

PIC16C925/926 Data Sheet

64/68-Pin CMOS Microcontrollers with LCD Driver

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64/68-Pin CMOS Microcontrollers with LCD Driver

High Performance RISC CPU:

- · Only 35 single word instructions to learn
- All single cycle instructions except for program branches which are two-cycle
- Operating speed: DC 20 MHz clock input DC - 200 ns instruction cycle
- Up to 8K x 14-bit words of EPROM program memory, 336 bytes general purpose registers (SRAM), 60 special function registers
- Pinout compatible with PIC16C923/924

Peripheral Features:

- 25 I/O pins with individual direction control and 25-27 input only pins
- Timer0 module: 8-bit timer/counter with programmable 8-bit prescaler
- Timer1 module: 16-bit timer/counter, can be incremented during SLEEP via external crystal/clock
- Timer2 module: 8-bit timer/counter with 8-bit period register, prescaler, and postscaler
- One Capture, Compare, PWM module
- Synchronous Serial Port (SSP) module with two modes of operation:
 - 3-wire SPI™ (supports all 4 SPI modes)
 - I²C™ Slave mode
- Programmable LCD timing module:
 - Multiple LCD timing sources available
 - Can drive LCD panel while in SLEEP mode
 - Static, 1/2, 1/3, 1/4 multiplex
 - Static drive and 1/3 bias capability
 - 16 bytes of dedicated LCD RAM
 - Up to 32 segments, up to 4 commons

Common	Segment	Pixels
1	32	32
2	31	62
3	30	90
4	29	116

Analog Features:

- 10-bit 5-channel Analog-to-Digital Converter (A/D)
- Brown-out Reset (BOR)

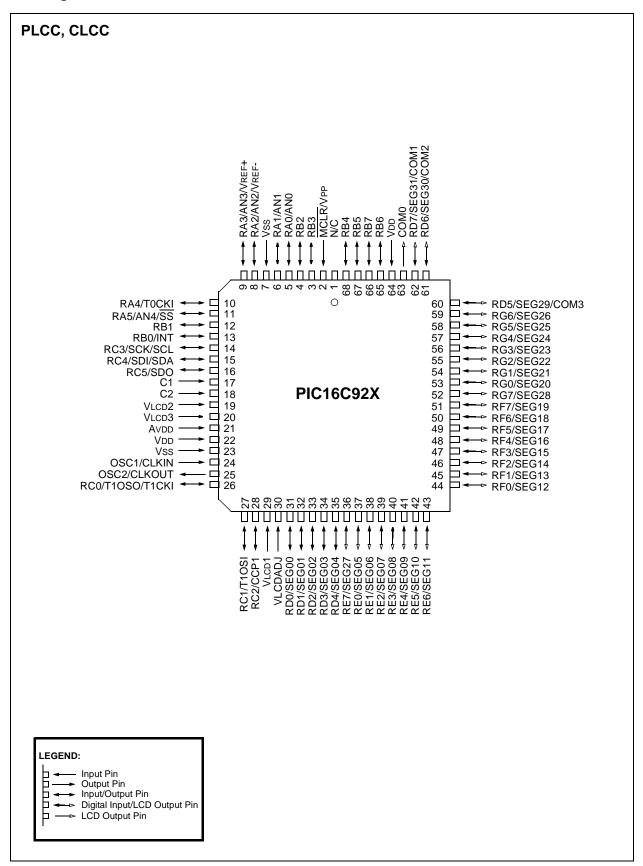
Special Microcontroller Features:

- Power-on Reset (POR)
- Power-up Timer (PWRT) and Oscillator Start-up Timer (OST)
- Watchdog Timer (WDT) with its own on-chip RC oscillator for reliable operation
- · Programmable code protection
- · Selectable oscillator options
- In-Circuit Serial Programming[™] (ICSP[™]) via two pins
- Processor read access to program memory

CMOS Technology:

- Low power, high speed CMOS/EPROM technology
- · Fully static design
- Wide operating voltage range: 2.5V to 5.5V
- · Commercial and Industrial temperature ranges
- Low power consumption

Pin Diagrams



Pin Diagrams (Continued)

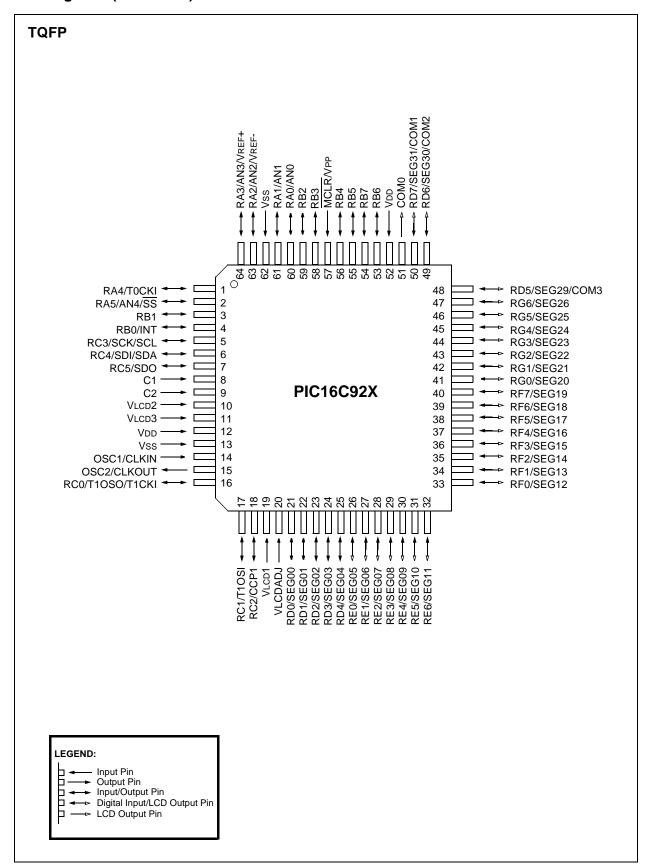


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1.0 DEVICE OVERVIEW

This document contains device-specific information for the following devices:

- 1. PIC16C925
- 2. PIC16C926

The PIC16C925/926 series is a family of low cost, high performance, CMOS, fully static, 8-bit microcontrollers with an integrated LCD Driver module, in the PIC16CXXX mid-range family.

For the PIC16C925/926 family, there are two device "types" as indicated in the device number:

- 1. **C**, as in PIC16C926. These devices operate over the standard voltage range.
- 2. **LC**, as in PIC16**LC**926. These devices operate over an extended voltage range.

These devices come in 64-pin and 68-pin packages, as well as die form. Both configurations offer identical peripheral devices and other features. The only difference between the PIC16C925 and PIC16C926 is the additional EPROM and data memory offered in the latter. An overview of features is presented in Table 1-1.

A UV-erasable, CERQUAD packaged version (compatible with PLCC) is also available for both the PIC16C925 and PIC16C926. This version is ideal for cost effective code development.

A block diagram for the PIC16C925/926 family architecture is presented in Figure 1-1.

TABLE 1-1: PIC16C925/926 DEVICE FEATURES

Features	PIC16C925	PIC16C926		
Operating Frequency	DC-20 MHz	DC-20 MHz		
EPROM Program Memory (words)	4K	8K		
Data Memory (bytes)	176	336		
Timer Module(s)	TMR0,TMR1,TMR2	TMR0,TMR1,TMR2		
Capture/Compare/PWM Module(s)	1	1		
Serial Port(s) (SPI/I ² C, USART)	SPI/I ² C	SPI/I ² C		
Parallel Slave Port	_	_		
A/D Converter (10-bit) Channels	5	5		
LCD Module	4 Com, 32 Seg	4 Com, 32 Seg		
Interrupt Sources	9	9		
I/O Pins	25	25		
Input Pins	27	27		
Voltage Range (V)	2.5-5.5	2.5-5.5		
In-Circuit Serial Programming	Yes	Yes		
Brown-out Reset	Yes	Yes		
Packages	64-pin TQFP 68-pin PLCC 68-pin CLCC (CERQUAD) Die	64-pin TQFP 68-pin PLCC 68-pin CLCC (CERQUAD) Die		

PORTA Data Bus Program Counter **RA0**/AN0 **EPROM** RA1/AN1 RA2/AN2 Program RAM RA3/AN3/VREF 8 Level Stack Memory RA4/T0CKI File (13-bit) RA5/AN4/SS Registers Program 9 RAM Addr **PORTB** Addr MUX Instruction reg RB0/INT Direct Addr 8 Indirect 8 Addr ⊳⊠ RB1-RB7 FSR reg STATUS reg PORTC 8 RC0/T1OSO/T1CKI RC1/T1OSI RC2/CCP1 MUX Power-up RC3/SCK/SCL Timer RC4/SDI/SDA Instruction Oscillator RC5/SDO Start-up Timer Decode & Control ALU Power-on Reset PORTD 8 Timing Generation Watchdog Timer W reg RD0-RD4/SEGnn OSC1/CLKIN OSC2/CLKOUT RD5-RD7/SEGnn/COMn X MCLR VDD, VSS **PORTE** RE0-RE7/SEGnn PORTF RF0-RF7/SEGnn **PORTG** RG0-RG7/SEGnn Timer1, Timer2, A/D Timer0 CCP1 COM0 Synchronous VLCD1 Serial Port VLCD2 VLCD3 LCD C1 C2 VLCDADJ

PIC16C925/926 BLOCK DIAGRAM FIGURE 1-1:

TABLE 1-2: PIC16C925/926 PINOUT DESCRIPTION

Pin Name	PLCC, CLCC Pin#	TQFP Pin#	Pin Type	Buffer Type	Description
OSC1/CLKIN	24	14	I	ST/CMOS	Oscillator crystal input or external clock source input. This buffer is a Schmitt Trigger input when configured in RC oscillator mode and a CMOS input otherwise.
OSC2/CLKOUT	25	15	0	_	Oscillator crystal output. Connects to crystal or resonator in crystal oscillator mode. In RC mode, OSC2 pin outputs CLKOUT, which has 1/4 the frequency of OSC1 and denotes the instruction cycle rate.
MCLR/VPP	2	57	I/P	ST	Master Clear (Reset) input or programming voltage input. This pin is an active low RESET to the device.
					PORTA is a bi-directional I/O port.
RA0/AN0	5	60	I/O	TTL	RA0 can also be Analog input0.
RA1/AN1	6	61	I/O	TTL	RA1 can also be Analog input1.
RA2/AN2	8	63	I/O	TTL	RA2 can also be Analog input2.
RA3/AN3/VREF	9	64	I/O	TTL	RA3 can also be Analog input3 or A/D Voltage Reference.
RA4/T0CKI	10	1	I/O	ST	RA4 can also be the clock input to the Timer0 timer/counter. Output is open drain type.
RA5/AN4/SS	11	2	I/O	TTL	RA5 can be the slave select for the synchronous serial port or Analog input4.
					PORTB is a bi-directional I/O port. PORTB can be software programmed for internal weak pull-ups on all inputs.
RB0/INT	13	4	I/O	TTL/ST	RB0 can also be the external interrupt pin. This buffer is a Schmitt Trigger input when configured as an external interrupt.
RB1	12	3	I/O	TTL	
RB2	4	59	I/O	TTL	
RB3	3	58	I/O	TTL	
RB4	68	56	I/O	TTL	Interrupt-on-change pin.
RB5	67	55	I/O	TTL	Interrupt-on-change pin.
RB6	65	53	I/O	TTL/ST	Interrupt-on-change pin. Serial programming clock. This buffer is a Schmitt Trigger input when used in Serial Programming mode.
RB7	66	54	I/O	TTL/ST	Interrupt-on-change pin. Serial programming data. This buffer is a Schmitt Trigger input when used in Serial Programming mode.
					PORTC is a bi-directional I/O port.
RC0/T1OSO/T1CKI	26	16	I/O	ST	RC0 can also be the Timer1 oscillator output or Timer1 clock input.
RC1/T1OSI	27	17	I/O	ST	RC1 can also be the Timer1 oscillator input.
RC2/CCP1	28	18	I/O	ST	RC2 can also be the Capture1 input/Compare1 output/PWM1 output.
RC3/SCK/SCL	14	5	I/O	ST	RC3 can also be the synchronous serial clock input/output for both SPI and I ² C modes.
RC4/SDI/SDA	15	6	I/O	ST	RC4 can also be the SPI Data In (SPI mode) or data I/O (I ² C mode).
RC5/SDO	16	7	I/O	ST	RC5 can also be the SPI Data Out (SPI mode).
C1	17	8	Р		LCD Voltage Generation.
C2	18	9	Р		LCD Voltage Generation.
COM0	63	51	L		Common Driver0.

Legend: I = input — = Not used O = output TTL = TTL input P = power ST = Schmitt Trigger input

L = LCD Driver

TABLE 1-2: PIC16C925/926 PINOUT DESCRIPTION (CONTINUED)

Pin Name	PLCC, CLCC Pin#	TQFP Pin#	Pin Type	Buffer Type	Description
					PORTD is a digital input/output port. These pins are also used as LCD Segment and/or Common Drivers.
RD0/SEG00	31	21	I/O/L	ST	Segment Driver 00/Digital input/output.
RD1/SEG01	32	22	I/O/L	ST	Segment Driver 01/Digital input/output.
RD2/SEG02	33	23	I/O/L	ST	Segment Driver 02/Digital input/output.
RD3/SEG03	34	24	I/O/L	ST	Segment Driver 03/Digital input/output.
RD4/SEG04	35	25	I/O/L	ST	Segment Driver04/Digital input/output.
RD5/SEG29/COM3	60	48	I/L	ST	Segment Driver29/Common Driver 3/Digital input.
RD6/SEG30/COM2	61	49	I/L	ST	Segment Driver30/Common Driver 2/Digital input.
RD7/SEG31/COM1	62	50	I/L	ST	Segment Driver31/Common Driver 1/Digital input.
					PORTE is a Digital input or LCD Segment Driver port.
RE0/SEG05	37	26	I/L	ST	Segment Driver 05.
RE1/SEG06	38	27	I/L	ST	Segment Driver 06.
RE2/SEG07	39	28	I/L	ST	Segment Driver 07.
RE3/SEG08	40	29	I/L	ST	Segment Driver 08.
RE4/SEG09	41	30	I/L	ST	Segment Driver 09.
RE5/SEG10	42	31	I/L	ST	Segment Driver 10.
RE6/SEG11	43	32	I/L	ST	Segment Driver 11.
RE7/SEG27	36	_	I/L	ST	Segment Driver 27 (not available on 64-pin devices).
			-,-		PORTF is a Digital input or LCD Segment Driver port.
RF0/SEG12	44	33	I/L	ST	Segment Driver 12.
RF1/SEG13	45	34	I/L	ST	Segment Driver 13.
RF2/SEG14	46	35	I/L	ST	Segment Driver 14.
RF3/SEG15	47	36	I/L	ST	Segment Driver 15.
RF4/SEG16	48	37	I/L	ST	Segment Driver 16.
RF5/SEG17	49	38	I/L	ST	Segment Driver 17.
RF6/SEG18	50	39	I/L	ST	Segment Driver 18.
RF7/SEG19	51	40	I/L	ST	Segment Driver 19.
102010	31	70	1/ -	- 01	PORTG is a Digital input or LCD Segment Driver port.
RG0/SEG20	53	41	I/L	ST	Segment Driver 20.
RG1/SEG21	54	42	I/L	ST	Segment Driver 21.
RG2/SEG22	55	43	I/L	ST	Segment Driver 22.
RG3/SEG23	56	44	I/L	ST	Segment Driver 23.
RG4/SEG24	57	45	I/L	ST	Segment Driver 24.
RG5/SEG25	58	46	I/L	ST	Segment Driver 25.
RG6/SEG26	59	47	I/L	ST	Segment Driver 26.
RG7/SEG28	52	47	I/L	ST	Segment Driver 28 (not available on 64-pin devices).
VLCDADJ	30	20	P	31	LCD Voltage Generation.
					Analog Power (PLCC and CLCC packages only).
AVDD VLCD1	21 29	10	P P		LCD Voltage.
		19	P	_	
VLCD2	19	10	P		LCD Voltage.
VLCD3	20	11			LCD Voltage.
VDD	22, 64	12, 52	Р	_	Digital power.
Vss	7, 23	13, 62	Р		Ground reference.
NC	1	_			These pins are not internally connected. These pins should be left unconnected.

Legend: I = input

O = output

P = power

L = LCD Driver

— = Not used TTL = TTL input

ST = Schmitt Trigger input

1.1 Clocking Scheme/Instruction Cycle

The clock input (from OSC1) is internally divided by four to generate four non-overlapping quadrature clocks, namely Q1, Q2, Q3 and Q4. Internally, the program counter (PC) is incremented every Q1, the instruction is fetched from the program memory and latched into the instruction register in Q4. The instruction is decoded and executed during the following Q1 through Q4. The clocks and instruction execution flow are shown in Figure 1-2.

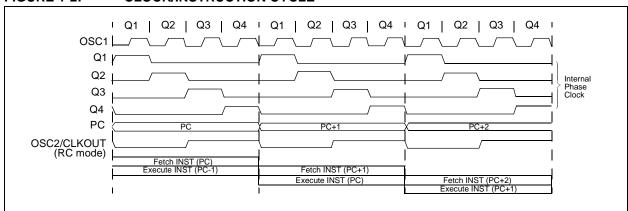
1.2 Instruction Flow/Pipelining

An "Instruction Cycle" consists of four Q cycles (Q1, Q2, Q3 and Q4). The instruction fetch and execute are pipelined, such that fetch takes one instruction cycle, while decode and execute takes another instruction cycle. However, due to the pipelining, each instruction effectively executes in one cycle. If an instruction causes the program counter to change (e.g. GOTO), then two cycles are required to complete the instruction (Example 1-1).

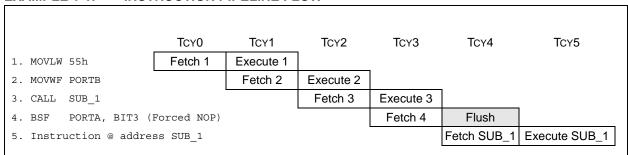
A fetch cycle begins with the program counter (PC) incrementing in Q1.

In the execution cycle, the fetched instruction is latched into the "Instruction Register" in cycle Q1. This instruction is then decoded and executed during the Q2, Q3, and Q4 cycles. Data memory is read during Q2 (operand read) and written during Q4 (destination write).





EXAMPLE 1-1: INSTRUCTION PIPELINE FLOW



All instructions are single cycle, except for any program branches. These take two cycles, since the fetch instruction is "flushed" from the pipeline while the new instruction is being fetched and then executed.

NOTES:

2.0 MEMORY ORGANIZATION

2.1 Program Memory Organization

The PIC16C925/926 family has a 13-bit program counter capable of addressing an 8K \times 14 program memory space.

For the PIC16C925, only the first 4K x 14 (0000h-0FFFh) are physically implemented. Accessing a location above the physically implemented addresses will cause a wraparound. The RESET vector is at 0000h and the interrupt vector is at 0004h.

FIGURE 2-1: PROGRAM MEMORY MAP
AND STACK FOR
PIC16C925

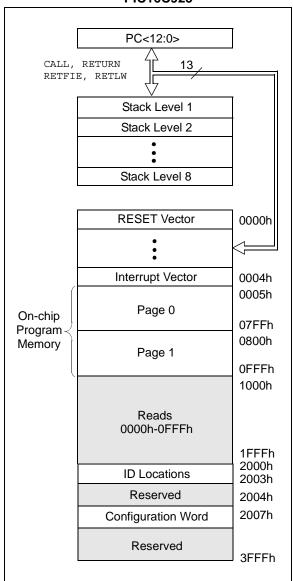
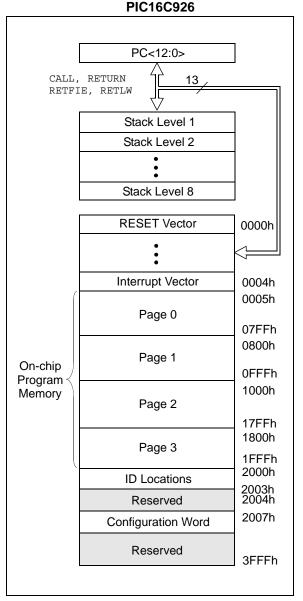


FIGURE 2-2: PROGRAM MEMORY MAP
AND STACK FOR



2.2 Data Memory Organization

The data memory is partitioned into four banks which contain the General Purpose Registers and the Special Function Registers. Bits RP1 and RP0 are the bank select bits.

RP1:RP0 (STATUS<6:5>)	Bank
11	3 (180h-1FFh)
10	2 (100h-17Fh)
01	1 (80h-FFh)
00	0 (00h-7Fh)

The lower locations of each Bank are reserved for the Special Function Registers. Above the Special Function Registers are General Purpose Registers implemented as static RAM. All four banks contain special function registers. Some "high use" special function registers are mirrored in other banks for code reduction and quicker access.

2.2.1 GENERAL PURPOSE REGISTER FILE

The register file can be accessed either directly, or indirectly through the File Select Register FSR (Section 2.6).

The following General Purpose Registers are not physically implemented:

- F0h-FFh of Bank 1
- 170h-17Fh of Bank 2
- 1F0h-1FFh of Bank 3

These locations are used for common access across banks.

FIGURE 2-3: REGISTER FILE MAP — PIC16C925

	File Address	,	File Address		File Address		File Addres
Indirect addr.	*) 00h	Indirect addr.(*)	80h	Indirect addr.(*)	100h	Indirect addr.(*)	180h
TMR0	01h	OPTION	81h	TMR0	101h	OPTION	181h
PCL	02h	PCL	82h	PCL	102h	PCL	182h
STATUS	03h	STATUS	83h	STATUS	103h	STATUS	183h
FSR	04h	FSR	84h	FSR	104h	FSR	184h
PORTA	05h	TRISA	85h		105h		185h
PORTB	06h	TRISB	86h	PORTB	106h	TRISB	186h
PORTC	07h	TRISC	87h	PORTF	107h	TRISF	187h
PORTD	08h	TRISD	88h	PORTG	108h	TRISG	188h
PORTE	09h	TRISE	89h		109h		189h
PCLATH	0Ah	PCLATH	8Ah	PCLATH	10Ah	PCLATH	18Al
INTCON	0Bh	INTCON	8Bh	INTCON	10Bh	INTCON	18BI
PIR1	0Ch	PIE1	8Ch	PMCON1	10Ch	PMDATA	18C
	0Dh		8Dh	LCDSE	10Dh	PMADR	18D
TMR1L	0Eh	PCON	8Eh	LCDPS	10Eh	PMDATH	18E
TMR1H	0Fh		8Fh	LCDCON	10Fh	PMADRH	18FI
T1CON	10h		90h	LCDD00	110h		190ł
TMR2	11h		91h	LCDD01	111h		191ŀ
T2CON	12h	PR2	92h	LCDD02	112h		192l
SSPBUF	13h	SSPADD	93h	LCDD03	113h		193
SSPCON	14h	SSPSTAT	94h	LCDD04	114h		1941
CCPR1L	15h		95h	LCDD05	115h		195l
CCPR1H	16h		96h	LCDD06	116h		196
CCP1CON	17h		97h	LCDD07	117h		197
	18h		98h	LCDD08	118h		198
	19h		99h	LCDD09	119h		199
	1Ah		9Ah	LCDD10	11Ah		19A
	1Bh		9Bh	LCDD11	11Bh		19B
	1Ch		9Ch	LCDD12	11Ch		19C
	1Dh		9Dh	LCDD13	11Dh		19D
ADRESH	1Eh	ADRESL	9Eh	LCDD14	11Eh		19EI
ADCON0	1Fh	ADCON1	9Fh	LCDD15	11Fh		19FI
	20h		A0h		120h		1A0
General Purpose		General Purpose Register					
Register			EFh		16Fh		1EF
		accesses	F0h	accesses	170h	accesses	1F0I
	754	70h - 7Fh		70h - 7Fh		70h - 7Fh	
Bank 0	7Fh	Bank 1	FFh	Bank 2	17Fh	Bank 3	1FF

FIGURE 2-4: REGISTER FILE MAP— PIC16C926

A	File Address	,	File Address		File Address		File Addres
Indirect addr.(*)	00h	Indirect addr.(*)	80h	Indirect addr.(*)	100h	Indirect addr.(*)	180h
TMR0	01h	OPTION	81h	TMR0	101h	OPTION	181h
PCL	02h	PCL	82h	PCL	102h	PCL	182h
STATUS	03h	STATUS	83h	STATUS	103h	STATUS	183h
FSR	04h	FSR	84h	FSR	104h	FSR	184h
PORTA	05h	TRISA	85h		105h		185h
PORTB	06h	TRISB	86h	PORTB	106h	TRISB	186h
PORTC	07h	TRISC	87h	PORTF	107h	TRISF	187h
PORTD	08h	TRISD	88h	PORTG	108h	TRISG	188h
PORTE	09h	TRISE	89h		109h		189h
PCLATH	0Ah	PCLATH	8Ah	PCLATH	10Ah	PCLATH	18Ał
INTCON	0Bh	INTCON	8Bh	INTCON	10Bh	INTCON	18Bh
PIR1	0Ch	PIE1	8Ch	PMCON1	10Ch	PMDATA	18Cł
	0Dh		8Dh	LCDSE	10Dh	PMADR	18Dł
TMR1L	0Eh	PCON	8Eh	LCDPS	10Eh	PMDATH	18Eł
TMR1H	0Fh		8Fh	LCDCON	10Fh	PMADRH	18Fh
T1CON	10h		90h	LCDD00	110h		190h
TMR2	11h		91h	LCDD01	111h		191h
T2CON	12h	PR2	92h	LCDD02	112h		192h
SSPBUF	13h	SSPADD	93h	LCDD03	113h		193h
SSPCON	14h	SSPSTAT	94h	LCDD04	114h		194h
CCPR1L	15h		95h	LCDD05	115h		195h
CCPR1H	16h		96h	LCDD06	116h		196h
CCP1CON	17h		97h	LCDD07	117h		197h
	18h		98h	LCDD08	118h		198h
	19h		99h	LCDD09	119h		199h
	1Ah		9Ah	LCDD10	11Ah		19Ał
	1Bh		9Bh	LCDD11	11Bh		19Bł
	1Ch		9Ch	LCDD12	11Ch		19Cł
	1Dh		9Dh	LCDD13	11Dh		19Dł
ADRESH	1Eh	ADRESL	9Eh	LCDD14	11Eh		19Eł
ADCON0	1Fh	ADCON1	9Fh	LCDD15	11Fh		19Fh
	20h		A0h		120h		1A0ł
General Purpose Register 96 Bytes		General Purpose Register 80 Bytes	BFh C0h	General Purpose Register 80 Bytes		General Purpose Register 80 Bytes	
	7Fh	accesses 70h - 7Fh	EFh F0h FFh	accesses 70h - 7Fh	16Fh 170h 17Fh	accesses 70h - 7Fh	1EFt 1F0t 1FFt
Bank 0	. ,,,,,	Bank 1		Bank 2		Bank 3	

2.3 Special Function Registers

The Special Function Registers (SFRs) are registers used by the CPU and Peripheral Modules for controlling the desired operation of the device. These registers are implemented as static RAM.

The special function registers can be classified into two sets, core and peripheral. Those registers associated with the "core" functions are described in this section. Those related to the operation of the peripheral features are described in the section of that peripheral feature.

TABLE 2-1: SPECIAL FUNCTION REGISTER SUMMARY

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Details on page
Bank 0											
00h	INDF	Addressing	this location	uses content	s of FSR to a	ddress data	memory (no	t a physical r	egister)	0000 0000	26
01h	TMR0	Timer0 Mod	dule Register							xxxx xxxx	41
02h	PCL	Program Co	ounter (PC) L	east Significa	ant Byte					0000 0000	25
03h	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	19
04h	FSR	Indirect Dat	a Memory Ad	Idress Pointe	r					xxxx xxxx	26
05h	PORTA	_	PORTA Data Latch when written: PORTA pins when read							0x 0000	29
06h	PORTB	PORTB Dat	ta Latch wher	n written: PO	RTB pins wh	en read				xxxx xxxx	31
07h	PORTC	_	_	PORTC Dat	a Latch wher	n written: PO	RTC pins wh	en read		xx xxxx	33
08h	PORTD	PORTD Da	ta Latch whe	n written: PO	RTD pins wh	en read				0000 0000	34
09h	PORTE	PORTE pin	s when read							0000 0000	36
0Ah	PCLATH	_	_	_	Write Buffer	for the uppe	r 5 bits of the	Program Co	ounter	0 0000	25
0Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	21
0Ch	PIR1	LCDIF	ADIF	_	_	SSPIF	CCP1IF	TMR2IF	TMR1IF	00 0000	23
0Dh	_	Unimpleme	nted							_	_
0Eh	TMR1L	Holding reg	Holding register for the Least Significant Byte of the 16-bit TMR1 Register							xxxx xxxx	47
0Fh	TMR1H	Holding reg	ister for the N	lost Significa	nt Byte of the	e 16-bit TMR	1 Register			xxxx xxxx	47
10h	T1CON	_	_	T1CKPS1	T1CKPS0	T10SCEN	T1SYNC	TMR1CS	TMR10N	00 0000	47
11h	TMR2	Timer2 Mod	dule Register							0000 0000	51
12h	T2CON	_	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	52
13h	SSPBUF	Synchronou	us Serial Port	Receive Buf	fer/Transmit	Register				xxxx xxxx	64, 72
14h	SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	60
15h	CCPR1L	Capture/Co	mpare/PWM	Register (LS	B)					xxxx xxxx	58
16h	CCPR1H	Capture/Co	mpare/PWM	Register (MS	SB)					xxxx xxxx	58
17h	CCP1CON	_	_	CCP1X	CCP1Y	CCP1M3	CCP1M2	CCP1M1	CCP1M0	00 0000	53
18h	1	Unimpleme	nted							_	_
19h	-	Unimpleme	nted							_	_
1Ah	_	Unimpleme	nted							_	_
1Bh	_	Unimpleme	nted							_	_
1Ch	_	Unimpleme	nted							_	_
1Dh		Unimpleme	nted								_
1Eh	ADRESH	A/D Result	Register High	1						xxxx xxxx	80, 81
1Fh	ADCON0	ADCS1	ADCS0	CHS2	CHS1	CHS0	GO/DONE	_	ADON	0000 0000	75

Legend: x = unknown, u = unchanged, q = value depends on condition, -= unimplemented, read as '0'. Shaded locations are unimplemented, read as '0'.

Note 1: These pixels do not display, but can be used as general purpose RAM.

SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED) **TABLE 2-1:**

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Details on page
Bank 1											
80h	INDF	Addressing	this location	uses conten	ts of FSR to a	address data	memory (no	t a physical r	egister)	0000 0000	26
81h	OPTION	RBPU	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	20
82h	PCL	Program Co	ounter (PC) L	east Signific	ant Byte					0000 0000	25
83h	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	19
84h	FSR	Indirect Dat	a Memory Ad	ddress Pointe	er					xxxx xxxx	26
85h	TRISA	_	PORTA Data Direction Register								29
86h	TRISB	PORTB Dat	ta Direction R	Register						1111 1111	31
87h	TRISC	_	_	PORTC Da	ta Direction F	Register				11 1111	33
88h	TRISD	PORTD Date	ta Direction F	Register						1111 1111	34
89h	TRISE	PORTE Dat	ta Direction R	Register						1111 1111	36
8Ah	PCLATH	_	_	_	Write Buffer	for the uppe	er 5 bits of the	PC		0 0000	25
8Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	21
8Ch	PIE1	LCDIE	ADIE	_	_	SSPIE	CCP1IE	TMR2IE	TMR1IE	00 0000	24
8Dh	_	Unimpleme	nted							_	_
8Eh	PCON	_	_	_	_	_	_	POR	BOR	0-	24
8Fh	_	Unimpleme	nted							_	_
90h	_	Unimpleme	nted							_	_
91h	_	Unimpleme	nted							_	_
92h	PR2		od Register							1111 1111	51
93h	SSPADD	Synchronou	ıs Serial Port	(I ² C mode)	Address Reg	ister				0000 0000	69, 72
94h	SSPSTAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	0000 0000	59
95h	_	Unimpleme	nted							_	_
96h	_	Unimpleme	nted							_	_
97h	_	Unimpleme	nted							_	_
98h	_	Unimpleme	nted							_	_
99h	_	Unimpleme	nted							_	_
9Ah	_	Unimpleme	nted							_	_
9Bh	_	Unimpleme	nted							_	_
9Ch	_	Unimpleme	nted							_	_
9Dh		Unimpleme	nted							_	_
9Eh	ADRESL	A/D Result	Register Low	'						xxxx xxxx	79
9Fh	ADCON1	_	_	_	_	_	PCFG2	PCFG1	PCFG0	000	76

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0'. Shaded locations are unimplemented, read as '0'.

Note 1: These pixels do not display, but can be used as general purpose RAM.

SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED) **TABLE 2-1:**

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Details on page
Bank 2											
100h	INDF	Addressing	this location	uses content	ts of FSR to a	address data	memory (no	t a physical r	egister)	0000 0000	26
101h	TMR0	Timer0 Mod	lule Register							xxxx xxxx	41
102h	PCL	Program Co	ounter (PC) L	east Significa	ant Byte					0000 0000	25
103h	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	19
104h	FSR	Indirect Data	a Memory Ad	dress Pointe	er					xxxx xxxx	26
105h	_	Unimpleme	nted							_	_
106h	PORTB	PORTB Dat	a Latch wher	n written: PO	RTB pins wh	en read				xxxx xxxx	31
107h	PORTF	PORTF pins	s when read							0000 0000	37
108h	PORTG	PORTG pin	s when read							0000 0000	38
109h	_	Unimpleme	nted							_	_
10Ah	PCLATH	_	_	_	Write Buffer	for the uppe	r 5 bits of the	PC		0 0000	25
10Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	21
10Ch	PMCON1	reserved	_	_	_	_	_	_	RD	10	27
10Dh	LCDSE	SE29	SE27	SE20	SE16	SE12	SE9	SE5	SE0	1111 1111	94
10Eh	LCDPS	_	_	_	_	LP3	LP2	LP1	LP0	0000	84
10Fh	LCDCON	LCDEN	SLPEN	_	VGEN	CS1	CS0	LMUX1	LMUX0	00-0 0000	83
110h	LCDD00	SEG07 COM0	SEG06 COM0	SEG05 COM0	SEG04 COM0	SEG03 COM0	SEG02 COM0	SEG01 COM0	SEG00 COM0	xxxx xxxx	92
111h	LCDD01	SEG15 COM0	SEG14 COM0	SEG13 COM0	SEG12 COM0	SEG11 COM0	SEG10 COM0	SEG09 COM0	SEG08 COM0	xxxx xxxx	92
112h	LCDD02	SEG23 COM0	SEG22 COM0	SEG21 COM0	SEG20 COM0	SEG19 COM0	SEG18 COM0	SEG17 COM0	SEG16 COM0	xxxx xxxx	92
113h	LCDD03	SEG31 COM0	SEG30 COM0	SEG29 COM0	SEG28 COM0	SEG27 COM0	SEG26 COM0	SEG25 COM0	SEG24 COM0	xxxx xxxx	92
114h	LCDD04	SEG07 COM1	SEG06 COM1	SEG05 COM1	SEG04 COM1	SEG03 COM1	SEG02 COM1	SEG01 COM1	SEG00 COM1	xxxx xxxx	92
115h	LCDD05	SEG15 COM1	SEG14 COM1	SEG13 COM1	SEG12 COM1	SEG11 COM1	SEG10 COM1	SEG09 COM1	SEG08 COM1	xxxx xxxx	92
116h	LCDD06	SEG23 COM1	SEG22 COM1	SEG21 COM1	SEG20 COM1	SEG19 COM1	SEG18 COM1	SEG17 COM1	SEG16 COM1	xxxx xxxx	92
117h	LCDD07	SEG31 COM1 ⁽¹⁾	SEG30 COM1	SEG29 COM1	SEG28 COM1	SEG27 COM1	SEG26 COM1	SEG25 COM1	SEG24 COM1	xxxx xxxx	92
118h	LCDD08	SEG07 COM2	SEG06 COM2	SEG05 COM2	SEG04 COM2	SEG03 COM2	SEG02 COM2	SEG01 COM2	SEG00 COM2	xxxx xxxx	92
119h	LCDD09	SEG15 COM2	SEG14 COM2	SEG13 COM2	SEG12 COM2	SEG11 COM2	SEG10 COM2	SEG09 COM2	SEG08 COM2	xxxx xxxx	92
11Ah	LCDD10	SEG23 COM2	SEG22 COM2	SEG21 COM2	SEG20 COM2	SEG19 COM2	SEG18 COM2	SEG17 COM2	SEG16 COM2	xxxx xxxx	92
11Bh	LCDD11	SEG31 COM2 ⁽¹⁾	SEG30 COM2 ⁽¹⁾	SEG29 COM2	SEG28 COM2	SEG27 COM2	SEG26 COM2	SEG25 COM2	SEG24 COM2	xxxx xxxx	92
11Ch	LCDD12	SEG07 COM3	SEG06 COM3	SEG05 COM3	SEG04 COM3	SEG03 COM3	SEG02 COM3	SEG01 COM3	SEG00 COM3	xxxx xxxx	92
11Dh	LCDD13	SEG15 COM3	SEG14 COM3	SEG13 COM3	SEG12 COM3	SEG11 COM3	SEG10 COM3	SEG09 COM3	SEG08 COM3	xxxx xxxx	92
11Eh	LCDD14	SEG23 COM3	SEG22 COM3	SEG21 COM3	SEG20 COM3	SEG19 COM3	SEG18 COM3	SEG17 COM3	SEG16 COM3	xxxx xxxx	92
11Fh	LCDD15	SEG31 COM3 ⁽¹⁾	SEG30 COM3 ⁽¹⁾	SEG29 COM3 ⁽¹⁾	SEG28 COM3	SEG27 COM3	SEG26 COM3	SEG25 COM3	SEG24 COM3	xxxx xxxx	92

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0'. Shaded locations are unimplemented, read as '0'.

Note 1: These pixels do not display, but can be used as general purpose RAM.

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SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED) **TABLE 2-1:**

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Details on page
Bank 3											
180h	INDF	Addressing	this location	uses conten	ts of FSR to	address data	memory (no	t a physical r	egister)	0000 0000	26
181h	OPTION	RBPU	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	20
182h	PCL	Program Co	unter's (PC)	Least Signif	icant Byte					0000 0000	25
183h	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	19
184h	FSR	Indirect Dat	a Memory Ad	Idress Pointe	er					xxxx xxxx	26
185h	_	Unimpleme	nted		_	_					
186h	TRISB	PORTB Dat	a Direction R	egister						1111 1111	31
187h	TRISF	PORTF Dat	a Direction R	egister						1111 1111	37
188h	TRISG	PORTG Da	ta Direction R	Register						1111 1111	38
189h	_	Unimpleme	nted							_	_
18Ah	PCLATH	_	_	_	Write Buffer	for the uppe	er 5 bits of the	PC		0 0000	25
18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	21
18Ch	PMDATA	Data Regist	er Low Byte							xxxx xxxx	27
18Dh	PMADR	Address Re	gister Low By	/te						xxxx xxxx	27
18Eh	PMDATH	_	_	Data Regis	ter High Byte					xxxx xxxx	27
18Fh	PMADRH	_	_	_	Address Re	gister High E	Byte			xxxx xxxx	27
190h	_	Unimpleme	nted							_	_
191h	_	Unimpleme	nted							_	_
192h	_	Unimpleme	nted							_	_
193h	_	Unimpleme	nted							_	_
194h	_	Unimpleme	nted							_	_
195h	_	Unimpleme	nted							_	_
196h	_	Unimpleme	nted							_	_
197h	1	Unimpleme	nted							_	_
198h		Unimpleme	nted							_	_
199h	_	Unimpleme	nted							_	_
19Ah	1	Unimpleme	nted							_	_
19Bh		Unimpleme	nted								_
19Ch		Unimpleme	nted								_
19Dh	_	Unimpleme	nted								_
19Eh	_	Unimpleme	nted							_	_
19Fh		Unimpleme	nted								_

 $\begin{array}{ll} \mbox{Legend:} & x = \mbox{u-known}, \ u = \mbox{u-known}, \ q = \mbox{value depends on condition, -= unimplemented, read as '0'.} \\ \mbox{Note 1:} & \mbox{These pixels do not display, but can be used as general purpose RAM.} \\ \end{array}$

2.3.1 STATUS REGISTER

The STATUS register, shown in Register 2-1, contains the arithmetic status of the ALU, the RESET status and the bank select bits for data memory.

The STATUS register can be the destination for any instruction, as with any other register. If the STATUS register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the $\overline{\text{TO}}$ and $\overline{\text{PD}}$ bits are not writable. Therefore, the result of an instruction with the STATUS register as destination may be different than intended.

For example, CLRF STATUS will clear the upper-three bits and set the Z bit. This leaves the STATUS register as 000u uluu (where u = unchanged).

It is recommended, therefore, that only BCF, BSF, SWAPF and MOVWF instructions are used to alter the STATUS register because these instructions do not affect the Z, C or DC bits from the STATUS register. For other instructions, not affecting any status bits, see the "Instruction Set Summary."

Note: The C and DC bits operate as a borrow and digit borrow bit, respectively, in subtraction. See the SUBLW and SUBWF instructions for examples.

REGISTER 2-1: STATUS REGISTER (ADDRESS 03h, 83h, 103h, 183h)

R/W-0	R/W-0	R/W-0	R-1	R-1	R/W-x	R/W-x	R/W-x
IRP	RP1	RP0	TO	PD	Z	DC	С
hit 7							hit 0

bit 7 IRP: Register Bank Select bit (used for indirect addressing)

1 = Bank 2, 3 (100h - 1FFh) 0 = Bank 0, 1 (00h - FFh)

bit 6-5 RP1:RP0: Register Bank Select bits (used for direct addressing)

11 = Bank 3 (180h - 1FFh)

10 = Bank 2 (100h - 17Fh)

01 = Bank 1 (80h - FFh)

00 = Bank 0 (00h - 7Fh)

bit 4 **TO**: Time-out bit

1 = After power-up, CLRWDT instruction, or SLEEP instruction

0 = A WDT time-out occurred

bit 3 **PD**: Power-down bit

1 = After power-up or by the CLRWDT instruction

0 = By execution of the SLEEP instruction

bit 2 Z: Zero bit

1 = The result of an arithmetic or logic operation is zero

0 = The result of an arithmetic or logic operation is not zero

bit 1 DC: Digit carry/borrow bit (ADDWF, ADDLW, SUBLW, SUBWF instructions)

(for borrow the polarity is reversed)

1 = A carry-out from the 4th low order bit of the result occurred

0 = No carry-out from the 4th low order bit of the result

bit 0 C: Carry/borrow bit (ADDWF, ADDLW, SUBLW, SUBWF instructions)

(for borrow the polarity is reversed)

1 = A carry-out from the Most Significant bit of the result occurred

0 = No carry-out from the Most Significant bit of the result occurred

Note: A subtraction is executed by adding the two's complement of the second operand. For rotate (RRF, RLF) instructions, this bit is loaded with either the high or low order bit of the source register.

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

n = Value at POR

'1' = Bit is set

'0' = Bit is cleared x = Bit is unknown

2.3.2 OPTION REGISTER

The OPTION register is a readable and writable register, which contains various control bits to configure the TMR0/WDT prescaler, the external RB0/INT pin interrupt, TMR0, and the weak pull-ups on PORTB.

Note: To achieve a 1:1 prescaler assignment for the TMR0 register, assign the prescaler to the Watchdog Timer.

REGISTER 2-2: OPTION REGISTER (ADDRESS 81h, 181h)

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
RBPU	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0
bit 7							bit 0

bit 7 RBPU: PORTB Pull-up Enable bit

1 = PORTB pull-ups are disabled

0 = PORTB pull-ups are enabled by individual port latch values

bit 6 INTEDG: Interrupt Edge Select bit

1 = Interrupt on rising edge of RB0/INT pin0 = Interrupt on falling edge of RB0/INT pin

bit 5 TOCS: TMR0 Clock Source Select bit

1 = Transition on RA4/T0CKI pin

0 = Internal instruction cycle clock (CLKOUT)

bit 4 **T0SE**: TMR0 Source Edge Select bit

1 = Increment on high-to-low transition on RA4/T0CKI pin

0 = Increment on low-to-high transition on RA4/T0CKI pin

bit 3 **PSA**: Prescaler Assignment bit

1 = Prescaler is assigned to the WDT

0 = Prescaler is assigned to the Timer0 module

bit 2-0 **PS2:PS0**: Prescaler Rate Select bits

Bit Value	TMR0 Rate	WDT Rate
000	1:2	1:1
001	1:4	1:2
010	1:8	1:4
011	1:16	1:8
100	1:32	1 : 16
101	1:64	1:32
110	1:128	1 : 64
111	1:256	1 : 128

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented	bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

2.3.3 INTCON REGISTER

The INTCON Register is a readable and writable register which contains various enable and flag bits for the TMR0 register overflow, RB Port change and external RB0/INT pin interrupts.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the global enable bit, GIE (INTCON<7>).

REGISTER 2-3: INTCON REGISTER (ADDRESS 0Bh, 8Bh, 10Bh, 18Bh)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-x
GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF
bit 7							bit 0

bit 7 GIE: Global Interrupt Enable bit

1 = Enables all unmasked interrupts

0 = Disables all interrupts

bit 6 **PEIE/GEIL:** Peripheral Interrupt Enable bit

1 = Enables all unmasked peripheral interrupts

0 = Disables all peripheral interrupts

bit 5 **TMR0IE:** TMR0 Overflow Interrupt Enable bit

1 = Enables the TMR0 overflow interrupt

0 = Disables the TMR0 overflow interrupt

bit 4 INTE: RB0/INT0 External Interrupt Enable bit

1 = Enables the RB0/INT external interrupt

0 = Disables the RB0/INT external interrupt

bit 3 RBIE: RB Port Change Interrupt Enable bit

1 = Enables the RB port change interrupt

0 = Disables the RB port change interrupt

bit 2 TMR0IF: TMR0 Overflow Interrupt Flag bit

1 = TMR0 register has overflowed (must be cleared in software)

0 = TMR0 register did not overflow

bit 1 INTF: RB0/INT0 External Interrupt Flag bit

1 = The RB0/INT external interrupt occurred (must be cleared in software)

0 = The RB0/INT external interrupt did not occur

bit 0 RBIF: RB Port Change Interrupt Flag bit

1 = At least one of the RB7:RB4 pins changed state (must be cleared in software)

0 = None of the RB7:RB4 pins have changed state

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'

- n = Value at POR reset '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

2.3.4 PIE1 REGISTER

This register contains the individual enable bits for the peripheral interrupts.

Note: Bit PEIE (INTCON<6>) must be set to enable any peripheral interrupt.

REGISTER 2-4: PIE1 REGISTER (ADDRESS 8Ch)

R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	
LCDIE	ADIE	_	_	SSPIE	CCP1IE	TMR2IE	TMR1IE	l
bit 7							bit 0	•

bit 7 LCDIE: LCD Interrupt Enable bit

1 = Enables the LCD interrupt
0 = Disables the LCD interrupt

bit 6 ADIE: A/D Converter Interrupt Enable bit

1 = Enables the A/D interrupt 0 = Disables the A/D interrupt Unimplemented: Read as '0'

bit 5-4 Unimplemented: Read as '0'

bit 3 **SSPIE**: Synchronous Serial Port Interrupt Enable bit 1 = Enables the SSP interrupt

0 = Disables the SSP interrupt

bit 2 **CCP1IE**: CCP1 Interrupt Enable bit

1 = Enables the CCP1 interrupt0 = Disables the CCP1 interrupt

bit 1 TMR2IE: TMR2 to PR2 Match Interrupt Enable bit

1 = Enables the TMR2 to PR2 match interrupt

0 = Disables the TMR2 to PR2 match interrupt

bit 0 TMR1IE: TMR1 Overflow Interrupt Enable bit

1 = Enables the TMR1 overflow interrupt

0 = Disables the TMR1 overflow interrupt

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' - n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

2.3.5 PIR1 REGISTER

This register contains the individual flag bits for the peripheral interrupts.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the global enable bit, GIE (INTCON<7>). User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

REGISTER 2-5: PIR1 REGISTER (ADDRESS 0Ch)

R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0
LCDIF	ADIF	_	_	SSPIF	CCP1IF	TMR2IF	TMR1IF
bit 7							bit 0

bit 7 LCDIF: LCD Interrupt Flag bit

1 = LCD interrupt has occurred (must be cleared in software)

0 = LCD interrupt did not occur

bit 6 ADIF: A/D Converter Interrupt Flag bit

1 = An A/D conversion completed (must be cleared in software)

0 = The A/D conversion is not complete

bit 5-4 Unimplemented: Read as '0'

bit 3 SSPIF: Master Synchronous Serial Port Interrupt Flag bit

1 = The transmission/reception is complete (must be cleared in software)

0 = Waiting to transmit/receive

bit 2 **CCP1IF**: CCP1 Interrupt Flag bit

Capture mode:

1 = A TMR1 register capture occurred (must be cleared in software)

0 = No TMR1 register capture occurred

Compare mode:

1 = A TMR1 register compare match occurred (must be cleared in software)

0 = No TMR1 register compare match occurred

PWM mode:

Unused in this mode

bit 1 TMR2IF: TMR2 to PR2 Match Interrupt Flag bit

1 = TMR2 to PR2 match occurred (must be cleared in software)

0 = No TMR2 to PR2 match occurred

bit 0 TMR1IF: TMR1 Overflow Interrupt Flag bit

1 = TMR1 register overflowed (must be cleared in software)

0 = TMR1 register did not overflow

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'

- n = Value at POR reset '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

2.3.6 PCON REGISTER

The Power Control (PCON) register contains a flag bit to allow differentiation between a Power-on Reset (POR) to an external MCLR Reset or WDT Reset.

For various RESET conditions, see Table 12-4 and Table 12-5.

REGISTER 2-6: PCON REGISTER (ADDRESS 8Eh)

U-0	U-0	U-0	U-0	U-0	U-0	R/W-0	R/W-1
_	_	_	_	_	_	POR	BOR
bit 7							bit 0

bit 7-2 Unimplemented: Read as '0'
bit 1 POR: Power-on Reset Status bit

1 = No Power-on Reset occurred

0 = A Power-on Reset occurred (must be set in software after a Power-on Reset occurs)

bit 0 BOR: Brown-out Reset Status bit

1 = No Brown-out Reset occurred

0 = A Brown-out Reset occurred (must be set in software after a Brown-out Reset occurs)

2.4 PCL and PCLATH

The program counter (PC) is 13-bits wide. The low byte comes from the PCL register, which is a readable and writable register. The upper bits (PC<12:8>) are not readable, but are indirectly writable through the PCLATH register. On any RESET, the upper bits of the PC will be cleared. Figure 2-5 shows the two situations for the loading of the PC. The upper example in the figure shows how the PC is loaded on a write to PCL (PCLATH<4:0> \rightarrow PCH). The lower example in the figure shows how the PC is loaded during a CALL or GOTO instruction (PCLATH<4:3> \rightarrow PCH).

FIGURE 2-5: LOADING OF PC IN DIFFERENT SITUATIONS

2.5 Program Memory Paging

PIC16C925/926 devices are capable of addressing a continuous 8K word block of program memory. The CALL and GOTO instructions provide only 11-bits of address to allow branching within any 2K program memory page. When doing a CALL or GOTO instruction, the upper 2-bits of the address are provided by PCLATH<4:3>. When doing a CALL or GOTO instruction, the user must ensure that the page select bits are programmed so that the desired program memory page is addressed. If a return from a CALL instruction (or interrupt) is executed, the entire 13-bit PC is pushed onto the stack. Therefore, manipulation of the PCLATH<4:3> bits is not required for the RETURN instructions (which POPs the address from the stack).

Example 2-1 shows the calling of a subroutine in

2.4.1 COMPUTED GOTO

A computed GOTO is accomplished by adding an offset to the program counter (ADDWF PCL). When doing a table read using a computed GOTO method, care should be exercised if the table location crosses a PCL memory boundary (each 256 byte block). Refer to the application note "Implementing a Table Read" (AN556).

2.4.2 STACK

The PIC16CXXX family has an 8-level deep x 13-bit wide hardware stack. The stack space is not part of either program or data space and the stack pointer is not readable or writable. The PC is PUSHed onto the stack when a CALL instruction is executed, or an interrupt causes a branch. The stack is POPed in the event of a RETURN, RETLW or a RETFIE instruction execution. PCLATH is not affected by a PUSH or POP operation.

The stack operates as a circular buffer. This means that after the stack has been PUSHed eight times, the ninth push overwrites the value that was stored from the first push. The tenth push overwrites the second push (and so on).

2.6 Indirect Addressing, INDF and FSR Registers

The INDF register is not a physical register. Addressing the INDF register will cause indirect addressing.

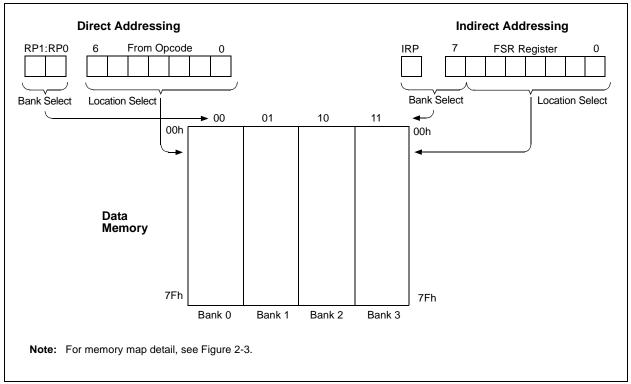
Indirect addressing is possible by using the INDF register. Any instruction using the INDF register actually accesses the register pointed to by the File Select Register (FSR). Reading the INDF register itself, indirectly (FSR = '0'), will produce 00h. Writing to the INDF register indirectly results in a no operation (although status bits may be affected). An effective 9-bit address is obtained by concatenating the 8-bit FSR register and the IRP bit (STATUS<7>), as shown in Figure 2-6.

A simple program to clear RAM locations 20h-2Fh using indirect addressing is shown in Example 2-2.

EXAMPLE 2-2: INDIRECT ADDRESSING

	MOVLW	0x20	;initialize pointer
	MOVWF	FSR	;to RAM
NEXT	CLRF	INDF	;clear INDF register
	INCF	FSR,F	;inc pointer
	BTFSS	FSR,4	;all done?
	GOTO	NEXT	;no clear next
CONTINUE			
			; yes continue
	•		, yes concinue

FIGURE 2-6: DIRECT/INDIRECT ADDRESSING



3.0 READING PROGRAM MEMORY

The Program Memory is readable during normal operation over the entire VDD range. It is indirectly addressed through Special Function Registers (SFR). Up to 14-bit numbers can be stored in memory for use as calibration parameters, serial numbers, packed 7-bit ASCII, etc. Executing a program memory location containing data that forms an invalid instruction results in a NOP.

There are five SFRs used to read the program and memory. These registers are:

- PMCON1
- PMDATA
- PMDATH
- PMADR
- PMADRH

The program memory allows word reads. Program memory access allows for checksum calculation and reading calibration tables.

When interfacing to the program memory block, the PMDATH:PMDATA registers form a two-byte word, which holds the 14-bit data for reads. The PMADRH:PMADR registers form a two-byte word, which holds the 13-bit address of the location being accessed. These devices can have from 4K words to 8K words of program memory, with an address range from 0h to 3FFFh.

The unused upper bits in both the PMDATH and PMADRH registers are not implemented and read as "0's".

3.1 PMADR

The address registers can address up to a maximum of 8K words of program memory.

When selecting a program address value, the MSByte of the address is written to the PMADRH register and the LSByte is written to the PMADR register. The upper MSbits of PMADRH must always be clear.

3.2 PMCON1 Register

PMCON1 is the control register for memory accesses.

The control bit RD initiates read operations. This bit cannot be cleared, only set, in software. It is cleared in hardware at the completion of the read operation.

REGISTER 3-1: PMCON1 REGISTER (ADDRESS 10Ch)

R-1	U-0	U-0	U-0	U-x	U-0	U-0	R/S-0
r	_	_	_	_	_	_	RD
bit 7							bit 0

bit 7 Reserved: Read as '1'

bit 6-1 **Unimplemented:** Read as '0'

bit 0 RD: Read Control bit

1 = Initiates a read, RD is cleared in hardware. The RD bit can only be set (not cleared) in software.

0 = Does not initiate a read

3.3 Reading the Program Memory

A program memory location may be read by writing two bytes of the address to the PMADR and PMADRH registers, and then setting control bit RD (PMCON1<0>). Once the read control bit is set, the microcontroller will use the next two instruction cycles to read the data. The

data is available in the PMDATA and PMDATH registers after the \mathtt{NOP} instruction. Therefore, it can be read as two bytes in the following instructions. The PMDATA and PMDATH registers will hold this value until another read operation.

EXAMPLE 3-1: PROGRAM READ

```
BSF
        STATUS, RP1
                                ;
        STATUS, RP0
                                ; Bank 3
BSF
        MS PROG PM ADDR
MOVLW
MOVWF
        PMADRH
                                ; MS Byte of Program Address to read
MOVLW
        LS_PROG_PM_ADDR
MOVWF
        PMADR
                                ; LS Byte of Program Address to read
        STATUS, RP0
BCF
                                ; Bank 2
BSF
        PMCON1, RD
                                ; PM Read
                                ; First instruction after BSF PMCON1, RD executes normally
BSF
        STATUS, RP0
                                ; Bank 3
NOP
                                ; Any instructions here are ignored as program
                                 ; memory is read in second cycle after BSF PMCON1,RD
MOVF
        PMDATA, W
                                ; W = LS Byte of Program PMDATA
                                ; W = MS Byte of Program PMDATA
        PMDATH, W
MOVE
```

3.4 Operation During Code Protect

If the program memory is not code protected, the program memory control can read anywhere within the program memory.

If the entire program memory is code protected, the program memory control can read anywhere within the program memory.

If only part of the program memory is code protected, the program memory control can read the unprotected segment and cannot read the protected segment. The protected area cannot be read, because it may be possible to write a downloading routine into the unprotected segment.

TABLE 3-1: REGISTERS ASSOCIATED WITH PROGRAM MEMORY

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other RESETS
10Ch	PMCON1	(1)	_	_	_	_	_	_	RD	10	10
18Ch	PMDATA	Data Regist	Pata Register Low Byte							xxxx xxxx	uuuu uuuu
18Dh	PMADR	Address Re	gister Lov	v Byte						xxxx xxxx	uuuu uuuu
18Eh	PMDATH	_	_	Data Reg	Data Register High Byte					xxxx xxxx	uuuu uuuu
18Fh	PMADRH	_	1		Address Register High Byte					xxxx xxxx	uuuu uuuu

Legend: x = unknown, u = unchanged, r = reserved, - = unimplemented, read as '0'. Shaded cells are not used during FLASH access.

Note 1: This bit always reads as a '1'.

4.0 I/O PORTS

Some pins for these ports are multiplexed with an alternate function for the peripheral features on the device. In general, when a peripheral is enabled, that pin may not be used as a general purpose I/O pin.

4.1 PORTA and TRISA Register

The RA4/T0CKI pin is a Schmitt Trigger input and an open drain output. All other RA port pins have TTL input levels and full CMOS output drivers. All RA pins have data direction bits (TRISA register), which can configure these pins as output or input.

Setting a bit in the TRISA register puts the corresponding output driver in a Hi-Impedance mode. Clearing a bit in the TRISA register puts the contents of the output latch on the selected pin.

Reading the PORTA register reads the status of the pins, whereas writing to it will write to the port latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified, and then written to the port data latch.

Pin RA4 is multiplexed with the Timer0 module clock input to become the RA4/T0CKI pin. The other PORTA pins are multiplexed with analog inputs and the analog VREF input. The operation of each pin is selected by clearing/setting the control bits in the ADCON1 register (A/D Control Register1).

Note: On a Power-on Reset, these pins are configured as analog inputs and read as '0'.

The TRISA register controls the direction of the RA pins, even when they are being used as analog inputs. The user must ensure the bits in the TRISA register are maintained set when using them as analog inputs.

EXAMPLE 4-1: INITIALIZING PORTA

FIGURE 4-1: BLOCK DIAGRAM OF PINS RA3:RA0 AND RA5

FIGURE 4-2: BLOCK DIAGRAM OF RA4/T0CKI PIN

TABLE 4-1: PORTA FUNCTIONS

Name	Bit#	Buffer	Function
RA0/AN0	bit0	TTL	Input/output or analog input.
RA1/AN1	bit1	TTL	Input/output or analog input.
RA2/AN2	bit2	TTL	Input/output or analog input.
RA3/AN3/VREF	bit3	TTL	Input/output or analog input or VREF.
RA4/T0CKI	bit4	ST	Input/output or external clock input for Timer0. Output is open drain type.
RA5/AN4/SS	bit5	TTL	Input/output or analog input or slave select input for synchronous serial port.

Legend: TTL = TTL input, ST = Schmitt Trigger input

TABLE 4-2: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other RESETS
05h	PORTA	_	_	RA5	RA4	RA3	RA2	RA1	RA0	0x 0000	0x 0000
85h	TRISA	_	_	PORTA D	PORTA Data Direction Control Register						11 1111
9Fh	ADCON1	_	_	_	1	-	PCFG2	PCFG1	PCFG0	000	000

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTA.

4.2 **PORTB and TRISB Register**

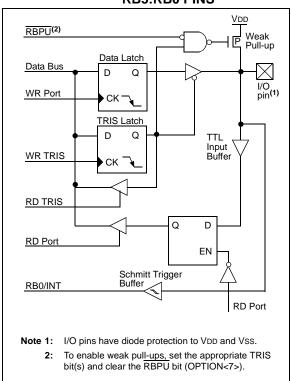
PORTB is an 8-bit wide, bi-directional port. The corresponding data direction register is TRISB. Setting a bit in the TRISB register puts the corresponding output driver in a Hi-Impedance Input mode. Clearing a bit in the TRISB register puts the contents of the output latch on the selected pin(s).

EXAMPLE 4-2: INITIALIZING PORTB

BCF	STATUS, RP0	; Select Bank0
BCF	STATUS, RP1	
CLRF	PORTB	; Initialize PORTB
BSF	STATUS, RP0	; Select Bank1
MOVLW	0xCF	; Value used to
		; initialize data
		; direction
MOVWF	TRISB	; Set RB<3:0> as inputs
		; RB<5:4> as outputs
		; RB<7:6> as inputs

Each of the PORTB pins has a weak internal pull-up. A single control bit can turn on all the pull-ups. This is performed by clearing bit RBPU (OPTION<7>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are also disabled on a Power-on Reset.

FIGURE 4-3: BLOCK DIAGRAM OF RB3:RB0 PINS



Four of the PORTB pins (RB7:RB4) have an interrupt-on-change feature. Only pins configured as inputs can cause this interrupt to occur (i.e., any RB7:RB4 pin configured as an output is excluded from the interrupt-on-change comparison). The input pins (of RB7:RB4) are compared with the old value latched on the last read of PORTB. The "mismatch" outputs of RB7:RB4 are OR'ed together to generate the RB Port Change Interrupt with flag bit RBIF (INTCON<0>).

This interrupt can wake the device from SLEEP. The user, in the Interrupt Service Routine, can clear the interrupt in the following manner:

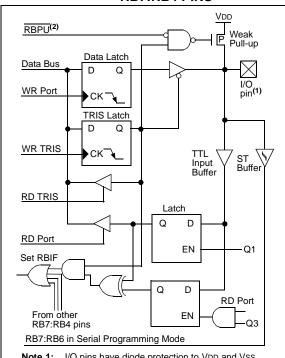
- Any read or write of PORTB. This will end the mismatch condition.
- Clear flag bit RBIF.

A mismatch condition will continue to set flag bit RBIF. Reading PORTB will end the mismatch condition, and allow flag bit RBIF to be cleared.

This interrupt-on-mismatch feature, together with software configurable pull-ups on these four pins, allow easy interface to a keypad and make it possible for wake-up on key depression. Refer to the Embedded Control Handbook, "Implementing Wake-Up on Key Stroke" (AN552).

The interrupt-on-change feature is recommended for wake-up on key depression operation and operations where PORTB is only used for the interrupt-on-change feature. Polling of PORTB is not recommended while using the interrupt-on-change feature.

BLOCK DIAGRAM OF FIGURE 4-4: RB7:RB4 PINS



Note 1: I/O pins have diode protection to VDD and Vss.

To enable weak pull-ups, set the appropriate TRIS bit(s) and clear the RBPU bit (OPTION<7>).

TABLE 4-3: PORTB FUNCTIONS

Name	Bit#	Buffer	Function
RB0/INT	bit0	TTL/ST	Input/output pin or external interrupt input. Internal software programmable weak pull-up. This buffer is a Schmitt Trigger input when configured as the external interrupt.
RB1	bit1	TTL	Input/output pin. Internal software programmable weak pull-up.
RB2	bit2	TTL	Input/output pin. Internal software programmable weak pull-up.
RB3	bit3	TTL	Input/output pin. Internal software programmable weak pull-up.
RB4	bit4	TTL	Input/output pin (with interrupt-on-change). Internal software programmable weak pull-up.
RB5	bit5	TTL	Input/output pin (with interrupt-on-change). Internal software programmable weak pull-up.
RB6	bit6	TTL/ST	Input/output pin (with interrupt-on-change). Internal software programmable weak pull-up. Serial programming clock. This buffer is a Schmitt Trigger input when used in Serial Programming mode.
RB7	bit7	TTL/ST	Input/output pin (with interrupt-on-change). Internal software programmable weak pull-up. Serial programming data. This buffer is a Schmitt Trigger input when used in Serial Programming mode.

Legend: TTL = TTL input, ST = Schmitt Trigger input

TABLE 4-4: SUMMARY OF REGISTERS ASSOCIATED WITH PORTB

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other RESETS
06h, 106h	PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	xxxx xxxx	uuuu uuuu
86h, 186h	TRISB	PORTB	PORTB Data Direction Control Register								1111 1111
81h, 181h	OPTION	RBPU	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111

Legend: x = unknown, u = unchanged. Shaded cells are not used by PORTB.

4.3 **PORTC and TRISC Register**

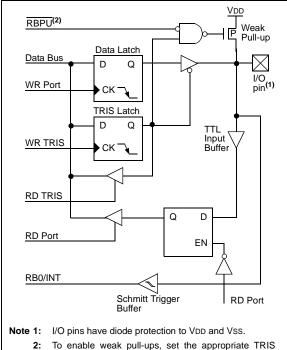
PORTC is a 6-bit, bi-directional port. Each pin is individually configurable as an input or output through the TRISC register. PORTC is multiplexed with several peripheral functions (Table 4-5). PORTC pins have Schmitt Trigger input buffers.

When enabling peripheral functions, care should be taken in defining TRIS bits for each PORTC pin. Some peripherals override the TRIS bit to make a pin an output, while other peripherals override the TRIS bit to make a pin an input. Since the TRIS bit override is in effect while the peripheral is enabled, readmodify-write instructions (BSF, BCF, XORWF) with TRISC as destination should be avoided. The user should refer to the corresponding peripheral section for the correct TRIS bit settings.

EXAMPLE 4-3: INITIALIZING PORTC

-/\/\		•••	WITH VEHENICO I OIVIO
BCF	STATUS, RP0	;	Select Bank0
BCF	STATUS, RP1		
CLRF	PORTC	;	Initialize PORTC
BSF	STATUS, RP0	;	Select Bank1
MOVLW	0xCF	;	Value used to
		;	initialize data
		;	direction
MOVWF	TRISC	;	Set RC<3:0> as inputs
		;	RC<5:4> as outputs
		;	RC<7:6> always read 0

FIGURE 4-5: PORTC BLOCK DIAGRAM (PERIPHERAL OUTPUT **OVERRIDE)**



To enable weak pull-ups, set the appropriate TRIS bit(s) and clear the \overline{RBPU} bit (OPTION<7>).

TABLE 4-5: PORTC FUNCTIONS

Name	Bit#	Buffer Type	Function
RC0/T1OSO/T1CKI	bit0	ST	Input/output port pin or Timer1 oscillator output or Timer1 clock input.
RC1/T1OSI	bit1	ST	Input/output port pin or Timer1 oscillator input.
RC2/CCP1	bit2	ST	Input/output port pin or Capture input/Compare output/PWM output.
RC3/SCK/SCL	bit3	ST	Input/output port pin or the synchronous serial clock for both SPI and I ² C modes.
RC4/SDI/SDA	bit4	ST	Input/output port pin or the SPI Data In (SPI mode) or data I/O (I ² C mode).
RC5/SDO	bit5	ST	Input/output port pin or Synchronous Serial Port data out.

Legend: ST = Schmitt Trigger input

TABLE 4-6: SUMMARY OF REGISTERS ASSOCIATED WITH PORTC

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other RESETS
07h	PORTC	_	_	RC5	RC4	RC3	RC2	RC1	RC0	xx xxxx	uu uuuu
87h	TRISC		1	PORTC I	Data Direc	tion Contr	11 1111	11 1111			

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by PORTC.



TABLE 4-7: PORTD FUNCTIONS

Name	Bit#	Buffer Type	Function
RD0/SEG00	bit0	ST	Input/output port pin or Segment Driver00.
RD1/SEG01	bit1	ST	Input/output port pin or Segment Driver01.
RD2/SEG02	bit2	ST	Input/output port pin or Segment Driver02.
RD3/SEG03	bit3	ST	Input/output port pin or Segment Driver03.
RD4/SEG04	bit4	ST	Input/output port pin or Segment Driver04.
RD5/SEG29/COM3	bit5	ST	Digital input pin or Segment Driver29 or Common Driver3.
RD6/SEG30/COM2	bit6	ST	Digital input pin or Segment Driver30 or Common Driver2.
RD7/SEG31/COM1	bit7	ST	Digital input pin or Segment Driver31 or Common Driver1.

Legend: ST = Schmitt Trigger input

TABLE 4-8: SUMMARY OF REGISTERS ASSOCIATED WITH PORTD

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other RESETS
08h	PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0	0000 0000	0000 0000
88h	TRISD	PORTD [PORTD Data Direction Control Register								1111 1111
10Dh	LCDSE	SE29	SE27	SE20	SE16	SE12	SE9	SE5	SE0	1111 1111	1111 1111

Legend: Shaded cells are not used by PORTD.

