



# STM7007

## Single-chip hardware accelerated encryption engine for computer and peripherals applications

Data brief

### Features

- World class encryption engine
  - HardCache™ Crypto module
  - FIPS 140-2 security level 3
  - NIST Certificate #1599
- Symmetric algorithms (NIST FIPS approved)
  - AES Rijndael block cipher: Key length 128/192/256 bits, ECB/CBC/CBC-FVE modes, NIST Certificate #1068
  - Triple DES: Key length 112/168 bits, ECB/CBC modes, NIST Certificate #798
  - SHA-256, SHA-384 and SHA-512 with associated HMAC: NIST SHS Certificate #1015, NIST HMAC Certificate #606
- Asymmetric Algorithms
  - ECDSA, 256- and 384-bit elliptic curves
  - RSA PKCS#1 v2.1 padding scheme
  - 2048 bit operands
- Algorithm control policy
  - Easy algorithm configuration
  - Export control
- PCI Express® x1 interface
- Advanced 65 nm process technology
- Windows XP, Windows 2000, Windows Vista, Windows 7<sup>(a)</sup>
- ECOPACK® ROHS compliant

### System benefits

- Off-loads intensive cryptographic calculations (asymmetric and symmetric) from the system CPU
- Out performs SW based solutions and 3 GHz processors
- Full duplex DMA operation



- Best in class performance / power ratio
- Secure key management protected inside a tamper resistant cryptographic boundary
- Resists attacks that use software and/or instrumentation monitors
- Enables secure remote administration
- Shielded locations and protected operations

### Applications

- Desktop PC clients
- Laptop and mobile PC
- Server and workstation
- Entire volume encryption
- Data loss protection
- Cryptographic service provider
- Secure messaging

### Description

The STM7007 is a PCI Express encryption engine built on the STMicroelectronics' HardCache™ cryptographic channel controller (a high performance core computing block, with FIPS 140-2 security level 3 certification). The STM7007 provides hardware acceleration for commonly used algorithms, including AES, 3DES, SHA, HMAC, RSA, and ECC.

a. Windows is a trademark of Microsoft Corporation.

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# 1 Device pinout and package

Device pinout, top view

Device pins

Signal types

Signal groups

STM7007 pins

Device package

3

4

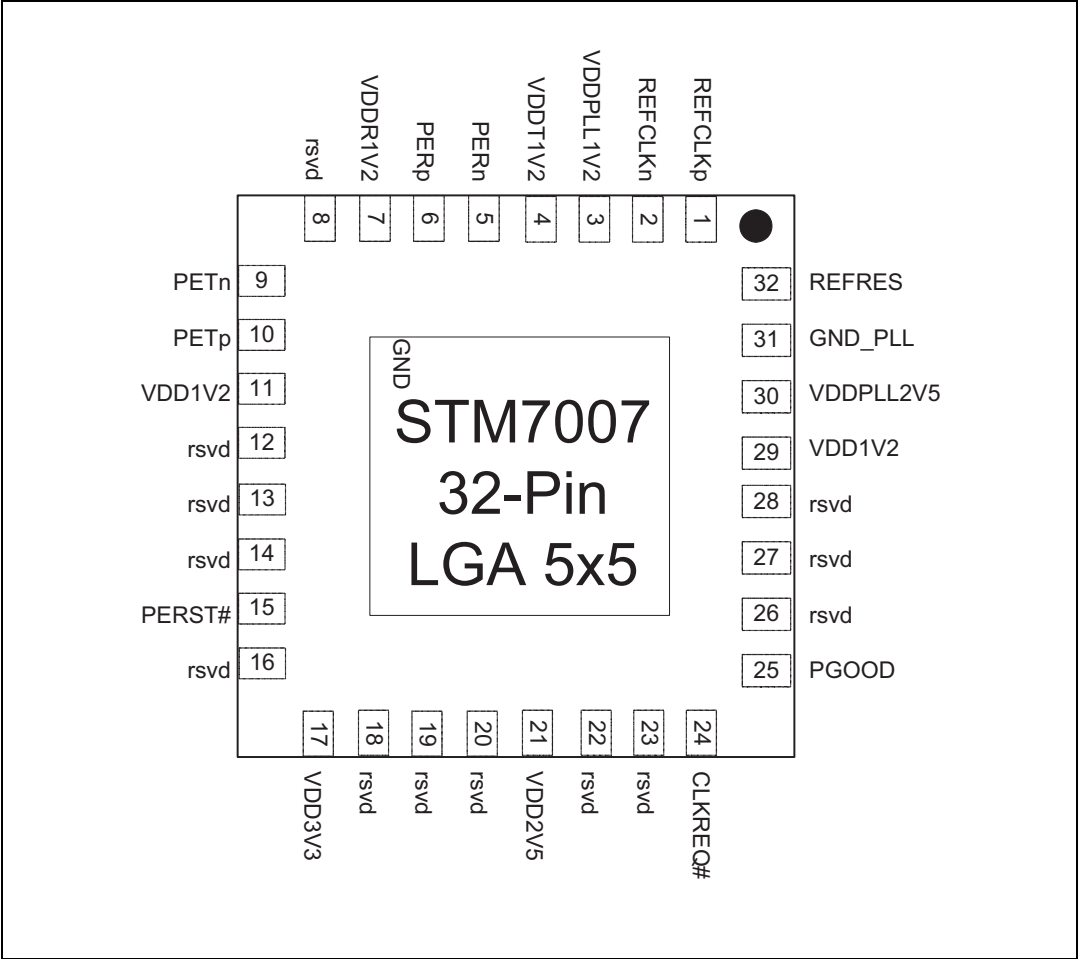
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4

4

7

Figure 1. Device pinout, top view



## 1.1 Device pins

- [Table 1](#) lists the abbreviations used for the signal type names in [Table 3](#).  
Note that abbreviations may be combined, as with APWR (Analog power supply).
- [Table 2](#) lists device signals and how they are grouped, color coding each group.
- [Table 3](#) lists device pins, and uses the same color coding as [Table 2](#) to indicate pin grouping.

