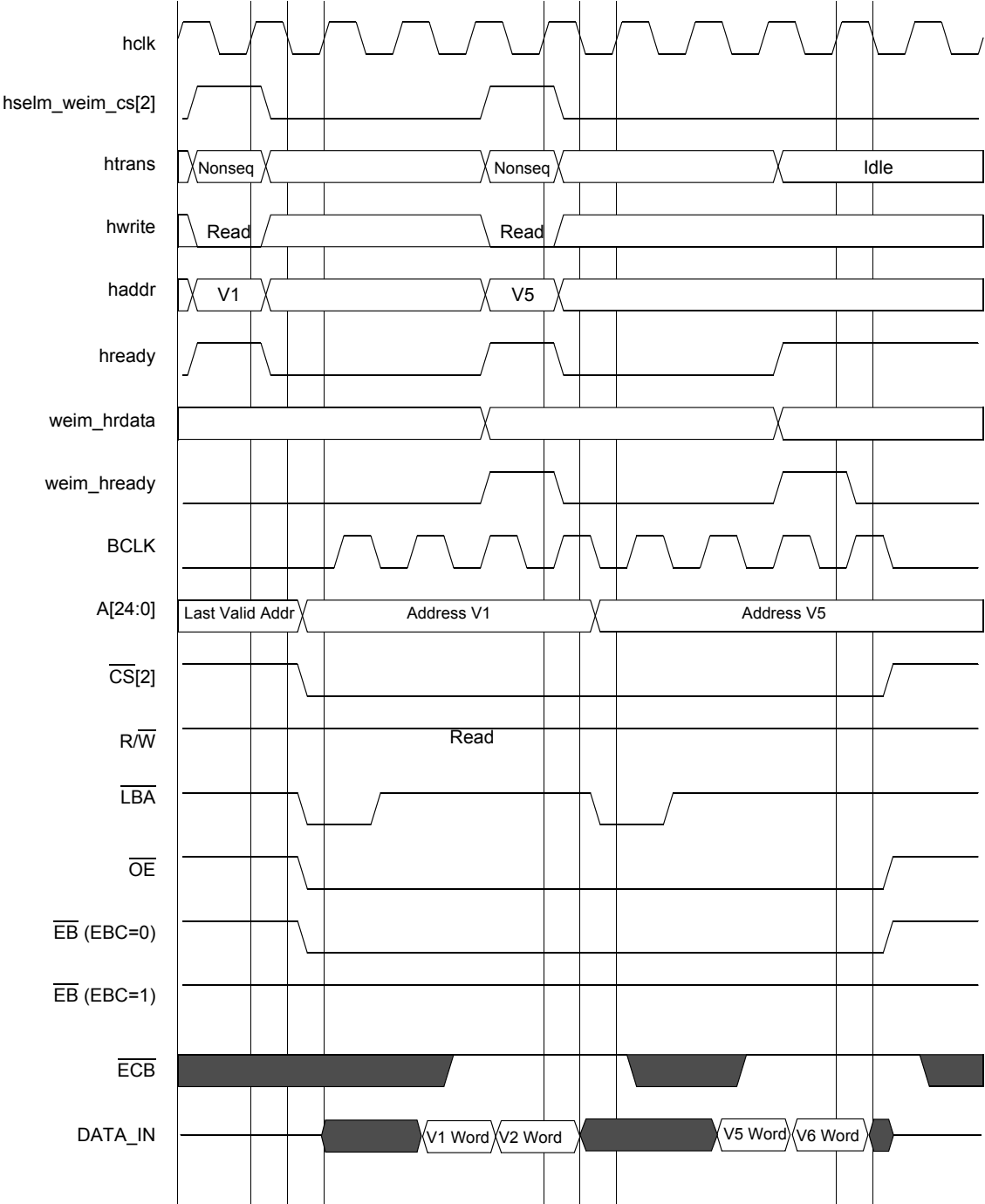


Figure 72. WSC = 3, SYNC = 1, A.HALF/E.HALF



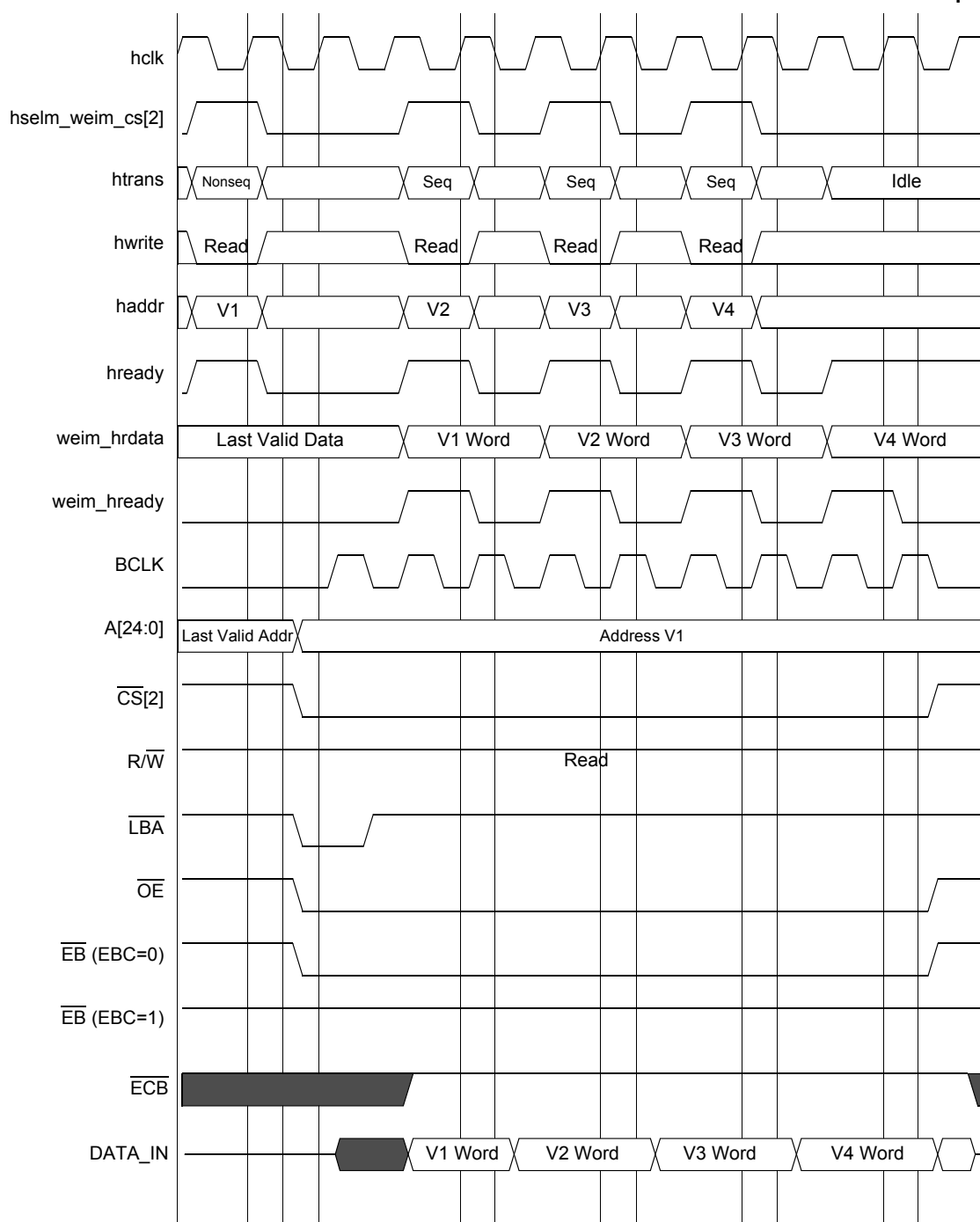


Figure 73. WSC = 2, SYNC = 1, DOL = [1/0], A.WORD/E.WORD

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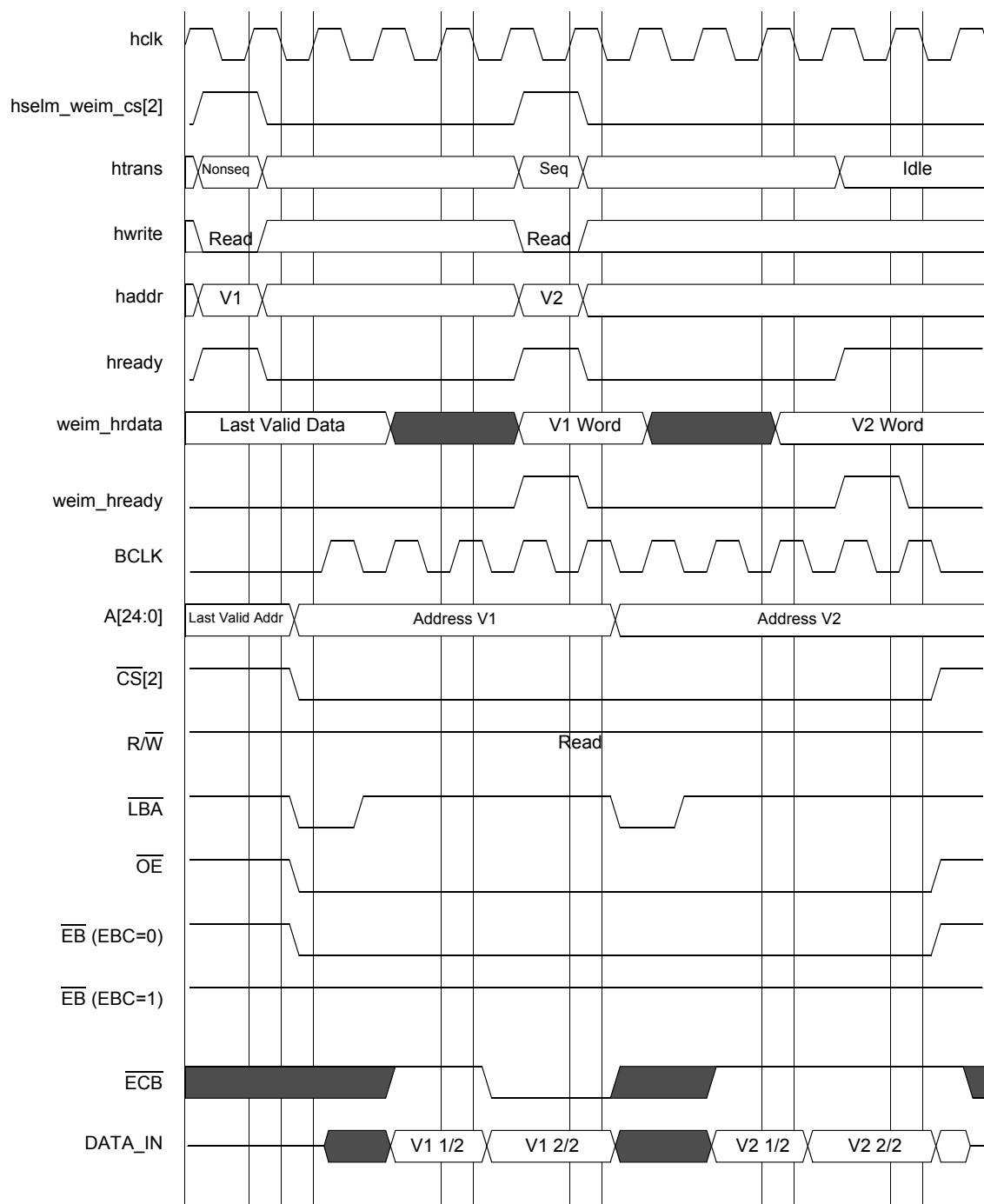
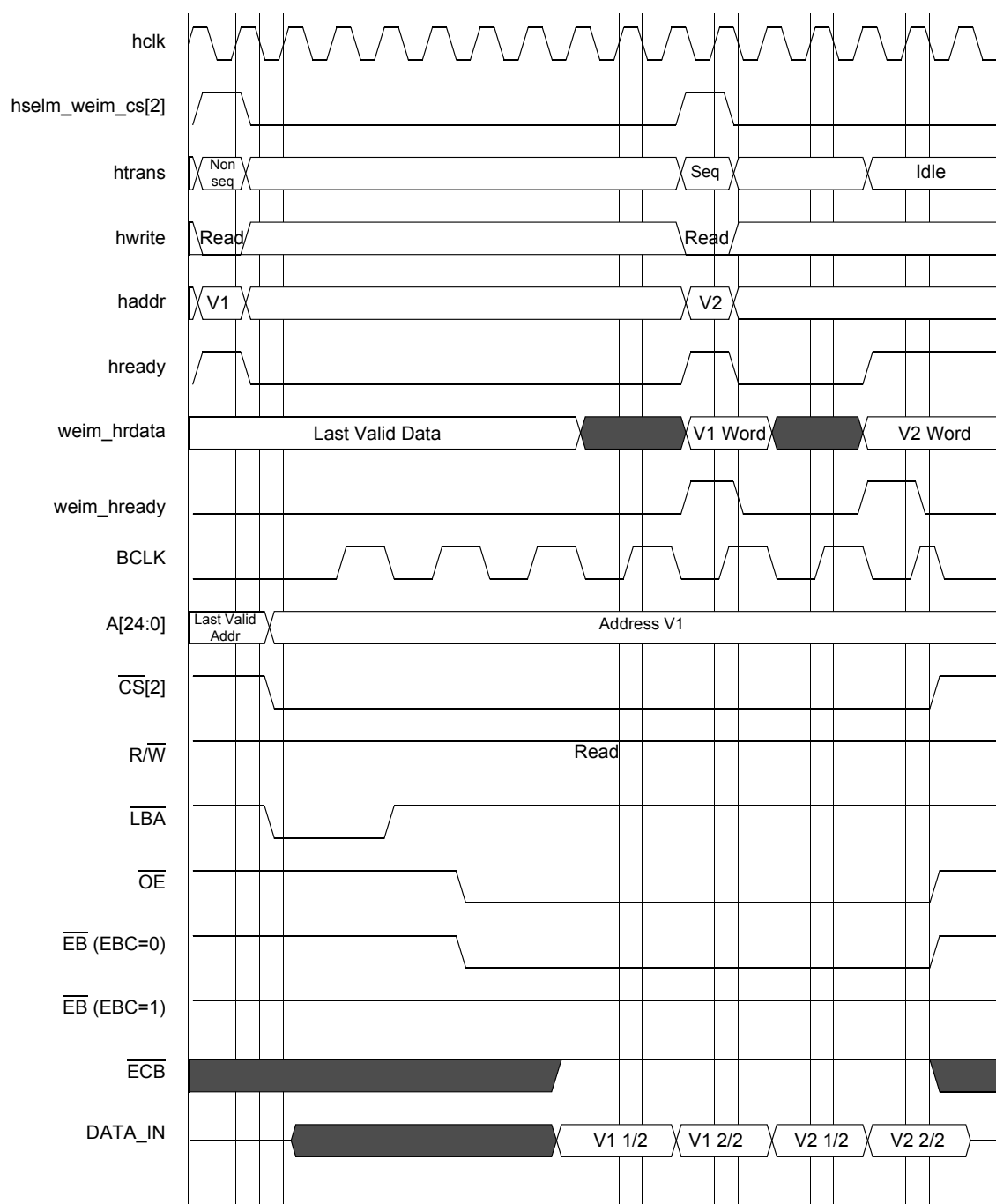