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4.5 PORTE and TRISE Register

PORTE is a digital input only port. Each pin is multiplexed with an LCD segment driver. These pins have Schmitt Trigger input buffers.

- **Note 1:** On a Power-on Reset, these pins are configured as LCD segment drivers.
 - 2: To configure the pins as a digital port, the corresponding bits in the LCDSE register must be cleared. Any bit set in the LCDSE register overrides any bit settings in the corresponding TRIS register.

EXAMPLE 4-5: INITIALIZING PORTE

BCF S	STATUS,	RP0	;Select Bank2
BSF S	STATUS,	RP1	;
BCF I	LCDSE,	SE27	;Make all PORTE
BCF I	LCDSE,	SE5	;and PORTG<7>
BCF I	LCDSE,	SE9	digital inputs;

FIGURE 17-1: PORTE BLOCK DIAGRAM

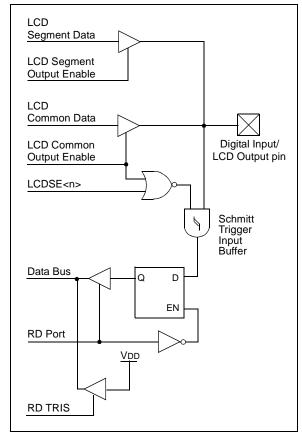


TABLE 4-9: PORTE FUNCTIONS

Name	Bit#	Buffer Type	Function
RE0/SEG05	bit0	ST	Digital input or Segment Driver05.
RE1/SEG06	bit1	ST	Digital input or Segment Driver06.
RE2/SEG07	bit2	ST	Digital input or Segment Driver07.
RE3/SEG08	bit3	ST	Digital input or Segment Driver08.
RE4/SEG09	bit4	ST	Digital input or Segment Driver09.
RE5/SEG10	bit5	ST	Digital input or Segment Driver10.
RE6/SEG11	bit6	ST	Digital input or Segment Driver11.
RE7/SEG27	bit7	ST	Digital input or Segment Driver27 (not available on 64-pin devices).

Legend: ST = Schmitt Trigger input

TABLE 4-10: SUMMARY OF REGISTERS ASSOCIATED WITH PORTE

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other RESETS
09h	PORTE	RE7	RE6	RE5	RE4	RE3	RE2	RE1	RE0	0000 0000	0000 0000
89h	TRISE	PORTE [PORTE Data Direction Control Register								1111 1111
10Dh	LCDSE	SE29	SE27	SE20	SE16	SE12	SE9	SE5	SE0	1111 1111	1111 1111

Legend: Shaded cells are not used by PORTE.

4.6 PORTF and TRISF Register

PORTF is a digital input only port. Each pin is multiplexed with an LCD segment driver. These pins have Schmitt Trigger input buffers.

- **Note 1:** On a Power-on Reset, these pins are configured as LCD segment drivers.
 - 2: To configure the pins as a digital port, the corresponding bits in the LCDSE register must be cleared. Any bit set in the LCDSE register overrides any bit settings in the corresponding TRIS register.

EXAMPLE 4-6: INITIALIZING PORTF

BCF STATUS,	RP0	;Select Bank2
BSF STATUS,	RP1	;
BCF LCDSE,	SE16	;Make all PORTF
BCF LCDSE,	SE12	digital inputs;

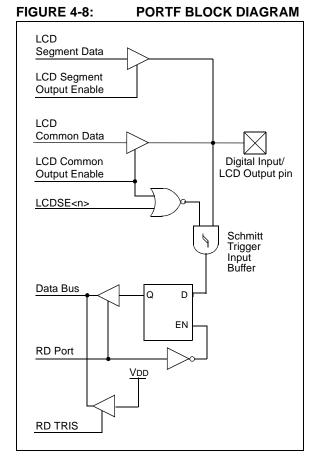


TABLE 4-11: PORTF FUNCTIONS

Name	Bit#	Buffer Type	Function
RF0/SEG12	bit0	ST	Digital input or Segment Driver12.
RF1/SEG13	bit1	ST	Digital input or Segment Driver13.
RF2/SEG14	bit2	ST	Digital input or Segment Driver14.
RF3/SEG15	bit3	ST	Digital input or Segment Driver15.
RF4/SEG16	bit4	ST	Digital input or Segment Driver16.
RF5/SEG17	bit5	ST	Digital input or Segment Driver17.
RF6/SEG18	bit6	ST	Digital input or Segment Driver18.
RF7/SEG19	bit7	ST	Digital input or Segment Driver19.

Legend: ST = Schmitt Trigger input

TABLE 4-12: SUMMARY OF REGISTERS ASSOCIATED WITH PORTF

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other RESETS
107h	PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	RF0	0000 0000	0000 0000
187h	TRISF	PORTF [PORTF Data Direction Control Register							1111 1111	1111 1111
10Dh	LCDSE	SE29	SE27	SE20	SE16	SE12	SE9	SE5	SE0	1111 1111	1111 1111

Legend: Shaded cells are not used by PORTF.

4.7 PORTG and TRISG Register

PORTG is a digital input only port. Each pin is multiplexed with an LCD segment driver. These pins have Schmitt Trigger input buffers.

- **Note 1:** On a Power-on Reset, these pins are configured as LCD segment drivers.
 - 2: To configure the pins as a digital port, the corresponding bits in the LCDSE register must be cleared. Any bit set in the LCDSE register overrides any bit settings in the corresponding TRIS register.

EXAMPLE 4-7: INITIALIZING PORTG

BCF	STATUS,	RP0	;Select Bank2
BSF	STATUS,	RP1	;
BCF	LCDSE,	SE27	;Make all PORTG
BCF	LCDSE,	SE20	;and PORTE<7>
			;digital inputs

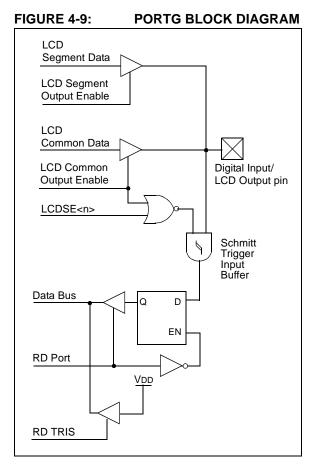


TABLE 4-13: PORTG FUNCTIONS

Name	Bit#	Buffer Type	Function
RG0/SEG20	bit0	ST	Digital input or Segment Driver20.
RG1/SEG21	bit1	ST	Digital input or Segment Driver21.
RG2/SEG22	bit2	ST	Digital input or Segment Driver22.
RG3/SEG23	bit3	ST	Digital input or Segment Driver23.
RG4/SEG24	bit4	ST	Digital input or Segment Driver24.
RG5/SEG25	bit5	ST	Digital input or Segment Driver25.
RG6/SEG26	bit6	ST	Digital input or Segment Driver26.
RG7/SEG28	bit7	ST	Digital input or Segment Driver28 (not available on 64-pin devices).

Legend: ST = Schmitt Trigger input

TABLE 4-14: SUMMARY OF REGISTERS ASSOCIATED WITH PORTG

4.8 I/O Programming Considerations

4.8.1 BI-DIRECTIONAL I/O PORTS

Any instruction which writes, operates internally as a read followed by a write operation. The BCF and BSF instructions, for example, read the register into the CPU, execute the bit operation and write the result back to the register. Caution must be used when these instructions are applied to a port with both inputs and outputs defined. For example, a BSF operation on bit5 of PORTB will cause all eight bits of PORTB to be read into the CPU. Then the BSF operation takes place on bit5 and PORTB is written to the output latches. If another bit of PORTB is used as a bi-directional I/O pin (e.g., bit0) and it is defined as an input at this time, the input signal present on the pin itself would be read into the CPU and rewritten to the data latch of this particular pin, overwriting the previous content. As long as the pin stays in the input mode, no problem occurs. However, if bit0 is switched into output mode later on, the contents of the data latch may now be unknown.

Reading the port register reads the values of the port pins. Writing to the port register, writes the value to the port latch. When using read-modify-write instructions (e.g. BCF, BSF) on a port, the value of the port pins is read, the desired operation is done to this value, and this value is then written to the port latch.

Example 4-8 shows the effect of two sequential read-modify-write instructions on an I/O port. A pin actively outputting a Low or High should not be driven from external devices at the same time, in order to change the level on this pin ("wired-or", "wired-and"). The resulting high output currents may damage the chip.

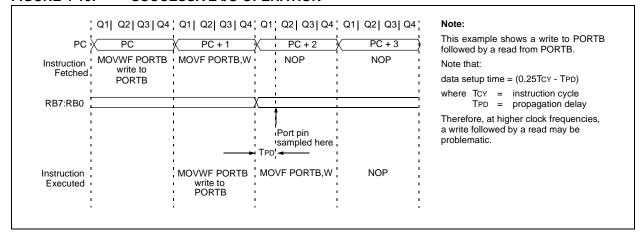
EXAMPLE 4-8: READ-MODIFY-WRITE INSTRUCTIONS ON AN I/O PORT

```
; Initial PORT settings: PORTB<7:4> Inputs
                         PORTB<3:0> Outputs
;PORTB<7:6> have external pull-ups and are
;not connected to other circuitry
                                 PORT pins
                    PORT latch
                  ; 01pp pppp
 BCF PORTB,
                                 11pp pppp
 BCF PORTB.
             6
                  ; 10pp pppp
                                 11pp pppp
 BCF STATUS, RP1 ; Select Bank1
 BSF STATUS, RPO ;
 BCF TRISB,
              7
                  ; 10pp pppp
                                  11pp pppp
 BCF TRISB,
              6
                  ; 10pp pppp
                                  10pp pppp
; Note that the user may have expected the
;pin values to be 00pp ppp. The 2nd BCF
; caused RB7 to be latched as the pin value
: (high).
```

4.8.2 SUCCESSIVE OPERATIONS ON I/O PORTS

The actual write to an I/O port happens at the end of an instruction cycle, whereas for reading, the data must be valid at the beginning of the instruction cycle (Figure 4-10). Therefore, care must be exercised if a write followed by a read operation is carried out on the same I/O port. The sequence of instructions should be such to allow the pin voltage to stabilize (load dependent) before the next instruction, which causes that file to be read into the CPU, is executed. Otherwise, the previous state of that pin may be read into the CPU, rather than the new state. When in doubt, it is better to separate these instructions with a NOP, or another instruction not accessing this I/O port.

FIGURE 4-10: SUCCESSIVE I/O OPERATION



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NOTES:

5.0 TIMERO MODULE

The Timer0 module has the following features:

- · 8-bit timer/counter
- · Readable and writable
- 8-bit software programmable prescaler
- · Internal or external clock select
- Interrupt-on-overflow from FFh to 00h
- · Edge select for external clock

Figure 5-1 is a simplified block diagram of the Timer0 module.

Timer mode is selected by clearing bit TOCS (OPTION<5>). In Timer mode, the Timer0 module will increment every instruction cycle (without prescaler). If the TMR0 register is written, the increment is inhibited for the following two instruction cycles (Figure 5-2 and Figure 5-3). The user can work around this by writing an adjusted value to the TMR0 register.

Counter mode is selected by setting bit TOCS (OPTION<5>). In Counter mode, Timer0 will increment either on every rising, or falling edge of pin RA4/TOCKI. The incrementing edge is determined by the Timer0 Source Edge Select bit TOSE (OPTION<4>). Clearing

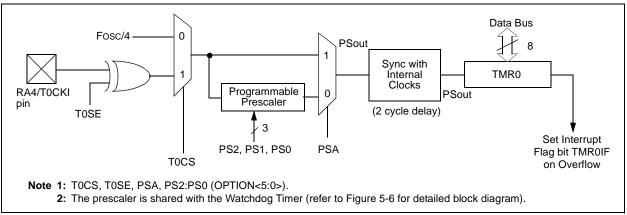
bit T0SE selects the rising edge. Restrictions on the external clock input are discussed in detail in Section 5.2.

The prescaler is mutually exclusively shared between the Timer0 module and the Watchdog Timer. The prescaler assignment is controlled in software by control bit PSA (OPTION<3>). Clearing bit PSA will assign the prescaler to the Timer0 module. The prescaler is not readable or writable. When the prescaler is assigned to the Timer0 module, prescale values of 1:2, 1:4,..., 1:256 are selectable. Section 5.3 details the operation of the prescaler.

5.1 Timer0 Interrupt

The TMR0 interrupt is generated when the TMR0 register overflows from FFh to 00h. This overflow sets bit TMR0IF (INTCON<2>). The interrupt can be masked by clearing bit T0IE (INTCON<5>). Bit TMR0IF must be cleared in software by the Timer0 module Interrupt Service Routine before re-enabling this interrupt. The TMR0 interrupt cannot awaken the processor from SLEEP, since the timer is shut-off during SLEEP. Figure 5-4 displays the Timer0 interrupt timing.

FIGURE 5-1: TIMERO BLOCK DIAGRAM



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FIGURE 5-2: TIMERO TIMING: INTERNAL CLOCK/NO PRESCALE

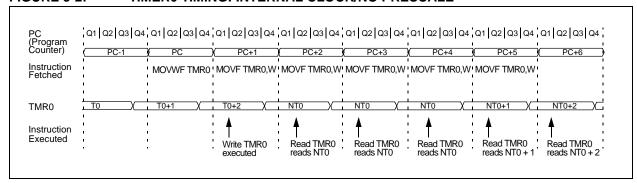


FIGURE 5-3: TIMERO TIMING: INTERNAL CLOCK/PRESCALE 1:2

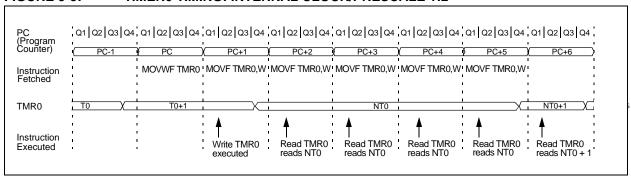
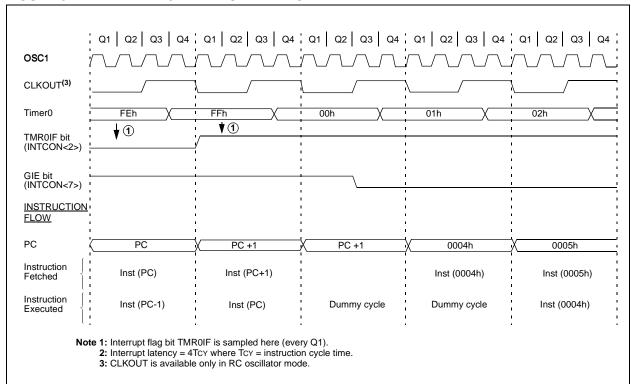


FIGURE 5-4: TIMERO INTERRUPT TIMING



5.2 Using Timer0 with an External Clock

When an external clock input is used for Timer0, it must meet certain requirements. The requirements ensure the external clock can be synchronized with the internal phase clock (Tosc). Also, there is a delay in the actual incrementing of Timer0 after synchronization.

5.2.1 EXTERNAL CLOCK SYNCHRONIZATION

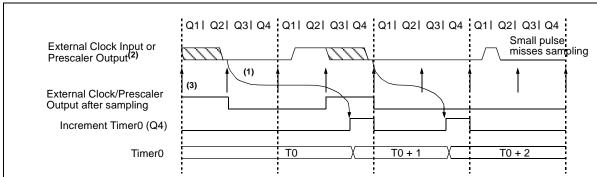
When no prescaler is used, the external clock input is the same as the prescaler output. The synchronization of T0CKI with the internal phase clocks is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks (Figure 5-5). Therefore, it is necessary for T0CKI to be high for at least 2Tosc (and a small RC delay of 20 ns) and low for at least 2Tosc (and a small RC delay of 20 ns). Refer to the electrical specification of the desired device.

When a prescaler is used, the external clock input is divided by the asynchronous ripple counter type prescaler, so that the prescaler output is symmetrical. For the external clock to meet the sampling requirement, the ripple counter must be taken into account. Therefore, it is necessary for TOCKI to have a period of at least 4Tosc (and a small RC delay of 40 ns) divided by the prescaler value. The only requirement on TOCKI high and low time is that they do not violate the minimum pulse width requirement of 10 ns. Refer to parameters 40, 41 and 42 in the electrical specification of the desired device.

5.2.2 TMR0 INCREMENT DELAY

Since the prescaler output is synchronized with the internal clocks, there is a small delay from the time the external clock edge occurs to the time the Timer0 module is actually incremented. Figure 5-5 shows the delay from the external clock edge to the timer incrementing.





- **Note 1:** Delay from clock input change to Timer0 increment is 3Tosc to 7Tosc. (Duration of Q = Tosc.) Therefore, the error in measuring the interval between two edges on Timer0 input = ±4Tosc max.
 - 2: External clock if no prescaler selected, prescaler output otherwise.
 - 3: The arrows indicate the points in time where sampling occurs.

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5.3 Prescaler

5.3.1 SWITCHING PRESCALER ASSIGNMENT

The prescaler assignment is fully under software control, i.e., it can be changed "on the fly" during program execution.

Note: To avoid an unintended device RESET, the following instruction sequence (shown in Example 5-1) must be executed when changing the prescaler assignment from Timer0 to the WDT. This precaution must be followed even if the WDT is disabled.

EXAMPLE 5-1: CHANGING PRESCALER (TIMER0→WDT)

```
STATUS, RPO
                              1)
                                   BSF
                                                          ;Select Bank1
Lines 2 and 3 do NOT have to
                              2)
                                  MOVLW b'xx0x0xxx'
                                                          ;Select clock source and prescale value of
be included if the final desired
                              3)
                                  MOVWF
                                         OPTION REG
                                                          ;other than 1:1
prescale value is other than 1:1.
                                  BCF
                                          STATUS, RPO
                                                          ;Select Bank0
If 1:1 is final desired value, then
                                                          ;Clear TMR0 and prescaler
                              5)
                                  CLRF
                                          TMR0
a temporary prescale value is
                                  BSF
                                          STATUS, RP1
                                                          ;Select Bank1
                              6)
set in lines 2 and 3 and the final
                              7)
                                  MOVLW b'xxxx1xxx'
                                                          ;Select WDT, do not change prescale value
prescale value will be set in lines
10 and 11.
                                  MOVWF OPTION REG
                              8)
                              9)
                                  CLRWDT
                                                          ;Clears WDT and prescaler
                              10) MOVLW b'xxxx1xxx'
                                                          ;Select new prescale value and WDT
                              11) MOVWF
                                         OPTION REG
                                          STATUS, RPO
                                                          ;Select Bank0
                              12) BCF
```

To change prescaler from the WDT to the Timer0 module use the precaution shown in Example 5-2.

EXAMPLE 5-2: CHANGING PRESCALER (WDT→TIMER0)

```
CLRWDT ;Clear WDT and precaler

BSF STATUS, RPO ;Select Bank1

MOVLW b'xxxx0xxx' ;Select TMRO,
 ;new prescale value and

MOVWF OPTION_REG ;clock source

BCF STATUS, RPO ;Select Bank0
```

TABLE 5-1: REGISTERS ASSOCIATED WITH TIMERO

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other RESETS
01h, 101h	TMR0	Timer0	Module Re		xxxx xxxx	uuuu uuuu					
0Bh, 8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000u
81h, 181h	OPTION	RBPU	INTEDG	INTEDG TOCS TOSE PSA PS2 PS1 PS0						1111 1111	1111 1111
85h	TRISA	_		PORTA D	PORTA Data Direction Control Register						11 1111

Legend: $\mathbf{x} =$ unknown, $\mathbf{u} =$ unchanged, $\mathbf{-} =$ unimplemented locations read as '0'. Shaded cells are not used by Timer0.

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NOTES:

6.0 TIMER1 MODULE

Timer1 is a 16-bit timer/counter consisting of two 8-bit registers (TMR1H and TMR1L), which are readable and writable. The TMR1 Register pair (TMR1H:TMR1L) increments from 0000h to FFFFh and rolls over to 0000h. The TMR1 Interrupt, if enabled, is generated on overflow, which is latched in interrupt flag bit, TMR1IF (PIR1<0>). This interrupt can be enabled/disabled by setting/clearing TMR1 interrupt enable bit, TMR1IE (PIE1<0>).

Timer1 can operate in one of two modes:

- · As a timer
- As a counter

The operating mode is determined by the clock select bit, TMR1CS (T1CON<1>).

In Timer mode, Timer1 increments every instruction cycle. In Counter mode, it increments on every rising edge of the external clock input.

Timer1 can be turned on and off using the control bit TMR1ON (T1CON<0>).

Timer1 also has an internal "RESET input". This RESET can be generated by the CCP module (Section 8.0). Register 6-1 shows the Timer1 control register.

When the Timer1 oscillator is enabled (T1OSCEN is set), the RC1/T1OSI and RC0/T1OSO/T1CKI pins become inputs, regardless of the TRISC<1:0>. RC1 and RC0 will be read as '0'.

REGISTER 6-1: T1CON: TIMER1 CONTROL REGISTER (ADDRESS 10h)

U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
_	_	T1CKPS1	T1CKPS0	T10SCEN	T1SYNC	TMR1CS	TMR10N
bit 7							bit 0

bit 7-6 Unimplemented: Read as '0'

bit 5-4 T1CKPS1:T1CKPS0: Timer1 Input Clock Prescale Select bits

11 = 1:8 Prescale value 10 = 1:4 Prescale value 01 = 1:2 Prescale value 00 = 1:1 Prescale value

bit 3 T10SCEN: Timer1 Oscillator Enable Control bit

1 = Oscillator is enabled0 = Oscillator is shut-off

Note: The oscillator inverter and feedback resistor are turned off to eliminate power drain.

bit 2 T1SYNC: Timer1 External Clock Input Synchronization Control bit

TMR1CS = 1:

1 = Do not synchronize external clock input

0 = Synchronize external clock input

TMR1CS = 0:

This bit is ignored. Timer1 uses the internal clock when TMR1CS = 0.

bit 1 TMR1CS: Timer1 Clock Source Select bit

1 = External clock from pin T1CKI (on the rising edge)

0 = Internal clock (Fosc/4)

bit 0 TMR10N: Timer1 On bit

1 = Enables Timer10 = Stops Timer1

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

n = Value at POR

'1' = Bit is set

'0' = Bit is cleared x = Bit is unknown

6.1 Timer1 Operation in Timer Mode

Timer mode is selected by clearing the TMR1CS (T1CON<1>) bit. In this mode, the input clock to the timer is Fosc/4. The synchronize control bit T1SYNC (T1CON<2>) has no effect since the internal clock is always in sync.

6.2 Timer1 Operation in Synchronized Counter Mode

Counter mode is selected by setting bit TMR1CS. In this mode, the timer increments on every rising edge of clock input on pin RC1/T1OSI when bit T1OSCEN is set, or pin RC0/T1OSO/T1CKI when bit T1OSCEN is cleared.

If T1SYNC is cleared, then the external clock input is synchronized with internal phase clocks. The synchronization is done after the prescaler stage. The prescaler is an asynchronous ripple counter.

In this configuration, during SLEEP mode, Timer1 will not increment even if the external clock is present, since the synchronization circuit is shut-off. The prescaler however will continue to increment.

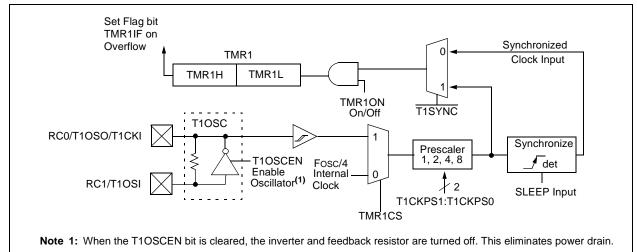
6.2.1 EXTERNAL CLOCK INPUT TIMING FOR SYNCHRONIZED COUNTER MODE

When an external clock input is used for Timer1 in Synchronized Counter mode, it must meet certain requirements. The external clock requirement is due to internal phase clock (Tosc) synchronization. Also, there is a delay in the actual incrementing of TMR1 after synchronization.

When the prescaler is 1:1, the external clock input is the same as the prescaler output. The synchronization of T1CKI with the internal phase clocks is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks. Therefore, it is necessary for T1CKI to be high for at least 2Tosc (and a small RC delay of 20 ns), and low for at least 2Tosc (and a small RC delay of 20 ns). Refer to the appropriate electrical specifications, parameters 45, 46, and 47.

When a prescaler other than 1:1 is used, the external clock input is divided by the asynchronous ripple counter type prescaler, so that the prescaler output is symmetrical. In order for the external clock to meet the sampling requirement, the ripple counter must be taken into account. Therefore, it is necessary for T1CKI to have a period of at least 4Tosc (and a small RC delay of 40 ns), divided by the prescaler value. The only requirement on T1CKI high and low time is that they do not violate the minimum pulse width requirements of 10 ns). Refer to the appropriate electrical specifications, parameters 40, 42, 45, 46, and 47.

FIGURE 6-1: TIMER1 BLOCK DIAGRAM



6.3 Timer1 Operation in Asynchronous Counter Mode

If control bit T1SYNC (T1CON<2>) is set, the external clock input is not synchronized. The timer continues to increment asynchronous to the internal phase clocks. The timer will continue to run during SLEEP and can generate an interrupt-on-overflow which will wake-up the processor. However, special precautions in software are needed to read from, or write to the Timer1 register pair (TMR1H:TMR1L) (Section 6.3.2).

In Asynchronous Counter mode, Timer1 cannot be used as a time-base for capture or compare operations.

6.3.1 EXTERNAL CLOCK INPUT TIMING WITH UNSYNCHRONIZED CLOCK

If control bit T1SYNC is set, the timer will increment completely asynchronously. The input clock must meet certain minimum high time and low time requirements, as specified in timing parameters 45, 46, and 47.

6.3.2 READING AND WRITING TMR1 IN ASYNCHRONOUS COUNTER MODE

Reading TMR1H or TMR1L, while the timer is running from an external asynchronous clock, will ensure a valid read (taken care of in hardware). However, the user should keep in mind that reading the 16-bit timer in two 8-bit values itself, poses certain problems, since the timer may overflow between the reads.

For writes, it is recommended that the user simply stop the timer and write the desired values. A write contention may occur by writing to the timer registers while the register is incrementing. This may produce an unpredictable value in the timer register.

Reading the 16-bit value requires some care. Example 6-1 is an example routine to read the 16-bit timer value. This is useful if the timer cannot be stopped.

EXAMPLE 6-1: READING A 16-BIT FREE-RUNNING TIMER

```
; All interrupts are disabled
          MOVF
                 TMR1H, W
                             ; Read high byte
          MOVWF TMPH
          MOVF
                 TMR1L, W
                             ;Read low byte
          MOVWF
                 TMPL
           MOVF
                 TMR1H, W
                             ;Read high byte
           SUBWF TMPH, W
                             ;Sub 1st read with 2nd read
                             ;Is result = 0
          BTFSC STATUS, Z
          GOTO
                 CONTINUE
                             ;Good 16-bit read
; TMR1L may have rolled over between the read of the high and low bytes.
; Reading the high and low bytes now will read a good value.
          MOVF
                 TMR1H, W
                             ; Read high byte
          MOVWF TMPH
          MOVF
                 TMR1L, W
                             ;Read low byte
          MOVWF TMPL
; Re-enable the Interrupt (if required)
CONTINUE
                             ;Continue with your code
```

6.4 Timer1 Oscillator

A crystal oscillator circuit is built-in between pins T1OSI (input) and T1OSO (amplifier output). It is enabled by setting control bit T1OSCEN (T1CON<3>). The oscillator is a low power oscillator rated up to 200 kHz. It will continue to run during SLEEP. It is primarily intended for a 32 kHz crystal. Table 6-1 shows the capacitor selection for the Timer1 oscillator.

The Timer1 oscillator is identical to the LP oscillator. The user must provide a software time delay to ensure proper oscillator start-up.

TABLE 6-1: CAPACITOR SELECTION FOR THE TIMER1 OSCILLATOR

Freq	C1							
	5	C2						
32 kHz	33 pF	33 pF						
100 kHz	100 kHz 15 pF 15 pF							
200 kHz 15 pF 15 pF								
These values are for design guidance only.								
Crystals	Tested:							
Epson C-00	1R32.768K-A	± 20 PPM						
100 kHz Epson C-2 100.00 KC-P ± 20 PPM								
STD XTL	STD XTL 200.000 kHz ± 20 PPM							
	100 kHz 200 kHz lues are for o Crystals Epson C-00 Epson C-2	100 kHz 15 pF 200 kHz 15 pF lues are for design guidar Crystals Tested: Epson C-001R32.768K-A Epson C-2 100.00 KC-P						

- **Note 1:** Higher capacitance increases the stability of the oscillator but also increases the start-up time.
 - 2: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.

6.5 Resetting Timer1 Using the CCP Trigger Output

If the CCP1 module is configured in Compare mode to generate a "special event trigger" (CCP1M3:CCP1M0 = 1011), this signal will reset Timer1.

Note: The special event trigger from the CCP1 module will not set interrupt flag bit TMR1IF (PIR1<0>).

Timer1 must be configured for either Timer or Synchronized Counter mode, to take advantage of this feature. If Timer1 is running in Asynchronous Counter mode, this reset operation may not work.

In the event that a write to Timer1 coincides with a special event trigger from CCP1, the write will take precedence.

In this mode of operation, the CCPR1H:CCPR1L registers pair effectively become the period register for Timer1.

6.6 Resetting of Timer1 Register Pair (TMR1H:TMR1L)

TMR1H and TMR1L registers are not reset on a POR or any other RESET, except by the CCP1 special event trigger.

T1CON register is reset to 00h on a Power-on Reset. In any other RESET, the register is unaffected.

6.7 Timer1 Prescaler

The prescaler counter is cleared on writes to the TMR1H or TMR1L registers.

TABLE 6-2: REGISTERS ASSOCIATED WITH TIMER1 AS A TIMER/COUNTER

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Valu Powe Re:	er-on	all c	e on other SETS
0Bh, 8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000	000x	0000	000u
0Ch	PIR1	LCDIF	ADIF	_	_	SSPIF	CCP1IF	TMR2IF	TMR1IF	00	0000	00	0000
8Ch	PIE1	LCDIE	ADIE	_	_	SSPIE	CCP1IE	TMR2IE	TMR1IE	00	0000	00	0000
0Eh	TMR1L	Holding	Holding register for the Least Significant Byte of the 16-bit TMR1 Register								xxxx	uuuu	uuuu
0Fh	TMR1H	Holding	lolding register for the Most Significant Byte of the 16-bit TMR1 Register								xxxx	uuuu	uuuu
10h	T1CON	_	1	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	00	0000	uu	uuuu

 $\label{eq:local_equation} \textbf{Legend:} \quad \textbf{x} = \textbf{unknown}, \textbf{u} = \textbf{unchanged}, \textbf{-} = \textbf{unimplemented}, \textbf{read as '0'}. \textbf{Shaded cells are not used by the Timer 1 module}.$

7.0 TIMER2 MODULE

Timer2 is an 8-bit timer with a prescaler and a postscaler. It can be used as the PWM time-base for the PWM mode of the CCP module. It can also be used as a time-base for the Master mode SPI clock. The TMR2 register is readable and writable, and is cleared on any device RESET.

The input clock (Fosc/4) has a prescale option of 1:1, 1:4, or 1:16 (selected by control bits T2CKPS1:T2CKPS0 (T2CON<1:0>)).

The Timer2 module has an 8-bit period register, PR2. TMR2 increments from 00h until it matches PR2 and then resets to 00h on the next increment cycle. PR2 is a readable and writable register. The PR2 register is set during RESET.

The match output of TMR2 goes through a 4-bit postscaler (which gives a 1:1 to 1:16 scaling inclusive) to generate a TMR2 interrupt (latched in flag bit TMR2IF, (PIR1<1>)).

Timer2 can be shut-off by clearing control bit TMR2ON (T2CON<2>) to minimize power consumption.

Figure 7-1 shows the Timer2 control register.

7.1 Timer2 Prescaler and Postscaler

The prescaler and postscaler counters are cleared when any of the following occurs:

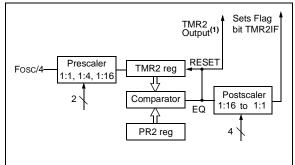
- · a write to the TMR2 register
- a write to the T2CON register
- any device RESET (Power-on Reset, MCLR Reset, or Watchdog Timer Reset)

TMR2 will not clear when T2CON is written.

7.2 Output of TMR2

The output of TMR2 (before the postscaler) is fed to the Synchronous Serial Port module, which optionally uses it to generate the shift clock.

FIGURE 7-1: TIMER2 BLOCK DIAGRAM



Note 1: TMR2 register output can be software selected by the SSP Module as the source clock.

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REGISTER 7-1: T2CON: TIMER2 CONTROL REGISTER (ADDRESS 12h)

U-0 R/W-0 R/W-0 R/W-0 R/W-0 R/W-0 R/W-0 R/W-0 TOUTPS2 TOUTPS3 TOUTPS1 TOUTPS0 TMR2ON T2CKPS1 T2CKPS0 bit 7 bit 0

bit 7 Unimplemented: Read as '0'

bit 6-3 TOUTPS3:TOUTPS0: Timer2 Output Postscale Select bits

0000 = 1:1 Postscale 0001 = 1:2 Postscale

•

1111 = 1:16 Postscale

bit 2 TMR2ON: Timer2 On bit

1 = Timer2 is on 0 = Timer2 is off

bit 1-0 T2CKPS1:T2CKPS0: Timer2 Clock Prescale Select bits

00 =Prescaler is 1 01 =Prescaler is 4 1x =Prescaler is 16

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'

- n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

TABLE 7-1: REGISTERS ASSOCIATED WITH TIMER2 AS A TIMER/COUNTER

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other RESETS
0Bh, 8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000u
0Ch	PIR1	LCDIF	ADIF	_	_	SSPIF	CCP1IF	TMR2IF	TMR1IF	00 0000	00 0000
8Ch	PIE1	LCDIE	ADIE	_	_	SSPIE	CCP1IE	TMR2IE	TMR1IE	00 0000	00 0000
11h	TMR2	Timer2 Mod	dule's Regist		0000 0000	0000 0000					
12h	T2CON	_	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	-000 0000
92h	PR2	Timer2 Per	ner2 Period Register								1111 1111

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the Timer2 module.

8.0 CAPTURE/COMPARE/PWM (CCP) MODULE

The CCP (Capture/Compare/PWM) module contains a 16-bit register which can operate as a 16-bit capture register, as a 16-bit compare register, or as a PWM master/slave duty cycle register. Table 8-1 shows the timer resources used by the CCP module.

The Capture/Compare/PWM Register1 (CCPR1) is comprised of two 8-bit registers: CCPR1L (low byte) and CCPR1H (high byte). The CCP1CON register controls the operation of CCP1. All three are readable and writable.

Register 8-1 shows the CCP1CON register.

For use of the CCP module, refer to the *Embedded Control Handbook*, "Using the CCP Modules" (AN594).