**Table 1. Signal types**

Abbreviation	Description
A	Analog
DI	Digital input
DO	Digital output
GND	Ground
I	Input
LVDSI	Low voltage differential input
LVDSO	Low voltage differential output
NC	No connection
O	Output
PWR	Power supply

**Table 2. Signal groups**

Signal	Group	Count (32-pin LGA 5x5 package)
VDD1V2, VDD2V5, VDD3V3, VDD1V2, VDDPLL1V2, VDDPLL2V5, VDDR1V2, VDDT1V2	Power (VDD)	8
VSS_PLL, VSS	Ground (VSS)	2 (GND pad slug on bottom of device)
PETp, PETn, PERp, PERn, REFCLKp, REFCLKn, CLKREQ#, PERST#	PCI Express	8
REFRES, PGOOD	MISC	3
RSVD	Not connected	3

**Table 3. STM7007 pins**

Pin number	Pin name	Signal type <sup>(1)</sup>	External component	Description
1	REFCLKp	LVDSI	—	Low voltage differential clock input
2	REFCLKn			
3	VDDPLL1V2	APWR	Decoupling capacitor	1.2 volt analog power input for the internal PLL block
4	VDDT1V2	APWR	Decoupling capacitor	1.2 volt analog power input for PCIe transmit channel

**Table 3. STM7007 pins (continued)**

Pin number	Pin name	Signal type <sup>(1)</sup>	External component	Description
5	PERn	LVDSI	—	Low voltage differential receive pair
6	PERp			
7	VDDR1V2	APWR	Decoupling capacitor	1.2 volt analog power input for PCIe receive channel
8	RSVD	NC	—	Reserved pins, do not connect
9	PETn	LVDSO	—	Low voltage differential transmit pair
10	PETp			
11	VDD1V2	PWR	Decoupling capacitor	1.2 volt power supply input for core
12	RSVD	NC	—	Reserved
13	RSVD	NC	—	Reserved
14	RSVD	NC	—	Reserved
15	PERST#	DI	—	Asynchronous device reset input from system. When asserted, STM7007 resets all volatile content and executes a reboot after de-assertion of signal.
16	RSVD	NC	—	Reserved
17	VDD3V3	PWR	Decoupling capacitor	3.3 volt power supply input for I/O logic
18	RSVD	NC	—	Reserved
19	RSVD	NC	—	Reserved
20	RSVD	NC	—	Reserved
21	VDD2V5	PWR	Decoupling capacitor	2.5 volt power supply
22	RSVD	NC	—	Reserved pins, do not connect
23				
24	CLKREQ#	DO	—	Informs the host controller that the reference clock input is required by the STM7007. It is an open drain active low output; when 0, the reference clock is required, otherwise the reference clock can be off.
25	PGOOD	DI	—	Power on reset, used to reset the core.
26	RSVD	NC	—	Reserved
27	RSVD	NC	—	Reserved
28	RSVD	NC	—	Reserved
29	VDD1V2	PWR	Decoupling capacitor	1.2 volt power supply input for core
30	VDDPLL2V5	PWR	—	
31	GND_PLL	GND	—	DC return pin for PLL block

**Table 3. STM7007 pins (continued)**

Pin number	Pin name	Signal type <sup>(1)</sup>	External component	Description
32	REFRES	AI	Resistor	Reference resistor input for PLL block. This pin is tied to 1.2V PLL using a 475_1% ohm resistor
—	GND	GND	-	DC return pad

1. For acronym definitions, see [Table 1: Signal types on page 4](#).

1.2 Device package

*Dimensions* ..... 8

*Recommended footprint* ..... 9

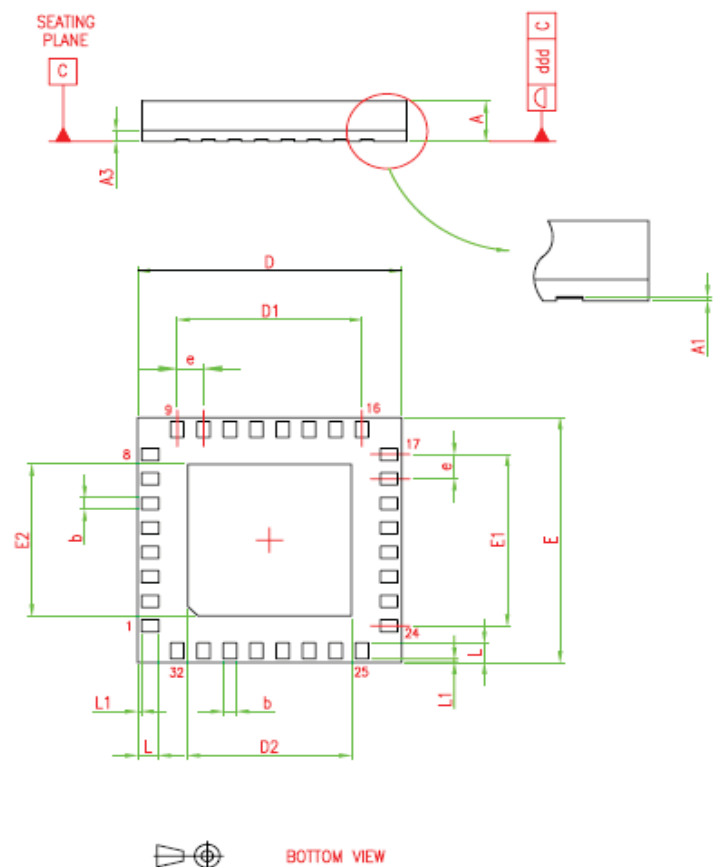
*Markings* ..... 10

This device uses a 32-Lead, 5 x 5 x 0.82 mm land grid array (LGA) package.

- Note:*
- 1 *This package is not yet defined in JEDEC publications.*
  - 2 *The exact shape of each corner is optional.*

## 1.2.1 Dimensions

Table 4. Package dimensions (mm)