Figure 74. WSC = 2, SYNC = 1, DOL = [1/0], A.WORD/E.HALF



**Figure 75. WSC = 7, OEA = 8, SYNC = 1, DOL = 1, BCD = 1, BCS = 2, A.WORD/E.HALF**

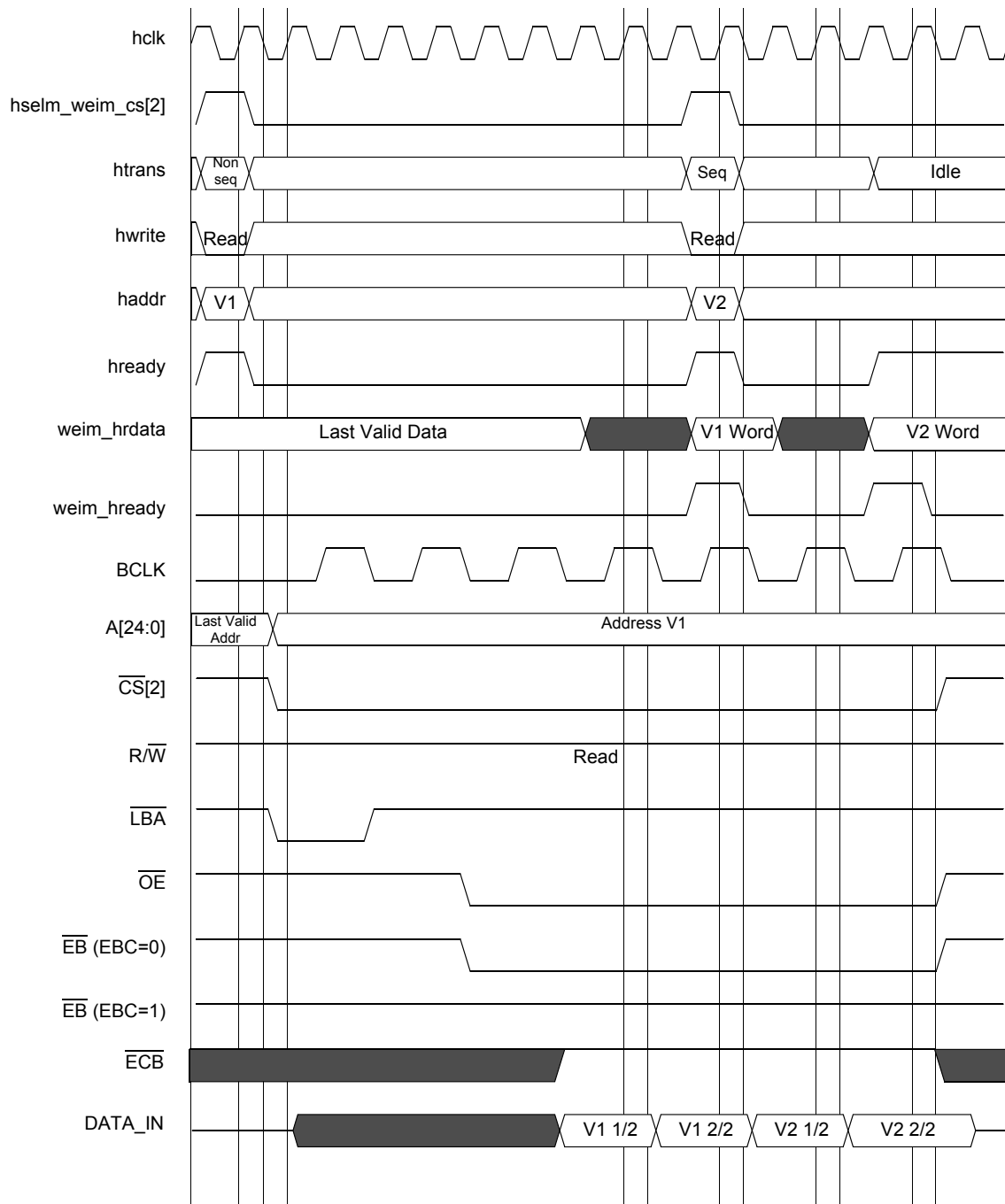


Figure 76. WSC = 7, OEA = 8, SYNC = 1, DOL = 1, BCD = 1, BCS = 1, A.WORD/E.HALF

### 3.20 DTACK Mode Memory Access Timing Diagrams

When enabled, the  $\overline{\text{DTACK}}$  input signal is used to externally terminate a data transfer. For  $\overline{\text{DTACK}}$  enabled operations, a bus time-out monitor generates a bus error when an external bus cycle is not terminated by the  $\overline{\text{DTACK}}$  input signal after 1024 HCLK clock cycles have elapsed, where HCLK is the internal system clock driven from the PLL module. For a 133 MHz HCLK setting, this time equates to 7.7  $\mu\text{s}$ . Refer to the Section 3.5, “DPLL Timing Specifications,” on page 18 for more information on how to generate different HCLK frequencies.

There are two modes of operation for the  $\overline{\text{DTACK}}$  input signal: rising edge detection or level sensitive detection with a programmable insensitivity time.  $\overline{\text{DTACK}}$  is only used during external asynchronous data transfers, thus the SYNC bit in the chip select control registers must be cleared.

During edge detection mode, the EIM will terminate an external data transfer following the detection of the  $\overline{\text{DTACK}}$  signal's rising edge, so long as it occurs within the 1024 HCLK cycle time. Edge detection mode is used for devices that follow the PCMCIA standard. Note that  $\overline{\text{DTACK}}$  rising edge detection mode can only be used for  $\overline{\text{CS}}[5]$  operations. To configure  $\overline{\text{CS}}[5]$  for  $\overline{\text{DTACK}}$  rising edge detection, the following bits must be programmed in the Chip Select 5 Control Register and EIM Configuration Register:

- WSC bit field set to 0x3F and CSA (or CSN) set to 1 or greater in the Chip Select 5 Control Register
- AGE bit set in the EIM Configuration Register

Other bits such as DSZ, OEA, OEN, and so on, may be set according to system and timing requirements of the external device. The requirement of setting CSA or CSN is required to allow the EIM to wait for the rising edge of  $\overline{\text{DTACK}}$  during back-to-back external transfers, such as during DMA transfers or an internal 32-bit access through an external 16-bit data port.

During level sensitive detection, the EIM will first hold off sampling the  $\overline{\text{DTACK}}$  signal for at least 2 HCLK cycles, and up to 5 HCLK cycles as programmed by the DCT bits in the Chip Select Control Register. After this insensitivity time, the EIM will sample  $\overline{\text{DTACK}}$  and if it detects that  $\overline{\text{DTACK}}$  is logic high, it will continue the data transfer at the programmed number of wait states. However, if the EIM detects that  $\overline{\text{DTACK}}$  is logic low, it will wait until  $\overline{\text{DTACK}}$  goes to logic high to continue the access, so long as this occurs within the 1024 HCLK cycle time. If at anytime during an external data transfer  $\overline{\text{DTACK}}$  goes to logic low, the EIM will wait until  $\overline{\text{DTACK}}$  returns to logic high to resume the data transfer. Level detection is often used for asynchronous devices such graphic controller chips. Level detection may be used with any chip select except CS[4] as it is multiplexed with the  $\overline{\text{DTACK}}$  signal. To configure a chip select for  $\overline{\text{DTACK}}$  level sensitive detection, the following bits must be programmed in the Chip Select Control Register and EIM Configuration Register:

- EW bit set, WSC set to > 1, and CSN set to < 3 in the Chip Select Control Register
- BCD/DCT set to desired “insensitivity time” in the Chip Select Control Register. The “insensitivity time” is dictated by the external device's timing requirements.
- AGE bit cleared in the EIM Configuration Register

Other bits such as DSZ, OEA, OEN, and so on, may be set according to system and timing requirements of the external device.