TABLE 8-1: CCP MODE - TIMER RESOURCE

CCP Mode	Timer Resource			
Capture	Timer1			
Compare	Timer1			
PWM	Timer2			

REGISTER 8-1: CCP1CON REGISTER (ADDRESS 17h)

U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
_	_	CCP1X	CCP1Y	CCP1M3	CCP1M2	CCP1M1	CCP1M0
bit 7							bit 0

bit 7-6 Unimplemented: Read as '0'

bit 5-4 CCP1X:CCP1Y: PWM Least Significant bits

Capture mode:

Unused

Compare mode:

Unused

PWM mode:

These bits are the two LSbs of the PWM duty cycle. The eight MSbs are found in CCPR1L.

bit 3-0 CCP1M3:CCP1M0: CCP1 Mode Select bits

0000 = Capture/Compare/PWM disabled (resets CCP1 module)

0100 = Capture mode, every falling edge

0101 = Capture mode, every rising edge

0110 = Capture mode, every 4th rising edge

0111 = Capture mode, every 16th rising edge

1000 = Compare mode, set output on match (bit CCP1IF is set)

1001 = Compare mode, clear output on match (bit CCP1IF is set)

1010 = Compare mode, generate software interrupt-on-match (bit CCP1IF is set, CCP1 pin is unaffected)

'0' = Bit is cleared

x = Bit is unknown

1011 = Compare mode, trigger special event (CCP1IF bit is set; CCP1 resets TMR1)

11xx = PWM mode

- n = Value at POR

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'

'1' = Bit is set

8.1 Capture Mode

In Capture mode, CCPR1H:CCPR1L captures the 16-bit value of the TMR1 register when an event occurs on pin RC2/CCP1 (Figure 8-1). An event can be selected to be one of the following:

- · Every falling edge
- · Every rising edge
- · Every 4th rising edge
- · Every 16th rising edge

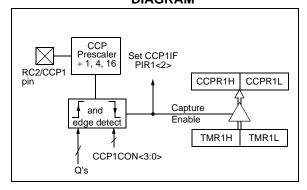
An event is selected by control bits CCP1M3:CCP1M0 (CCP1CON<3:0>). When a capture is made, the interrupt request flag bit CCP1IF (PIR1<2>) is set. It must be cleared in software. If another capture occurs before the value in register CCPR1 is read, the old captured value is overwritten with the new captured value.

8.1.1 CCP PIN CONFIGURATION

In Capture mode, the RC2/CCP1 pin should be configured as an input by setting the TRISC<2> bit.

Note: If the RC2/CCP1 pin is configured as an output, a write to the port can cause a capture condition.

FIGURE 8-1: CAPTURE MODE OPERATION BLOCK DIAGRAM



8.1.2 TIMER1 MODE SELECTION

Timer1 must be running in Timer mode, or Synchronized Counter mode, for the CCP module to use the capture feature. In Asynchronous Counter mode, the capture operation may not work.

8.1.3 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep enable bit CCP1IE (PIE1<2>) clear to avoid false interrupts and should clear flag bit CCP1IF following any such change in operating mode.

8.1.4 CCP PRESCALER

There are four prescaler settings, specified by bits CCP1M3:CCP1M0. Whenever the CCP module is turned off, or the CCP module is not in Capture mode, the prescaler counter is cleared. This means that any RESET will clear the prescaler counter.

Switching from one capture prescaler to another may generate an interrupt. Also, the prescaler counter will not be cleared, therefore, the first capture may be from a non-zero prescaler. Example 8-1 shows the recommended method for switching between capture prescalers. This example also clears the prescaler counter and will not generate the "false" interrupt.

EXAMPLE 8-1: CHANGING BETWEEN CAPTURE PRESCALERS

CLRF	CCP1CON	;	Turn CCP module off
MOVLW	NEW_CAPT_PS	;	Load the W reg with
		;	the new prescaler
		;	mode value and CCP ON
MOVWF	CCP1CON	;	Load CCP1CON with
		;	this value

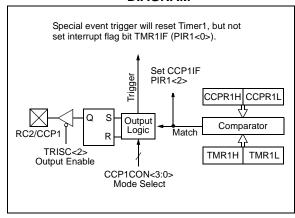
8.2 Compare Mode

In Compare mode, the 16-bit CCPR1 register value is constantly compared against the TMR1 register pair value. When a match occurs, the RC2/CCP1 pin is:

- · Driven high
- Driven low
- · Remains unchanged

The action on the pin is based on the value of control bits CCP1M3:CCP1M0 (CCP1CON<3:0>). At the same time, a compare interrupt is also generated.

FIGURE 8-2: COMPARE MODE OPERATION BLOCK DIAGRAM



8.2.1 CCP PIN CONFIGURATION

The user must configure the RC2/CCP1 pin as an output by clearing the TRISC<2> bit.

Note: Clearing the CCP1CON register will force the RC2/CCP1 compare output latch to the default low level. This is not the PORTC I/O data latch.

8.2.2 TIMER1 MODE SELECTION

Timer1 must be running in Timer mode, or Synchronized Counter mode, if the CCP module is using the compare feature. In Asynchronous Counter mode, the compare operation may not work.

8.2.3 SOFTWARE INTERRUPT MODE

When Generate Software Interrupt is chosen, the CCP1 pin is not affected. Only a CCP interrupt is generated (if enabled).

8.2.4 SPECIAL EVENT TRIGGER

In this mode, an internal hardware trigger is generated which may be used to initiate an action.

The special event trigger output of CCP1 resets the TMR1 register pair and starts an A/D conversion. This allows the CCPR1H:CCPR1L register pair to effectively be a 16-bit programmable period register for Timer1.

Note: The "special event trigger" from the CCP1 module will not set interrupt flag bit TMR1IF (PIR1<0>).

8.3 PWM Mode

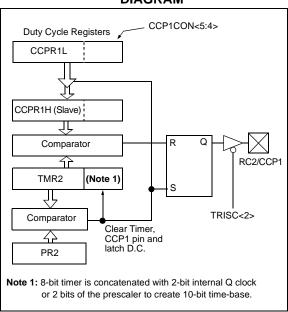
In Pulse Width Modulation (PWM) mode, the CCP1 pin produces up to a 10-bit resolution PWM output. Since the CCP1 pin is multiplexed with the PORTC data latch, the TRISC<2> bit must be cleared to make the CCP1 pin an output.

Note: Clearing the CCP1CON register will force the CCP1 PWM output latch to the default low level. This is not the PORTC I/O data latch.

Figure 8-3 shows a simplified block diagram of the CCP module in PWM mode.

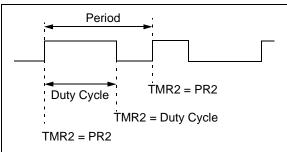
For a step-by-step procedure on how to set up the CCP module for PWM operation, see Section 8.3.3.

FIGURE 8-3: SIMPLIFIED PWM BLOCK DIAGRAM



A PWM output (Figure 8-4) has a time-base (period) and a time that the output stays high (duty cycle). The frequency of the PWM is the inverse of the period (1/period).

FIGURE 8-4: PWM OUTPUT



8.3.1 PWM PERIOD

The PWM period is specified by writing to the PR2 register. The PWM period can be calculated using the following formula:

PWM period =
$$[(PR2) + 1] \cdot 4 \cdot TOSC \cdot (TMR2 \text{ prescale value})$$

PWM frequency is defined as 1 / [PWM period].

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- · TMR2 is cleared
- The CCP1 pin is set (exception: if PWM duty cycle = 0%, the CCP1 pin will not be set)
- The PWM duty cycle is latched from CCPR1L into CCPR1H

Note: The Timer2 postscaler (Section 7.0) is not used in the determination of the PWM frequency. The postscaler could be used to have a servo update rate at a different frequency than the PWM output.

8.3.2 PWM DUTY CYCLE

The PWM duty cycle is specified by writing to the CCPR1L register and to the CCP1CON<5:4> bits. Up to 10-bit resolution is available; the CCPR1L contains the eight MSbs and CCP1CON<5:4> contains the two LSbs. This 10-bit value is represented by CCPR1L:CCP1CON<5:4>. The following equation is used to calculate the PWM duty cycle in time:

CCPR1L and CCP1CON<5:4> can be written to at any time, but the duty cycle value is not latched into CCPR1H until after a match between PR2 and TMR2 occurs (i.e., the period is complete). In PWM mode, CCPR1H is a read only register.

The CCPR1H register and a 2-bit internal latch are used to double buffer the PWM duty cycle. This double buffering is essential for glitchless PWM operation.

When the CCPR1H and 2-bit latch match TMR2, concatenated with an internal 2-bit Q clock, or 2 bits of the TMR2 prescaler, the CCP1 pin is cleared.

The maximum PWM resolution (bits) for a given PWM frequency is given by the equation:

PWM Resolution (max) =
$$\frac{\log\left(\frac{FOSC}{FPWM}\right)}{\log(2)} bits$$

Note: If the PWM duty cycle value is longer than the PWM period, the CCP1 pin will not be cleared.

EQUATION 8-1: EXAMPLES OF PWM PERIOD AND DUTY CYCLE CALCULATION

- 1. Find the value of the PR2 register, given:
 - Desired PWM frequency = 31.25 kHz
 - Fosc = 8 MHz
 - TMR2 prescale = 1

From the equation for PWM period in Section 8.3.1,

$$1/31.25 \text{ kHz} = [(PR2) + 1] \cdot 4 \cdot 1/8 \text{ MHz} \cdot 1$$

or

8.0

= $[(PR2) + 1] \cdot 4 \cdot 125 \text{ ns} \cdot 1 = [(PR2) + 1] \cdot 0.5 \mu s$ 32 µs

PR2 $= (32 \mu s / 0.5 \mu s) - 1$

PR2 = 63

2. Find the maximum resolution of the duty cycle that can be used with a 31.25 kHz frequency and 8 MHz oscillator.

From the equation from maximum PWM resolution in Section 8.3.2,

$$1/31.25 \text{ kHz} = 2^{\text{PWM RESOLUTION}} \bullet 1/8 \text{ MHz} \bullet 1$$
 or
$$32 \text{ } \mu \text{s} = 2^{\text{PWM RESOLUTION}} \bullet 125 \text{ ns} \bullet 1$$

$$256 = 2^{\text{PWM RESOLUTION}}$$

$$\log(256) = (\text{PWM Resolution}) \bullet \log(2)$$

$$8.0 = \text{PWM Resolution}$$

At most, an 8-bit resolution duty cycle can be obtained from a 31.25 kHz frequency and a 8 MHz oscillator, i.e., $0 \le CCPR1L:CCP1CON < 5:4 > \le 255$. Any value greater than 255 will result in a 100% duty cycle.

In order to achieve higher resolution, the PWM frequency must be decreased. In order to achieve higher PWM frequency, the resolution must be decreased.

Table 8-2 lists example PWM frequencies and resolutions for Fosc = 8 MHz. TMR2 prescaler and PR2 values are also shown.

8.3.3 SET-UP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for PWM operation:

- 1. Set the PWM period by writing to the PR2 register.
- 2. Set the PWM duty cycle by writing to the CCPR1L register and CCP1CON<5:4> bits.
- 3. Make the CCP1 pin an output by clearing the TRISC<2> bit.
- 4. Set the TMR2 prescale value and enable Timer2 by writing to T2CON.
- 5. Configure the CCP module for PWM operation.

TABLE 8-2: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS AT 8 MHz

PWM Frequency	488 Hz	1.95 kHz	7.81 kHz	31.25 kHz	62.5 kHz	250 kHz
Timer Prescaler (1, 4, 16)	16	4	1	1	1	1
PR2 Value	0xFF	0xFF	0xFF	0x3F	0x1F	0x07
Maximum Resolution (bits)	10	10	10	8	7	5

TABLE 8-3: REGISTERS ASSOCIATED WITH TIMER1, CAPTURE AND COMPARE

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Valu Powe Res	er-on	all o	e on other SETS
0Bh, 8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000	000x	0000	000u
0Ch	PIR1	LCDIF	ADIF	_	_	SSPIF	CCP1IF	TMR2IF	TMR1IF	00	0000	00	0000
8Ch	PIE1	LCDIE	ADIE		_	SSPIE	CCP1IE	TMR2IE	TMR1IE	00	0000	00	0000
87h	TRISC	_	PORTC Data Direction Control Register								1111	11	1111
0Eh	TMR1L	Holding	register f	or the Least	t Significant	Byte of the 1	6-bit TMR1	1 Register		xxxx	xxxx	uuuu	uuuu
0Fh	TMR1H	Holding	register f	or the Most	Significant I	Byte of the 1	6-bit TMR1	Register		xxxx	xxxx	uuuu	uuuu
10h	T1CON	_	1	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	00	0000	uu	uuuu
15h	CCPR1L	Capture	Capture/Compare/PWM1 (LSB)								xxxx	uuuu	uuuu
16h	CCPR1H	Capture	Capture/Compare/PWM1 (MSB)								xxxx	uuuu	uuuu
17h	CCP1CON		_	CCP1X	CCP1Y	CCP1M3	CCP1M2	CCP1M1	CCP1M0	00	0000	00	0000

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used in these modes.

TABLE 8-4: REGISTERS ASSOCIATED WITH PWM AND TIMER2

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other RESETS
0Bh, 8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000u
0Ch	PIR1	LCDIF	ADIF		_	SSPIF	CCP1IF	TMR2IF	TMR1IF	00 0000	00 0000
8Ch	PIE1	LCDIE	ADIE		_	SSPIE	CCP1IE	TMR2IE	TMR1IE	00 0000	00 0000
87h	TRISC	_	_	PORTC Da	ata Direction	Control Re	gister			11 1111	11 1111
11h	TMR2	Timer2 N	Module Regi	ster						0000 0000	0000 0000
92h	PR2	Timer2 N	Module Perio	d Register						1111 1111	1111 1111
12h	T2CON	_	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	-000 0000
15h	CCPR1L	Capture/	Capture/Compare/PWM1 (LSB)								uuuu uuuu
16h	CCPR1H	Capture/	oture/Compare/PWM1 (MSB)								uuuu uuuu
17h	CCP1CON	_	_	CCP1X	CCP1Y	CCP1M3	CCP1M2	CCP1M1	CCP1M0	00 0000	00 0000

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used in this mode.

9.0 SYNCHRONOUS SERIAL PORT (SSP) MODULE

The Synchronous Serial Port (SSP) module is a serial interface useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be serial EEPROMs, shift registers, display drivers, A/D converters, etc. The SSP module can operate in one of two modes:

- Serial Peripheral Interface (SPITM)
- Inter-Integrated Circuit (I²C[™])

Refer to Application Note AN578, "Use of the SSP Module in the I²C Multi-Master Environment."

REGISTER 9-1: SSPSTAT: SERIAL PORT STATUS REGISTER (ADDRESS 94h)

R/W-0	R/W-0	R-0	R-0	R-0	R-0	R-0	R-0
SMP	CKE	D/A	Р	S	R/W	UA	BF
bit 7							bit 0

bit 7 SMP: SPI Data Input Sample Phase bit

SPI Master mode:

- 1 = Input data sampled at end of data output time
- 0 = Input data sampled at middle of data output time

SPI Slave mode:

SMP must be cleared when SPI is used in Slave mode

bit 6 CKE: SPI Clock Edge Select bit (see Figure 9-3, Figure 9-4, and Figure 9-5)

CKP = 0:

- 1 = Data transmitted on rising edge of SCK
- 0 = Data transmitted on falling edge of SCK

CKP = 1:

- 1 = Data transmitted on falling edge of SCK
- 0 = Data transmitted on rising edge of SCK
- bit 5 **D/A**: Data/Address bit (I²C mode only)
 - 1 = Indicates that the last byte received or transmitted was data
 - 0 = Indicates that the last byte received or transmitted was address
- bit 4 **P:** STOP bit (I²C mode only. This bit is cleared when the SSP module is disabled, or when the START bit was detected last.)
 - 1 = Indicates that a STOP bit has been detected last (this bit is '0' on RESET)
 - 0 = STOP bit was not detected last
- bit 3 **S:** START bit (I²C mode only. This bit is cleared when the SSP module is disabled, or when the STOP bit was detected last.)
 - 1 = Indicates that a START bit has been detected last (this bit is '0' on RESET)
 - 0 = START bit was not detected last
- bit 2 **R/W**: Read/Write bit Information (I²C mode only)

This bit holds the R/W bit information following the last address match. This bit is only valid from the address match to the next START bit, STOP bit, or \overline{ACK} bit.

- 1 = Read
- 0 = Write
- bit 1 **UA**: Update Address (10-bit I²C mode only)
 - 1 = Indicates that the user needs to update the address in the SSPADD register
 - 0 = Address does not need to be updated
- bit 0 BF: Buffer Full Status bit

Receive (SPI and I²C modes):

- 1 = Receive complete, SSPBUF is full
- 0 = Receive not complete, SSPBUF is empty

Transmit (I²C mode only)

- 1 = Transmit in progress, SSPBUF is full
- 0 = Transmit complete, SSPBUF is empty

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

REGISTER 9-2: SSPCON: SYNC SERIAL PORT CONTROL REGISTER (ADDRESS 14h)

| R/W-0 |
|-------|-------|-------|-------|-------|-------|-------|-------|
| WCOL | SSPOV | SSPEN | CKP | SSPM3 | SSPM2 | SSPM1 | SSPM0 |
| | | | | | | | |

bit 7

bit 7 WCOL: Write Collision Detect bit

1 = SSPBUF register is written while still transmitting the previous word (must be cleared in software)

0 = No collision

bit 6 SSPOV: Receive Overflow Indicator bit

In SPI mode:

1 = A new byte is received while SSPBUF is holding previous data. Data in SSPSR is lost on overflow. Overflow only occurs in Slave mode. The user must read the SSPBUF, even if only transmitting data, to avoid setting overflows. In Master mode, the overflow bit is not set since each operation is initiated by writing to the SSPBUF register. (Must be cleared in software.)

0 =No overflow

In I²C mode:

- 1 = A byte is received while the SSPBUF is holding the previous byte. SSPOV is a "don't care" in transmit mode. (Must be cleared in software.)
- 0 = No overflow
- bit 5 SSPEN: Synchronous Serial Port Enable bit

In SPI mode:

When enabled, these pins must be properly configured as input or output.

- 1 = Enables serial port and configures SCK, SDO, SDI, and \overline{SS} as the source of the serial port pins
- 0 = Disables serial port and configures these pins as I/O port pins

In I²C mode:

When enabled, these pins must be properly configured as input or output.

- 1 = Enables the serial port and configures the SDA and SCL pins as the source of the serial port pins
- 0 = Disables serial port and configures these pins as I/O port pins
- bit 4 **CKP**: Clock Polarity Select bit

In SPI mode:

- 1 = Idle state for clock is a high level
- 0 = Idle state for clock is a low level

In I²C mode:

SCK release control

- 1 = Enable clock
- 0 = Holds clock low (clock stretch). (Used to ensure data setup time.)
- bit 3-0 SSPM3:SSPM0: Synchronous Serial Port Mode Select bits

0000 = SPI Master mode, clock = Fosc/4

0001 = SPI Master mode, clock = Fosc/16

0010 = SPI Master mode, clock = Fosc/64

0011 = SPI Master mode, clock = TMR2 output/2

0100 = SPI Slave mode, clock = SCK pin (SS pin control enabled)

0101 = SPI Slave mode, clock = SCK pin (SS pin control disabled, SS can be used as I/O pin)

 $0110 = I^2C$ Slave mode, 7-bit address

 $0111 = I^2C$ Slave mode, 10-bit address

 $1011 = I^2C$ firmware controlled Master mode (slave idle)

1110 = I²C firmware controlled Master mode, 7-bit address with START and STOP bit interrupts enabled

 $1111 = I^2C$ firmware controlled Master mode, 10-bit address with START and STOP bit interrupts enabled

1000, 1001, 1010, 1100, 1101 = reserved

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'

- n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

9.1 SPI Mode

The SPI mode allows 8-bits of data to be synchronously transmitted and received simultaneously. To accomplish communication, typically three pins are used:

- Serial Data Out (SDO) RC5/SDO
- · Serial Data In (SDI) RC4/SDI
- Serial Clock (SCK) RC3/SCK

Additionally, a fourth pin may be used when in a Slave mode of operation:

Slave Select (SS) RA5/AN4/SS

When initializing the SPI, several options need to be specified. This is done by programming the appropriate control bits in the SSPCON register (SSPCON<5:0>) and SSPSTAT<7:6>. These control bits allow the following to be specified:

- Master mode (SCK is the clock output)
- Slave mode (SCK is the clock input)
- Clock Polarity (Idle state of SCK)
- Clock Edge (output data on rising/falling edge of SCK)
- Clock Rate (Master mode only)
- Slave Select mode (Slave mode only)

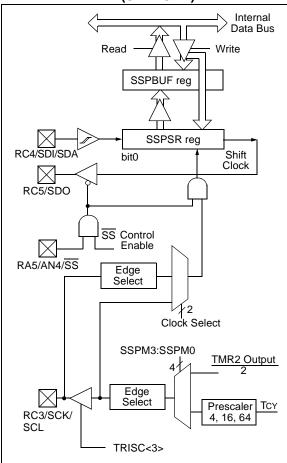
The SSP consists of a transmit/receive shift register (SSPSR) and a buffer register (SSPBUF). The SSPSR shifts the data in and out of the device. MSb first. The SSPBUF holds the data that was written to the SSPSR. until the received data is ready. Once the 8-bits of data have been received, that byte is moved to the SSPBUF register. Then, the buffer full detect bit, BF (SSPSTAT<0>), and interrupt flag bit, SSPIF (PIR1<3>), are set. This double buffering of the received data (SSPBUF) allows the next byte to start reception before reading the data that was just received. Any write to the SSPBUF register during transmission/reception of data will be ignored, and the write collision detect bit, WCOL (SSPCON<7>), will be set. User software must clear the WCOL bit so that it can be determined if the following write(s) to the SSPBUF register completed successfully. When the application software is expecting to receive valid data, the SSPBUF should be read before the next byte of data to transfer is written to the SSPBUF. Buffer full bit, BF (SSPSTAT<0>), indicates when SSPBUF has been loaded with the received data (transmission is complete). When the SSPBUF is read, bit BF is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally, the SSP interrupt is used to determine when the transmission/reception has completed. The SSPBUF must be read and/or written. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur. Example 9-1 shows the loading of the SSPBUF (SSPSR) for data transmission. The MOVWF RXDATA instruction (shaded) is only required if the received data is meaningful.

EXAMPLE 9-1: LOADING THE SSPBUF (SSPSR) REGISTER

	BCF BSF	STATUS, RP1 STATUS, RP0	;Select Bank1
	DOL	SIAIUS, RPU	i
LOOP	BTFSS	SSPSTAT, BF	;Has data been
			;received
			;(transmit
			;complete)?
	GOTO	LOOP	;No
	BCF	STATUS, RP0	;Select Bank0
	MOVF	SSPBUF, W	;W reg = contents
			;of SSPBUF
	MOVWF	RXDATA	;Save in user RAM
	MOVF	TXDATA, W	;W reg = contents ; of TXDATA
	MOVWF	SSPBUF	;New data to xmit

The block diagram of the SSP module, when in SPI mode (Figure 9-1), shows that the SSPSR is not directly readable or writable, and can only be accessed from addressing the SSPBUF register. Additionally, the SSP status register (SSPSTAT) indicates the various status conditions.

FIGURE 9-1: SSP BLOCK DIAGRAM (SPI MODE)



PIC16C925/926

To enable the serial port, SSP enable bit, SSPEN (SSPCON<5>) must be set. To reset or reconfigure SPI mode, clear bit SSPEN, re-initialize the SSPCON register, and then set bit SSPEN. This configures the SDI, SDO, SCK, and SS pins as serial port pins. For the pins to behave as the serial port function, they must have their data direction bits (in the TRISC register) appropriately programmed. That is:

- SDI must have TRISC<4> set
- SDO must have TRISC<5> cleared
- SCK (Master mode) must have TRISC<3> cleared
- SCK (Slave mode) must have TRISC<3> set
- SS must have TRISA<5> set and ADCON must be configured such that RA5 is a digital I/O

Any serial port function that is not desired may be overridden by programming the corresponding data direction (TRIS) register to the opposite value. An example would be in Master mode, where you are only sending data (to a display driver), then both SDI and SS could be used as general purpose outputs by clearing their corresponding TRIS register bits.

Figure 9-2 shows a typical connection between two microcontrollers. The master controller (Processor 1) initiates the data transfer by sending the SCK signal. Data is shifted out of both shift registers on their programmed clock edge, and latched on the opposite edge of the clock. Both processors should be programmed to same Clock Polarity (CKP), then both controllers would send and receive data at the same time. Whether the data is meaningful (or dummy data), depends on the application software. This leads to three scenarios for data transmission:

- Master sends data Slave sends dummy data
- Master sends data Slave sends data

• Master sends dummy data — Slave sends data

The master can initiate the data transfer at any time because it controls the SCK. The master determines when the slave (Processor 2) is to broadcast data by the firmware protocol.

In Master mode, the data is transmitted/received as soon as the SSPBUF register is written to. If the SPI is only going to receive, the SCK output could be disabled (programmed as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPBUF register as if a normal received byte (interrupts and status bits appropriately set). This could be useful in receiver applications as a "line activity monitor" mode.

In Slave mode, the data is transmitted and received as the external clock pulses appear on SCK. When the last bit is latched, the interrupt flag bit SSPIF (PIR1<3>) is set.

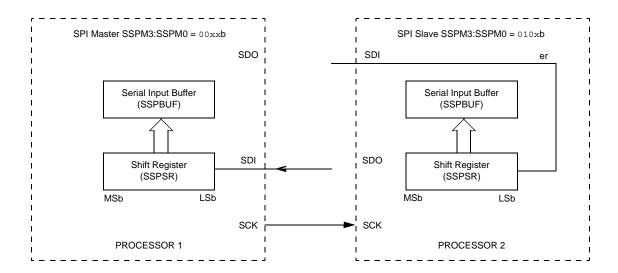
The clock polarity is selected by appropriately programming bit CKP (SSPCON<4>). This then, would give waveforms for SPI communication as shown in Figure 9-3, Figure 9-4, and Figure 9-5, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- Fosc/4 (or Tcy)
- Fosc/16 (or 4 Tcy)
- Fosc/64 (or 16 Tcy)
- Timer2 output/2

This allows a maximum bit clock frequency (at 8 MHz) of 2 MHz. When in Slave mode, the external clock must meet the minimum high and low times.

In SLEEP mode, the slave can transmit and receive data and wake the device from SLEEP.

FIGURE 9-2: SPI MASTER/SLAVE CONNECTION



The \overline{SS} pin allows a Synchronous Slave mode. The SPI must be in Slave mode (SSPCON<3:0> = 04h) and the TRISA<5> bit must be set for the Synchronous Slave mode to be enabled. When the \overline{SS} pin is low, transmission and reception are enabled and the SDO pin is driven. When the \overline{SS} pin goes high, the SDO pin is no longer driven, even if in the middle of a transmitted byte and becomes a floating output. External pull-up/pull-down resistors may be desirable, depending on the application.

- Note 1: When the SPI is in Slave mode with SS pin control enabled (SSPCON<3:0> = 0100), the SPI module will reset if the SS pin is set to VDD.
 - 2: If the SPI is used in Slave mode with CKE = '1', then the SS pin control must be enabled.

To emulate two-wire communication, the SDO pin can be connected to the SDI pin. When the SPI needs to operate as a receiver, the SDO pin can be configured as an input. This disables transmissions from the SDO. The SDI can always be left as an input (SDI function) since it cannot create a bus conflict.

FIGURE 9-3: SPI MODE TIMING, MASTER MODE

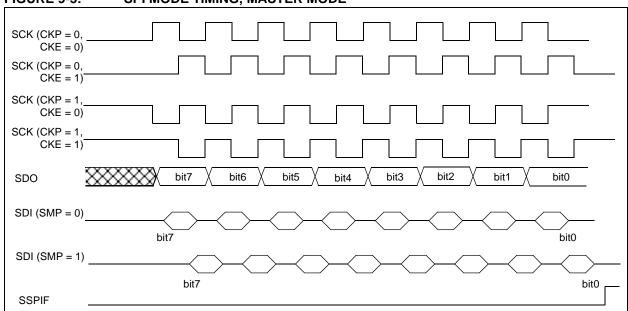
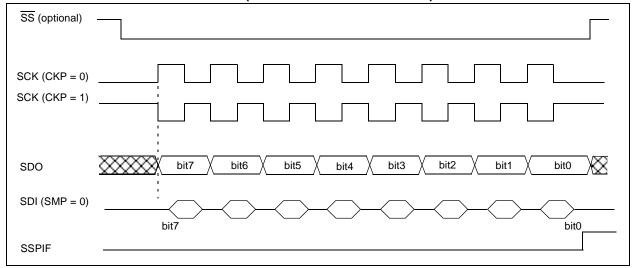


FIGURE 9-4: SPI MODE TIMING (SLAVE MODE WITH CKE = 0)



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FIGURE 9-5: SPI MODE TIMING (SLAVE MODE WITH CKE = 1)

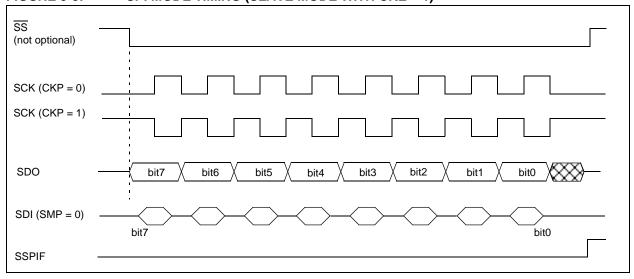


TABLE 9-1: REGISTERS ASSOCIATED WITH SPI OPERATION

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset		Power-on other	
0Bh, 8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000	000x	0000	000u
0Ch	PIR1	LCDIF	ADIF	_	_	SSPIF	CCP1IF	TMR2IF	TMR1IF	00	0000	00 0	0000
8Ch	PIE1	LCDIE	ADIE	_	_	SSPIE	CCP1IE	TMR2IE	TMR1IE	00	0000	00 0	0000
13h	SSPBUF	BUF Synchronous Serial Port Receive Buffer/Transmit Register			xxxx	xxxx	uuuu ı	uuuu					
14h	SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000	0000	0000	0000
85h	TRISA	_	_	PORTA Data Direction Control Register					11	1111	11 1	1111	
87h	TRISC	_	_	PORTC Data Direction Control Register					11	1111	11 1	1111	
94h	SSPSTAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	0000	0000	0000	0000

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the SSP in SPI mode.

9.2 I²C Overview

This section provides an overview of the Inter-Integrated Circuit (I^2C) bus, with Section 9.3 discussing the operation of the SSP module in I^2C mode.

The I²C bus is a two-wire serial interface developed by the Philips Corporation. The original specification, or standard mode, was for data transfers of up to 100 Kbps. An enhanced specification, or fast mode is not supported. This device will communicate with fast mode devices if attached to the same bus.

The I²C interface employs a comprehensive protocol to ensure reliable transmission and reception of data. When transmitting data, one device is the "master" which initiates transfer on the bus and generates the clock signals to permit that transfer, while the other device(s) acts as the "slave." All portions of the slave protocol are implemented in the SSP module's hardware, except general call support, while portions of the master protocol need to be addressed in the PIC16CXXX software. Table 9-2 defines some of the I²C bus terminology. For additional information on the I²C interface specification, refer to the Philips document #939839340011, "The I²C bus and how to use it", which can be obtained from the Philips Corporation.

In the I²C interface protocol, each device has an address. When a master wishes to initiate a data transfer, it first transmits the address of the device that it wishes to "talk" to. All devices "listen" to see if this is their address. Within this address, a bit specifies if the master wishes to read from/write to the slave device. The master and slave are always in opposite modes (transmitter/receiver) of operation during a data transfer. That is, they can be thought of as operating in either of these two relations:

- Master-transmitter and Slave-receiver
- · Slave-transmitter and Master-receiver

In both cases, the master generates the clock signal.

The output stages of the clock (SCL) and data (SDA) lines must have an open drain or open collector, in order to perform the wired-AND function of the bus. External pull-up resistors are used to ensure a high level when no device is pulling the line down. The number of devices that may be attached to the $\rm I^2C$ bus is limited only by the maximum bus loading specification of 400 pF.

9.2.1 INITIATING AND TERMINATING DATA TRANSFER

During times of no data transfer (idle time), both the clock line (SCL) and the data line (SDA) are pulled high through the external pull-up resistors. The START and STOP conditions determine the start and stop of data transmission. The START condition is defined as a high to low transition of the SDA when the SCL is high. The STOP condition is defined as a low to high transition of the SDA when the SCL is high. Figure 9-6 shows the START and STOP conditions. The master generates these conditions for starting and terminating data transfer. Due to the definition of the START and STOP conditions, when data is being transmitted, the SDA line can only change state when the SCL line is low.

FIGURE 9-6: START AND STOP CONDITIONS

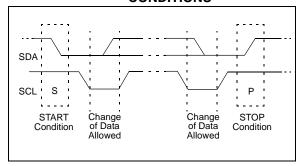


TABLE 9-2: I²C BUS TERMINOLOGY

Term	Description			
Transmitter	The device that sends the data to the bus.			
Receiver	The device that receives the data from the bus.			
Master	The device which initiates the transfer, generates the clock and terminates the transfer.			
Slave	The device addressed by a master.			
Multi-master	More than one master device in a system. These masters can attempt to control the bus at the same time without corrupting the message.			
Arbitration	Procedure that ensures that only one of the master devices will control the bus. This ensures that the transfer data does not get corrupted.			
Synchronization	Procedure where the clock signals of two or more devices are synchronized.			

9.2.2 ADDRESSING I²C DEVICES

There are two address formats. The simplest is the 7-bit address format with a R/\overline{W} bit (Figure 9-7). The more complex is the 10-bit address with a R/\overline{W} bit (Figure 9-8). For 10-bit address format, two bytes must be transmitted with the first five bits specifying this to be a 10-bit address.

FIGURE 9-7: 7-BIT ADDRESS FORMAT

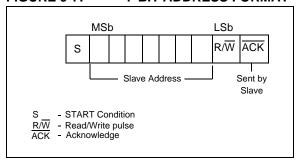
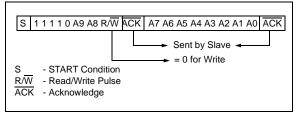


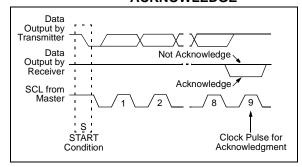
FIGURE 9-8: I²C 10-BIT ADDRESS FORMAT



9.2.3 TRANSFER ACKNOWLEDGE

All data must be transmitted per byte, with no limit to the number of bytes transmitted per data transfer. After each byte, the slave-receiver generates an Acknowledge bit (ACK) (see Figure 9-9). When a slave-receiver doesn't acknowledge the slave address or received data, the master must abort the transfer. The slave must leave SDA high so that the master can generate the STOP condition (Figure 9-6).

FIGURE 9-9: SLAVE-RECEIVER ACKNOWLEDGE



If the master is receiving the data (master-receiver), it generates an Acknowledge signal for each received byte of data, except for the last byte. To signal the end of data to the slave-transmitter, the master does not generate an Acknowledge (Not Acknowledge). The slave then releases the SDA line so the master can generate the STOP condition. The master can also generate the STOP condition during the Acknowledge pulse for valid termination of data transfer.