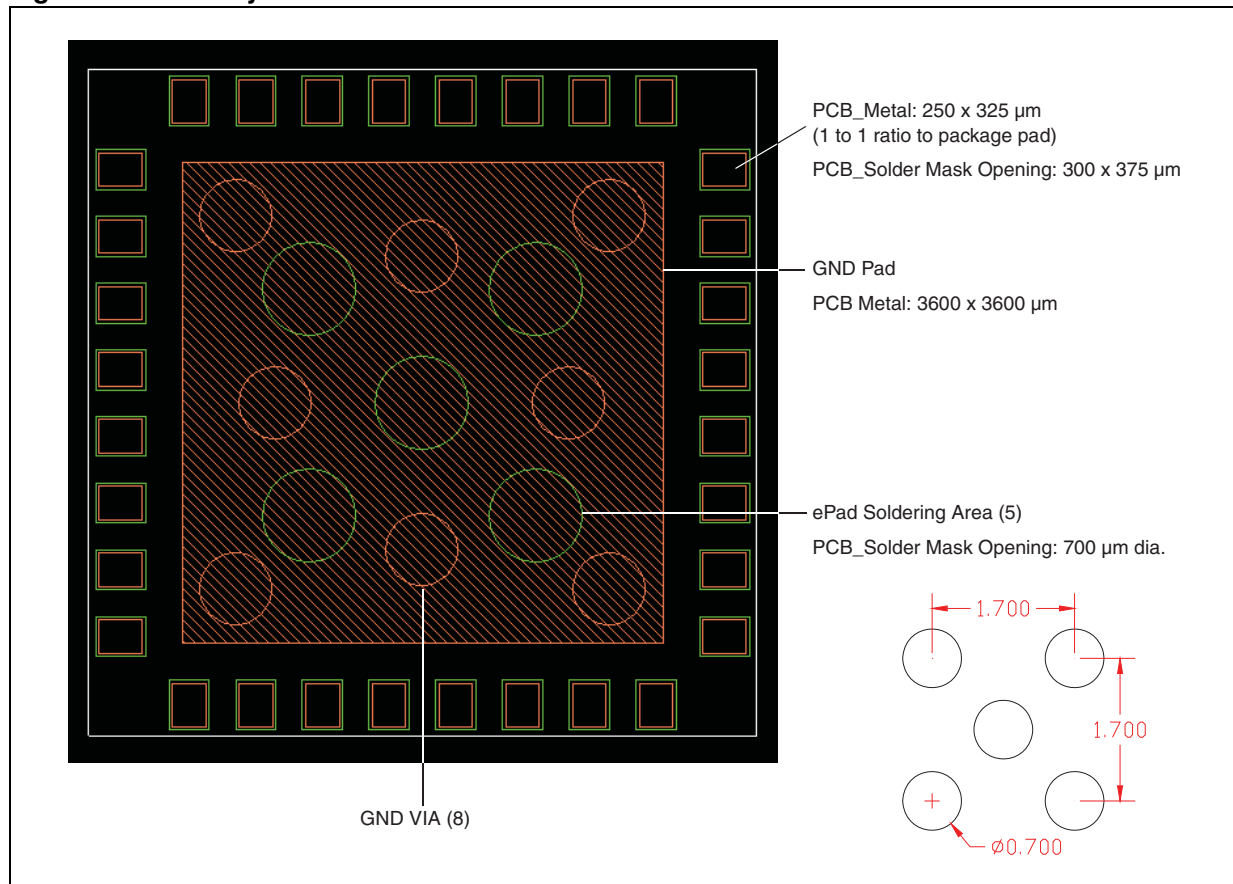
The diagram illustrates the mechanical dimensions of the STM7007 package. The top view shows a square package with a central square area. Dimensions include overall width (D), overall height (E), and various internal offsets (D1, D2, E1, E2, e, b, L, L1). A callout shows a detail of the seating plane with dimensions A, A1, and A3. The bottom view shows the package from below, with dimensions L and L1. The dimensions are defined as follows:

- A: Seating plane offset
- A1: Package thickness
- A3: Seating plane offset
- b: Pin pitch
- D: Overall width
- D1: Internal width offset
- D2: Internal width offset
- E: Overall height
- E1: Internal height offset
- E2: Internal height offset
- e: Pin pitch
- L: Pin length
- L1: Pin length
- ddd: Pin length

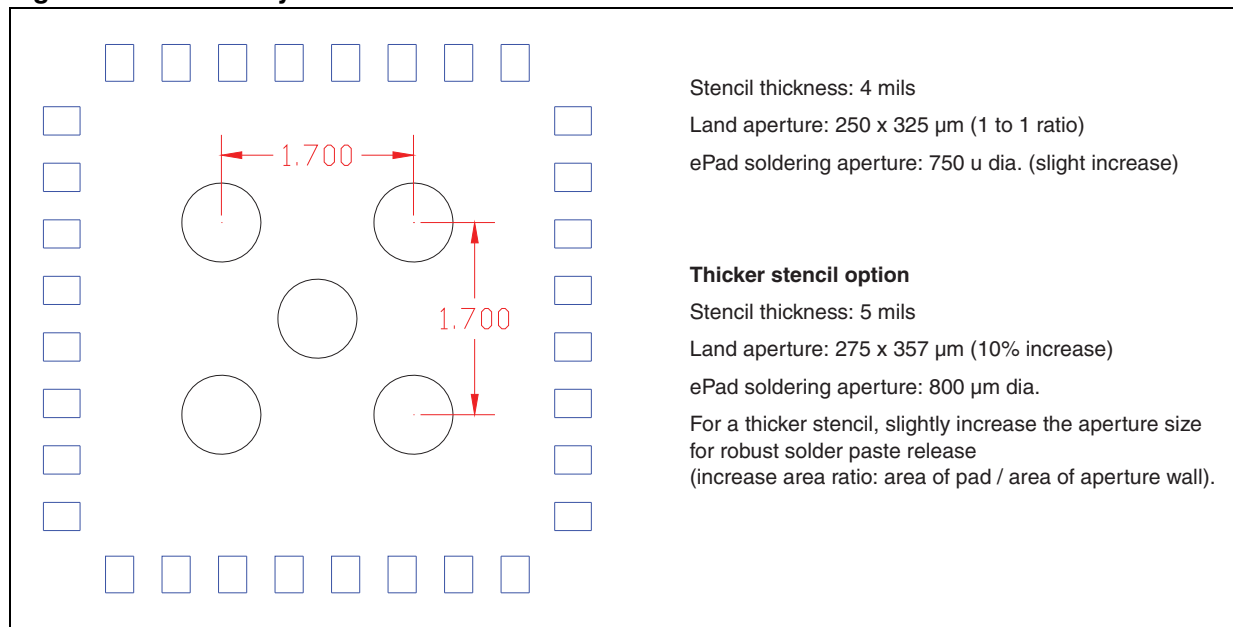
Reference	Minimum	Typical	Maximum
A	0.74	0.82	0.88
A1	0.00		0.035
A3		0.22	
b	0.18	0.25	0.30
D	4.90	5.00	5.10
D1		3.50	
D2	2.65	2.70	2.75
E	4.90	5.00	5.10
E1		3.50	
E2	2.65	2.70	2.75
e		0.50	
L	0.35		0.45
L1		0.075	
ddd			0.100

