

© L6377 0.5A HIGH-SIDE DRIVER INTELLIGENT POWER SWITCH

PRODUCT PREVIEW

- 0.5 A OUTPUT CURRENT
- 8 TO 35 V SUPPLY VOLTAGE RANGE
- EXTERNALLY PROGRAMMABLE CURRENT LIMIT
- NON-DISSIPATIVE OVER-CURRENT PRO-TECTION
- THERMAL SHUTDOWN
- UNDER VOLTAGE LOCKOUT WITH HYS-TERESYS
- DIAGNOSTIC OUTPUT FOR UNDER VOLT-AGE, OVER TEMPERATURE AND OVER CURRENT
- EXTERNAL ASYNCHRONOUS RESET INPUT
- PRESETTABLE DELAY FOR OVERCUR-RENT DIAGNOSTIC
- OPEN GROUND PROTECTION
- PROTECTION AGAINST SURGE TRAN-SIENT (IEC 801-5)
- IMMUNITY AGAINST BURST TRANSIENT (IEC 801-4)
- ESD PROTECTION (HUMAN BODY MODEL ±2KV)



DESCRIPTION

This device is a monolithic Intelligent Power Switch in Multipower BCD Technology for driving inductive, capacitive or resistive loads.

Diagnostic for CPU feedback and extensive use of electrical protections make this device inherently indistructible and suitable for general purpose industrial applications.



BLOCK DIAGRAM

This is advanced information on a new product now in development or undergoing evaluation. Details are subject to change without notice.

Δ	BS	OL	UTF	MΔ	XIMI	JM	RAT	INGS
						J 111	1171	1100

Symbol	Pin	Parameter	Value	Unit
V	2	Supply Voltage ($t_w \leq 10ms$)	50	V
VS	5	Supply Voltage (DC)	40	V
Vs - Vout	3 vs 2	Supply to Output Differential voltage	internally limited	
V _{od}	5	Externally Forced Voltage	-0.3 to 7	V
lod	5	Externally Forced Current	±1	mA
IRESET	4	Reset Input Current (forced)	±2	mA
Vreset		Reset Input Voltage	-0.3 to 40	V
lout		Output Current (see also I _{sc})	internally limited	
Vout	2	Output Voltage	internally limited	
E _{il}		Total Energy Inductive Load (T _j = 125°C)	50	mJ
P _{tot}		Power Dissipation	internally limited	
V _{diag}	6	External voltage	-0.3 to 40	V
I _{diag}		Externally forced current	-10 to 10	mA
li	7	Input Current	20	mA
Vi		Input Voltage	-10 to V _s +0.3	V
Top		Ambient temperature, operating range	-25 to 85	°C
Tj		Junction temperature, operating range (see Overtemperature Protection)	-25 to 125	°C
T _{stg}		Storage temperature	-55 to 150	°C

PIN CONNECTION (Top view)



PIN DESCRIPTION

No	Pins	Description
1	GND	Ground pin.
2	OUT	High side output. Controlled output with current limitation.
3	Vs	Supply voltage. Range with under voltage monitoring
4	Rsc	Current limiting setting.
5	ON DELAY	Delay setting for overcurrent diagnostic
6	DIAG	Diagnostic open drain output for over temperature, under voltage and overcurrent
7	IN+	Comparator non inverting input
8	RESET	Asynchronous reset input

THERMAL DATA

Symbol	Parameter		MINIDIP	SO14	Unit
Rth j-amb	Thermal Resistance, Junction Ambient Ma	ax.	100	150	°C/W



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ELECTRICAL CHARACTERISTICS (Vs = 24V; Tj = -25 to 125°C; unless otherwise specified.)