If the slave needs to delay the transmission of the next byte, holding the SCL line low will force the master into a wait state. Data transfer continues when the slave releases the SCL line. This allows the slave to move the received data, or fetch the data it needs to transfer before allowing the clock to start. This wait state technique can also be implemented at the bit level, Figure 9-10. The slave will inherently stretch the clock when it is a transmitter, but will not when it is a receiver. The slave will have to clear the SSPCON<4> bit to enable clock stretching when it is a receiver.

FIGURE 9-10: DATA TRANSFER WAIT STATE

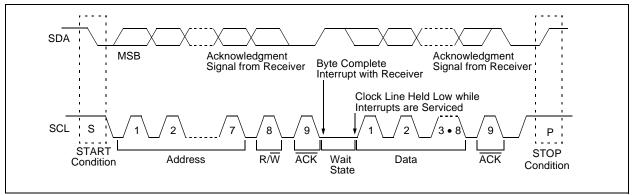


Figure 9-11 and Figure 9-12 show master-transmitter and Master-receiver data transfer sequences.

When a master does not wish to relinquish the bus (by generating a STOP condition), a Repeated START condition (Sr) must be generated. This condition is identical to the START condition (SDA goes high-to-low

while SCL is high), but occurs after a data transfer Acknowledge pulse (not the bus-free state). This allows a master to send "commands" to the slave and then receive the requested information, or to address a different slave device. This sequence is shown in Figure 9-13.

FIGURE 9-11: MASTER-TRANSMITTER SEQUENCE

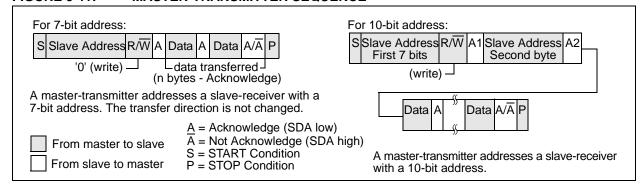


FIGURE 9-12: MASTER-RECEIVER SEQUENCE

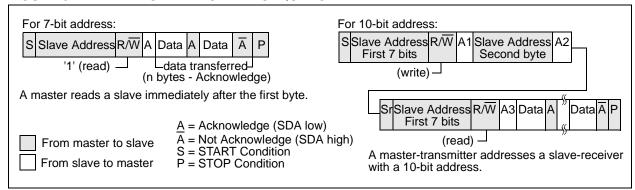
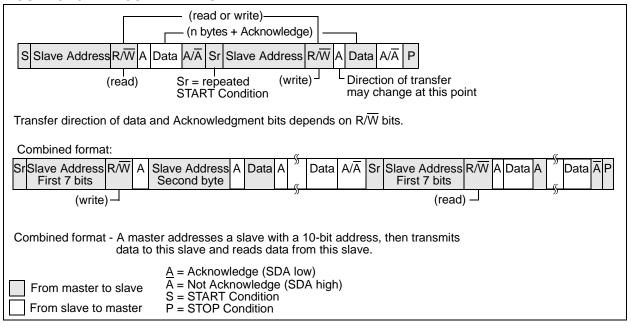


FIGURE 9-13: COMBINED FORMAT



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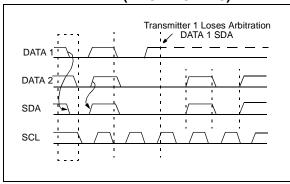
9.2.4 MULTI-MASTER

The I²C protocol allows a system to have more than one master. This is called multi-master. When two or more masters try to transfer data at the same time, arbitration and synchronization occur.

9.2.4.1 Arbitration

Arbitration takes place on the SDA line, while the SCL line is high. The master, which transmits a high when the other master transmits a low, loses arbitration (Figure 9-14) and turns off its data output stage. A master, which lost arbitration can generate clock pulses until the end of the data byte where it lost arbitration. When the master devices are addressing the same device, arbitration continues into the data.

FIGURE 9-14: MULTI-MASTER ARBITRATION (TWO MASTERS)



Masters that also incorporate the slave function and have lost arbitration, must immediately switch over to Slave-Receiver mode. This is because the winning master-transmitter may be addressing it.

Arbitration is not allowed between:

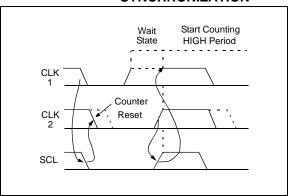
- · A Repeated START condition
- · A STOP condition and a data bit
- A Repeated START condition and a STOP condition

Care needs to be taken to ensure that these conditions do not occur.

9.2.4.2 Clock Synchronization

Clock synchronization occurs after the devices have started arbitration. This is performed using a wired-AND connection to the SCL line. A high to low transition on the SCL line causes the concerned devices to start counting off their low period. Once a device clock has gone low, it will hold the SCL line low until its SCL high state is reached. The low to high transition of this clock may not change the state of the SCL line, if another device clock is still within its low period. The SCL line is held low by the device with the longest low period. Devices with shorter low periods enter a high wait state, until the SCL line comes high. When the SCL line comes high, all devices start counting off their high periods. The first device to complete its high period will pull the SCL line low. The SCL line high time is determined by the device with the shortest high period, Figure 9-15.

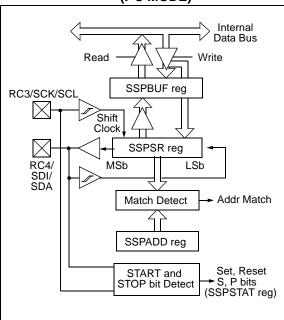
FIGURE 9-15: CLOCK SYNCHRONIZATION



9.3 SSP I²C Operation

The SSP module in I²C mode fully implements all slave functions, except general call support, and provides interrupts on START and STOP bits in hardware to facilitate firmware implementations of the master functions. The SSP module implements the standard mode specifications as well as 7-bit and 10-bit addressing. Two pins are used for data transfer. These are the RC3/SCK/SCL pin, which is the clock (SCL), and the RC4/SDI/SDA pin, which is the data (SDA). The user must configure these pins as inputs or outputs through the TRISC<4:3> bits. The SSP module functions are enabled by setting SSP enable bit, SSPEN (SSPCON<5>).

FIGURE 9-16: SSP BLOCK DIAGRAM (I²C MODE)



The SSP module has five registers for I²C operation. These are the:

- SSP Control Register (SSPCON)
- SSP Status Register (SSPSTAT)
- Serial Receive/Transmit Buffer (SSPBUF)
- SSP Shift Register (SSPSR) Not directly accessible
- SSP Address Register (SSPADD)

The SSPCON register allows control of the I²C operation. Four mode selection bits (SSPCON<3:0>) allow one of the following I²C modes to be selected:

- I²C Slave mode (7-bit address)
- I²C Slave mode (10-bit address)
- I²C Slave mode (7-bit address), with START and STOP bit interrupts enabled
- I²C Slave mode (10-bit address), with START and STOP bit interrupts enabled
- I²C Firmware controlled Master mode, slave is idle

Selection of any I²C mode, with the SSPEN bit set, forces the SCL and SDA pins to be open drain, provided these pins are programmed to inputs by setting the appropriate TRISC bits.

The SSPSTAT register gives the status of the data transfer. This information includes detection of a START or STOP bit, specifies if the received byte was data or address, if the next byte is the completion of 10-bit address, and if this will be a read or write data transfer. The SSPSTAT register is read only.

The SSPBUF is the register to which transfer data is written to or read from. The SSPSR register shifts the data in or out of the device. In receive operations, the SSPBUF and SSPSR create a doubled buffered receiver. This allows reception of the next byte to begin before reading the last byte of received data. When the complete byte is received, it is transferred to the SSPBUF register and flag bit SSPIF is set. If another complete byte is received before the SSPBUF register is read, a receiver overflow has occurred and bit SSPOV (SSPCON<6>) is set and the byte in the SSPSR is lost.

The SSPADD register holds the slave address. In 10-bit mode, the user needs to write the high byte of the address (1111 0 A9 A8 0). Following the high byte address match, the low byte of the address needs to be loaded (A7:A0).

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9.3.1 SLAVE MODE

In Slave mode, the SCL and SDA pins must be configured as inputs (TRISC<4:3> set). The SSP module will override the input state with the output data when required (slave-transmitter).

When an address is matched or the data transfer after an address match is received, the hardware automatically will generate the Acknowledge (ACK) pulse, and then load the SSPBUF register with the received value currently in the SSPSR register.

There are certain conditions that will cause the SSP module not to give this \overline{ACK} pulse. These are if either (or both):

- The buffer full bit BF (SSPSTAT<0>) was set before the transfer was received.
- The overflow bit SSPOV (SSPCON<6>) was set before the transfer was received.

In this case, the SSPSR register value is not loaded into the SSPBUF, but bit SSPIF (PIR1<3>) is set. Table 9-3 shows what happens when a data transfer byte is received, given the status of bits BF and SSPOV. The shaded cells show the condition where user software did not properly clear the overflow condition. Flag bit BF is cleared by reading the SSPBUF register while bit SSPOV is cleared through software.

The SCL clock input must have a minimum high and low time for proper operation. The high and low times of the I²C specification as well as the requirement of the SSP module is shown in timing parameter #100 and parameter #101.

9.3.1.1 Addressing

Once the SSP module has been enabled, it waits for a START condition to occur. Following the START condition, the 8-bits are shifted into the SSPSR register. All incoming bits are sampled with the rising edge of the clock (SCL) line. The value of register SSPSR<7:1> is compared to the value of the SSPADD register. The

address is compared on the falling edge of the eighth clock (SCL) pulse. If the addresses match, and the BF and SSPOV bits are clear, the following events occur:

- The SSPSR register value is loaded into the SSPBUF register.
- b) The buffer full bit, BF is set.
- c) An ACK pulse is generated.
- d) SSP interrupt flag bit, SSPIF (PIR1<3>) is set (interrupt is generated if enabled) - on the falling edge of the ninth SCL pulse.

In 10-bit Address mode, two address bytes need to be

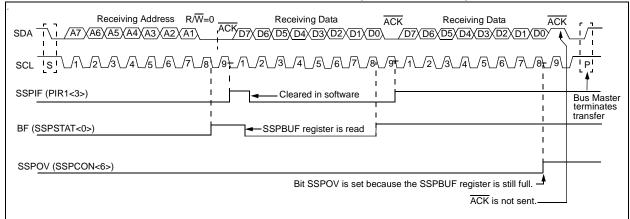
9.3.1.2 Reception

When the R/\overline{W} bit of the address byte is clear and an address match occurs, the R/\overline{W} bit of the SSPSTAT register is cleared. The received address is loaded into the SSPBUF register.

When the address byte overflow condition exists, then no Acknowledge (ACK) pulse is given. An overflow condition is defined as either bit BF (SSPSTAT<0>) is set, or bit SSPOV (SSPCON<6>) is set.

An SSP interrupt is generated for each data transfer byte. Flag bit SSPIF (PIR1<3>) must be cleared in software. The SSPSTAT register is used to determine the status of the byte.

FIGURE 9-17: I²C WAVEFORMS FOR RECEPTION (7-BIT ADDRESS)



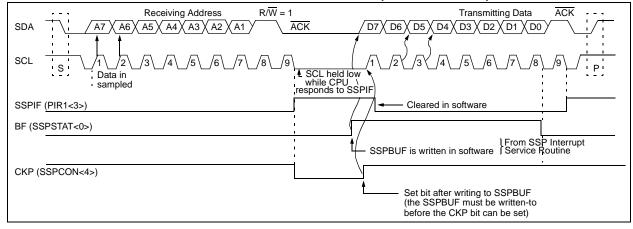
9.3.1.3 Transmission

When the R/\overline{W} bit of the incoming address byte is set and an address match occurs, the R/\overline{W} bit of the SSPSTAT register is set. The received address is loaded into the SSPBUF register. The \overline{ACK} pulse will be sent on the ninth bit, and pin RC3/SCK/SCL is held low. The transmit data must be loaded into the SSPBUF register, which also loads the SSPSR register. Then, pin RC3/SCK/SCL should be enabled by setting bit CKP (SSPCON<4>). The master must monitor the SCL pin prior to asserting another clock pulse. The slave devices may be holding off the master by stretching the clock. The eight data bits are shifted out on the falling edge of the SCL input. This ensures that the SDA signal is valid during the SCL high time (Figure 9-18).

An SSP interrupt is generated for each data transfer byte. Flag bit SSPIF must be cleared in software, and the SSPSTAT register is used to determine the status of the byte. Flag bit SSPIF is set on the falling edge of the ninth clock pulse.

As a slave-transmitter, the \overline{ACK} pulse from the master-receiver is latched on the rising edge of the ninth SCL input pulse. If the SDA line was high (not \overline{ACK}), then the data transfer is complete. When the \overline{ACK} is latched by the slave, the slave logic is reset and the slave then monitors for another occurrence of the START bit. If the SDA line was low (\overline{ACK}) , the transmit data must be loaded into the SSPBUF register, which also loads the SSPSR register. Then, pin RC3/SCK/SCL should be enabled by setting bit CKP.

FIGURE 9-18: I²C WAVEFORMS FOR TRANSMISSION (7-BIT ADDRESS)



9.3.2 MASTER MODE

Master mode of operation is supported, in firmware, using interrupt generation on the detection of the START and STOP conditions. The STOP (P) and START (S) bits are cleared from a RESET, or when the SSP module is disabled. The STOP and START bits will toggle based on the START and STOP conditions. Control of the $\rm I^2C$ bus may be taken when the P bit is set, or the bus is idle with both the S and P bits clear.

In Master mode, the SCL and SDA lines are manipulated by clearing the corresponding TRISC<4:3> bit(s). The output level is always low, irrespective of the value(s) in PORTC<4:3>. So when transmitting data, a '1' data bit must have the TRISC<4> bit set (input) and a '0' data bit must have the TRISC<4> bit cleared (output). The same scenario is true for the SCL line with the TRISC<3> bit.

The following events will cause SSP Interrupt Flag bit, SSPIF, to be set (SSP Interrupt if enabled):

- · START condition
- · STOP condition
- · Data transfer byte transmitted/received

Master mode of operation can be done with either the Slave mode idle (SSPM3:SSPM0 = 1011), or with the slave active. When both Master and Slave modes are enabled, the software needs to differentiate the source(s) of the interrupt.

9.3.3 MULTI-MASTER MODE

In Multi-Master mode, the interrupt generation on the detection of the START and STOP conditions allows the determination of when the bus is free. The STOP (P) and START (S) bits are cleared from a RESET or when the SSP module is disabled. The STOP and START bits will toggle based on the START and STOP conditions. Control of the I²C bus may be taken when bit P (SSPSTAT<4>) is set, or the bus is idle, with both the S and P bits clear. When the bus is busy, enabling the SSP interrupt will generate the interrupt when the STOP condition occurs.

In multi-master operation, the SDA line must be monitored to see if the signal level is the expected output level. This check only needs to be done when a high level is output. If a high level is expected and a low level is present, the device needs to release the SDA and SCL lines (set TRISC<4:3>). There are two stages where this arbitration can be lost, they are:

- Address Transfer
- Data Transfer

When the slave logic is enabled, the slave continues to receive. If arbitration was lost during the address transfer stage, communication to the device may be in progress. If addressed, an ACK pulse will be generated. If arbitration was lost during the data transfer stage, the device will need to re-transfer the data at a later time.

TABLE 9-4: REGISTERS ASSOCIATED WITH I²C OPERATION

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other RESETS
0Bh, 8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000u
0Ch	PIR1	LCDIF	ADIF	_	_	SSPIF	CCP1IF	TMR2IF	TMR1IF	00 0000	00 0000
8Ch	PIE1	LCDIE	ADIE	_	_	SSPIE	CCP1IE	TMR2IE	TMR1IE	00 0000	00 0000
13h	SSPBUF	Synchrono	us Serial F	Port Receiv	e Buffer/T	ransmit R	egister			xxxx xxxx	uuuu uuuu
93h	SSPADD	Synchrono	us Serial F	Port (I ² C m	ode) Addr	ess Regis	ter			0000 0000	0000 0000
14h	SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	0000 0000
94h	SSPSTAT	SMP	CKE	D/Ā	Р	S	R/W	UA	BF	0000 0000	0000 0000
87h	TRISC	_	_	PORTC D	ata Direct	ion Contro	ol Registe	r		11 1111	11 1111

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by SSP in I²C mode.

FIGURE 9-19: OPERATION OF THE I²C MODULE IN IDLE_MODE, RCV_MODE OR XMIT_MODE

```
IDLE_MODE (7-bit):
if (Addr_match)
                                {
                                        Set interrupt;
                                        if (R/\overline{W} = 1)
                                                               Send \overline{ACK} = 0;
                                                               set XMIT_MODE;
                                        else if (R/\overline{W} = 0) set RCV_MODE;
RCV_MODE:
if ((SSPBUF = Full) OR (SSPOV = 1))
                 Set SSPOV;
                 Do not acknowledge;
                  transfer SSPSR → SSPBUF;
else
                 send \overline{ACK} = 0;
Receive 8-bits in SSPSR;
Set interrupt;
XMIT_MODE:
While ((SSPBUF = Empty) AND (CKP=0)) Hold SCL Low;
Send byte;
Set interrupt;
if (\overline{ACK} Received = 1)
                                {
                                        End of transmission;
                                        Go back to IDLE_MODE;
else if ( ACK Received = 0) Go back to XMIT_MODE;
IDLE_MODE (10-Bit):
If (High_byte_addr_match AND (R/\overline{W} = 0))
                 PRIOR_ADDR_MATCH = FALSE;
                 Set interrupt;
                 if ((SSPBUF = Full) OR ((SSPOV = 1))
                                 Set SSPOV:
                                Do not acknowledge;
                                Set UA = 1;
                 else
                                Send \overline{ACK} = 0;
                                While (SSPADD not updated) Hold SCL low;
                                Clear UA = 0;
                                Receive Low_addr_byte;
                                Set interrupt;
                                Set UA = 1;
                                If (Low_byte_addr_match)
                                               PRIOR_ADDR_MATCH = TRUE;
                                               Send \overline{ACK} = 0;
                                               while (SSPADD not updated) Hold SCL low;
                                               Clear UA = 0;
                                               Set RCV_MODE;
                                        }
                         }
else if (High_byte_addr_match AND (R/\overline{W} = 1))
                 if (PRIOR_ADDR_MATCH)
                                send \overline{ACK} = 0;
                                set XMIT_MODE;
          else PRIOR_ADDR_MATCH = FALSE;
          }
```

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NOTES:

10.0 ANALOG-TO-DIGITAL CONVERTER (A/D) MODULE

The Analog-to-Digital (A/D) Converter module has five inputs.

The analog input charges a sample and hold capacitor. The output of the sample and hold capacitor is the input into the converter. The converter then generates a digital result of this analog level via successive approximation. The A/D conversion of the analog input signal results in a corresponding 10-bit digital number. The A/D module has high and low voltage reference input, that is software selectable to some combination of VDD, Vss, RA2 or RA3.

The A/D converter has a unique feature of being able to operate while the device is in SLEEP mode. To operate in SLEEP, the A/D clock must be derived from the A/D's internal RC oscillator.

The A/D module has four registers. These registers are:

- A/D Result High Register (ADRESH)
- A/D Result Low Register (ADRESL)
- A/D Control Register0 (ADCON0)
- A/D Control Register1 (ADCON1)

REGISTER 10-2: ADCON1 REGISTER (ADDRESS 9Fh)

U-0	U-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0
ADFM	_	_	_	PCFG3	PCFG2	PCFG1	PCFG0
bit 7					bit 0		

bit 7 ADFM: A/D Result Format Select bit

1 = Right justified. 6 Most Significant bits of ADRESH are read as '0'.

0 = Left justified. 6 Least Significant bits of ADRESL are read as '0'.

bit 6-4 Unimplemented: Read as '0'

bit 3-0 **PCFG<3:0>**: A/D Port Configuration Control bits:

PCFG<3:0>	AN4 RA5	AN3 RA3	AN2 RA2	AN1 RA1	AN0 RA0	VREF+	VREF-	CHAN/ Refs ⁽¹⁾
0000	Α	Α	Α	Α	Α	Vdd	Vss	5/0
0001	Α	VREF+	Α	Α	Α	RA3	Vss	4/1
0010	Α	Α	Α	Α	Α	Vdd	Vss	5/0
0011	Α	VREF+	Α	Α	Α	RA3	Vss	4/1
0100	D	Α	D	Α	Α	Vdd	Vss	3/0
0101	D	VREF+	D	Α	Α	RA3	Vss	2/1
011x	D	D	D	D	D	Vdd	Vss	0/0
1000	Α	VREF+	VREF-	Α	Α	RA3	RA2	3/2
1001	Α	Α	Α	Α	Α	Vdd	Vss	5/0
1010	Α	VREF+	Α	Α	Α	RA3	Vss	4/1
1011	Α	VREF+	VREF-	Α	Α	RA3	RA2	3/2
1100	Α	VREF+	VREF-	Α	Α	RA3	RA2	3/2
1101	D	VREF+	VREF-	Α	Α	RA3	RA2	2/2
1110	D	D	D	D	Α	Vdd	Vss	1/0
1111	D	VREF+	VREF-	D	Α	RA3	RA2	1/2

A = Analog input D = Digital I/O

Note 1: This column indicates the number of analog channels available as A/D inputs and the number of analog channels used as voltage reference inputs.

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented I	bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

The ADRESH:ADRESL registers contain the 10-bit result of the A/D conversion. When the A/D conversion is complete, the result is loaded into this A/D result register pair, the GO/DONE bit (ADCON0<2>) is cleared and the A/D interrupt flag bit ADIF is set. The block diagram of the A/D module is shown in Figure 10-1.

After the A/D module has been configured as desired, the selected channel must be acquired before the conversion is started. The analog input channels must have their corresponding TRIS bits selected as inputs. To determine sample time, see Section 10.1. After this acquisition time has elapsed, the A/D conversion can be started.

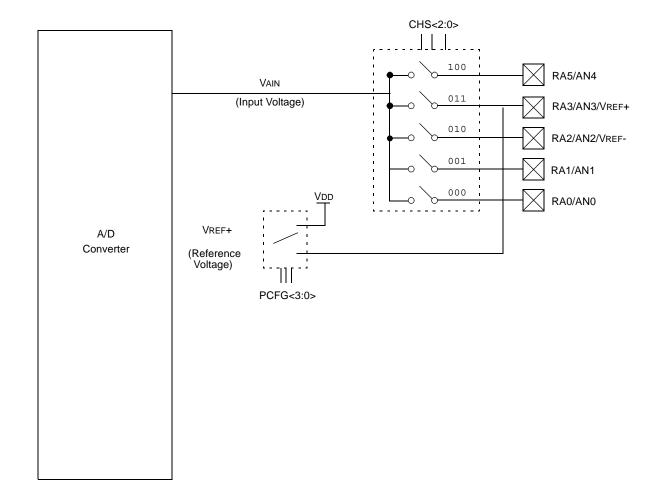
The following steps should be followed for doing an A/D conversion:

- 1. Configure the A/D module:
 - Configure analog pins/voltage reference/ and digital I/O (ADCON1)
 - Select A/D input channel (ADCON0)
 - Select A/D conversion clock (ADCON0)
 - Turn on A/D module (ADCON0)
- 2. Configure A/D interrupt (if desired):
 - · Clear ADIF bit
 - Set ADIE bit
 - · Set PEIE bit
 - · Set GIE bit
- FIGURE 10-1: A/D BLOCK DIAGRAM

- 3. Wait the required acquisition time.
- 4. Start conversion:
 - Set GO/DONE bit (ADCON0)
- 5. Wait for A/D conversion to complete, by either:
 - Polling for the GO/DONE bit to be cleared (interrupts disabled)

OR

- · Waiting for the A/D interrupt
- 6. Read A/D Result register pair (ADRESH:ADRESL), clear bit ADIF if required.
- 7. For next conversion, go to step 1 or step 2 as required. The A/D conversion time per bit is defined as TAD. A minimum wait of 2TAD is required before next acquisition starts.



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10.1 A/D Acquisition Requirements

For the A/D converter to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The analog input model is shown in Figure 10-2. The source impedance (Rs) and the internal sampling switch (Rss) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (Rss) impedance varies over the device voltage (VDD), see Figure 10-2. The maximum recommended impedance for analog sources is 10 kΩ. As the impedance is decreased, the acquisition time may

be decreased. After the analog input channel is selected (changed), this acquisition must be done before the conversion can be started.

To calculate the minimum acquisition time, Equation 10-1 may be used. This equation assumes that 1/2 LSb error is used (1024 steps for the A/D). The 1/2 LSb error is the maximum error allowed for the A/D to meet its specified resolution.

To calculate the minimum acquisition time, TACQ, see the PICmicro™ Mid-Range Reference Manual (DS33023).

EQUATION 10-1: ACQUISITION TIME EXAMPLE

TACQ = Amplifier Settling Time +

Hold Capacitor Charging Time +

Temperature Coefficient

TAMP + TC + TCOFF

= $2\mu S + TC + [(Temperature -25^{\circ}C)(0.05\mu S/^{\circ}C)]$

TC = CHOLD (RIC + RSS + RS) In(1/2047)

= $-120 \text{pF} (1 \text{k}\Omega + 7 \text{k}\Omega + 10 \text{k}\Omega) \text{In}(0.0004885)$

16.47μS

TACQ = $2\mu S + 16.47\mu S + [(50^{\circ}C - 25^{\circ}C)(0.05\mu S/^{\circ}C)]$

 $= 19.72 \mu S^{\circ}$

FIGURE 10-2: ANALOG INPUT MODEL

10.2 Selecting the A/D Conversion Clock

The A/D conversion time per bit is defined as TAD. The A/D conversion requires a minimum 12TAD per 10-bit conversion. The source of the A/D conversion clock is software selected. The four possible options for TAD are:

- 2Tosc
- 8Tosc
- 32Tosc
- Internal A/D module RC oscillator

For correct A/D conversions, the A/D conversion clock (TAD) must be selected to ensure a minimum TAD time of 1.6 $\mu s.$

Table 10-1 shows the resultant TAD times derived from the device operating frequencies and the A/D clock source selected.

TABLE 10-1: TAD VS. MAXIMUM DEVICE OPERATING FREQUENCIES (STANDARD DEVICES (C))

AD Clock	AD Clock Source (TAD)			
Operation	ADCS<1:0>	Max.		
2Tosc	00	1.25 MHz		
8Tosc	01	5 MHz		
32Tosc	10	20 MHz		
RC ^(1, 2, 3)	11	(Note 1)		

- **Note 1:** The RC source has a typical TAD time of 4 μ s, but can vary between 2-6 μ s.
 - 2: When the device frequencies are greater than 1 MHz, the RC A/D conversion clock source is only recommended for SLEEP operation.
 - **3:** For extended voltage devices (LC), please refer to the Electrical Specifications section.

10.3 Configuring Analog Port Pins

The ADCON1 and TRIS registers control the operation of the A/D port pins. The port pins that are desired as analog inputs must have their corresponding TRIS bits set (input). If the TRIS bit is cleared (output), the digital output level (VOH or VOL) will be converted.

The A/D operation is independent of the state of the CHS<2:0> bits and the TRIS bits.

- Note 1: When reading the port register, any pin configured as an analog input channel will read as cleared (a low level). Pins configured as digital inputs will convert an analog input. Analog levels on a digitally configured input will not affect the conversion accuracy.
 - 2: Analog levels on any pin that is defined as a digital input (including the AN<4:0> pins), may cause the input buffer to consume current that is out of the device specifications.

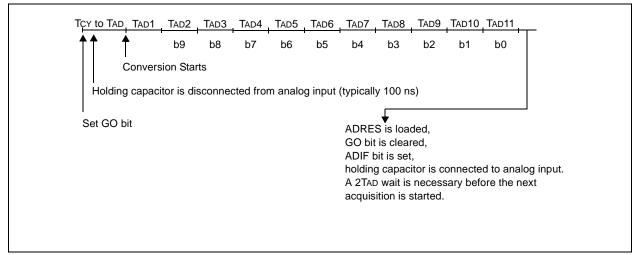
10.4 A/D Conversions

Clearing the GO/DONE bit during a conversion will abort the current conversion. The A/D result register pair will NOT be updated with the partially completed A/D conversion sample. That is, the ADRESH:ADRESL registers will continue to contain the value of the last completed conversion (or the last value written to the ADRESH:ADRESL registers). After the A/D conversion is aborted, a 2TAD wait is required before the next acquisition is started. After this 2TAD wait, acquisition on the selected channel is automatically started. After this, the GO/DONE bit can be set to start the conversion.

In Figure 10-3, after the GO bit is set, the first time segment has a minimum of TCY and a maximum of TAD.

Note: The GO/DONE bit should **NOT** be set in the same instruction that turns on the A/D.

FIGURE 10-3: A/D CONVERSION TAD CYCLES



10.4.1 A/D RESULT REGISTERS

The ADRESH:ADRESL register pair is the location where the 10-bit A/D result is loaded at the completion of the A/D conversion. This register pair is 16-bits wide. The A/D module gives the flexibility to left or right justify the 10-bit result in the 16-bit result register. The A/D Format Select bit (ADFM) controls this justification. Figure 10-4 shows the operation of the A/D result justification. The extra bits are loaded with '0's'. When an A/D result will not overwrite these locations (A/D disable), these registers may be used as two general purpose 8-bit registers.

FIGURE 10-4: A/D RESULT JUSTIFICATION

10.5 A/D Operation During SLEEP

The A/D module can operate during SLEEP mode. This requires that the A/D clock source be set to RC (ADCS<1:0> = 11). When the RC clock source is selected, the A/D module waits one instruction cycle before starting the conversion. This allows the SLEEP instruction to be executed, which eliminates all digital switching noise from the conversion. When the conversion is completed, the GO/DONE bit will be cleared and the result loaded into the ADRES register. If the A/D interrupt is enabled, the device will wake-up from SLEEP. If the A/D interrupt is not enabled, the A/D module will then be turned off, although the ADON bit will remain set.

When the A/D clock source is another clock option (not RC), a SLEEP instruction will cause the present conversion to be aborted and the A/D module to be turned off, though the ADON bit will remain set.

Turning off the A/D places the A/D module in its lowest current consumption state.

Note:	For the A/D module to operate in SLEEP,
	the A/D clock source must be set to RC
	(ADCS<1:0> = 11). To allow the conver-
	sion to occur during SLEEP, ensure the
	SLEEP instruction immediately follows the
	instruction that sets the GO/DONE bit.

10.6 Effects of a RESET

A device RESET forces all registers to their RESET state. This forces the A/D module to be turned off, and any conversion is aborted. All A/D input pins are configured as analog inputs.

The value that is in the ADRESH:ADRESL registers is not modified for a Power-on Reset. The ADRESH:ADRESL registers will contain unknown data after a Power-on Reset.

TABLE 10-2:	REGISTERS/BITS ASSOCIATED WITH A/D

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	POR, BOR	MCLR, WDT
0Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000u
0Ch	PIR1	LCDIF	ADIF	(1)	(1)	SSPIF	CCP1IF	TMR2IF	TMR1IF	r0rr 0000	r0rr 0000
8Ch	PIE1	LCDIE	ADIE	(1)	(1)	SSPIE	CCP1IE	TMR2IE	TMR1IE	r0rr 0000	r0rr 0000
1Eh	ADRESH	A/D Resu	It Registe	r High Byt	е					xxxx xxxx	uuuu uuuu
9Eh	ADRESL	A/D Resu	It Registe	r Low Byte)					xxxx xxxx	uuuu uuuu
1Fh	ADCON0	ADCS1	ADCS0	CHS2	CHS1	CHS0	GO/DONE	_	ADON	0000 00-0	0000 00-0
9Fh	ADCON1	ADFM	_	_	_	PCFG3	PCFG2	PCFG1	PCFG0	0- 0000	0- 0000
85h	TRISA	_	_	PORTA D	ORTA Data Direction Register					11 1111	11 1111
05h	PORTA	_		PORTA D	ata Latch	when wri	tten: PORTA	pins wher	n read	0x 0000	0u 0000

 $\label{eq:local_equation} \textbf{Legend:} \quad \textbf{x} = \textbf{unknown}, \ \textbf{u} = \textbf{unchanged}, \ \textbf{-} = \textbf{unimplemented}, \ \textbf{read as '0'}. \ \textbf{Shaded cells are not used for A/D conversion}.$

Note 1: These bits are reserved; always maintain these bits clear.

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NOTES:

11.0 LCD MODULE

The LCD module generates the timing control to drive a static or multiplexed LCD panel, with support for up to 32 segments multiplexed with up to four commons. It also provides control of the LCD pixel data.

The interface to the module consists of 3 control registers (LCDCON, LCDSE, and LCDPS), used to define the timing requirements of the LCD panel and up to 16 LCD data registers (LCD00-LCD15) that represent the array of the pixel data. In normal operation, the control registers are configured to match the LCD panel being used. Primarily, the initialization information consists of

selecting the number of commons required by the LCD panel, and then specifying the LCD frame clock rate to be used by the panel.

Once the module is initialized for the LCD panel, the individual bits of the LCD data registers are cleared/set to represent a clear/dark pixel, respectively.

Once the module is configured, the LCDEN (LCDCON<7>) bit is used to enable or disable the LCD module. The LCD panel can also operate during SLEEP by clearing the SLPEN (LCDCON<6>) bit.

Figure 11-2 through Figure 11-5 provides waveforms for static, half-duty cycle, one-third-duty cycle, and quarter-duty cycle drives.

REGISTER 11-1: LCDCON REGISTER (ADDRESS 10Fh)

R/W-0								
LCDEN	SLPEN	WERR	BIAS	CS1	CS0	LMUX1	LMUX0	
bit 7							bit 0	

bit 7 LCDEN: Module Drive Enable bit

1 = LCD drive enabled 0 = LCD drive disabled

bit 6 SLPEN: LCD Display Enabled to SLEEP bit

1 = LCD module will stop driving in SLEEP

0 = LCD module will continue driving in SLEEP

bit 5 WERR: Write Failed Error bit

1 = System tried to write LCDD register during disallowed time. (Must be reset in software.)