## 1.2.2 Recommended footprint

**Figure 2. PCB layout recommendation**



**Figure 3. Stencil layout recommendation**



### 1.2.3 Markings

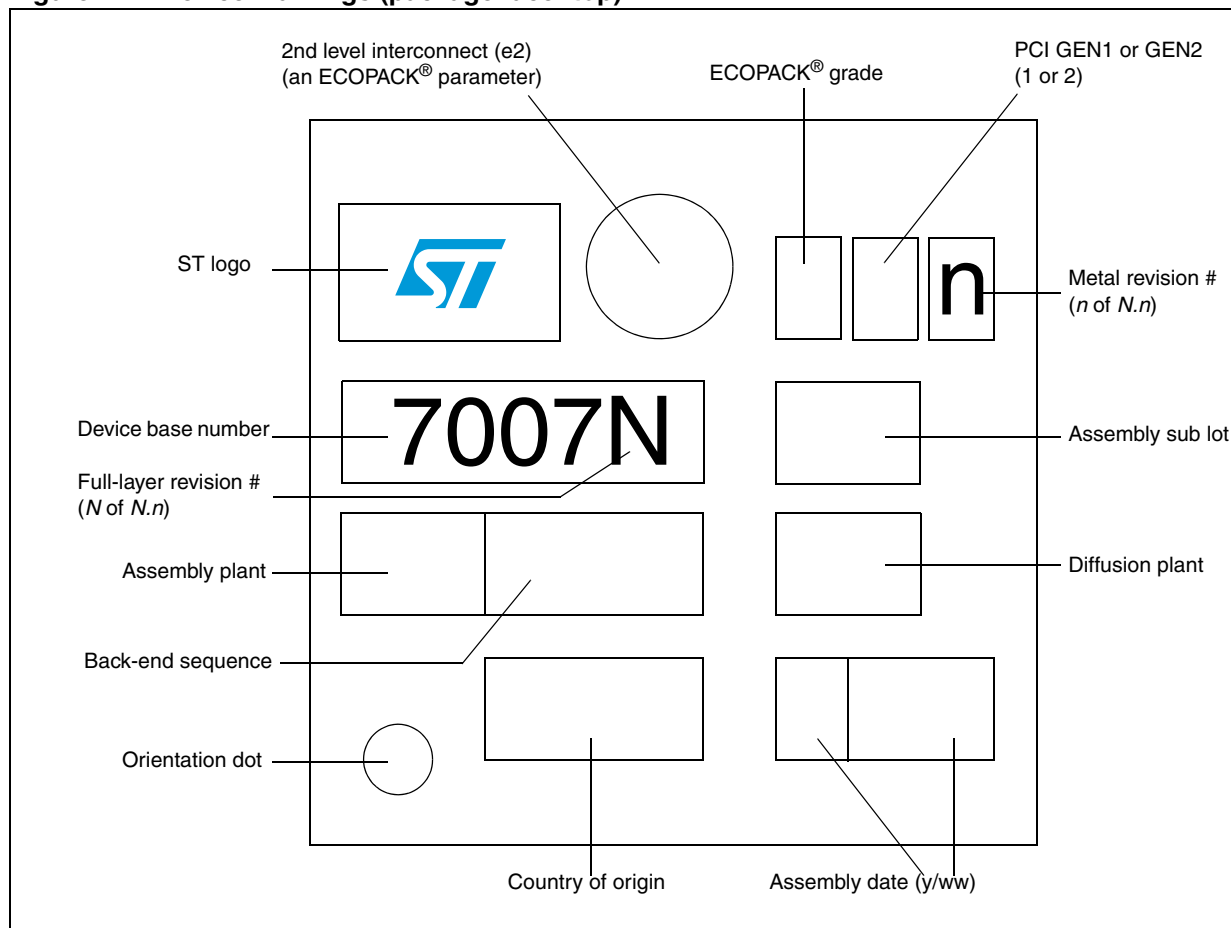
Device base number: 7007 (four digits)

Device revision number: *N.n*

*N* = Full-layer revision number (one digit)

*n* = Metal revision number (one digit)

**Figure 4. Device markings (package face: top)**



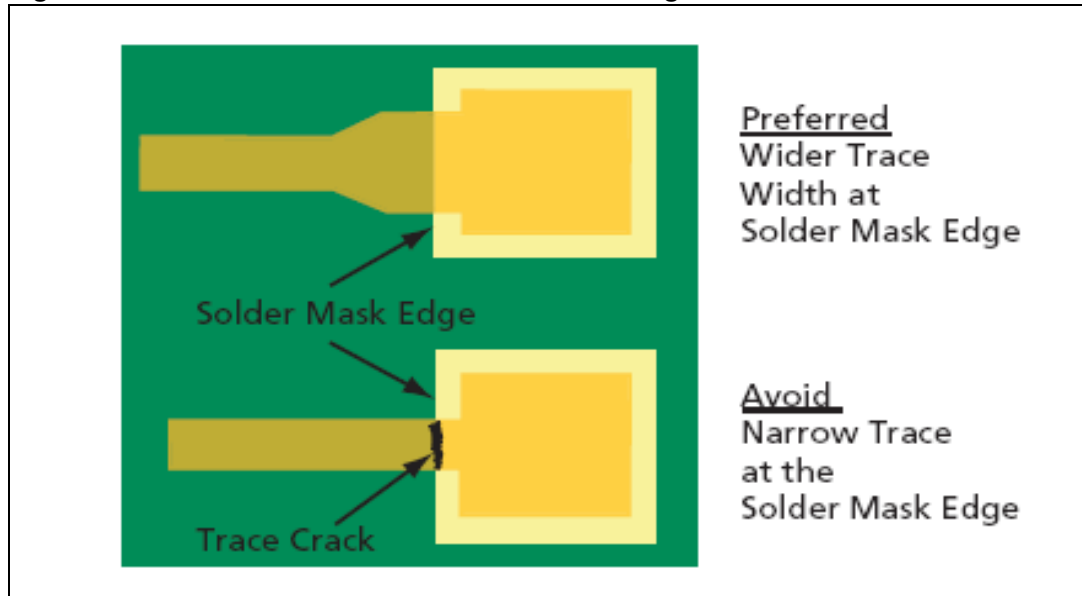
In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: [www.st.com](http://www.st.com). ECOPACK® is an ST trademark.

### 1.2.4 Recommended land pad layout and solder reflow profile

Pay special attention to the traces connecting to the metal pads on the board. Trace cracking can occur during normal handling of the board. This trace cracking usually occurs at the edge of the solder mask opening, around the metal pad.

To avoid this mode of failure, make the trace under the solder mask edge wider than the rest of the trace, as shown in [Figure 5](#). Depending on the reliability requirements of the connection, the wider part of the trace might need to be as wide as 50 to 75% of the metal pad width.

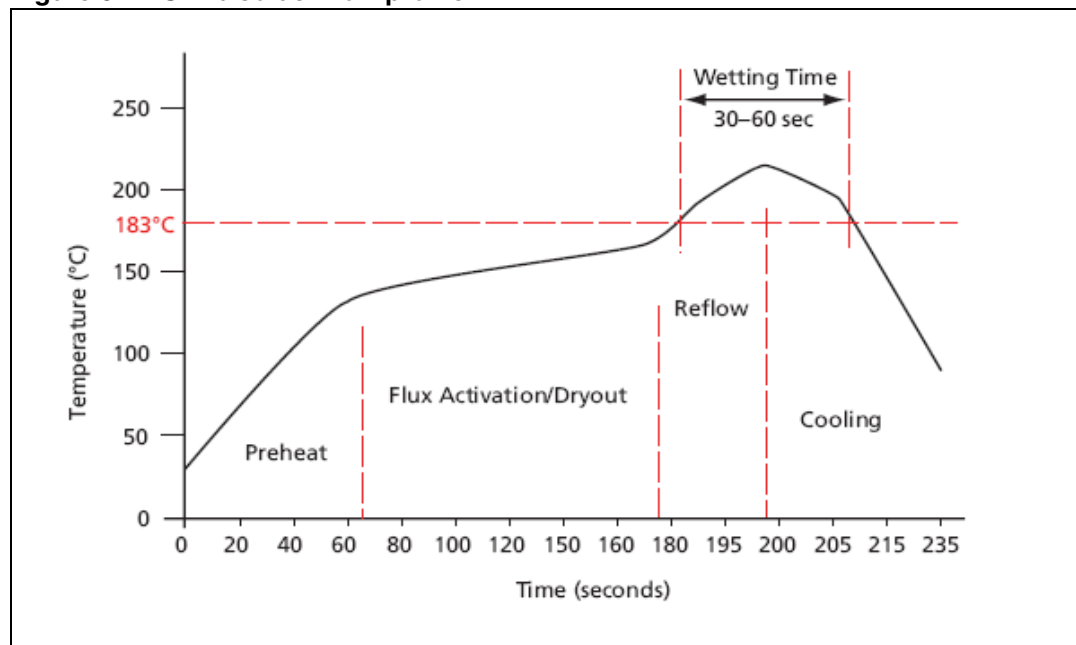
**Figure 5. Wider trace connection to avoid cracking**