The waveforms in the following section provide examples of the  $\overline{\text{DTACK}}$  signal operation.

### 3.20.1 DTACK Example Waveforms: Internal ARM AHB Word Accesses to Word-Width (32-bit) Memory

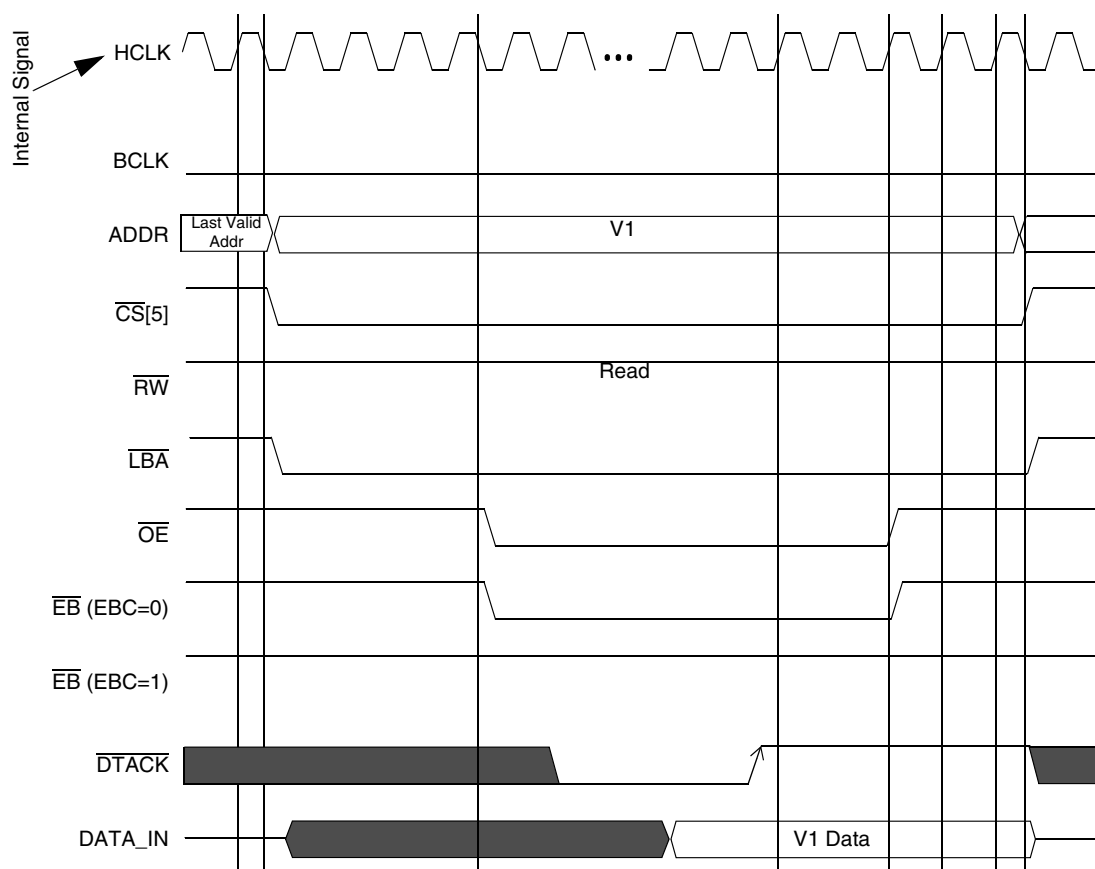
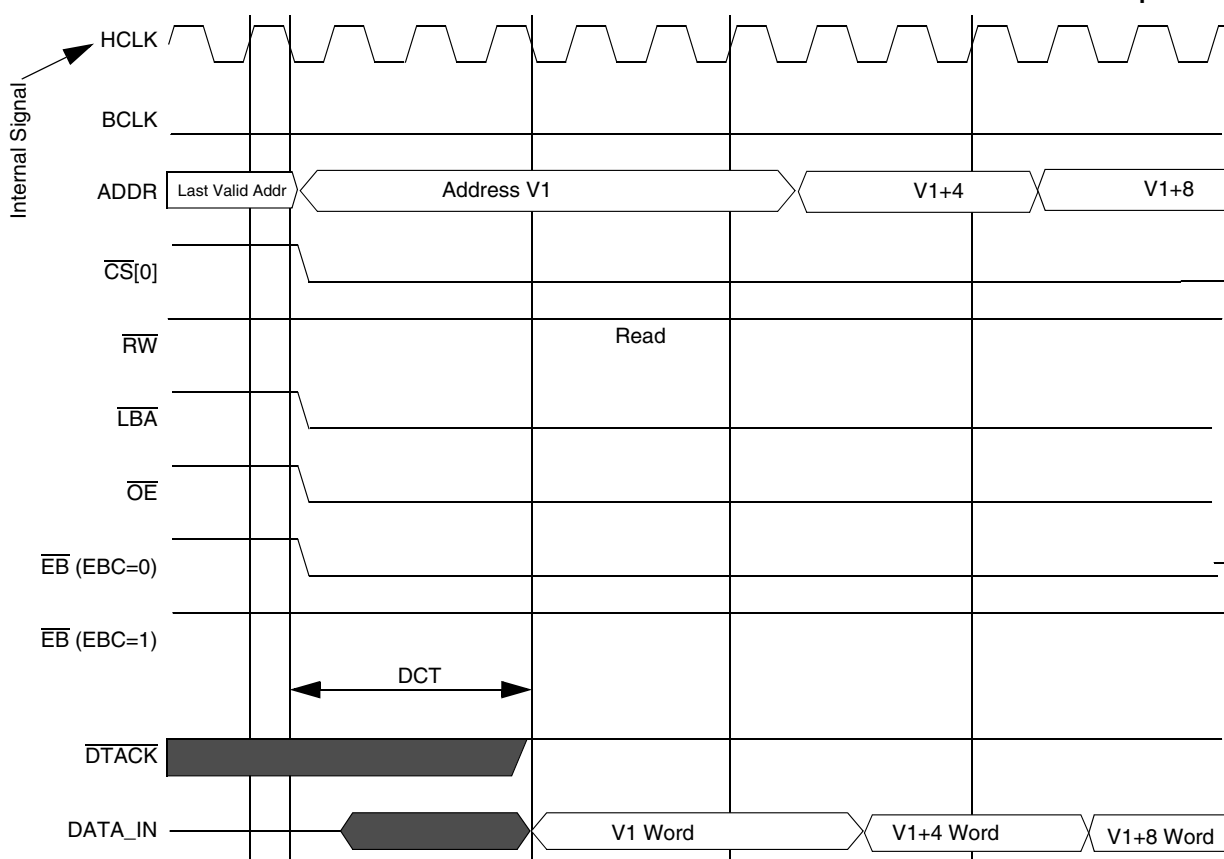
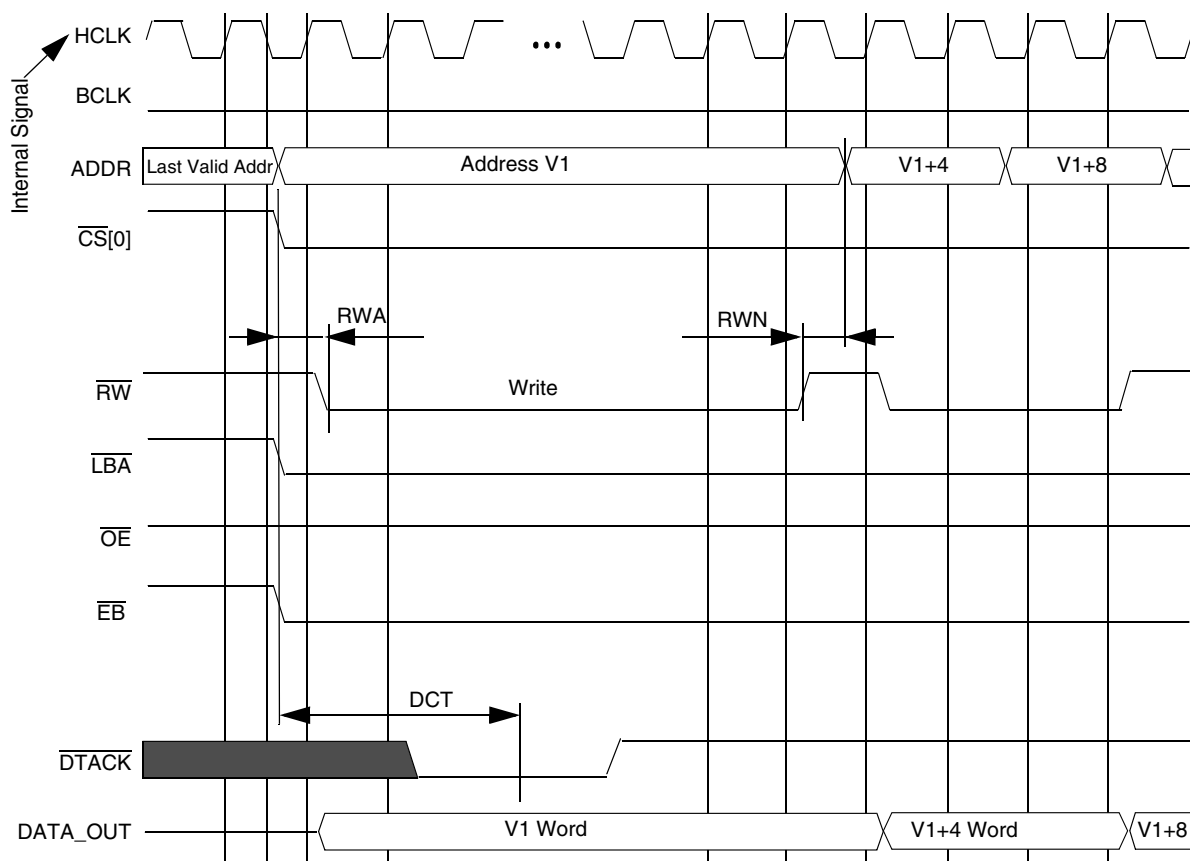