0 = No error

bit 4 BIAS: Bias Generator Enable bit

0 = Internal bias generator powered down, bias is expected to be provided externally

1 = Internal bias generator enabled, powered up

bit 3-2 CS<1:0>: Clock Source bits

00 = Fosc/256

01 = T1CKI (Timer1)

1x = Internal RC oscillator

bit 1-0 LMUX<1:0>: Common Selection bits

Specifies the number of commons

00 = Static(COM0)

01 = 1/2 (COM0, 1)

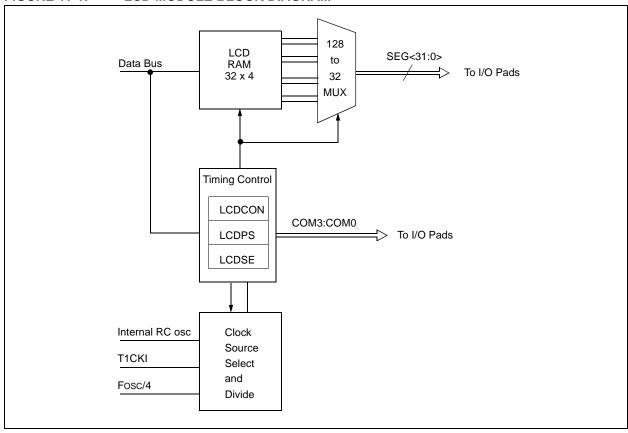
10 = 1/3 (COM0, 1, 2)

11 = 1/4 (COM0, 1, 2, 3)

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' - n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

FIGURE 11-1: LCD MODULE BLOCK DIAGRAM



REGISTER 11-2: LCDPS REGISTER (ADDRESS 10Eh)

U-0	U-0	U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0
_	_	_	_	LP3	LP2	LP1	LP0
bit 7							bit 0

bit 7-4 Unimplemented: Read as '0'

bit 3-0 LP<3:0>: Frame Clock Prescale Selection bits (see Section 11.1.2)

LMUX1:LMUX0	Multiplex	Frame Frequency
0.0	Static	Clock source/(128 * (LP3:LP0 + 1))
01	1/2	Clock source/(128 * (LP3:LP0 + 1))
10	1/3	Clock source/(96 * (LP3:LP0 + 1))
11	1/4	Clock source/(128 * (LP3:LP0 + 1))

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared x = Bit is unknown	

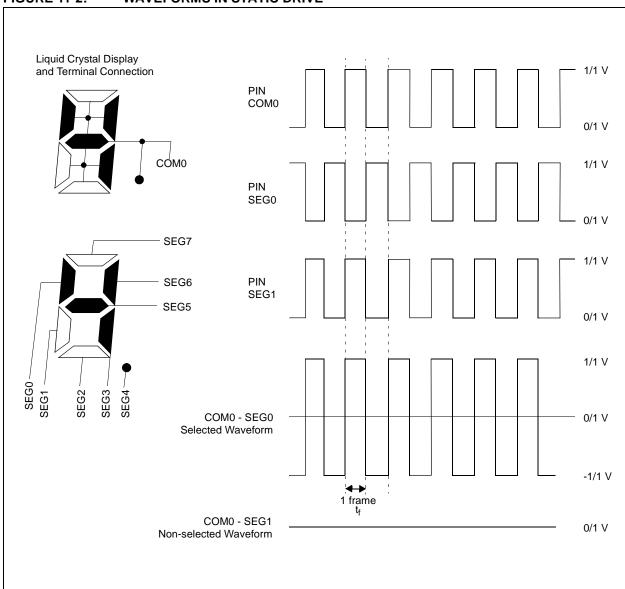
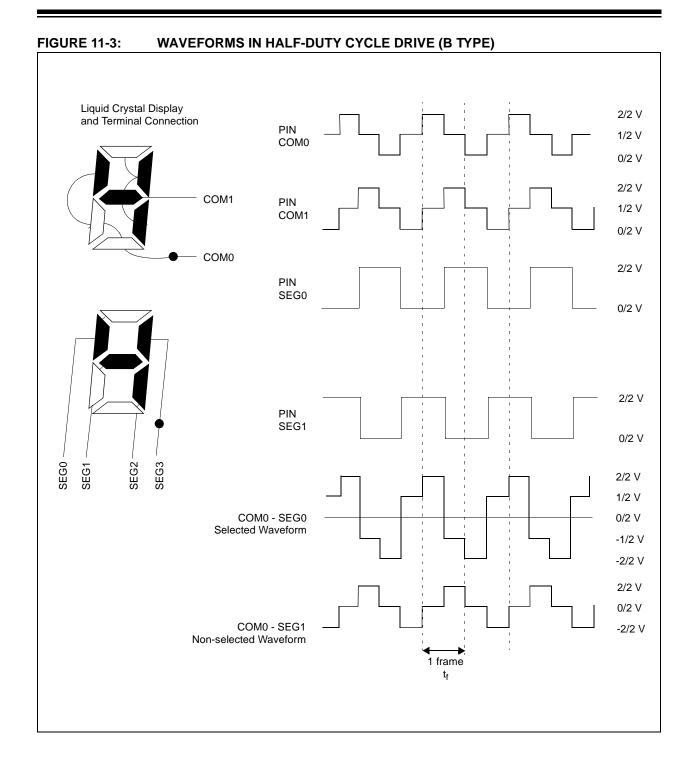


FIGURE 11-2: WAVEFORMS IN STATIC DRIVE



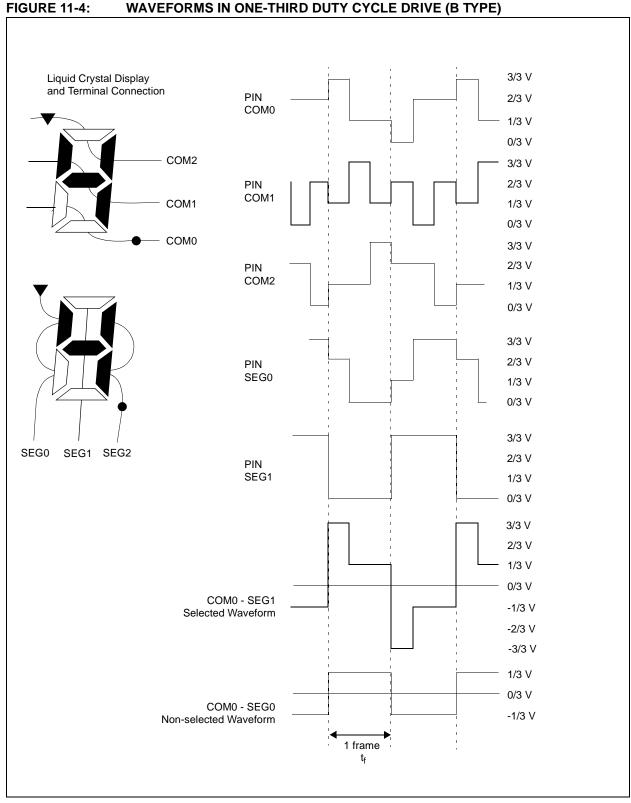
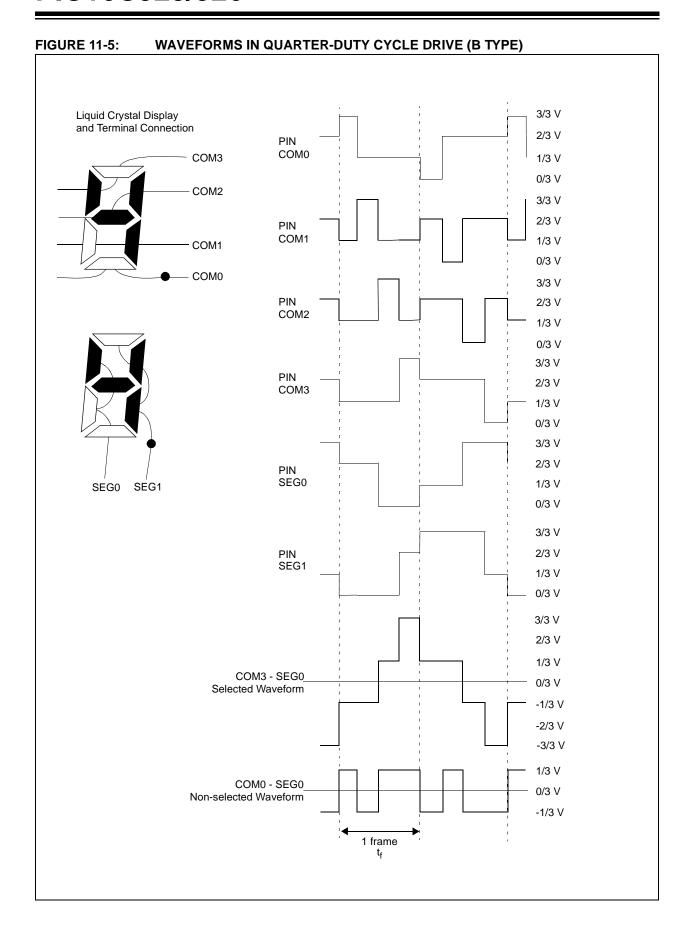


FIGURE 11-4:

Preliminary DS39544A-page 87 © 2001 Microchip Technology Inc.



11.1 LCD Timing

The LCD module has 3 possible clock source inputs and supports static, 1/2, 1/3, and 1/4 multiplexing.

11.1.1 TIMING CLOCK SOURCE SELECTION

The clock sources for the LCD timing generation are:

- · Internal RC oscillator
- Timer1 oscillator
- · System clock divided by 256

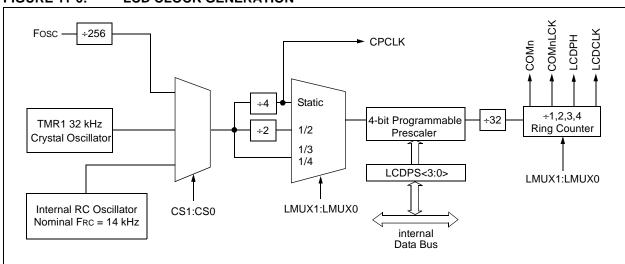
The first timing source is an internal RC oscillator which runs at a nominal frequency of 14 kHz. This oscillator provides a lower speed clock which may be used to continue running the LCD while the processor is in SLEEP. The RC oscillator will power-down when it is not selected or when the LCD module is disabled.

The second source is the Timer1 external oscillator. This oscillator provides a lower speed clock which may be used to continue running the LCD while the processor is in SLEEP. It is assumed that the frequency provided on this oscillator will be 32 kHz. To use the Timer1 oscillator as a LCD module clock source, it is only necessary to set the T1OSCEN (T1CON<3>) bit.

The third source is the system clock divided by 256. This divider ratio is chosen to provide about 32 kHz output when the external oscillator is 8 MHz. The divider is not programmable. Instead the LCDPS register is used to set the LCD frame clock rate.

All of the clock sources are selected with bits CS1:CS0 (LCDCON<3:2>). Refer to Register 11-1 for details of the register programming.

FIGURE 11-6: LCD CLOCK GENERATION



11.1.2 MULTIPLEX TIMING GENERATION

The timing generation circuitry will generate one to four common clocks based on the display mode selected. The mode is specified by bits LMUX1:LMUX0 (LCDCON<1:0>). Table 11-1 shows the formulas for calculating the frame frequency.

TABLE 11-1: FRAME FREQUENCY FORMULAS

Multiplex	Frame Frequency =
Static	Clock source/(128 * (LP3:LP0 + 1))
1/2	Clock source/(128 * (LP3:LP0 + 1))
1/3	Clock source/(96 * (LP3:LP0 + 1))
1/4	Clock source/(128 * (LP3:LP0 + 1))

TABLE 11-2: APPROXIMATE FRAME
FREQUENCY (IN Hz)
USING TIMER1 @ 32.768 kHz
OR Fosc @ 8 MHz

LP3:LP0	Static	1/2	1/3	1/4
2	85	85	114	85
3	64	64	85	64
4	51	51	68	51
5	43	43	57	43
6	37	37	49	37
7	32	32	43	32

TABLE 11-3: APPROXIMATE FRAME FREQUENCY (IN Hz)
USING INTERNAL RC OSC
@ 14 kHz

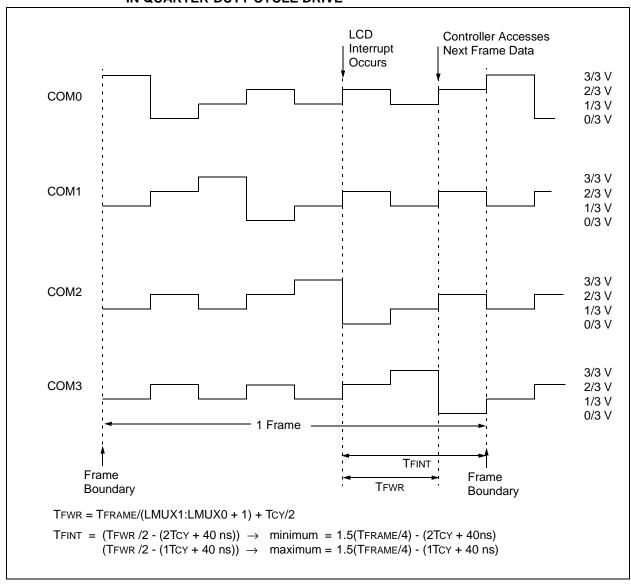
LP3:LP0	Static	1/2	1/3	1/4
0	109	109	146	109
1	55	55	73	55
2	36	36	49	36
3	27	27	36	27

11.2 LCD Interrupts

The LCD timing generation provides an interrupt that defines the LCD frame timing. This interrupt can be used to coordinate the writing of the pixel data with the start of a new frame. Writing pixel data at the frame boundary allows a visually crisp transition of the image. This interrupt can also be used to synchronize external events to the LCD. For example, the interface to an external segment driver, such as a Microchip AY0438, can be synchronized for segment data update to the LCD frame.

A new frame is defined to begin at the leading edge of the COM0 common signal. The interrupt will be set immediately after the LCD controller completes accessing all pixel data required for a frame. This will occur at a fixed interval before the frame boundary (TFINT), as shown in Figure 11-7. The LCD controller will begin to access data for the next frame within the interval from the interrupt to when the controller begins to access data after the interrupt (TFWR). New data must be written within TFWR, as this is when the LCD controller will begin to access the data for the next frame.

FIGURE 11-7: EXAMPLE WAVEFORMS AND INTERRUPT TIMING IN QUARTER-DUTY CYCLE DRIVE



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11.3 Pixel Control

11.3.1 LCDD (PIXEL DATA) REGISTERS

The pixel registers contain bits which define the state of each pixel. Each bit defines one unique pixel.

Table 11-4 shows the correlation of each bit in the LCDD registers to the respective common and segment signals.

Any LCD pixel location not being used for display can be used as general purpose RAM.

REGISTER 11-3: GENERIC LCDD REGISTER LAYOUT

| R/W-x |
|-------|-------|-------|-------|-------|-------|-------|-------|
| SEGs |
| COMc |
| bit 7 | | | | | | | bit 0 |

bit 7-0 **SEGsCOMc**: Pixel Data bit for Segment S and Common C

1 = Pixel on (dark)
0 = Pixel off (clear)

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented	l bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

11.4 Operation During SLEEP

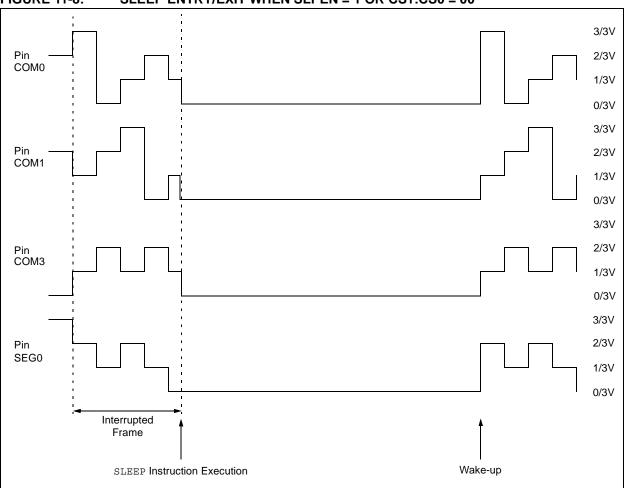
The LCD module can operate during SLEEP. The selection is controlled by bit SLPEN (LCDCON<6>). Setting the SLPEN bit allows the LCD module to go to SLEEP. Clearing the SLPEN bit allows the module to continue to operate during SLEEP.

If a SLEEP instruction is executed and SLPEN = '1', the LCD module will cease all functions and go into a very low current consumption mode. The module will stop operation immediately and drive the minimum LCD voltage on both segment and common lines. Figure 11-8 shows this operation. To ensure that the LCD completes the frame, the SLEEP instruction should be executed immediately after a LCD frame boundary. The LCD interrupt can be used to determine the frame boundary. See Section 11.2 for the formulas to calculate the delay.

If a SLEEP instruction is executed and SLPEN = '0', the module will continue to display the current contents of the LCDD registers. To allow the module to continue operation while in SLEEP, the clock source must be either the internal RC oscillator or Timer1 external oscillator. While in SLEEP, the LCD data cannot be changed. The LCD module current consumption will not decrease in this mode, however, the overall consumption of the device will be lower due to shut-down of the core and other peripheral functions.

Note: The internal RC oscillator or external Timer1 oscillator must be used to operate the LCD module during SLEEP.





11.4.1 SEGMENT ENABLES

The LCDSE register is used to select the pin function for groups of pins. The selection allows each group of pins to operate as either LCD drivers or digital only pins. To configure the pins as a digital port, the corresponding bits in the LCDSE register must be cleared.

If the pin is a digital I/O the corresponding TRIS bit controls the data direction. Any bit set in the LCDSE register overrides any bit settings in the corresponding TRIS register.

- **Note 1:** On a Power-on Reset, these pins are configured as LCD drivers.
 - 2: The LMUX1:LMUX0 takes precedence over the LCDSE bit settings for pins RD7, RD6 and RD5.

EXAMPLE 11-1: STATIC MUX WITH 32 SEGMENTS

BCF STATUS,RPO ;Select Bank 2
BSF STATUS,RP1 ;
BCF LCDCON,LMUX1 ;Select Static MUX
BCF LCDCON,LMUX0 ;
MOVLW 0xFF ;Make PortD,E,F,G
MOVWF LCDSE ;LCD pins
. . . ;configure rest of LCD

EXAMPLE 11-2: ONE-THIRD DUTY CYCLE WITH 13 SEGMENTS

BCF	STATUS, RP0	;Select Bank 2
BSF	STATUS, RP1	;
BSF	LCDCON, LMUX1	;Select 1/3 MUX
BCF	LCDCON, LMUX0	;
MOVLW	0x87	;Make PORTD<7:0> &
MOVWF	LCDSE	;PORTE<6:0> LCD pins
		;configure rest of LCD

REGISTER 11-4: LCDSE REGISTER (ADDRESS 10Dh)

| R/W-1 |
|-------|-------|-------|-------|-------|-------|-------|-------|
| SE29 | SE27 | SE20 | SE16 | SE12 | SE9 | SE5 | SE0 |
| bit 7 | | | | | | | bit 0 |

- bit 7 SE29: Pin Function Select RD7/COM1/SEG31 RD5/COM3/SEG29
 - 1 = Pins have LCD drive function
 - 0 = Pins have digital Input function
- bit 6 SE27: Pin Function Select RG7/SEG28 and RE7/SEG27
 - 1 = Pins have LCD drive function
 - 0 = Pins have LCD drive function
- bit 5 SE20: Pin Function Select RG6/SEG26 RG0/SEG20
 - 1 = Pins have LCD drive function
 - 0 = Pins have digital Input function
- bit 4 SE16: Pin Function Select RF7/SEG19 RF4/SEG16
 - 1 = Pins have LCD drive function
 - 0 = Pins have digital Input function
- bit 3 SE12: Pin Function Select RF3/SEG15 RF0/SEG12
 - 1 = Pins have LCD drive function
 - 0 = Pins have digital Input function
- bit 2 SE9: Pin Function Select RE6/SEG11 RE4/SEG09
 - 1 = Pins have LCD drive function
 - 0 = Pins have digital Input function
- bit 1 SE5: Pin Function Select RE3/SEG08 RE0/SEG05
 - 1 = Pins have LCD drive function
 - 0 = Pins have digital Input function
- bit 0 SE0: Pin Function Select RD4/SEG04 RD0/SEG00
 - 1 = Pins have LCD drive function
 - 0 = Pins have digital Input function

Legend	l:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' - n = Value at POR '1' = Bit is set '0' = Bit is cleared <math>x = Bit is unknown

11.5 Voltage Generation

There are two methods for LCD voltage generation: internal charge pump, or external resistor ladder.

11.5.1 CHARGE PUMP

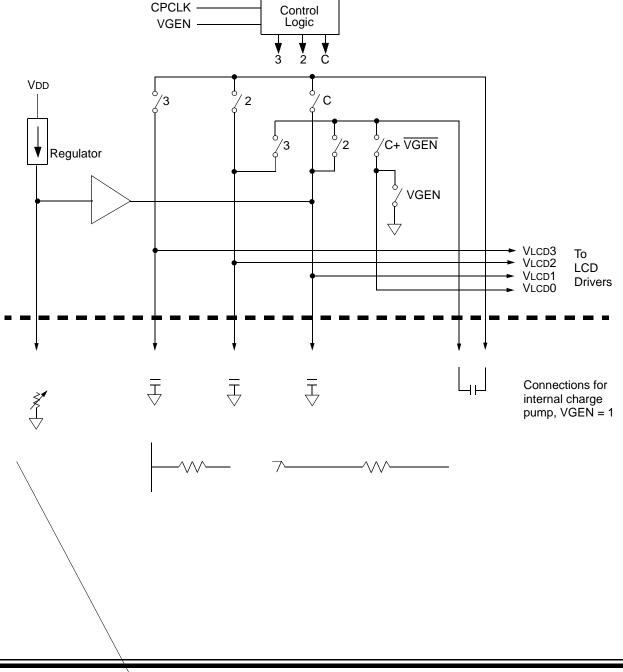
The LCD charge pump is shown in Figure 11-9. The 1.0V - 2.3V regulator will establish a stable base voltage from the varying battery voltage. This regulator is adjustable through the range by connecting a variable external resistor from VLCDADJ to ground. The potentiometer provides contrast adjustment for the LCD. This base voltage is connected to VLCD1 on the charge

pump. The charge pump boosts VLCD1 into VLCD2 = 2*VLCD1 and VLCD3 = 3*VLCD1. When the charge pump is not operating, Vlcd3 will be internally tied to VDD. See the Electrical Specifications section for charge pump capacitor and potentiometer values.

11.5.2 EXTERNAL R-LADDER

The LCD module can also use an external resistor ladder (R-Ladder) to generate the LCD voltages. Figure 11-9 shows external connections for static and 1/3 bias. The VGEN (LCDCON<4>) bit must be cleared to use an external R-Ladder.

FIGURE 11-9: CHARGE PUMP AND RESISTOR LADDER



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11.6 Configuring the LCD Module

The following is the sequence of steps to follow to configure the LCD module.

- 1. Select the frame clock prescale using bits LP3:LP0 (LCDPS<3:0>).
- 2. Configure the appropriate pins to function as segment drivers using the LCDSE register.
- 3. Configure the LCD module for the following using the LCDCON register:
 - Multiplex mode and Bias, bits LMUX1:LMUX0
 - Timing source, bits CS1:CS0
 - Voltage generation, bit VGEN
 - SLEEP mode, bit SLPEN

- Write initial values to pixel data registers, LCDD00 through LCDD15.
- 5. Clear LCD interrupt flag, LCDIF (PIR1<7>), and if desired, enable the interrupt by setting bit LCDIE (PIE1<7>).
- Enable the LCD module, by setting bit LCDEN (LCDCON<7>).

TABLE 11-4: SUMMARY OF REGISTERS ASSOCIATED WITH THE LCD MODULE

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other RESETS
0Bh, 8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000u
0Ch	PIR1	LCDIF	ADIF	_	_	SSPIF	CCP1IF	TMR2IF	TMR1IF	00 0000	00 0000
8Ch	PIE1	LCDIE	ADIE	1	_	SSPIE	CCP1IE	TMR2IE	TMR1IE	00 0000	00 0000
10h	T1CON	_	1	T1CKPS1	T1CKPS0	T10SCEN	T1SYNC	TMR1CS	TMR10N	00 0000	uu uuuu
110h	LCDD00	SEG07 COM0	SEG06 COM0	SEG05 COM0	SEG04 COM0	SEG03 COM0	SEG02 COM0	SEG01 COM0	SEG00 COM0	xxxx xxxx	uuuu uuuu
111h	LCDD01	SEG15 COM0	SEG14 COM0	SEG13 COM0	SEG12 COM0	SEG11 COM0	SEG10 COM0	SEG09 COM0	SEG08 COM0	xxxx xxxx	uuuu uuuu
112h	LCDD02	SEG23 COM0	SEG22 COM0	SEG21 COM0	SEG20 COM0	SEG19 COM0	SEG18 COM0	SEG17 COM0	SEG16 COM0	xxxx xxxx	uuuu uuuu
113h	LCDD03	SEG31 COM0	SEG30 COM0	SEG29 COM0	SEG28 COM0	SEG27 COM0	SEG26 COM0	SEG25 COM0	SEG24 COM0	xxxx xxxx	uuuu uuuu
114h	LCDD04	SEG07 COM1	SEG06 COM1	SEG05 COM1	SEG04 COM1	SEG03 COM1	SEG02 COM1	SEG01 COM1	SEG00 COM1	xxxx xxxx	uuuu uuuu
115h	LCDD05	SEG15 COM1	SEG14 COM1	SEG13 COM1	SEG12 COM1	SEG11 COM1	SEG10 COM1	SEG09 COM1	SEG08 COM1	xxxx xxxx	uuuu uuuu
116h	LCDD06	SEG23 COM1	SEG22 COM1	SEG21 COM1	SEG20 COM1	SEG19 COM1	SEG18 COM1	SEG17 COM1	SEG16 COM1	xxxx xxxx	uuuu uuuu
117h	LCDD07	SEG31 COM1 ⁽¹⁾	SEG30 COM1	SEG29 COM1	SEG28 COM1	SEG27 COM1	SEG26 COM1	SEG25 COM1	SEG24 COM1	xxxx xxxx	uuuu uuuu
118h	LCDD08	SEG07 COM2	SEG06 COM2	SEG05 COM2	SEG04 COM2	SEG03 COM2	SEG02 COM2	SEG01 COM2	SEG00 COM2	xxxx xxxx	uuuu uuuu
119h	LCDD09	SEG15 COM2	SEG14 COM2	SEG13 COM2	SEG12 COM2	SEG11 COM2	SEG10 COM2	SEG09 COM2	SEG08 COM2	xxxx xxxx	uuuu uuuu
11Ah	LCDD10	SEG23 COM2	SEG22 COM2	SEG21 COM2	SEG20 COM2	SEG19 COM2	SEG18 COM2	SEG17 COM2	SEG16 COM2	xxxx xxxx	uuuu uuuu
11Bh	LCDD11	SEG31 COM2 ⁽¹⁾	SEG30 COM2 ⁽¹⁾	SEG29 COM2	SEG28 COM2	SEG27 COM2	SEG26 COM2	SEG25 COM2	SEG24 COM2	xxxx xxxx	uuuu uuuu
11Ch	LCDD12	SEG07 COM3	SEG06 COM3	SEG05 COM3	SEG04 COM3	SEG03 COM3	SEG02 COM3	SEG01 COM3	SEG00 COM3	xxxx xxxx	uuuu uuuu
11Dh	LCDD13	SEG15 COM3	SEG14 COM3	SEG13 COM3	SEG12 COM3	SEG11 COM3	SEG10 COM3	SEG09 COM3	SEG08 COM3	xxxx xxxx	uuuu uuuu
11Eh	LCDD14	SEG23 COM3	SEG22 COM3	SEG21 COM3	SEG20 COM3	SEG19 COM3	SEG18 COM3	SEG17 COM3	SEG16 COM3	xxxx xxxx	uuuu uuuu
11Fh	LCDD15	SEG31 COM3 ⁽¹⁾	SEG30 COM3 ⁽¹⁾	SEG29 COM3 ⁽¹⁾	SEG28 COM3	SEG27 COM3	SEG26 COM3	SEG25 COM3	SEG24 COM3	xxxx xxxx	uuuu uuuu
10Dh	LCDSE	SE29	SE27	SE20	SE16	SE12	SE9	SE5	SE0	1111 1111	1111 1111
10Eh	LCDPS	_	_	_	_	LP3	LP2	LP1	LP0	0000	0000
10Fh	LCDCON	LCDEN	SLPEN	_	VGEN	CS1	CS0	LMUX1	LMUX0	00-0 0000	00-0 0000

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the LCD module.

Note 1: These pixels do not display, but can be used as general purpose RAM.

12.0 SPECIAL FEATURES OF THE CPU

What sets a microcontroller apart from other processors are special circuits to deal with the needs of real time applications. The PIC16CXXX family has a host of such features, intended to maximize system reliability, minimize cost through elimination of external components, provide power saving operating modes and offer code protection. These are:

- · Oscillator Selection
- RESET
 - Power-on Reset (POR)
 - Power-up Timer (PWRT)
 - Oscillator Start-up Timer (OST)
 - Brown-out Reset (BOR)
- Interrupts
- Watchdog Timer (WDT)
- SLEEP
- Code Protection
- ID Locations
- · In-Circuit Serial Programming

The PIC16CXXX has a Watchdog Timer which can be shut-off only through configuration bits. It runs off its own RC oscillator for added reliability.

There are two timers that offer necessary delays on power-up. One is the Oscillator Start-up Timer (OST), intended to keep the chip in RESET until the crystal oscillator is stable. The other is the Power-up Timer (PWRT), which provides a fixed delay of 72 ms (nominal) on power-up only, designed to keep the part in

RESET while the power supply stabilizes. With these two timers on-chip, most applications need no external RESET circuitry.

SLEEP mode is designed to offer a very low current power-down mode. The user can wake-up from SLEEP through external RESET, Watchdog Timer Wake-up, or through an interrupt.

Several oscillator options are also made available to allow the part to fit the application. The RC oscillator option saves system cost, while the LP crystal option saves power. A set of configuration bits are used to select various options.

12.1 Configuration Bits

The configuration bits can be programmed (read as '0'), or left unprogrammed (read as '1'), to select various device configurations. These bits are mapped in program memory location 2007h.

The user will note that address 2007h is beyond the user program memory space and can be accessed only during programming.

REGISTER 12-1: **CONFIGURATION WORD (ADDRESS 2007h)**

—	_	1		1	BOREN	CP1	CP0	PWRTE	WDTE	F0SC1	F0SC0	
bit13											bit0	

bit0

bit 13-7 Unimplemented

bit 6 BOREN: Brown-out Reset Enable bit

> 1 = BOR enabled 0 = BOR disabled

bit 5-4 CP1:CP0: Program Memory Code Protection bits

PIC16C926 (8K program memory):

11 = Code protection off

10 = 0000h to 0FFFh code protected (1/2 protected)

01 = 0000h to 1EFFh code protected (all but last 256 protected)

00 = 0000h to 1FFFh code protected (all protected)

PIC16C925 (4K program memory):

11 = Code protection off

10 = 0000h to 07FFh code protected (1/2 protected)

01 = 0000h to 0EFFh code protected (all but last 256 protected)

00 = 0000h to 0FFFh code protected (all protected) 1000h to 1FFFh wraps around to 0000h to 0FFFh

PWRTE: Power-up Timer Enable bit bit 3

> 1 = PWRT disabled 0 = PWRT enabled

WDTE: Watchdog Timer Enable bit bit 2

> 1 = WDT enabled 0 = WDT disabled

FOSC1:FOSC0: Oscillator Selection bits bit 1-0

> 11 = RC oscillator 10 = HS oscillator 01 = XT oscillator 00 = LP oscillator

12.2 Oscillator Configurations

12.2.1 OSCILLATOR TYPES

The PIC16CXXX can be operated in four different oscillator modes. The user can program two configuration bits (FOSC1 and FOSC0) to select one of these four modes:

LP Low Power CrystalXT Crystal/Resonator

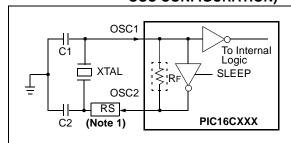
HS High Speed Crystal/Resonator

• RC Resistor/Capacitor

12.2.2 CRYSTAL OSCILLATOR/CERAMIC RESONATORS

In XT, LP, or HS modes a crystal or ceramic resonator is connected to the OSC1/CLKIN and OSC2/CLKOUT pins to establish oscillation (Figure 12-1). The PIC16CXXX oscillator design requires the use of a parallel cut crystal. Use of a series cut crystal may give a frequency out of the crystal manufacturers specifications. When in XT, LP, or HS modes, the device can have an external clock source to drive the OSC1/CLKIN pin (Figure 12-2).

FIGURE 12-1: CRYSTAL/CERAMIC
RESONATOR OPERATION
(HS, XT OR LP
OSC CONFIGURATION)



See Table 12-1 and Table 12-2 for recommended values of C1 and C2.

Note 1: A series resistor may be required for AT strip cut crystals.

FIGURE 12-2: EXTERNAL CLOCK INPUT OPERATION (HS, XT OR LP OSC CONFIGURATION)

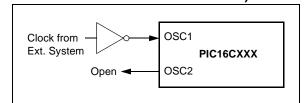


TABLE 12-1: CERAMIC RESONATORS

Ranges Tested:											
Mode	Freq.	C1	\\[C2								
XT	455 kHz 2.0 MHz	68 - 100 pF 15 - 68 pF 15 - 68 pF	68 - 100 pF 15 - 68 pF 15 - 68 pF								
#\$	80 MHz	10 - 68 pF	10 - 68 pF								
These values are for design guidance only. See notes following Table 12-2.											

TABLE 12-2: CAPACITOR SELECTION FOR CRYSTAL OSCILLATOR

Osc Type	Crystal Freq.	Freq. C1		
LP	32 kHz	33 pF	33°pF	
	200 kHz	15 pF	\\\ ph	
	200 kHz	47-68 pt	[≥] 47-68 pF	
XT	1 MHz	\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\	15 pF	
	AMH2	15 pF	15 pF	
	4 MHz	15 pF	15 pF	
\J\0	8 MHz	15-33 pF	15-33 pF	

These values are for design guidance only. See notes following this table.

- **Note 1:** Recommended ranges of C1 and C2 are depicted in Table 12-1.
 - **2:** Higher capacitance increases the stability of the oscillator, but also increases the start-up time.
 - 3: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
 - **4:** Rs may be required in HS mode, as well as XT mode, to avoid overdriving crystals with low drive level specification.

12.2.3 EXTERNAL CRYSTAL OSCILLATOR CIRCUIT

Either a prepackaged oscillator can be used, or a simple oscillator circuit with TTL gates can be built. Prepackaged oscillators provide a wide operating range and better stability. A well designed crystal oscillator will provide good performance with TTL gates. Two types of crystal oscillator circuits can be used: one with series resonance, or one with parallel resonance.

Figure 12-3 shows implementation of a parallel resonant oscillator circuit. The circuit is designed to use the fundamental frequency of the crystal. The 74AS04 inverter performs the 180-degree phase shift that a parallel oscillator requires. The 4.7 k Ω resistor provides the negative feedback for stability. The 10 k Ω potentiometer biases the 74AS04 in the linear region. This could be used for external oscillator designs.

FIGURE 12-3: EXTERNAL PARALLEL RESONANT CRYSTAL OSCILLATOR CIRCUIT

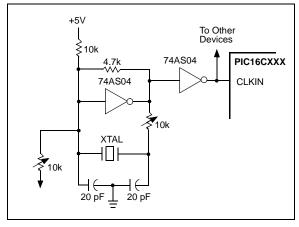
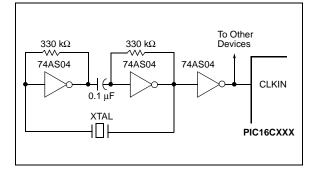


Figure 12-4 shows a series resonant oscillator circuit. This circuit is also designed to use the fundamental frequency of the crystal. The inverter performs a 180-degree phase shift in a series resonant oscillator circuit. The 330 $k\Omega$ resistors provide the negative feedback to bias the inverters in their linear region.