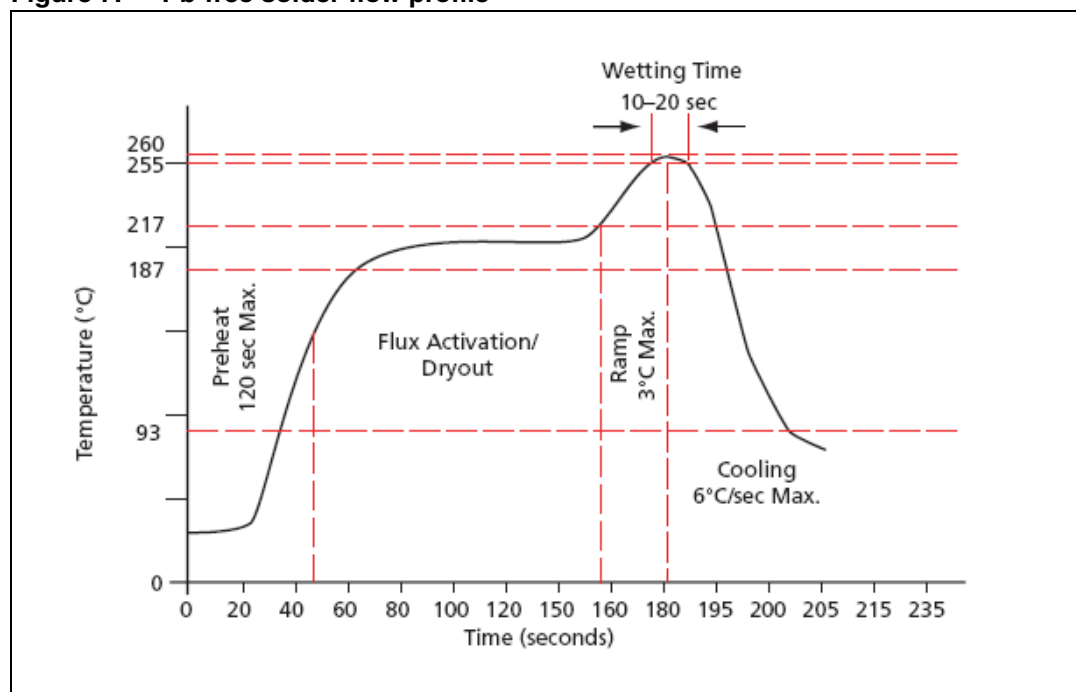
## Solder flow profile

Reflow profile and peak temperature have a strong influence on void formation. Follow the profile recommendation of the paste suppliers, because this is specific to the requirements of flux formation. The following two profiles can serve as reference for fine tuning the final profile that works for your application.

**Figure 6. SnPb solder flow profile**



**Figure 7. Pb-free solder flow profile**



### 1.2.5 Rework guidelines

Because solder joints are not fully exposed with QFN packages, any touch-up is limited. For defects underneath the package, the whole package must be removed. Because reflow of adjacent parts is not desirable during rework, the proximity may complicate the rework process. Because of the product-dependent complexities, the following provides only a guideline and a starting point for the development of a successful rework process for this package.

The rework process comprises the following steps:

#### 1. Component removal

The first step in removal of the component is the reflow of solder joints attaching the component to the PCB board. Ideally the reflow profile for part removal and attachment should be the same, but this is not always possible.

- Heat the board from the top side of the component.
- Use a special nozzle to direct the heating in the component area.
- Minimize the heating of adjacent components.
- Start with an air velocity of 15 to 20 liters per minute.
- Avoid excess heating and pad liftoff.

#### 2. Site redress

After removing the component, clean the site properly.

- Use de-soldering braid to remove excess solder.
- Once the residual solder is removed, clean the lands with solvent. The solvent is usually specific to the type of paste used in the original assembly.

#### 3. Solder paste printing

To achieve a uniform and precise deposition, use a miniature stencil specific to the component.

Avoid tinning the lands, because the amount of solder applied to the lands cannot be controlled.

#### 4. Component placement and attachment

Although this type of package does display self-centering abilities, use care in its placement.

Ideally, the reflow profile for part removal and attachment should be the same, but this is not always possible.

Heat the board from the top side of the component.

## 2 Electrical characteristics

*Note:* The values in this section are preliminary, and subject to change.

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<i>PHY 1.2V and 2.5V interfaces</i> .....	16
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### 2.1 Recommended operating conditions

*Note:* Operation beyond recommended conditions is neither recommended nor guaranteed.

**Table 5. Recommended operating conditions**

Parameter	Minimum	Typical	Maximum	Units	Notes
VDD1V2	1.14	1.2	1.26	V	Core voltage
VDD2V5	2.375	2.5	2.625	V	Antifuse charge-pump
VDD3V3	3.0	3.3	3.6	V	I/O voltage for PCIe interface
VDDT1V2	1.14	1.2	1.26	V	Transmit channel VDD
VDDR1V2	1.14	1.2	1.26	V	Receive channel VDD
VDDPLL1V2	1.14	1.2	1.26	V	PLL Core
VDDPLL2V5	2.375	2.5	2.625	V	PLL
Ta	0	-	70	C	Ambient temperature

## 2.2 Absolute maximum ratings

**Caution:** Stresses beyond those listed in [Table 6](#) may cause permanent damage to the device; because these are *stress* ratings, functional operation is not implied at these or any other conditions beyond those indicated in [Table 5 on page 14](#). Exposure to the conditions listed in [Table 6](#) for extended periods may affect device reliability.