Figure 77. DTACK Edge Triggered Read Access, WSC=3F, OEA=8, OEN=5, AGE=1.



**Figure 78. DTACK Level Sensitive Sequential Read Accesses, WSC=2, EW=1, DCT=1, AGE=0  
(Example of DTACK staying high)**

## Specifications



**Figure 79. DTACK Level Sensitive Sequential Write Accesses, WSC=2, EW=1, RWA=1, RWN=1, DCT=1, AGE=0 (Example of DTACK Asserting)**



## 3.21 I<sup>2</sup>C Module

The I<sup>2</sup>C communication protocol consists of seven elements: START, Data Source/Recipient, Data Direction, Slave Acknowledge, Data, Data Acknowledge, and STOP.

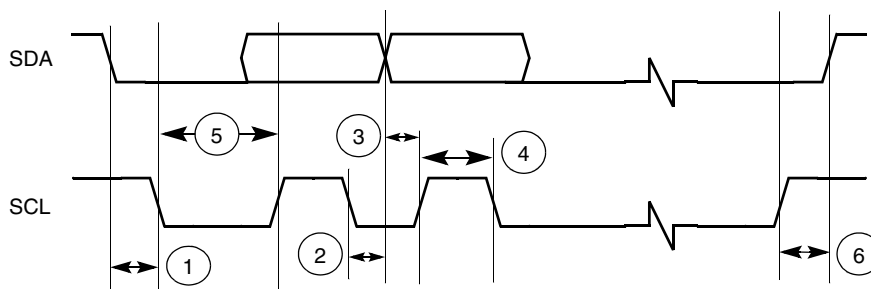


Figure 80. Definition of Bus Timing for I<sup>2</sup>C

Table 42. I<sup>2</sup>C Bus Timing Parameter Table

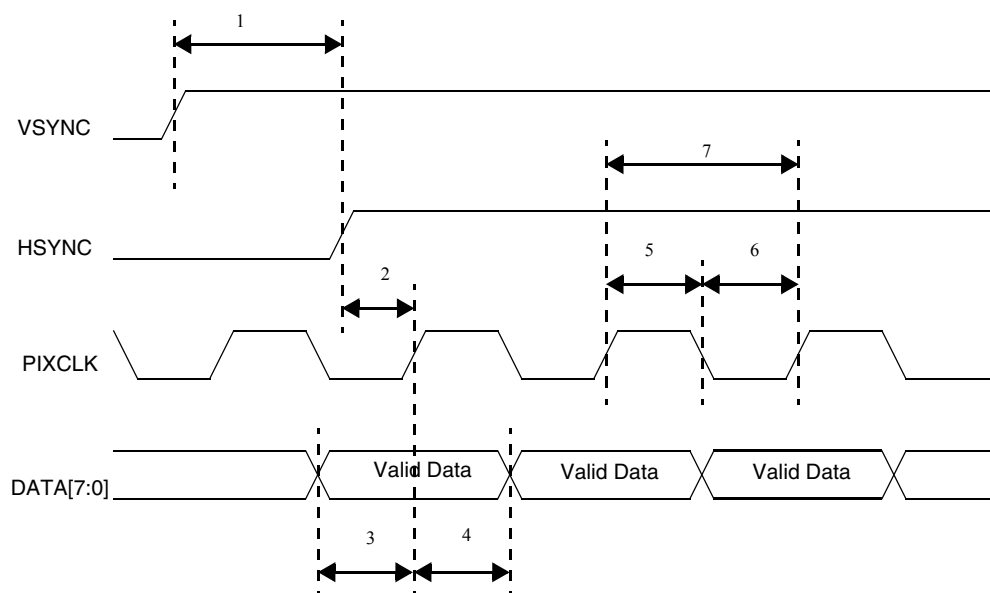
Ref No.	Parameter	1.8V +/- 0.10V		3.0V +/- 0.30V		Unit
		Minimum	Maximum	Minimum	Maximum	
	SCL Clock Frequency	0	100	0	100	kHz
1	Hold time (repeated) START condition	114.8	–	111.1	–	ns
2	Data hold time	0	69.7	0	72.3	ns
3	Data setup time	3.1	–	1.76	–	ns
4	HIGH period of the SCL clock	69.7	–	68.3	–	ns
5	LOW period of the SCL clock	336.4	–	335.1	–	ns
6	Setup time for STOP condition	110.5	–	111.1	–	ns