FIGURE 12-4: EXTERNAL SERIES RESONANT CRYSTAL OSCILLATOR CIRCUIT



12.2.4 RC OSCILLATOR

For timing insensitive applications, the "RC" device option offers additional cost savings. The RC oscillator frequency is a function of the supply voltage, the resistor (REXT) and capacitor (CEXT) values, and the operating temperature. In addition to this, the oscillator frequency will vary from unit to unit due to normal process parameter variation. Furthermore, the difference in lead frame capacitance between package types will also affect the oscillation frequency, especially for low CEXT values. The user also needs to take into account variation due to tolerance of external R and C components used. Figure 12-5 shows how the R/C combination is connected to the PIC16CXXX. For REXT values below 2.2 k Ω , the oscillator operation may become unstable, or stop completely. For very high REXT values (e.g. 1 $M\Omega$), the oscillator becomes sensitive to noise, humidity and leakage. Thus, we recommend to keep Rext between 3 k Ω and 100 k Ω .

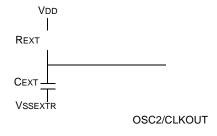
Although the oscillator will operate with no external capacitor (CEXT = 0 pF), we recommend using values above 20 pF for noise and stability reasons. With no or small external capacitance, the oscillation frequency can vary dramatically due to changes in external capacitances, such as PCB trace capacitance, or package lead frame capacitance.

See characterization data for desired device for RC frequency variation from part to part, due to normal process variation. The variation is larger for larger R (since leakage current variation will affect RC frequency more for large R) and for smaller C (since variation of input capacitance will affect RC frequency more).

See characterization data for desired device for variation of oscillator frequency, due to VDD for given REXT/CEXT values, as well as frequency variation due to operating temperature for given R, C, and VDD values.

The oscillator frequency, divided by 4, is available on the OSC2/CLKOUT pin, and can be used for test purposes or to synchronize other logic (see Figure 1-2 for waveform).

FIGURE 12-5: RC OSCILLATOR MODE



12.3 RESET

The PIC16C9XX differentiates between various kinds of RESET:

- Power-on Reset (POR)
- MCLR Reset during normal operation
- MCLR Reset during SLEEP
- WDT Reset (normal operation)
- Brown-out Reset (BOR)

Some registers are not affected in any RESET condition; their status is unknown on POR and unchanged in any other RESET. Most other registers are reset to a "RESET state" on Power-on Reset (POR), on the MCLR and WDT Reset, and on MCLR Reset during

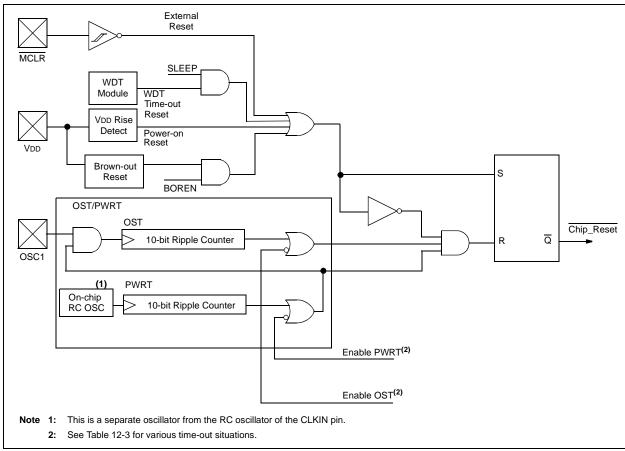
SLEEP. They are not affected by a WDT Wake-up, which is viewed as the resumption of normal operation. The $\overline{\text{TO}}$ and $\overline{\text{PD}}$ bits are set or cleared differently in different RESET situations, as indicated in Table 12-4. These bits are used in software to determine the nature of the RESET. See Table 12-6 for a full description of RESET states of all registers.

A simplified block diagram of the On-Chip Reset Circuit is shown in Figure 12-6.

The devices all have a $\overline{\text{MCLR}}$ noise filter in the $\overline{\text{MCLR}}$ Reset path. The filter will detect and ignore small pulses.

It should be noted that a WDT Reset does not drive $\overline{\text{MCLR}}$ pin low.

FIGURE 12-6: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



12.4 Power-on Reset (POR), Power-up Timer (PWRT), Brown-out Reset (BOR) and Oscillator Start-up Timer (OST)

12.4.1 POWER-ON RESET (POR)

A Power-on Reset pulse is generated on-chip when VDD rise is detected (in the range of 1.5V - 2.1V). To take advantage of the POR, just tie the MCLR pin directly (or through a resistor) to VDD. This will eliminate external RC components usually needed to create a Power-on Reset. A maximum rise time for VDD is specified. See Electrical Specifications for details.

When the device starts normal operation (exits the RESET condition), device operating parameters (voltage, frequency, temperature, etc.) must be met to ensure operation. If these conditions are not met, the device must be held in RESET until the operating conditions are met.

For additional information, refer to Application Note AN607, "Power-up Trouble Shooting."

12.4.2 POWER-UP TIMER (PWRT)

The Power-up Timer provides a fixed 72 ms nominal time-out on power-up only, from the POR. The Power-up Timer operates on an internal RC oscillator. The chip is kept in RESET as long as the PWRT is active. The PWRT's time delay allows VDD to rise to an acceptable level. A configuration bit is provided to enable/disable the PWRT.

The power-up time delay will vary from chip to chip due to VDD, temperature, and process variation. See DC parameters for details.

12.4.3 OSCILLATOR START-UP TIMER (OST)

The Oscillator Start-up Timer (OST), if enabled, provides 1024 oscillator cycle (from OSC1 input) delay after the PWRT delay (if the PWRT is enabled). This helps to ensure that the crystal oscillator or resonator has started and stabilized.

The OST time-out is invoked only for XT, LP and HS modes and only on Power-on Reset or wake-up from SLEEP.

12.4.4 BROWN-OUT RESET (BOR)

The configuration bit, BOREN, can enable or disable the Brown-out Reset circuit. If VDD falls below VBOR (parameter D005, about 4V) for longer than TBOR (parameter #35, about 100 μ S), the brown-out situation will reset the device. If VDD falls below VBOR for less than TBOR, a RESET may not occur.

Once the brown-out occurs, the device will remain in Brown-out Reset until VDD rises above VBOR. The Power-up Timer, if enabled, then keeps the device in RESET for TPWRT (parameter #33, about 72mS). If VDD should fall below VBOR during TPWRT, the Brown-out Reset process will restart when VDD rises above VBOR with the Power-up Timer Reset. The Power-up Timer is enabled separately from Brown-out Reset.

12.4.5 TIME-OUT SEQUENCE

On power-up, the time-out sequence is as follows: First, PWRT time-out is invoked after the POR time delay has expired. Then, OST is activated. The total time-out will vary based on oscillator configuration and the status of the PWRT. For example, in RC mode with the PWRT disabled, there will be no time-out at all. Figure 12-7, Figure 12-8, and Figure 12-9 depict time-out sequences on power-up.

Since the time-outs occur from the POR pulse, if MCLR is kept low long enough, the time-outs will expire. Then bringing MCLR high will begin execution immediately (Figure 12-8). This is useful for testing purposes or to synchronize more than one PIC16CXXX device operating in parallel.

Table 12-5 shows the RESET conditions for some special function registers, while Table 12-6 shows the RESET conditions for all the registers.

12.4.6 POWER CONTROL/STATUS REGISTER (PCON)

The Power Control/Status Register, PCON, has up to two bits depending upon the device.

Bit0 is Brown-out Reset Status bit, BOR. Bit BOR is unknown on a Power-on Reset. It must then be set by the user and checked on subsequent RESETS to see if bit BOR cleared, indicating a BOR occurred. When the Brown-out Reset is disabled, the state of the BOR bit is unpredictable and is, therefore, not valid at any time.

Bit1 is Power-on Reset Status bit POR. It is cleared on a Power-on Reset and unaffected otherwise. The user must set this bit following a Power-on Reset.

TABLE 12-3: TIME-OUT IN VARIOUS SITUATIONS

Oscillator Configuration	Power-up		Woke up from SI EED	
Oscillator Configuration	PWRTE = 1	PWRTE = 0	Wake-up from SLEEP	
XT, HS, LP	1024Tosc	72 ms + 1024Tosc	1024 Tosc	
RC	_	72 ms	_	

TABLE 12-4: STATUS BITS AND THEIR SIGNIFICANCE

POR	BOR	ТО	PD	Condition	
0	х	1	1	Power-on Reset	
0	х	0	х	Illegal, TO is set on POR	
0	х	х	0	Illegal, PD is set on POR	
1	0	1	1	Brown-out Reset	
1	1	0	1	WDT Reset	
1	1	0	0	WDT Wake-up	
1	1	u	u	MCLR Reset during normal operation	
1	1	1	0	MCLR Reset during SLEEP or interrupt wake-up from SLEEP	

TABLE 12-5: RESET CONDITION FOR SPECIAL REGISTERS

Condition	Program Counter	STATUS Register	PCON Register
Power-on Reset	000h	0001 1xxx	0x
MCLR Reset during normal operation	000h	000u uuuu	uu
MCLR Reset during SLEEP	000h	0001 0uuu	uu
WDT Reset	000h	0000 1uuu	uu
WDT Wake-up	PC + 1	uuu0 0uuu	uu
Brown-out Reset	000h	0001 1uuu	u0
Interrupt wake-up from SLEEP	PC + 1 ⁽¹⁾	uuu1 0uuu	uu

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0'.

Note 1: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

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TABLE 12-6: INITIALIZATION CONDITIONS FOR ALL REGISTERS

Register	Power-on Reset	MCLR Resets WDT Reset	Wake-up via WDT or Interrupt
W	xxxx xxxx	uuuu uuuu	uuuu uuuu
INDF	N/A	N/A	N/A
TMR0	xxxx xxxx	uuuu uuuu	uuuu uuuu
PCL	0000h	0000h	PC + 1 ⁽²⁾
STATUS	0001 1xxx	000q quuu ⁽³⁾	uuuq quuu ⁽³⁾
FSR	xxxx xxxx	uuuu uuuu	uuuu uuuu
PORTA	0x 0000	0u 0000	uu uuuu
PORTB	xxxx xxxx	uuuu uuuu	uuuu uuuu
PORTC	xx xxxx	uu uuuu	uu uuuu
PORTD	0000 0000	0000 0000	uuuu uuuu
PORTE	0000 0000	0000 0000	uuuu uuuu
PCLATH	0 0000	0 0000	u uuuu
INTCON	0000 000x	0000 000u	uuuu uuuu(1)
PIR1	00 0000	00 0000	uu uuuu ⁽¹⁾
TMR1L	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR1H	xxxx xxxx	uuuu uuuu	uuuu uuuu
T1CON	00 0000	uu uuuu	uu uuuu
TMR2	0000 0000	0000 0000	uuuu uuuu
T2CON	-000 0000	-000 0000	-uuu uuuu
SSPBUF	xxxx xxxx	uuuu uuuu	uuuu uuuu
SSPCON	0000 0000	0000 0000	uuuu uuuu
CCPR1L	xxxx xxxx	uuuu uuuu	uuuu uuuu
CCPR1H	xxxx xxxx	uuuu uuuu	uuuu uuuu
CCP1CON	00 0000	00 0000	uu uuuu
ADRES	xxxx xxxx	uuuu uuuu	uuuu uuuu
ADCON0	0000 00-0	0000 00-0	uuuu uu-u
OPTION_REG	1111 1111	1111 1111	uuuu uuuu
TRISA	11 1111	11 1111	uu uuuu
TRISB	1111 1111	1111 1111	uuuu uuuu
TRISC	11 1111	11 1111	uu uuuu
TRISD	1111 1111	1111 1111	uuuu uuuu
TRISE	1111 1111	1111 1111	uuuu uuuu
PIE1	00 0000	00 0000	uu uuuu
PCON	0-	u-	u-

 $\label{eq:local_local_local_local_local} \mbox{Legend: } \mbox{u = unchanged, x = unknown, $-$ = unimplemented bit, read as '0', q = value depends on condition}$

Note 1: One or more bits in INTCON and/or PIR1 will be affected (to cause wake-up).

^{2:} When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

^{3:} See Table 12-5 for RESET value for specific condition.

TABLE 12-6: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Register	Power-on Reset	MCLR Resets WDT Reset	Wake-up via WDT or Interrupt
PR2	1111 1111	1111 1111	1111 1111
SSPADD	0000 0000	0000 0000	uuuu uuuu
SSPSTAT	0000 0000	0000 0000	uuuu uuuu
ADCON1	000	000	uuu
PORTF	0000 0000	0000 0000	uuuu uuuu
PORTG	0000 0000	0000 0000	uuuu uuuu
LCDSE	1111 1111	1111 1111	uuuu uuuu
LCDPS	0000	0000	uuuu
LCDCON	00-0 0000	00-0 0000	uu-u uuuu
LCDD00 to LCDD15	xxxx xxxx	uuuu uuuu	uuuu uuuu
TRISF	1111 1111	1111 1111	uuuu uuuu
TRISG	1111 1111	1111 1111	uuuu uuuu

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition

Note 1: One or more bits in INTCON and/or PIR1 will be affected (to cause wake-up).

^{2:} When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

^{3:} See Table 12-5 for RESET value for specific condition.

FIGURE 12-7: TIME-OUT SEQUENCE ON POWER-UP (MCLR NOT TIED TO VDD): CASE 1

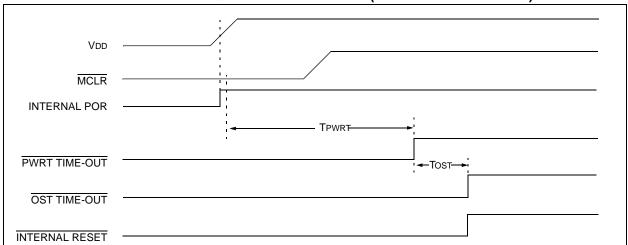


FIGURE 12-8: TIME-OUT SEQUENCE ON POWER-UP (MCLR NOT TIED TO VDD): CASE 2

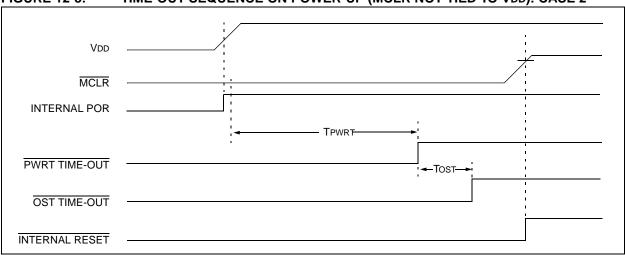
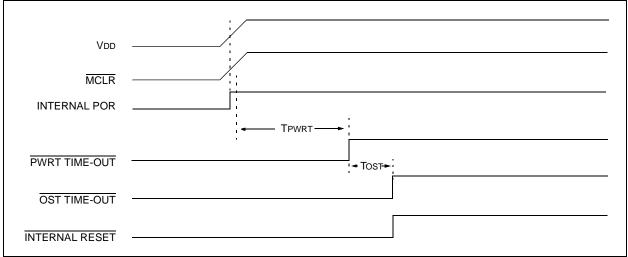


FIGURE 12-9: TIME-OUT SEQUENCE ON POWER-UP (MCLR TIED TO VDD)



12.5 Interrupts

The PIC16C925/926 family has nine sources of interrupt:

- External interrupt RB0/INT
- TMR0 overflow interrupt
- PORTB change interrupts (pins RB7:RB4)
- A/D Interrupt
- TMR1 overflow interrupt
- · TMR2 matches period interrupt
- CCP1 interrupt
- · Synchronous serial port interrupt
- · LCD module interrupt

The interrupt control register (INTCON) records individual interrupt requests in flag bits. It also has individual and global interrupt enable bits.

Note: Individual interrupt flag bits are set, regardless of the status of their corresponding mask bit, or the GIE bit.

A global interrupt enable bit, GIE (INTCON<7>), enables (if set) all unmasked interrupts, or disables (if cleared) all interrupts. When bit GIE is enabled, and an interrupt's flag bit and mask bit are set, the interrupt will vector immediately. Individual interrupts can be disabled through their corresponding enable bits in various registers. Individual interrupt bits are set, regardless of the status of the GIE bit. The GIE bit is cleared on RESET.

The "return from interrupt" instruction, RETFIE, exits the interrupt routine as well as sets the GIE bit, which re-enables interrupts.

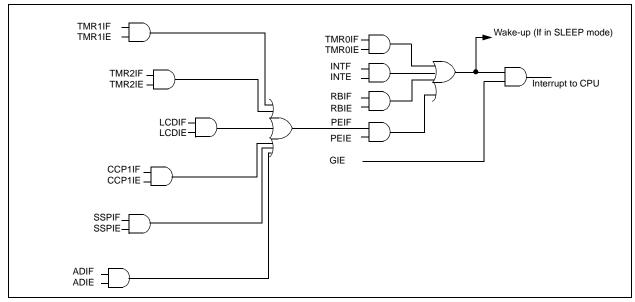
The RB0/INT pin interrupt, the RB port change interrupt and the TMR0 overflow interrupt flags are contained in the INTCON register.

The peripheral interrupt flags are contained in the special function register, PIR1. The corresponding interrupt enable bits are contained in special function register, PIE1, and the peripheral interrupt enable bit is contained in special function register, INTCON.

When an interrupt is responded to, the GIE bit is cleared to disable any further interrupts, the return address is pushed onto the stack and the PC is loaded with 0004h. Once in the Interrupt Service Routine the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid recursive interrupts.

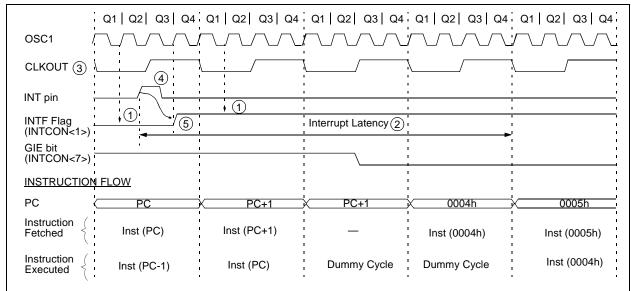
For external interrupt events, such as the RB0/INT pin or RB Port change interrupt, the interrupt latency will be three or four instruction cycles. The exact latency depends when the interrupt event occurs (Figure 12-11). The latency is the same for one or two cycle instructions. Individual interrupt flag bits are set, regardless of the status of their corresponding mask bit, PEIE bit, or the GIE bit.

FIGURE 12-10: INTERRUPT LOGIC



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FIGURE 12-11: INT PIN INTERRUPT TIMING



Note 1: INTF flag is sampled here (every Q1).

- 2: Interrupt latency = 3-4 TCY where TCY = instruction cycle time. Latency is the same whether Inst (PC) is a single cycle or a 2-cycle instruction.
- 3: CLKOUT is available only in RC oscillator mode.
- 4: For minimum width of INT pulse, refer to AC specs.
- 5: INTF can be set any time during the Q4-Q1 cycles.

12.5.1 INT INTERRUPT

External interrupt on RB0/INT pin is edge triggered: either rising if bit INTEDG (OPTION_REG<6>) is set, or falling, if the INTEDG bit is clear. When a valid edge appears on the RB0/INT pin, flag bit INTF (INTCON<1>) is set. This interrupt can be disabled by clearing enable bit INTE (INTCON<4>). Flag bit INTF must be cleared in software in the Interrupt Service Routine before re-enabling this interrupt. The INT interrupt can wake-up the processor from SLEEP, if bit INTE was set prior to going into SLEEP. The status of global interrupt enable bit, GIE, decides whether or not the processor branches to the interrupt vector following wake-up. See Section 12.8 for details on SLEEP mode.

12.5.2 TMR0 INTERRUPT

An overflow (FFh \rightarrow 00h) in the TMR0 register will set flag bit, TMR0IF (INTCON<2>). The interrupt can be enabled/disabled by setting/clearing enable bit, TMR0IE (INTCON<5>) (Section 5.0).

12.5.3 PORTB INTCON CHANGE

An input change on PORTB<7:4> sets flag bit RBIF (INTCON<0>). The interrupt can be enabled/disabled by setting/clearing enable bit, RBIE (INTCON<4>) (Section 4.2).

12.6 Context Saving During Interrupts

During an interrupt, only the return PC value is saved on the stack. Typically, users may wish to save key registers during an interrupt, i.e., the W and STATUS registers. This will have to be implemented in software.

Example 12-1 stores and restores the STATUS, W, and PCLATH registers. The register, W_TEMP, must be defined in each bank and must be defined at the same offset from the bank base address (i.e., if W_TEMP is defined at 0x20 in bank 0, it must also be defined at 0xA0 in bank 1).

The code in the example:

- e) Stores the W register.
- f) Stores the STATUS register in bank 0.
- g) Stores the PCLATH register.
- h) Executes the ISR code.
- Restores the STATUS register (and bank select bit).
- j) Restores the W and PCLATH registers.

EXAMPLE 12-1: SAVING STATUS. W. AND PCLATH REGISTERS IN RAM

```
MOVWF
         W TEMP
                           ; Copy W to TEMP register, could be bank one or zero
                           ;Swap status to be saved into W
SWAPF
         STATUS, W
CLRF
        STATUS
                           ; bank 0, regardless of current bank, Clears IRP, RP1, RP0
MOVWF
      STATUS TEMP
                           ; Save status to bank zero STATUS TEMP register
MOVF
        PCLATH, W
                           ;Only required if using pages 1, 2 and/or 3
MOVWF
        PCLATH TEMP
                          ;Save PCLATH into W
CLRF
        PCLATH
                          ; Page zero, regardless of current page
BCF
        STATUS, IRP
                           ;Return to Bank 0
        FSR, W
                           ;Copy FSR to W
MOVF
        FSR TEMP
MOVWF
                           ;Copy FSR from W to FSR TEMP
   :
   : (ISR)
                           ;Insert user code here
MOVF
         PCLATH TEMP, W
                           ;Restore PCLATH
MOVWF
         PCLATH
                           ; Move W into PCLATH
SWAPF
         STATUS TEMP, W
                           ;Swap STATUS TEMP register into W
                           ; (sets bank to original state)
MOVWF
         STATUS
                           ; Move W into STATUS register
                           ;Swap W TEMP
SWAPF
         W TEMP, F
SWAPF
         W TEMP, W
                           ;Swap W_TEMP into W
```

12.7 Watchdog Timer (WDT)

The Watchdog Timer is a free running on-chip RC oscillator, which does not require any external components. This RC oscillator is separate from the RC oscillator of the OSC1/CLKIN pin. That means that the WDT will run, even if the clock on the OSC1/CLKIN and OSC2/CLKOUT pins of the device has been stopped, for example, by execution of a SLEEP instruction. During normal operation, a WDT time-out generates a device RESET (Watchdog Timer Reset). If the device is in SLEEP mode, a WDT time-out causes the device to wake-up and continue with normal operation (Watchdog Timer Wake-up). The WDT can be permanently disabled by clearing configuration bit WDTE (Section 12.1).

12.7.1 WDT PERIOD

The WDT has a nominal time-out period of 18 ms (with no prescaler). The time-out periods vary with temperature, VDD and process variations from part to part (see DC specs). If longer time-out periods are desired, a prescaler with a division ratio of up to 1:128 can be

assigned to the WDT under software control, by writing to the OPTION register. Thus, time-out periods up to 2.3 seconds can be realized.

The CLRWDT and SLEEP instructions clear the WDT and the postscaler, if assigned to the WDT, prevent it from timing out and generating a device RESET condition.

The TO bit in the STATUS register will be cleared upon a Watchdog Timer time-out.

12.7.2 WDT PROGRAMMING CONSIDERATIONS

It should also be taken into account that under worst case conditions (VDD = Min., Temperature = Max., and Max. WDT prescaler) it may take several seconds before a WDT time-out occurs.

Note: When a CLRWDT instruction is executed and the prescaler is assigned to the WDT, the prescaler count will be cleared, but the prescaler assignment is not changed.

FIGURE 12-12: WATCHDOG TIMER BLOCK DIAGRAM

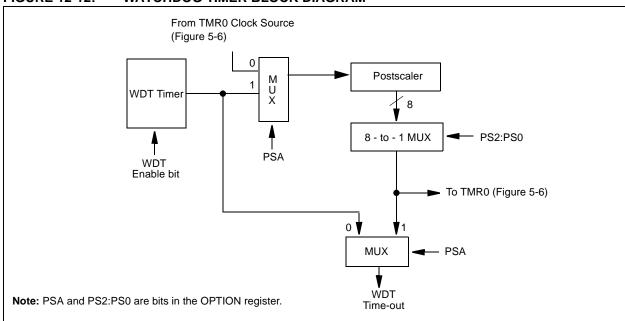


FIGURE 12-13: SUMMARY OF WATCHDOG TIMER REGISTERS

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
2007h	Config. bits	(1)	BOREN ⁽¹⁾	CP1	CP0	PWRTE ⁽¹⁾	WDTE	FOSC1	FOSC0
81h, 181h	OPTION	RBPU	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0

Legend: Shaded cells are not used by the Watchdog Timer.

Note 1: See Register 12-1 for operation of these bits.

12.8 Power-down Mode (SLEEP)

Power-down mode is entered by executing a SLEEP instruction.

If enabled, the Watchdog Timer will be cleared but keeps running, the \overline{PD} bit (STATUS<3>) is cleared, the \overline{TO} (STATUS<4>) bit is set, and the oscillator driver is turned off. The I/O ports maintain the status they had, before the SLEEP instruction was executed (driving high, low, or hi-impedance).

For lowest current consumption in this mode, place all I/O pins at either VDD, or Vss, ensure no external circuitry is drawing current from the I/O pin, power-down the A/D, disable external clocks. Pull all I/O pins that are hi-impedance inputs, high or low externally, to avoid switching currents caused by floating inputs. The TOCKI input should also be at VDD or Vss for lowest current consumption. The contribution from on-chip pull-ups on PORTB should also be considered.

The \overline{MCLR} pin must be at a logic high level (VIHMC).

12.8.1 WAKE-UP FROM SLEEP

The device can wake-up from SLEEP through one of the following events:

- 1. External RESET input on $\overline{\text{MCLR}}$ pin.
- Watchdog Timer Wake-up (if WDT was enabled).
- Interrupt from RB0/INT pin, RB port change, or peripheral interrupt.

External MCLR Reset will cause a device RESET. All other events are considered a continuation of program execution and cause a "wake-up". The TO and PD bits in the STATUS register can be used to determine the cause of device RESET. The PD bit, which is set on power-up is cleared when SLEEP is invoked. The TO bit is cleared if a WDT time-out occurred (and caused wake-up).

The following peripheral interrupts can wake the device from SLEEP:

- TMR1 interrupt. Timer1 must be operating as an asynchronous counter.
- 2. SSP (START/STOP) bit detect interrupt.
- 3. SSP transmit or receive in Slave mode (SPI/ I^2 C).
- 4. CCP Capture mode interrupt.
- 5. A/D conversion (when A/D clock source is RC).
- Special event trigger (Timer1 in Asynchronous mode using an external clock).
- 7. LCD module.

Other peripherals can not generate interrupts since during SLEEP, no on-chip Q clocks are present.

When the SLEEP instruction is being executed, the next instruction (PC + 1) is pre-fetched. For the device to wake-up through an interrupt event, the corresponding interrupt enable bit must be set (enabled). Wake-up is regardless of the state of the GIE bit. If the GIE bit is clear (disabled), the device continues execution at the instruction after the SLEEP instruction. If the GIE bit is set (enabled), the device executes the instruction after the SLEEP instruction and then branches to the interrupt address (0004h). In cases where the execution of the instruction following SLEEP is not desirable, the user should have a NOP after the SLEEP instruction.

12.8.2 WAKE-UP USING INTERRUPTS

When global interrupts are disabled (GIE cleared) and any interrupt source has both its interrupt enable bit and interrupt flag bit set, one of the following will occur:

- If the interrupt occurs before the execution of a SLEEP instruction, the SLEEP instruction will complete as a NOP. Therefore, the WDT and WDT postscaler will not be cleared, the TO bit will not be set and PD bits will not be cleared.
- If the interrupt occurs during or after the execution of a SLEEP instruction, the device will immediately wake-up from SLEEP. The SLEEP instruction will be completely executed before the wake-up. Therefore, the WDT and WDT postscaler will be cleared, the TO bit will be set and the PD bit will be cleared.

Even if the flag bits were checked before executing a SLEEP instruction, it may be possible for flag bits to become set before the SLEEP instruction completes. To determine whether a SLEEP instruction executed, test the PD bit. If the PD bit is set, the SLEEP instruction was executed as a NOP.

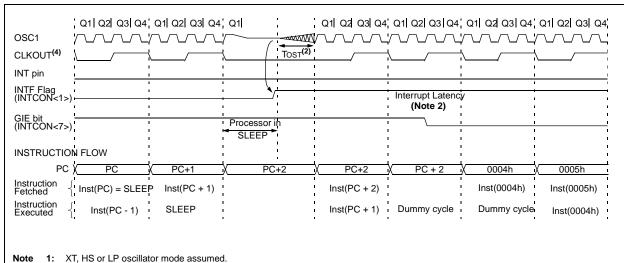
To ensure that the WDT is cleared, a CLRWDT instruction should be executed before a SLEEP instruction.



TOST = 1024Tosc (drawing not to scale) This delay will not be there for RC osc mode.

CLKOUT is not available in these osc modes, but shown here for timing reference.

GIE = '1' assumed. In this case after wake-up, the processor jumps to the interrupt routine.



12.9 Program Verification/Code Protection

If the code protection bit(s) have not been programmed, the on-chip program memory can be read out for verification purposes.

If GIE = '0', execution will continue in-line.

Note: Microchip does not recommend code protecting windowed devices.

12.10 ID Locations

3:

Four memory locations (2000h - 2003h) are designated as ID locations, where the user can store checksum or other code identification numbers. These locations are not accessible during normal execution, but are readable and writable during program/verify. It is recommended that only the four Least Significant bits of the ID location are used.

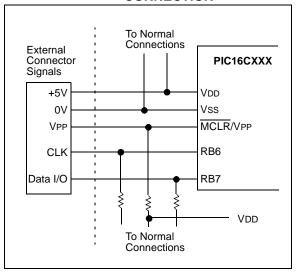
12.11 In-Circuit Serial Programming

PIC16CXXX microcontrollers can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data, and three other lines for power, ground, and the programming voltage. This allows customers to manufacture boards with unprogrammed devices, and then program the microcontroller just before shipping the product. This also allows the most recent firmware or a custom firmware to be programmed.

The device is placed into a Program/Verify mode by holding the RB6 and RB7 pins low, while raising the MCLR (VPP) pin from VIL to VIHH (see programming specification). RB6 becomes the programming clock and RB7 becomes the programming data. Both RB6 and RB7 are Schmitt Trigger inputs in this mode.

After RESET, to place the device into Program/Verify mode, the program counter (PC) is at location 00h. A 6-bit command is then supplied to the device. Depending on the command, 14-bits of program data are then supplied to or from the device, depending if the command was a load or a read. For complete details of serial programming, please refer to the PIC16C6X/7X Programming Specifications (Literature #DS30228).

FIGURE 12-15: TYPICAL IN-CIRCUIT
SERIAL PROGRAMMING
CONNECTION



13.0 INSTRUCTION SET SUMMARY

Each PIC16CXXX instruction is a 14-bit word divided into an OPCODE which specifies the instruction type and one or more operands which further specify the operation of the instruction. The PIC16CXXX instruction set summary in Table 13-2 lists byte-oriented, bitoriented, and literal and control operations. Table 13-1 shows the opcode field descriptions.

The instruction set is highly orthogonal and is grouped into three basic categories:

- Byte-oriented operations
- · Bit-oriented operations
- · Literal and control operations

For **byte-oriented** instructions, 'f' represents a file register designator and 'd' represents a destination designator. The file register designator specifies which file register is to be used by the instruction.

The destination designator specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the W register. If 'd' is one, the result is placed in the file register specified in the instruction.

For **bit-oriented** instructions, 'b' represents a bit field designator which selects the number of the bit affected by the operation, while 'f' represents the address of the file in which the bit is located.

For **literal and control** operations, 'k' represents an eight or eleven bit constant or literal value.

FIGURE 13-1: GENERAL FORMAT FOR INSTRUCTIONS

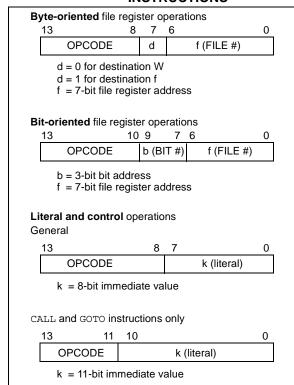


TABLE 13-1: OPCODE FIELD DESCRIPTIONS

Field	Description
f	Register file address (0x00 to 0x7F)
W	Working register (accumulator)
b	Bit address within an 8-bit file register
k	Literal field, constant data or label
х	Don't care location (= 0 or 1). The assembler will generate code with x = 0. It is the recommended form of use for compatibility with all Microchip software tools.
đ	Destination select; d = 0: store result in W, d = 1: store result in file register f. Default is d = 1.
label	Label name
TOS	Top-of-Stack
PC	Program Counter
PCLATH	Program Counter High Latch
GIE	Global Interrupt Enable bit
WDT	Watchdog Timer/Counter
TO	Time-out bit
PD	Power-down bit
dest	Destination either the W register or the specified register file location
[]	Options
()	Contents
\rightarrow	Assigned to
<>	Register bit field
€	In the set of
italics	User defined term (font is courier)

All instructions are executed within one single instruction cycle, unless a conditional test is true, or the program counter is changed, as a result of an instruction. In this case, the execution takes two instruction cycles, with the second cycle executed as a NOP. One instruction cycle consists of four oscillator periods. Thus, for an oscillator frequency of 4 MHz, the normal instruction execution time is 1 μs . If a conditional test is true, or the program counter is changed, as a result of an instruction, the instruction execution time is 2 μs .

Table 13-2 lists the instructions recognized by the MPASM TM assembler.

Figure 13-1 shows the general formats that the instructions can have.

Note: To maintain upward compatibility with future PIC16CXXX products, <u>do not use</u> the OPTION and TRIS instructions.

All examples use the format '0xnn' to represent a hexadecimal number.