**Table 6. Absolute maximum ratings<sup>(1)</sup>**

Parameter	Minimum(V)	Maximum (V)	Notes
VDD1V2	-0.5	1.5	Core voltage
VDD2V5	-0.5	3.0	Antifuse charge-pump
VDD3V3	-0.5	4.0	PCI IO Ring
VDDT1V2	-0.5	1.5	Transmit Channel VDD
VDDR1V2	-0.5	1.5	Receive channel VDD
VDDPLL1V2	-0.5	1.5	PLL core
VDDPLL2V5	-0.5	3.0	PLL

1. These are stress ratings, not functional operation ratings.

## 2.3 Environment maximum ratings

**Table 7. Environment maximum ratings**

Parameter	Minimum	Maximum	Units	Notes
ESD HBM	—	2000	V	Human body model
ESD CDM	—	500	V	Charged device model
Tstorage	-40	150	C	Storage temperature

## 2.4 Power dissipation

**Table 8. Power dissipation typical mode**

Symbol/mode	Current (mA) for Power rails (V)			Total (mW)	Notes
	3.3	2.5	1.2		
L0 - AF programming	2	48	216	385.8	
L0 - bulk	2	17	222	315.5	
L0 - average	2	17	173.4	257.18	Average power, assuming 10% duty cycle (0.1 * L0-Bulk + 0.9 * L0-Idle)
L0 - idle	2	17	168	250.7	
L0s	2	17	139	215.9	
L1 - CLKREQ# asserted	2	17	126	200.3	Clock power management disabled
L1 - CLKREQ# deasserted	1	3	20	34.8	Clock power management enabled

## 2.5 DC electrical specifications

### 2.5.1 PCI, miscellaneous 3.3V interfaces

Note: 1 While I/O is in high impedance state  
2 Does not include current flowing through termination resistors.

**Table 9. 3.3V interface pin DC specifications**

Parameter	Symbol	Test condition	Minimum	Typical	Maximum	Units	Notes
Input low level	VIL		-0.3	—	0.8	V	
Input high level	VIH		2.0	—	0.3 + VDD3V3	V	
Output low level	VOL	8 mA	—	—	0.4	V	
Output high level	VOH	8 mA	VDD3V3 - 0.4	—	—	V	
Input leakage current	IIL	0 < VIN < VDD3V3	< 1 (25° C)	—	2 (125° C)	μA	1;2
Pin capacitance	CPin			30	—	pF	

### 2.5.2 PHY 1.2V and 2.5V interfaces

**Table 10. 1.2 and 2.5 PHY interface pin dc specifications**

Parameter	Symbol	Minimum	Typical	Maximum	Units
1.2V supply voltage range	VDD1V2	1.14	1.2	1.26	V
Supply ripple (1 MHz to 3 GHz)	VDD1V2ripple	—	—	50	mV (pk-pk)
2.5V supply voltage range	VDD2V5	2.375	2.5	2.625	V
Supply ripple (1 GHz to 3 GHz)	VDD2V5ripple	—	—	50	mV (pk-pk)
Maximum reference clock input jitter. RJ component bounded at +/- 7 sigma	TJrefclk	—	—	30	ps (pk-pk)

## 2.6 AC electrical specification

**Table 11. PCI Express clock input**

Parameter	Description	Minimum	Typical	Maximum	Units
Fref	PCIe pin RefClkp, RefClkn The clock must be compliant with the LVDS input driver level.	99.97	100	100.03	MHz

## 2.7 Differential interface electrical characteristics

**Table 12. PCIe Interface driver and receiver characteristics**

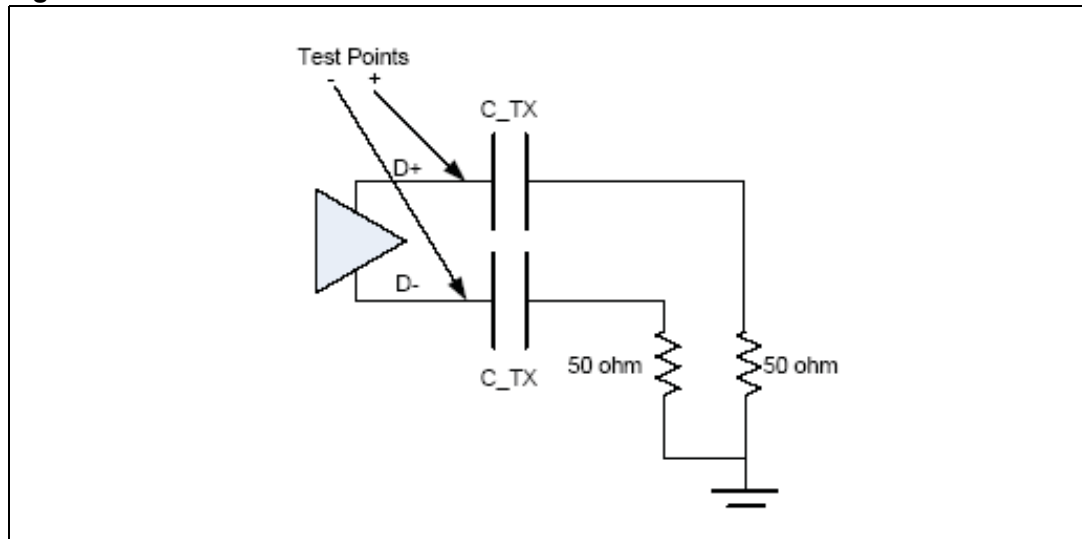
Description	Symbol	Minimum	Maximum	Units
Baud rate	BR	2.5		Gbps
Unit interval	UI	399.88	400.12	ps
Baud rate tolerance	BRtol	-300	300	ppm
<b>Driver parameters</b>				
Differential peak to peak output voltage	VTXpp	800	1200	mV
Minimum Tx eye width	TTxeye	0.750		UI
Differential return loss	TRLdiff	10		dB
Common mode return loss	TRLcomm	6		dB
DC differential TX impedance	ZTXdiff	85	115	$\Omega$
<b>Receiver parameters</b>				
Differential peak to peak voltage	VRXpp	0.175	1.2	V
Minimum Rx eye width	TRxeye	0.4		UI
Differential return loss	TRLdiff	10		dB
Common mode return loss	RRLcomm	6		dB
DC differential RX impedance	ZRXdiff	85	115	$\Omega$
DC single-ended input impedance	ZRXDC	40	60	$\Omega$

### 2.7.1 PCIe test circuit

When measuring transmitter output parameters, C\_TX is an optional portion of the test/measurement load.

When used, the value of C\_TX must be in the range of 75 to 200 nF. C\_TX must *not* be used when the test/measurement load is placed in the receiver package reference plane.

**Figure 8. PCIe test circuit**



### 3 PCB layout guidelines

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<i>PCB power and ground decoupling guidelines</i> .....	22
<i>Power sequencing</i> .....	23

The STM7007 uses high speed, low voltage differential signal pairs for the PCIe interface. For a successful board design, follow the placement and layout guidelines in this chapter for critical signals. Additional application notes may be provided for more detailed requirements.