## 3.22 CMOS Sensor Interface

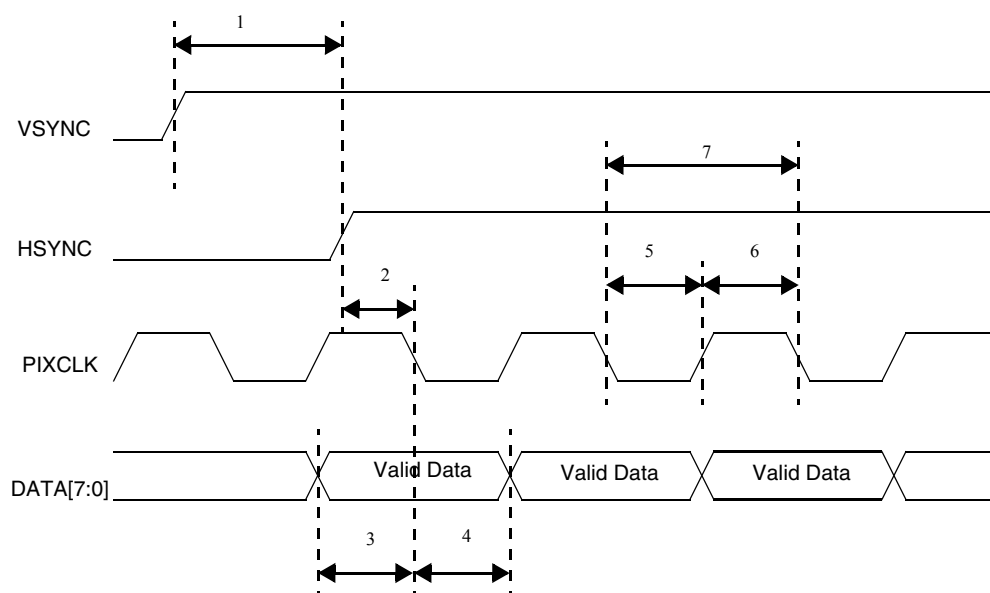
The CSI module consists of a control register to configure the interface timing, a control register for statistic data generation, a status register, interface logic, a 32 × 32 image data receive FIFO, and a 16 × 32 statistic data FIFO.

### 3.22.1 Gated Clock Mode

Figure 81 shows the timing diagram when the CMOS sensor output data is configured for negative edge and the CSI is programmed to received data on the positive edge. Figure 82 on page 98 shows the timing diagram when the CMOS sensor output data is configured for positive edge and the CSI is programmed to received data in negative edge. The parameters for the timing diagrams are listed in Table 43 on page 98. The formula for calculating the pixel clock rise and fall time is located in Section 3.22.3, “Calculation of Pixel Clock Rise/Fall Time,” on page 101.



**Figure 81. Sensor Output Data on Pixel Clock Falling Edge  
CSI Latches Data on Pixel Clock Rising Edge**



**Figure 82. Sensor Output Data on Pixel Clock Rising Edge  
CSI Latches Data on Pixel Clock Falling Edge**

**Table 43. Gated Clock Mode Timing Parameters**

Number	Parameter	Minimum	Maximum	Unit
1	csi_vsync to csi_hsync	$9 * T_{HCLK}$	—	ns
2	csi_hsync to csi_pixclk	3	$(T_P/2) - 3$	ns
3	csi_d setup time	1	—	ns
4	csi_d hold time	1	—	ns

**Table 43. Gated Clock Mode Timing Parameters**

Number	Parameter	Minimum	Maximum	Unit
5	csi_pixclk high time	$T_{HCLK}$	–	ns
6	csi_pixclk low time	$T_{HCLK}$	–	ns
7	csi_pixclk frequency	0	$HCLK / 2$	MHz

HCLK = AHB System Clock

$T_{HCLK}$  = Period for HCLK

$T_P$  = Period of CSI\_PIXCLK

The limitation on pixel clock rise time/fall time is not specified. It should be calculated from the hold time and setup time based on the following assumptions:

Rising-edge latch data

max rise time allowed = (positive duty cycle - hold time)

max fall time allowed = (negative duty cycle - setup time)

In most of case, duty cycle is 50 / 50, therefore

max rise time = (period / 2 - hold time)

max fall time = (period / 2 - setup time)

For example: Given pixel clock period = 10ns, duty cycle = 50 / 50, hold time = 1ns, setup time = 1ns.

positive duty cycle =  $10 / 2 = 5\text{ns}$

$\geq$  max rise time allowed =  $5 - 1 = 4\text{ns}$

negative duty cycle =  $10 / 2 = 5\text{ns}$

$\geq$  max fall time allowed =  $5 - 1 = 4\text{ns}$

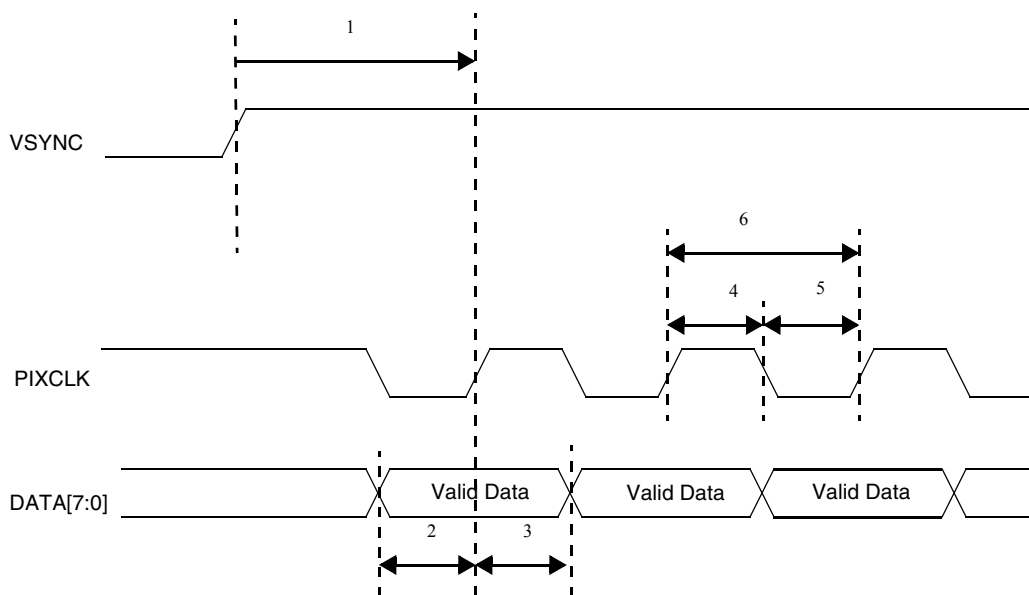
Falling-edge latch data

max fall time allowed = (negative duty cycle - hold time)

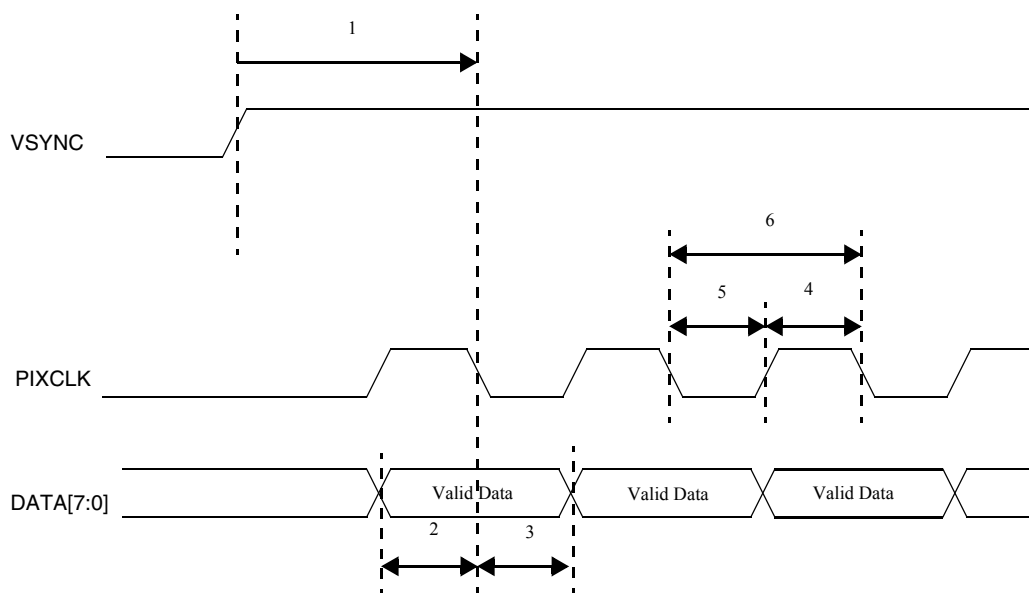
max rise time allowed = (positive duty cycle - setup time)

### 3.22.2 Non-Gated Clock Mode

Figure 83 shows the timing diagram when the CMOS sensor output data is configured for negative edge and the CSI is programmed to received data on the positive edge. Figure 84 on page 100 shows the timing diagram when the CMOS sensor output data is configured for positive edge and the CSI is programmed to received data in negative edge. The parameters for the timing diagrams are listed in Table 44 on page 100. The formula for calculating the pixel clock rise and fall time is located in Section 3.22.3, “Calculation of Pixel Clock Rise/Fall Time,” on page 101.