DC OPERATION

Symbol	Pin	Parameter	Test Condition	Min.	Тур.	Max.	Unit
Vsmin		Supply Voltage for Valid Diagnostic	$I_{diag} = >0.5mA;$ $V_{diag} = 1.5V;$	4		35	V
Vs		Operative Supply Voltage		8	24	35	V
Vsth	3	Under Voltage Lower Threshold		7		8	V
V _{shys}	-	Under Voltage Hysteresis		300	500	700	mV
lq		Quiescent Current	Output Open		800		μA
Iqo		Quiescent Current	Output On		1.6		mA
Vith		Input Threshold Voltage		0.8	1.3	2	V
V _{iths}	7	Input Threshold Hysteresis		50		400	mV
V _{il}		Input Low Level Voltage		-7		0.8	V
Vih		Input High Level Voltage	Vs< 18V	2		V _s -3	V
			Vs> 18V	2		15	V
l _{ib}		Input Bias Current	Vi = -7 to 15V	-250		250	μΑ
Vrth		Reset Threshold Voltage		0.8	1.3	2	V
V _{rl}	8	Reset Low Level Voltage		0		0.8	V
V _{rh}		Reset High Level Voltage		2		40	V
I _{rb}		Reset Pull Down Current			5		μA
Idch	5	Delay Capacitor Charging Current	ON DELAY pin shorted to Ground		2.5		μA
V _{rsc}	Λ	Output Voltage on R _{sc} pin	Rsc pin floating		1.25		V
I _{rsc}	7	Output Current on R _{sc} pin	Rsc pin shorted to GND			300	μA
Idlkg	6	Diagnostic Output Leakage Curr.	Diagnostic Off			25	μA
Vdiag	0	Diagnostic Output Voltage Drop	Idiag =5mA;			1.5	V
V _{don}		Output Voltage Drop	lout =625mA; T _j =25°C		250	350	mV
			I _{out} =625mA; T _j =125°C		400	550	mV
I _{olk}		Output Leakage Current	V _i = LOW; V _{out} =0			100	μA
Vol	2	Output Low State Voltage	V _i = HIGH; pin floating		0.8	1.5	V
V _{cl}	2	Internal Voltage Clamp (Vs-Vout)	l₀=200mA single pulsed =300µs	48	53	58	V
60		Short Circuit Output Current	$V_{s}=8 \text{ to } 35\text{V}; \text{R}_{\text{I}}=2\Omega; \\ \text{R}_{\text{sc}}=5 \text{ to } 30\text{K}\Omega$	Ę	$5/R_{sc} = K\Omega$		А
100				0.75	1.1	1.5	А
T _{max}		Over Temperature Upper Threshold			150		°C
T _{hys}		Over Temperature Hysteresis			20		°C



AC OPERATION

Symbol	I Pin Parameter Test Condition		Min.	Тур.	Max.	Unit	
tr -tr		Rise or Fall Time	Vs = 24V; RI = 70Ω		20		μs
ta	2	Delay Time	Ri to ground		5		μs
dV/dt		Slew Rate (Rise and Fall Edge)	$V_s = 24V; R_l = 70\Omega$ R _l to ground	0.7	1	1.5	V/µs
t _{ON}	5	On time during Short Circuit Condition	50pF < C _{DON} < 2nF		1.28		μs/pF
toff		Off time during Short Circuit Condition			64		t _{ON}
f _{max}		Maximum Operating Frequency			25		kHz

SOURCE DRAIN NDMOS DIODE

Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Unit
V _{fsd}	Forward On Voltage	I _{fsd} = 625mA		1	1.5	V
I _{fp}	Forward Peak Current	t _p = 10ms; duty cycle = 20%			1.5	А
t _{rr}	Reverse Recovery Time	I_{fsd} = 500mA; dI_{fsd}/dt = 25A/µs		200		ns
t _{fr}	Forward Recovery Time			50		ns

Figure 1: Undervoltage Comparator Hysteresis









INPUT SECTION

An Input and Asynchronous RESET, both TTL/CMOS compatible with wide voltage range and high noise immunity (thanks to a built in hysteresis) are available.

OVER TEMPERATURE PROTECTION (OVT)

An on-chip Over Temperature Protection providse an excellent protection of the device in extreme conditions. Whenever the temperature - measured on a central portion of the chip- exceeds Tmax=150 C (typical value) the device is shut off, and the DIAG output goes LOW.

Normal operation is resumed as the chip temperature (normally after few seconds) falls below Tmax-Thys= 130 C (typical value). The hysteresis avoid thats an intermittent behaviour take place.