The STM7007 is designed for optimized wiring for both a motherboard or PCI Express mini-card form factor. Pin locations of the critical high speed LVDS signals are designed to support a seamless flow from the STM7007 to the upstream host controller.

See also, *Figure 1: Device pinout, top view*.

### 3.1 PCB Routing of RX/TX differential signals

Ensure that the board design meets the following criteria:

- To avoid skew, make signal pair lengths equal from the ball to the connector:  
TXP length = TXN length, and RXP length = RXN length  
Lengths may differ if required to compensate for length mismatch due to poor package substrate routing.
- Route signal pairs in parallel
- Ensure a 100Ω differential impedance trace:
  - Route signals on an external layer with a full or partial ground plane layer as the next internal adjacent layer. A partial ground plane must cover and exceed the area below the TX and RX signal routing.
  - Between the package pin and the PCI-Express connectors, ensure equal signal trace interspace distance on each pair along the entire trace length.
- To avoid crosstalk:
  - Separate TX and RX pairs a minimum of 5 mm, with no other signals interleaved.
  - If several ports are routed, do not route the TX and RX traces of each port one above the other on different layers.
- Do *not* use:
  - Right angles or 45-degree trace angles; curved traces are preferred
  - PCB vias
- Make all traces:
  - As direct and straight as possible
  - As short as possible. Signal line attenuation may be compensated for by selecting a higher buffer swing, but if the largest swing is specified for the application, there is no capacity for swing compensation.
- If components (such as AC coupling capacitors) are required on the high speed signal traces, choose the smallest SMD packages and place them close to connectors.
- PCB trace crossing: if two differential traces from different ports *must* cross, cross the traces at 90° (perpendicular) to each other.

### 3.1.1 Example: PCB impedance calculation

The PCB manufacturer should provide the board designer with all relevant parameters for calculating trace impedance:

- trace thickness
- trace width
- PCB material, giving  $E_r$
- distance between layer 1 and internal layer 2
- coating thickness

#### Example calculation

Differential pair routed on top layer (1)

Trace width: 7 mils

Inter-space trace distance: 7 mils

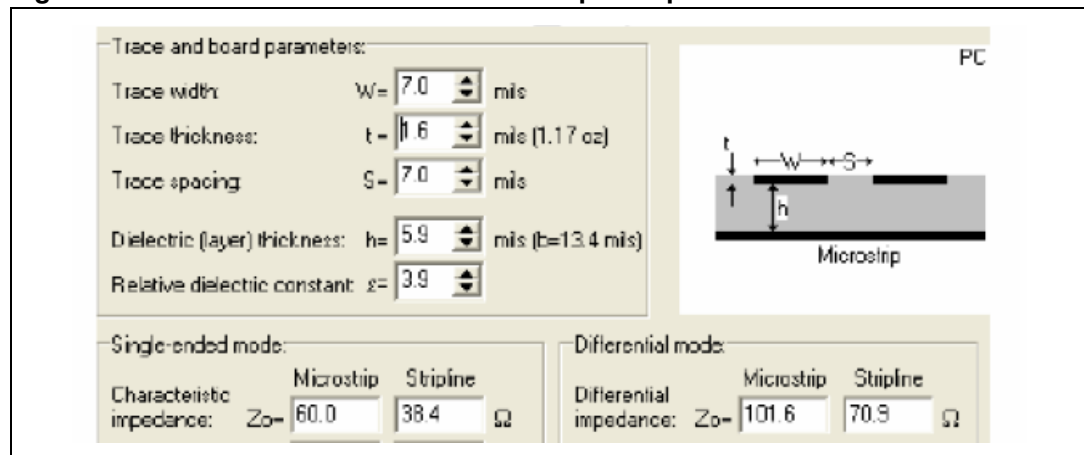
Ground plane on internal layer 2, giving 5.9 mils dielectric height ( $E_r = 3.9$ )

Trace thickness:  $40\ \mu\text{m} = 1.6\ \text{mils}$ .

The calculator shown in [Figure 9](#) gives a theoretical  $Z_{\text{diff}} = 101.6\ \Omega$

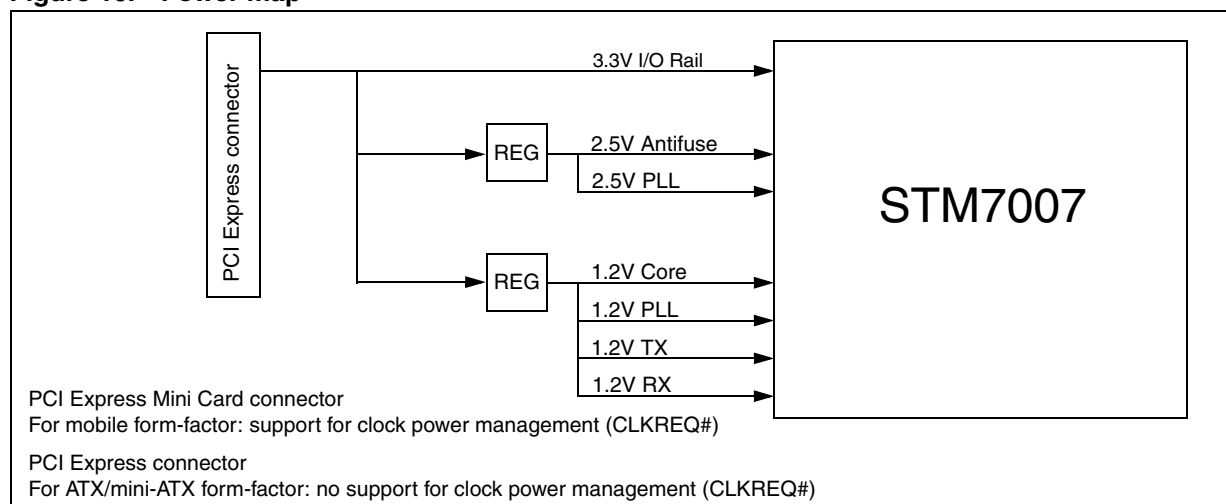
The PCB manufacturer double-checked with its tools and a more complex calculator, taking into account the cross-section form of the top layer trace and the coating used. The PCB manufacturer's calculation gives a theoretical  $Z_{\text{diff}} = 100.2\ \Omega$

**Figure 9. UltraZ calculation for differential pair impedance**



## 3.2 PCB power and ground decoupling guidelines

Figure 10. Power map



**All grounds:** Connect directly to the PCB common ground plane

**Power pins VDDR1V2 and VDDT1V2:**

- Directly connect to the board's 1.2V common supply dedicated to the PHY
- Decouple locally using:
  - Murata Ferrite BLM15AG121 or equivalent
  - 1.0  $\mu$ F ceramic capacitor
  - 0.1  $\mu$ F RF capacitor (low series access resistor)
  - 0.01  $\mu$ F RF capacitor (low series access resistor)