**Figure 83. Sensor Output Data on Pixel Clock Falling Edge  
CSI Latches Data on Pixel Clock Rising Edge**



**Figure 84. Sensor Output Data on Pixel Clock Rising Edge  
CSI Latches Data on Pixel Clock Falling Edge**

**Table 44. Non-Gated Clock Mode Parameters**

Number	Parameter	Minimum	Maximum	Unit
1	csi_vsync to csi_pixclk	$9 * T_{HCLK}$	—	ns
2	csi_d setup time	1	—	ns
3	csi_d hold time	1	—	ns
4	csi_pixclk high time	$T_{HCLK}$	—	ns

**Table 44. Non-Gated Clock Mode Parameters (Continued)**

Number	Parameter	Minimum	Maximum	Unit
5	csi_pixclk low time	$T_{HCLK}$	—	ns
6	csi_pixclk frequency	0	$HCLK / 2$	MHz

HCLK = AHB System Clock

 $T_{HCLK}$  = Period of HCLK

### 3.22.3 Calculation of Pixel Clock Rise/Fall Time

The limitation on pixel clock rise time/fall time is not specified. It should be calculated from the hold time and setup time based on the following assumptions:

Rising-edge latch data

- max rise time allowed = (positive duty cycle - hold time)
- max fall time allowed = (negative duty cycle - setup time)

In most of case, duty cycle is 50 / 50, therefore:

- max rise time = (period / 2 - hold time)
- max fall time = (period / 2 - setup time)

For example: Given pixel clock period = 10ns, duty cycle = 50 / 50, hold time = 1ns, setup time = 1ns.

positive duty cycle =  $10 / 2 = 5\text{ns}$

$\geq$  max rise time allowed =  $5 - 1 = 4\text{ns}$

negative duty cycle =  $10 / 2 = 5\text{ns}$

$\geq$  max fall time allowed =  $5 - 1 = 4\text{ns}$

Falling-edge latch data

- max fall time allowed = (negative duty cycle - hold time)
- max rise time allowed = (positive duty cycle - setup time)