UNDER VOLTAGE PROTECTION (UV)

The supply voltage is expected to range from 8 to 35 V. In this range the device operates correctly. Below 8 V the overall system has to be considered not reliable. To avoid any misfunctioning the supply voltage is continuously monitored to provide an under voltage protection. As Vs falls be-low Vsth-Vshys (typically 7.5 V, see fig.1) the output power MOS is switched off and DIAG output goes LOW. Normal operation is resumed as soon as Vs exceeds Vsth. The hysteretic behaviour prevents intermittent operation at low supply voltage.

OVER CURRENT OPERATION

In order to implement a short circuit protection the output power MOS is driven in linear mode to limit the output current to the Isc value. This Isc limit is externally settable by means of an external 1/4 W resistor connected from Rsc pin and GND. The value of the resistor must be chosen according to the following formula:

$$Isc(A) = 5/Rsc(kohm)$$

with

For

Rsc < 5 (kohm)

Isc is limited to Isc=1.1 A (typical value).

This condition (current limited to the lsc value) lasts for a Ton time interval, that can be set by means of a capacitor (Cdon) connected to the ON DELAY pin according to the following formula:

for

50pF<Cdon<2nF

After the Ton interval has expired the output power MOS is switched off for the Toff time inter-. val with:

Toff = 64*Ton.



MICROELECTRONICS

Figure 3: Short Circuit Operation Waveforms

When also the Toff interval has expired, the output power MOS is switched ON.

Now two conditions may occur

- the overload is still present. In this case the output power MOS is again driven in linear mode (limiting the output current to lsc) for another Ton, starting a new cycle, or

- the overload condition is removed, and the output power MOS is no longer driven in linear mode.

All these occurrences are presented on the DIAG pin (see fig 2). We call this unique feature **Non Dissipative Short Circuit Protection** and it ensures a very safe operation even in permanent overload conditions. Note that, of course, choosing the most appropriate value for the Ton interval (i.e. the value of the Cdon capacitor) a delay (the Ton itself) will prevent that a misleading Short Circuit information is presented on the DIAG output, when driving capacitive loads (that acts like short circuit in the very beginning) or Incandescent Lamp (a cold filament has a very low resistive value).

The Non Dissipative Short Circuit Protection can be disabled (keeping Ton = 0 but with the output current still limited to Isc, and Diagnostic disabled) simply shorting to ground the the ON DE-LAY pin.

DEMAGNETISATION OF INDUCTIVE LOADS

The L6377 has an internal clamping zener diode able to demagnetise inductive loads. Note that the limitation comes from the peak power that the package can handle. Attention must be paid to a proper thermal design of the board. If, for whatever reason (load current or inductive value too big) the peak power dissipation is too high, an external Zener plus Diode arrangement, can perform a demagnetisation versus Ground or versus Vs (see fig 5 and 6). The breakdown voltage of the external Zener Diode must be chosen considering the internal clamping voltage (Vcl) and the supply voltage (Vs) according to:

Vz< Vcl(min)-Vs(max)

for demagnetisation versus Ground or

Vs(max)<Vz<Vcl(min)

for demagnetisation versus Vs.

Figure 4: Input Comparator Hysteresis









Figure 6: External Demagnetisation Circuit (versus V_S)





514		mm		inch				
DIM.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.		
А		3.32			0.131			
a1	0.51			0.020				
В	1.15		1.65	0.045		0.065		
b	0.356		0.55	0.014		0.022		
b1	0.204		0.304	0.008		0.012		
D			10.92			0.430		
E	7.95		9.75	0.313		0.384		
е		2.54			0.100			
e3		7.62			0.300			
e4		7.62			0.300			
F			6.6			0.260		
I			5.08			0.200		
L	3.18		3.81	0.125		0.150		
Z			1.52			0.060		

MINIDIP PACKAGE MECHANICAL DATA



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016019	MIN	TYP	MAX	MIN	TYP	MAX
A			1.75			0.069
a1	0.1		0.2	0.004		0.008
a2			1.6			0.063
b	0.35		0.46	0.014		0.018
b1	0.19		0.25	0.007		0.010
С		0.5			0.020	
c1		45			1.772	
D		1	8.55		0.039	0.337
E	5.8		6.2	0.228		0.244
е		1.27			0.050	
e3		7.62			0.300	
F		1	3.8		0.039	0.150
G	4.6		5.3	0.181		0.209
L	0.5		1.27	0.020		0.050
М			0.68			0.027
S			8			0.315



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