**Power pin VDDPLL1V2:**

- Directly connect to the board's 1.2V common supply dedicated to the PHY
- Filter locally using:
  - 1 $\Omega$  resistor
  - 10  $\mu$ F ceramic capacitor
  - 0.1  $\mu$ F RF capacitor (low series access resistor)
  - 0.01  $\mu$ F RF capacitor (low series access resistor)

**Power pin VDD2V5:**

- Directly connect to the board's 2.5V common supply dedicated to the PHY
- Filter locally using:
  - 1 $\Omega$  resistor
  - 1.0  $\mu$ F ceramic capacitor
  - 100 pF RF capacitor (low series access resistor)
  - 10 nF RF capacitor (low series access resistor)

**Power pin VDDPLL2V5:**

- Directly connect to the board's 2.5V common supply dedicated to the PHY
- Filter locally using:
  - 30Ω resistor ( $\pm 5\%$  maximum tolerance)
  - 0.1  $\mu\text{F}$  RF capacitor (low series access resistor)
  - 0.01  $\mu\text{F}$  RF capacitor (low series access resistor)

### 3.3 Power sequencing

The STM7007 requires proper supply voltage and PGOOD (power good) sequencing for the internal power-on-self-test and the antifuse blocks. [Figure 11](#) shows the timing and level requirements.

The primary requirement for this device is PGOOD timing; PGOOD must provide a valid status of the supply voltages during power sequencing.

The PGOOD signal:

- remains inactive until all voltages are within their required tolerance bands during power up
- goes inactive before the supply voltages exit their required tolerance bands during power down

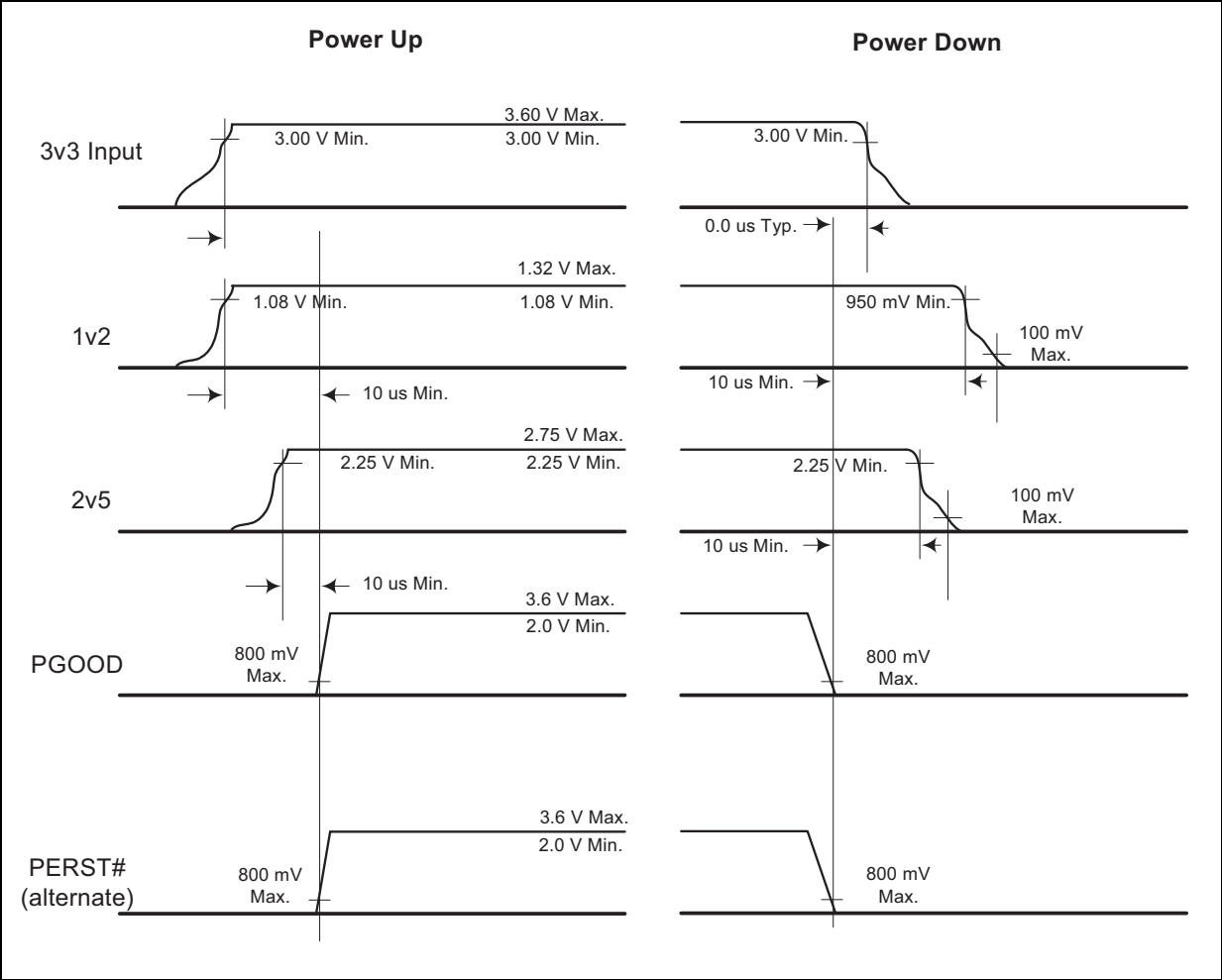
The first indication that power-down sequencing is occurring, is the dropping of the 3v3 input supply voltage.

The PGOOD input pin and the PERST# input pin are logically combined internally to ensure the proper reset state of the device during sequencing.

The 2v5 and 1v2 regulators must be sourced by the 3v3 supply such that the recommended order of supply sequencing at power up is 1v2 first and then 2v5, and at power down 2v5 drops first, and then 1v2.

Because of an internal ESD protection diode, it is expected that when the 1v2 is applied to the VDDPLL1V2 pin, the VDDPLL2V5 pin will source voltage at a level equal to the 1v2 voltage minus the forward voltage drop of a diode. To limit the diode current, and to provide lowpass filtering of the PLL 2V5 power source, place a 30Ω, 5% resistor between the VDDPLL2V5 pin and the 2V5 source. Position the bypass capacitors close to the VDDPLL2V5 pin (pad).

Figure 11. Power up / down timing requirements



## 4 Ordering information

When ordering parts, use the information below and contact your local STMicroelectronics sales office, field applications engineer, or manufacturer's representative for further information.

**Table 13. STM7007 ordering information**

Order code	Package type	Note
STM7007	32-Pin LGA 5x5	Commercial RoHS and halogen free package

## 5 Revision history

**Table 14. Document revision history**

Date	Revision	Changes
06-Oct-2011	1	Initial release.

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