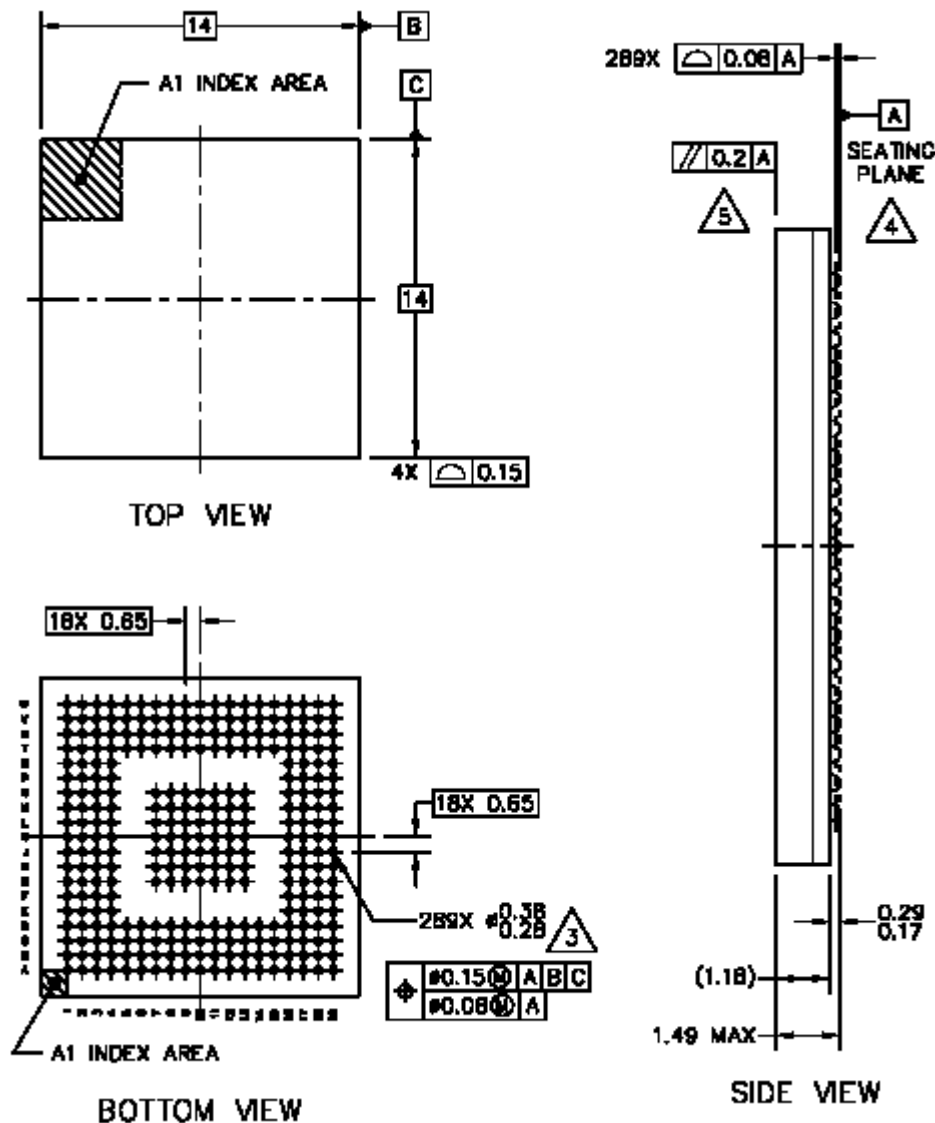
# 4 Pin-Out and Package Information

Table 45. i.MX21 Pin Assignments

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19
A	LD9	LD12	LD14	REV	HSYNC	OE_ACD	SD2_D2	CSI_D0	CSI_PIXCLK	CSI_VSYNC	USBH1_FS	USBH1_OE	USBG_FS	TOUT	SAP_TXDAT	SSI1_CLK	SSI2_RXDAT	SSI2_TXDAT	SSI3_FS
B	LD7	LD5	LD11	LD16	PS	CONTRAST	SD2_D0	SD2_CMD	CSI_D4	CSI_D6	USB_PWR	USBG_SCL	USBG_TXDM	SAP_FS	SSI1_FS	SSI2_FS	SSI3_TXDAT	I2C_DATA	CSPi2_SS2
C	LD1	LD3	LD6	LD10	LD17	VSYNC	SD2_D3	CSI_D1	CSI_MCLK	CSI_HSYNC	USB_OC	USBH1_RXDM	USBG_RXDM	TIN	SSI1_TXDAT	SSI3_RXDAT	SSI3_CLK	I2C_CLK	CSPi2_SS1
D	LD2	LD0	LD13	CLS	QVDD	QVSS	SD2_D1	SD2_CLK	CSI_D2	CSI_D7	USBH1_TXDM	USBH1_RXDP	USBG_ON	USBG_RXDP	SAP_RXDAT	SSI1_RXDAT	SSI2_CLK	CSPi2_SS0	CSPi2_SCLK
E	LD8	LD4	LD15	SPL_SPR												SAP_CLK	CSPi2_MISO	CSPi1_SS2	CSPi2_MOSI
F	A24_NFIO14	D31	A25_NFIO15	LSCLK												CSPi1_SS1	CSPi1_MISO	KP_ROW0	CSPi1_SS0
G	A22_NFIO12	D29	A23_NFIO13	D30			NVDD6	NVSS6	CSI_D3	USB_BYP	USBH_ON	USBG_SDA	USBG_TXDP			KP_ROW1	KP_ROW3	UART2_CTS	KP_ROW4
H	A20	D27	A21_NFIO11	D28			NVDD1	NVSS5	CSI_D5	CSPi1_SCLK	CSPi1_RDY	USBH1_TXDP	USBG_OE			TEST_WB4	TEST_WB2	TEST_WB3	PWMO
J	A19	A18	D25	D26			NVDD1	NVDD5	NVDD4	KP_ROW5	KP_ROW2	CSPi1_MOSI	TEST_WB0			UART2_RTS	KP_COL1	KP_COL0	TEST_WB1
K	A16	A17	D23	D24			NVSS1	NVSS4	QVDDX	UART1_RXD	TDO	QVDD	QVSS			KP_COL3	KP_COL5	KP_COL4	KP_COL2
L	A14_NFIO9	A15_NFIO10	D21	D22			NVSS1	NVDD3	QVDD	QVSS	NFIO2	NFWP	UART1_TXD			UART2_TXD	UART3_RTS	UART3_CTS	UART3_TXD
M	D19	A13_NFIO8	D20	D18			NVDD2	NVDD3	NVSS3	QVSS	NFIO7	NFRB	EXT_48M			UART2_RXD	UART3_RXD	UART1_RTS	UART1_CTS
N	A11	A12	D17	D16			LBA	NVSS3	SDCKE0	NVSS1	NVSS1	NVDD1	NVDD1			SD1_D0	TCK	SD1_D1	RTCK
P	A9	A10	D15	D14												SD1_D2	SD1_CMD	TDI	TMS
R	A7	A8	D13	D12												SD1_CLK	EXT_266M	NVSS2	TRST
T	A5	A6	EB3	D10	CS3	CS1	BCLK	MA11	RAS	CAS	NFIO5	NFIO3	NFWE	RESET_IN	NFCE	BOOT1	SD1_D3	CLKMODE1	CLKMODE0
U	D11	EB1	EB2	OE	CS4	D6	ECB	D3	MA10	PC_PWRON		NFIO4	NFIO1	NFALE	NFCLE	POR	BOOT2	BOOT3	XTAL32K
V	A4	EB0	D9	D8	CS5	D5	CS0	RW	D1	JTAG_CTRL	SDWE	CLKO	NFIO6	QVSS	RESET_OUT	BOOT0	OSC26M_TEST	VDDA	EXTAL32K
W	A3	A2	D7	A1	CS2	A0	D4	D2	D0	SDCLK	SDCKE1	NFIO0	NFRE	QVDD	QVSS	EXTAL26M	XTAL26M	QVDD	QVSS

## 4.1 MAPBGA Package Dimensions

Figure 85 illustrates the MAPBGA 14 mm × 14 mm × 1.41 mm package, which has 0.65 mm spacing between the pads.



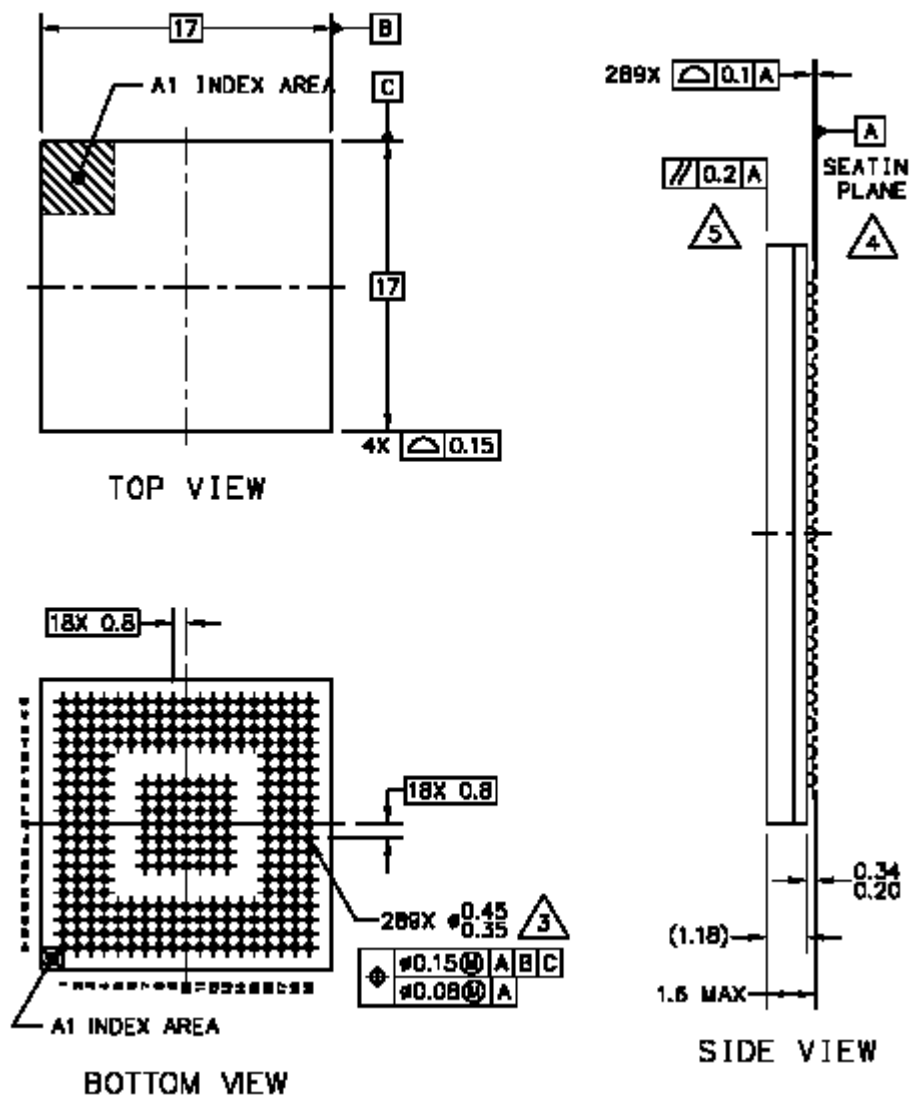
### NOTES:

1. ALL DIMENSIONS IN MILLIMETERS.
2. DIMENSIONING AND TOLERANCING PER ASME Y14.5M-1994.
3. MAXIMUM SOLDER BALL DIAMETER MEASURED PARALLEL TO DATUM A.
4. DATUM A, THE SEATING PLANE, IS DETERMINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.
5. PARALLELISM MEASUREMENT SHALL EXCLUDE ANY EFFECT OF MARK ON TOP SURFACE OF PACKAGE.

Figure 85. i.MX21 MAPBGA Mechanical Drawing

## 4.2 MAPBGA Package Dimensions

Figure 86 illustrates the MAPBGA 17 mm × 17 mm × 1.45 mm package, which has 0.8 mm spacing between the pads.



### NOTES:

1. ALL DIMENSIONS IN MILLIMETERS.

2. DIMENSIONING AND TOLERANCING PER ASME Y14.5M-1994.



3. MAXIMUM SOLDER BALL DIAMETER MEASURED PARALLEL TO DATUM A.



4. DATUM A, THE SEATING PLANE, IS DETERMINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.



5. PARALLELISM MEASUREMENT SHALL EXCLUDE ANY EFFECT OF MARK ON TOP SURFACE OF PACKAGE.

Figure 86. i.MX21 MAPBGA Mechanical Drawing



## 5 Document Revision History

This revision, Rev. 1.1, updates the functional block diagram, Figure 1 on page 2.

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