TABLE 13-2: PIC16CXXX INSTRUCTION SET

Mnemonic, Operands		Description	Cycles	14-Bit Opcode		Status	Notes		
		Description		MSb			LSb	Affected	notes
BYTE-ORI	BYTE-ORIENTED FILE REGISTER OPERATIONS								
ADDWF	f, d	Add W and f	1	00	0111	dfff	ffff	C,DC,Z	1,2
ANDWF	f, d	AND W with f	1	00	0101	dfff	ffff	Z	1,2
CLRF	f	Clear f	1	00	0001	lfff	ffff	Z	2
CLRW	-	Clear W	1	00	0001	0xxx	xxxx	Z	
COMF	f, d	Complement f	1	00	1001	dfff	ffff	Z	1,2
DECF	f, d	Decrement f	1	00	0011	dfff	ffff	Z	1,2
DECFSZ	f, d	Decrement f, Skip if 0	1(2)	00	1011	dfff	ffff		1,2,3
INCF	f, d	Increment f	1	00	1010	dfff	ffff	Z	1,2
INCFSZ	f, d	Increment f, Skip if 0	1(2)	00	1111	dfff	ffff		1,2,3
IORWF	f, d	Inclusive OR W with f	1	00	0100	dfff	ffff	Z	1,2
MOVF	f, d	Move f	1	00	1000	dfff	ffff	Z	1,2
MOVWF	f	Move W to f	1	00	0000	lfff	ffff		
NOP	-	No Operation	1	00	0000	0xx0	0000		
RLF	f, d	Rotate Left f through Carry	1	00	1101	dfff	ffff	С	1,2
RRF	f, d	Rotate Right f through Carry	1	00	1100	dfff	ffff	С	1,2
SUBWF	f, d	Subtract W from f	1	00	0010	dfff	ffff	C,DC,Z	1,2
SWAPF	f, d	Swap nibbles in f	1	00	1110	dfff	ffff		1,2
XORWF	f, d	Exclusive OR W with f	1	00	0110	dfff	ffff	Z	1,2
BIT-ORIEN	NTED FIL	E REGISTER OPERATIONS		•					
BCF	f, b	Bit Clear f	1	01	00bb	bfff	ffff		1,2
BSF	f, b	Bit Set f	1	01	01bb	bfff	ffff		1,2
BTFSC	f, b	Bit Test f, Skip if Clear	1 (2)	01	10bb	bfff	ffff		3
BTFSS	f, b	Bit Test f, Skip if Set	1 (2)	01	11bb	bfff	ffff		3
LITERAL	AND CO	NTROL OPERATIONS	I						
ADDLW	k	Add literal and W	1	11	111x	kkkk	kkkk	C,DC,Z	
ANDLW	k	AND literal with W	1	11	1001	kkkk	kkkk	Z	
CALL	k	Call subroutine	2	10	0kkk	kkkk	kkkk		
CLRWDT	-	Clear Watchdog Timer	1	00	0000	0110	0100	TO,PD	
GOTO	k	Go to address	2	10	1kkk	kkkk	kkkk		
IORLW	k	Inclusive OR literal with W	1	11	1000	kkkk	kkkk	Z	
MOVLW	k	Move literal to W	1	11	00xx	kkkk	kkkk		
RETFIE	-	Return from interrupt	2	00	0000	0000	1001		
RETLW	k	Return with literal in W	2	11	01xx	kkkk	kkkk		
RETURN	-	Return from Subroutine	2	00	0000	0000	1000		
SLEEP	-	Go into standby mode	1	00	0000	0110	0011	TO,PD	
SUBLW	k	Subtract W from literal	1	11	110x	kkkk		C,DC,Z	
XORLW	k	Exclusive OR literal with W	1	11	1010	kkkk	kkkk	Z	
L	147	1/0 :	l	1					

Note 1: When an I/O register is modified as a function of itself (e.g., MOVF PORTB, 1), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

^{2:} If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned to the Timer0 Module.

^{3:} If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

13.1 Instruction Descriptions

ADDLW	Add Literal and W					
Syntax:	[label] ADDLW k					
Operands:	$0 \le k \le 255$					
Operation:	(W) + k -	→ (W)				
Status Affected:	C, DC, Z					
Encoding:	11 111x kkkk kkkk					
Description:	The contents of the W register are added to the eight-bit literal 'k' and the result is placed in the W register.					
Words:	1					
Cycles:	1					
Q Cycle Activity:	Q1	Q2	Q3	Q4		
	Decode	Read literal 'k'	Process data	Write to W		

Example: ADDLW 0x15

Before Instruction: W = 0x10After Instruction: W = 0x25

ADDWF	Add W and f						
Syntax:	[label] ADDWF f [,d]						
Operands:	$0 \le f \le 127$ $d \in [0,1]$						
Operation:	(W) + (f) \rightarrow (destination)						
Status Affected:	C, DC, Z						
Encoding:	00	0111	dfff	ffff			
Description:	Add the contents of the W register with register 'f'. If 'd' is 0, the result is stored in the W register. If 'd' is 1, the result is stored back in register 'f'.						
Words:	1						
Cycles:	1						
Q Cycle Activity:	Q1	Q2	Q3	Q4			
	Decode	Read register 'f'	Process data	Write to destination			

Example ADDWF FSR, 0
Before Instruction:

W = 0x17 FSR = 0xC2

After Instruction:

W = 0xD9 FSR = 0xC2

ANDLW	AND Literal with W	ANDWF	AND W with f
Syntax:	[label] ANDLW k	Syntax:	[label] ANDWF f [,d]
Operands:	$0 \le k \le 255$	Operands:	0 ≤ f ≤ 127
Operation:	(W) .AND. (k) \rightarrow (W)	·	$d \in [0,1]$
Status Affected:	Z	Operation:	(W).AND. (f) \rightarrow (destination)
Encoding:	11 1001 kkkk kkkk	Status Affected:	Z
Description:	The contents of W register are AND'ed with the eight-bit literal 'k'.	Encoding:	00 0101 dfff ffff
Words:	The result is placed in the W register.	Description:	AND the W register with register 'f'. If 'd' is 0, the result is stored in the W register. If 'd' is 1, the result is
Cycles:	1		stored back in register 'f'.
Q Cycle Activity:	Q1 Q2 Q3 Q4	Words: Cycles:	1
	Decode Read Internal 'k' Process Write to data W	Q Cycle Activity:	Q1 Q2 Q3 Q4
Example	ANDLW 0x5F		Decode Read register Process data Write to destination
Before Instruction			
After Instruction:	0xA3	Example	ANDWF FSR, 1
W =	0x03	Before Instruction W FSR	: = 0x17 = 0xC2

After Instruction W

FSR

0x17

0x02

BCF Bit Clear f

Syntax: [label] BCF f [,b]

Operands: $0 \le f \le 127$ $0 \le b \le 7$

Operation: $0 \rightarrow (f < b >)$

Status Affected: None

Encoding: 01 00bb bfff ffff

Description: Bit 'b' in register 'f' is cleared.

Words: 1

Cycles: 1

Q Cycle Activity: Q1 Q2 Q3 Q4

Decode Read register 'f' Process data register 'f'

Example BCF FLAG_REG, 7

Before Instruction:

FLAG REG = 0xC7

After Instruction:

FLAG REG = 0x47

BSF Bit Set f

Syntax: [label] BSF f [,b]

Operands: $0 \le f \le 127$ $0 \le b \le 7$

Operation: $1 \rightarrow (f < b >)$

Status Affected: None

Encoding: 01 01bb bfff ffff

Description: Bit 'b' in register 'f' is set.

Words: 1

Cycles: 1

Q Cycle Activity: Q1 Q2 Q3 Q4

Decode Read Process Write register 'f'

Example BSF FLAG_REG, 7

Before Instruction:

 $FLAG_REG = 0x0A$

After Instruction:

 $FLAG_REG = 0x8A$

BTFSC Bit Test, Skip if Clear

Syntax: [label] BTFSC f [,b]

Operands: $0 \le f \le 127$ $0 \le b \le 7$

Operation: $skip\ if\ (f < b >) = 0$

Status Affected: None

Encoding: 01 10bb bfff ffff

Description: If bit 'b' in register 'f' is '1', then the

next instruction is executed.

If bit 'b' in register 'f' is '0', then the next instruction is discarded, and a NOP is executed instead, making this a

2Tcy instruction.

Words: 1

Cycles: 1(2)

Q Cycle Activity: Q1 Q2 Q3 Q4

Decode Read Process No Operation

If Skip: (2nd Cycle)

Q1 Q2 Q3 Q4

No No No No Operation Operation Operation

Example HERE BTFSC FLAG, 1

FALSE GOTO PROCESS_CODE TRUE •

•

Before Instruction:

PC = address HERE

After Instruction:

if FLAG<1> = 0,

PC = address TRUE

if FLAG<1> = 1,

PC = address FALSE

BTFSS	Bit Test f, Skip if Set	CALL	Call Subroutine
Syntax:	[label] BTFSS f [,b]	Syntax:	[label] CALL k
Operands:	$0 \le f \le 127$	Operands:	$0 \le k \le 2047$
Operation: Status Affected:	$0 \le b < 7$ skip if (f) = 1 None	Operation:	(PC)+ 1→ TOS, $k \rightarrow PC<10:0>$, (PCLATH<4:3>) $\rightarrow PC<12:11>$
	01 11bb bfff ffff	Status Affected:	None
Encoding: Description:	If bit 'b' in register 'f' is '0', then the	Encoding:	10 0kkk kkkk kkkk
·	next instruction is executed. If bit 'b' is '1', then the next instruction is discarded and a NOP is executed instead, making this a 2Tcy instruction.	Description:	Call Subroutine. First, return address (PC+1) is pushed onto the stack. The eleven-bit immediate address is loaded into PC bits <10:0>. The upper bits of the PC are loaded from PCLATH. CALL is a
Words:	1		two-cycle instruction.
Cycles:	1(2)	Words:	1
Q Cycle Activity:	Q1 Q2 Q3 Q4	Cycles:	2
	Decode Read Process No register 'f' data Operation	Q Cycle Activity:	Q1 Q2 Q3 Q4
If Skip:	(2nd Cycle) Q1 Q2 Q3 Q4 No No No No	1st Cycle	Decode Read literal 'k', Push PC to Stack Process Otack Write to PC
	Operation Operation Operation	2nd Cycle	No No No No Operation Operation
Example	HERE BTFSC FLAG, 1 FALSE GOTO PROCESS_CODE TRUE • •	Example Before Instru PC	HERE CALL THERE oction: = Address HERE
Before Instru PC After Instruct if FLAG< PC if FLAG<	= address HERE ion: <1> = 0,	After Instruct PC TOS	ion: = Address THERE = Address HERE+1
PC	= address TRUE		

CLRF	Clear f						
Syntax:	[label] CLRF f						
Operands:	$0 \le f \le 127$						
Operation:	$00h \rightarrow (f)$ $1 \rightarrow Z$						
Status Affected:	Z						
Encoding:	00	0001	1fff	ffff			
Description:	The contents of register 'f' are cleared and the Z bit is set.						
Words:	1						
Cycles:	1						
Q Cycle Activity:	Q1	Q2	Q3	Q4			
	Decode	Read register 'f'	Process data	Write register 'f'			
		•					

CLRW	Clear W					
Syntax:	[label]	CLRW				
Operands:	None					
Operation:	$\begin{array}{l} 00h \rightarrow (W) \\ 1 \rightarrow Z \end{array}$					
Status Affected:	Z					
Encoding:	00	0001	0xxx	xxxx		
Description:	W registe set.	er is cleare	d. Zero l	oit (Z) is		
Words:	1					
Cycles:	1					
Q Cycle Activity:	Q1	Q2	Q3	Q4		
	Decode	No Operation	Process data	Write to W		

Before Instruction:				
FLAG_REG =	0x5A			
After Instruction:	0.00			
FLAG_REG =	0x00			
Z =	1			

CLRF

FLAG_REG

Example

Example	CLRW				
Before Instruction:					
W	=	0x5A			
After Instruction:					
W	=	0x00			
Z	=	1			

CLRWDT	Clear Watchdog Timer						
Syntax:	[label]	CLRWDT					
Operands:	None						
Operation:	00h → WDT 0 → WDT prescaler, 1 → $\overline{10}$ 1 → \overline{PD}						
Status Affected:	TO, PD						
Encoding:	00	0000	0110	0100			
Description:	Watchdo	instructior g Timer. It of the WI are set.	also res	ets th <u>e</u>			
Words:	1						
Cycles:	1						
Q Cycle Activity:	Q1	Q2	Q3	Q4			
	Decode	No Operation	Process data	Clear WDT Counter			

Example CLRWDT

Before Instruction:

WDT counter = ?

After Instruction:

COMF	Compler	nent f		
Syntax:	[label]	COMF	f [,d]	
Operands:	$0 \le f \le 127$ $d \in [0,1]$			
Operation:	$(\bar{f}) \to (destination)$			
Status Affected:	Z			
Encoding:	00	1001	dfff	ffff
Description:	The content plemente stored in stored ba	d. If 'd' is W. If 'd' i	s 0, the r s 1, the	
Words:	1			
Cycles:	1			
Q Cycle Activity:	Q1	Q2	Q3	Q4
	Decode	Read register 'f'	Process data	Write to destination
Example	COMF	REG	1,0	
Before Instruc REG1	tion: = 0x13			
144	on: = 0x13 = 0xEC			

DECF	Decreme	ent f		
Syntax:	[label]	DECF f	[,d]	
Operands:	$0 \le f \le 12$ $d \in [0,1]$	27		
Operation:	(f) - 1 \rightarrow (destination)			
Status Affected:	Z			
Encoding:	00	0011	dfff	ffff
Description:	result is s	stored in result is		is 0, the gister. If 'd'
Words:	1			
Cycles:	1			
Q Cycle Activity:	Q1	Q2	Q3	Q4
	Decode	Read register 'f'	Process data	Write to destination
Example	DECF	CNT,	1	
Before Instruc CNT Z	etion: = 0x01 = 0			
After Instruction CNT Z	on: = 0x00 = 1			

DECFSZ	Decreme	ent f, Skip	o if 0		
Syntax:	[label]	DECFSZ	f [,d]		
Operands:	$0 \le f \le 12$ $d \in [0,1]$	27			
Operation:	(f) - 1 \rightarrow skip if res	(destinationsult = 0	on);		
Status Affected:	None				
Encoding:	0.0	00 1011 dfff ffff			
Description:	The contents of register 'f' are decremented. If 'd' is 0, the result is placed in the W register. If 'd' is 1, the result is placed back in register 'f'. If the result is 1, the next instruction is executed. If the result is 0, then a NOP is executed instead, making it a 2Tcy instruction.				
Words:	1				
Cycles:	1(2)				
Q Cycle Activity:	Q1	Q2	Q3	Q4	
	Decode	Read register 'f'	Process data	Write to destination	
If Skip: (2nd C	ycle)				
	Q1	Q2	Q3	Q4	
	No Operation	No Operation	No Operation	No Operation	
Example	HERE CONTINUI	DECFSZ GOTO • •	CNT, LOOP	1	
Before Instruc PC = 3	tion: address H	ERE			
if CNT = PC = if CNT =	= CNT - = 0, = addre ≠ 0,	· 1 ss CONTI ss HERE+			

GOTO	Unconditional Branch			
Syntax:	[label]	GOTO I	(
Operands:	$0 \le k \le 2047$			
Operation:	$k \rightarrow PC<10:0>$ PCLATH<4:3> \rightarrow PC<12:11>			
Status Affected:	None			
Encoding:	10	1kkk	kkkk	kkkk
Description:	eleven-bii into PC b PC are lo	t immedia its <10:0> aded fron	itional bra te value is The upp PCLATH instruction	s loaded er bits of I<4:3>.
Words:	1			
Cycles:	2			
Q Cycle Activity:	Q1	Q2	Q3	Q4
1st Cycle	Decode	Read literal 'k'	Process data	Write to PC
2nd Cycle	No Operation	No Operation	No Operation	No Operation

Example GOTO THERE

After Instruction:

PC = Address THERE

INCF	Increme	nt f			
Syntax:	[label]	INCF	f [,d]		
Operands:	$0 \le f \le 12$ $d \in [0,1]$	27			
Operation:	(f) + 1 →	(f) + 1 \rightarrow (destination)			
Status Affected:	Z	Z			
Encoding:	00	1010	dfff	ffff	
Description:	mented. in the W	lf 'd' is 0, register.	the result	are incre- t is placed the result '.	
Words:	1				
Cycles:	1				
Q Cycle Activity:	Q1	Q2	Q3	Q4	
	Decode	Read register 'f'	Process data	Write to destination	
Example	INCF	CNT,	1		
Before Instruc CNT Z	ction: = 0xFF = 0				
After Instructi	on:				

CNT

Ζ

= 0x00

INCFSZ	Increme	nt f, Skip	if O		
Syntax:	[label]	INCFSZ	f [,d]		
Operands:	$0 \le f \le 12$ $d \in [0,1]$	7			
Operation:	(f) + 1 \rightarrow skip if re	(destinations)	on),		
Status Affected:	None				
Encoding:	00	00 1111 dfff ffff			
Description:	mented. I in the W r placed ba If the resu executed	ents of reg f 'd' is 0, t egister. If ack in regi ult is 1, the . If the res instead, r	he result 'd' is 1, th ster 'f'. e next insi sult is 0, a	is placed e result is truction is	
Words:	1				
Cycles:	1(2)				
Q Cycle Activity:	Q1	Q2	Q3	Q4	
	Decode	Read register 'f'	Process data	Write to destination	
If Skip: (2nd	Cycle)				
	Q1	Q2	Q3	Q4	
	No Operation	No Operation	No Operation	No Operation	
Example	HERE CONTINUE	INCFSZ GOTO	CNT,	1	
Before Instruc	tion:				
PC	= addre	SS HERE			
if CNT PC if CNT	= CNT + = 0, = addre: ≠ 0,	+ 1 SS CONTI			

IORLW	Inclusive OR Literal with W			
Syntax:	[label] IORLW k			
Operands:	$0 \le k \le 255$			
Operation:	(W) .OR. $k \rightarrow$ (W)			
Status Affected:	Z			
Encoding:	11 1000 kkkk kkkk			
Description:	The contents of the W register is OR'ed with the eight-bit literal 'k'. The result is placed in the W register.			
Words:	1			
Cycles:	1			
Q Cycle Activity:	Q1 Q2 Q3 Q4			
	Decode Read Process Write to data W			
Example	IORLW 0x35			

Before Instruction: W = 0x9A After Instruction:

W = 0xBFZ = 0

IORWF	Inclusive	OR W	with f	
Syntax:	[label]	IORWF	f [,d]	
Operands:	$0 \le f \le 12$ $d \in [0,1]$	27		
Operation:	(W).OR. (f) \rightarrow (destination))
Status Affected:	Z			
Encoding:	0.0	0100	dfff	ffff
Description:	ister 'f'. If	'd' is 0, t register.	the result If 'd' is 1,	r with reg- is placed the result '.
Words:	1			
Cycles:	1			
Q Cycle Activity:	Q1	Q2	Q3	Q4
	Decode	Read register 'f'	Process data	Write to destination
Example	IORWF		RESULT,	0
Before Instruc RESULT W	= 0	x13 x91		

0x13

0x93

0

MOVF	Move f			
Syntax:	[label]	MOVF	f [,d]	
Operands:	$0 \le f \le 12$ $d \in [0,1]$	7		
Operation:	$(f) \rightarrow (destination)$			
Status Affected:	Z			
Encoding:	00	1000	dfff	ffff
Description:	W register	nation ded. If d = 0 er. If d = 1 er f itself register,	ependant), the des , the des . d = 1 is	upon the tination is tination is
Words:	1			
Cycles:	1			
Q Cycle Activity:	Q1	Q2	Q3	Q4
	Decode	Read register 'f'	Process data	Write to destination
		,	o er	

MOVLW	Move Lit	eral to V	v	
Syntax:	[label]	MOVLW	/ k	
Operands:	$0 \le k \le 25$	55		
Operation:	$k \to \text{(W)}$			
Status Affected:	None			
Encoding:	11	00xx	kkkk	kkkk
Description:	The eight W register assemble	r. The d		
Words:	1			
Cycles:	1			
Q Cycle Activity:	Q1	Q2	Q3	Q4
	Decode	Read literal 'k'	Process data	Write to W
Example	MOVLW	0x5A		
After Instruction W =	on: 0x5A			

After Instruction: RESULT

W

Ζ

MOVWF	Move W	to f		
Syntax:	[label]	MOVWI	F f	
Operands:	$0 \le f \le 127$			
Operation:	$(W) \rightarrow (f)$			
Status Affected:	None			
Encoding:	0.0	0000	1fff	ffff
Description:	Move dat register 'f		V register	to
Words:	1			
Cycles:	1			
Q Cycle Activity:	Q1	Q2	Q3	Q4
	Decode	Read register 'f'	Process data	Write register 'f'

Example MOVWF OPTION_REG

Before Instruction:

 $\begin{array}{lll} \mathsf{OPTION} &=& \mathsf{0xFF} \\ \mathsf{W} &=& \mathsf{0x4F} \end{array}$

After Instruction:

 $\begin{array}{lll} \mathsf{OPTION} &=& \mathsf{0x4F} \\ \mathsf{W} &=& \mathsf{0x4F} \end{array}$

NOP	No Operation			
Syntax:	[label]	NOP		
Operands:	None			
Operation:	No opera	ation		
Status Affected:	None			
Encoding:	00	0000	0xx0	0000
Description:	No operation.			
Words:	1			
Cycles:	1			
Q Cycle Activity:	Q1	Q2	Q3	Q4
	Decode	No Operation	No Operation	No Operation
Example	NOP			

OPTION	Load Option Register				
Syntax:	[label] OPTION				
Operands:	None				
Operation:	$(W) \rightarrow C$	PTION			
Status Affected:	None				
Encoding:	0.0	0000	0110	0010	
Description:	The conte loaded in This instr code com products. readable/ can direc	the OPT uction is npatibility Since C writable	FION regits supported with PICOPTION is register, to	ster. d for 16C5X s a	
Words:	1				
Cycles:	1				
Example					
	with futu	re PIC16	rd compa CXXX pro- struction.	•	

RETFIE	Return from Interrupt					
Syntax:	[label]	RETFIE				
Operands:	None					
Operation:	$ TOS \rightarrow PC, $					
Status Affected:	None					
Encoding:	00 0000 0000 1001					
Description:	Return from Interrupt. Stack is POPed and Top-of-Stack (TOS) is loaded in the PC. Interrupts are enabled by setting Global Interrupt Enable bit, GIE (INTCON<7>). This is a two-cycle instruction.					
Words:	1					
Cycles:	2					
Q Cycle Activity:	Q1	Q2	Q3	Q4		
1st Cycle	Decode	No Operation	Set the GIE bit	Pop from the Stack		
2nd Cycle	No Operation	No Operation	No Operation	No Operation		

- p	TOS → PC					
Status Affected:	None					
Encoding:	11	01xx	kkkk	kkkk		
Description:	The W register is loaded with the eight- bit literal 'k'. The program counter is loaded from the top of the stack (the return address). This is a two-cycle instruction.					
Words:	1					
Cycles:	2					
Q Cycle Activity:	Q1	Q2	Q3	Q4		
1st Cycle	Decode	Read literal 'k'	No Operation	Write to W, Pop from the Stack		
2nd Cycle	No Operation	No Operation	No Operation	No Operation		
Example	CALL TAB	;offs	ntains ta et value w has tab	able ole value		
TABLE	ADDWF PC RETLW k1 RETLW k2 •	-	offset n table			
	RETLW kn	; End	of table	е		
Before Instruc W = 0	tion: 0x07					
After Instruction: W = value of k8						

Return with Literal in W

[label] RETLW k

 $0 \le k \le 255$

 $k \to (W); \\$

RETLW

Syntax:

Operands:

Operation:

PC = TOSGIE = 1

RETURN	Return from Subroutine					
Syntax:	[label] R	RETURN				
Operands:	None					
Operation:	$TOS \to PC$;				
Status Affected:	None					
Encoding:	00	0000	0000	1000		
Description:	Return from subroutine. The stack is POPed and the top of the stack (TOS) is loaded into the program counter. This is a two-cycle instruction.					
Words:	1					
Cycles:	2					
Q Cycle Activity:	Q1	Q2	Q3	Q4		
1st Cycle	Decode	No Operation	No Operation	Pop from the Stack		
2nd Cycle	No No No No Operation Operation Operation					
Example	RETURN					

After Interrupt: PC = TOS

RLF	Rotate Left f through Carry						
Syntax:	[label] RLF f[,d]						
Operands:	$0 \le f \le 127$ $d \in [0,1]$						
Operation:	See desc	cription b	elow				
Status Affected:	С						
Encoding:	00	1101	dfff	ffff			
Description:	The contents of register 'f' are rotated one bit to the left through the Carry Flag. If 'd' is 0, the result is placed in the W register. If 'd' is 1, the result is stored back in register 'f'.						
Words:	1						
Cycles:	1						
Q Cycle Activity:	Q1	Q2	Q3	Q4			
	Decode	Read register 'f'	Process data	Write to destination			

REG1 = 1110 0110 C = 0

After Instruction:

REG1 = 1110 0110 W = 1100 1100 C = 1

RRF	Rotate Right f through Carry	SLEEP			
Syntax:	[label] RRF f[,d]	Syntax:	[label] SLEEP		
Operands:	$0 \le f \le 127$ $d \in [0,1]$	Operands: Operation:	None $00h \rightarrow WDT$.		
Operation:	See description below		$0 \rightarrow WDT$ prescaler,		
Status Affected:	С		$ \begin{array}{l} 1 \to \overline{\text{TO}}, \\ 0 \to \overline{\text{PD}} \end{array} $		
Encoding:	00 1100 dfff ffff	Status Affected:	$\overline{TO}, \overline{PD}$		
Description:	The contents of register 'f' are rotated one bit to the right through the Carry	Encoding:	00 0000 0110 0011		
		Description:	The power-down status bit, PD is cleared. Time-out status bit, TO is set. Watchdog Timer and its prescaler are cleared. The processor is put into SLEEP		
Words:	1		mode with the oscillator stopped. See Section 12.8 for more details.		
Cycles:	1	Words:	1		
Q Cycle Activity:	Q1 Q2 Q3 Q4	Cycles:	1		
	Decode register data Write to	Q Cycle Activity:	Q1 Q2 Q3 Q4		
	'f' data destination		Decode No No Go to Operation Operation Sleep		

Example:

SLEEP

Example REG1,0

Before Instruction:

REG1 = 1110 0110

С = 0

After Instruction:

REG1 = 1110 0110

W 0111 0011

С

SUBLW	Subtract W from Literal	SUBWF	Subtract W from f
Syntax:	[label] SUBLW k	Syntax:	[label] SUBWF f [,d]
Operands:	$0 \le k \le 255$	Operands:	$0 \le f \le 127$
Operation:	$k - (W) \rightarrow (W)$		d ∈ [0,1]
Status Affected:	C, DC, Z	Operation:	(f) - (W) \rightarrow (destination)
Encoding:	11 110x kkkk kkkk	Status Affected:	C, DC, Z
Description:	The W register is subtracted (2's	Encoding:	00 0010 dfff ffff
·	complement method) from the eight- bit literal 'k'. The result is placed in the W register.	Description:	Subtract (2's complement method) W register from register 'f'. If 'd' is 0, the result is stored in the W register. If 'd' is 1, the result is stored back in register 'f'.
Words:	1	Words:	1
Cycles:	1	Cycles:	1
Q Cycle Activity:	Q1 Q2 Q3 Q4	Q Cycle Activity:	Q1 Q2 Q3 Q4
	Decode Read Process data Write to W	<u> </u>	Decode Read Process Write to data destination
Example 1:	SUBLW 0x02		
Before Instru		Example 1:	SUBWF REG1,1
W =	1	Before Instru	
C = Z =	?	REG1 W	= 3 = 2
After Instruct	·	C	= 2 = ?
W =	1	Z	= ?
<u>C</u> =	1; result is positive	After Instruct	
Z =	0	REG1 W	= 1 = 2
Example 2:		Ċ	= 1; result is positive
Before Instru W =	ction: 2	Z	= 0
C =	?	Example 2:	
Z =	?	Before Instru	
After Instruct W =	ion: 0	REG1 W	= 2 = 2
	1; result is zero	С	= ?
Z =	1	Z	= ?
Example 3:		After Instruct REG1	ion: = 0
Before Instru		W	= 2
W = C =	3 ?	C	= 1; result is zero
Z =	?	Z	= 1
After Instruct	ion:	Example 3:	
W =	0xFF	Before Instru REG1	iction: = 1
C = Z =	0; result is negative 0	W	= 2
		C	= ?
		Z	= ?
		After Instruct REG1	ion: = 0xFF
		W	= 2
		C Z	= 0; result is negative = 0
		_	- •

SWAPF	Swap Nibbles in f							
Syntax:	[label]	[label] SWAPF f [,d]						
Operands:	$0 \le f \le 12$ $d \in [0,1]$	$0 \le f \le 127$ $d \in [0,1]$						
Operation:	,	$(f<3:0>) \rightarrow (destination<7:4>),$ $(f<7:4>) \rightarrow (destination<3:0>)$						
Status Affected:	None	None						
Encoding:	00 1110 dfff ffff							
Description:	The upper and lower nibbles of register 'f' are exchanged. If 'd' is 0, the result is placed in W register. If 'd' is 1, the result is placed in register 'f'.							
Words:	1							
Cycles:	1							
Q Cycle Activity:	Q1 Q2 Q3 Q4							
	Decode	Read register 'f'	Process data	Write to destination				

SWAPF REG,

0

TRIS	Load TRIS Register
Syntax:	[label] TRIS f
Operands:	$5 \le f \le 7$
Operation:	$(W) \rightarrow TRIS register f;$
Status Affected:	None
Encoding:	00 0000 0110 Offf
Words: Cycles: Example	The instruction is supported for code compatibility with the PIC16C5X products. Since TRIS registers are readable and writable, the user can directly address them. 1
	To maintain upward compatibility with future PIC16CXXX products, do not use this instruction.

Example

Before Instruction:

After Instruction:

REG1 = 0xA5

REG1 = 0xA5

= 0x5A

XORLW Exclusive OR Literal with W

Syntax: [label] XORLW k

Operands: $0 \le k \le 255$

Operation: (W) .XOR. $k \rightarrow (W)$

Status Affected: Z

Encoding: 11 1010 kkkk kkkk

Description: The contents of the W register are

XOR'ed with the eight-bit literal 'k'. The result is placed in the W

register.

Words: 1

Cycles: 1

Q Cycle Activity: Q1 Q2 Q3 Q4

Decode Read Process Write to data W

Example: XORLW 0xAF

Before Instruction:

W = 0xB5

After Instruction:

W = 0x1A

NOTES:

14.0 DEVELOPMENT SUPPORT

The PICmicro[®] microcontrollers are supported with a full range of hardware and software development tools:

- Integrated Development Environment
 - MPLAB® IDE Software
- · Assemblers/Compilers/Linkers
 - MPASM™ Assembler
 - MPLAB C17 and MPLAB C18 C Compilers
 - MPLINKTM Object Linker/ MPLIBTM Object Librarian
- Simulators
 - MPLAB SIM Software Simulator
- Emulators
 - MPLAB ICE 2000 In-Circuit Emulator
 - ICEPIC™ In-Circuit Emulator
- · In-Circuit Debugger
 - MPLAB ICD for PIC16F87X
- · Device Programmers
 - PRO MATE® II Universal Device Programmer
 - PICSTART® Plus Entry-Level Development Programmer
- Low Cost Demonstration Boards
 - PICDEM™ 1 Demonstration Board
 - PICDEM 2 Demonstration Board
 - PICDEM 3 Demonstration Board
 - PICDEM 17 Demonstration Board
 - KEELOQ® Demonstration Board

14.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8-bit microcontroller market. The MPLAB IDE is a Windows®-based application that contains:

- · An interface to debugging tools
 - simulator
 - programmer (sold separately)
 - emulator (sold separately)
 - in-circuit debugger (sold separately)
- · A full-featured editor
- · A project manager
- · Customizable toolbar and key mapping
- · A status bar
- · On-line help

The MPLAB IDE allows you to:

- Edit your source files (either assembly or 'C')
- One touch assemble (or compile) and download to PICmicro emulator and simulator tools (automatically updates all project information)
- Debug using:
 - source files
 - absolute listing file
 - machine code

The ability to use MPLAB IDE with multiple debugging tools allows users to easily switch from the cost-effective simulator to a full-featured emulator with minimal retraining.

14.2 MPASM Assembler

The MPASM assembler is a full-featured universal macro assembler for all PICmicro MCU's.

The MPASM assembler has a command line interface and a Windows shell. It can be used as a stand-alone application on a Windows 3.x or greater system, or it can be used through MPLAB IDE. The MPASM assembler generates relocatable object files for the MPLINK object linker, Intel[®] standard HEX files, MAP files to detail memory usage and symbol reference, an absolute LST file that contains source lines and generated machine code, and a COD file for debugging.

The MPASM assembler features include:

- · Integration into MPLAB IDE projects.
- User-defined macros to streamline assembly code
- Conditional assembly for multi-purpose source files
- Directives that allow complete control over the assembly process.

14.3 MPLAB C17 and MPLAB C18 C Compilers

The MPLAB C17 and MPLAB C18 Code Development Systems are complete ANSI 'C' compilers for Microchip's PIC17CXXX and PIC18CXXX family of microcontrollers, respectively. These compilers provide powerful integration capabilities and ease of use not found with other compilers.

For easier source level debugging, the compilers provide symbol information that is compatible with the MPLAB IDE memory display.

14.4 MPLINK Object Linker/ MPLIB Object Librarian

The MPLINK object linker combines relocatable objects created by the MPASM assembler and the MPLAB C17 and MPLAB C18 C compilers. It can also link relocatable objects from pre-compiled libraries, using directives from a linker script.

The MPLIB object librarian is a librarian for precompiled code to be used with the MPLINK object linker. When a routine from a library is called from another source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications. The MPLIB object librarian manages the creation and modification of library files.

The MPLINK object linker features include:

- Integration with MPASM assembler and MPLAB C17 and MPLAB C18 C compilers.
- Allows all memory areas to be defined as sections to provide link-time flexibility.

The MPLIB object librarian features include:

- Easier linking because single libraries can be included instead of many smaller files.
- Helps keep code maintainable by grouping related modules together.
- Allows libraries to be created and modules to be added, listed, replaced, deleted or extracted.

14.5 MPLAB SIM Software Simulator

The MPLAB SIM software simulator allows code development in a PC-hosted environment by simulating the PICmicro series microcontrollers on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a file, or user-defined key press, to any of the pins. The execution can be performed in single step, execute until break, or trace mode.

The MPLAB SIM simulator fully supports symbolic debugging using the MPLAB C17 and the MPLAB C18 C compilers and the MPASM assembler. The software simulator offers the flexibility to develop and debug code outside of the laboratory environment, making it an excellent multiproject software development tool.

14.6 MPLAB ICE High Performance Universal In-Circuit Emulator with MPLAB IDE

The MPLAB ICE universal in-circuit emulator is intended to provide the product development engineer with a complete microcontroller design tool set for PICmicro microcontrollers (MCUs). Software control of the MPLAB ICE in-circuit emulator is provided by the MPLAB Integrated Development Environment (IDE), which allows editing, building, downloading and source debugging from a single environment.

The MPLAB ICE 2000 is a full-featured emulator system with enhanced trace, trigger and data monitoring features. Interchangeable processor modules allow the system to be easily reconfigured for emulation of different processors. The universal architecture of the MPLAB ICE in-circuit emulator allows expansion to support new PICmicro microcontrollers.

The MPLAB ICE in-circuit emulator system has been designed as a real-time emulation system, with advanced features that are generally found on more expensive development tools. The PC platform and Microsoft® Windows environment were chosen to best make these features available to you, the end user.

14.7 ICEPIC In-Circuit Emulator

The ICEPIC low cost, in-circuit emulator is a solution for the Microchip Technology PIC16C5X, PIC16C6X, PIC16C7X and PIC16CXXX families of 8-bit One-Time-Programmable (OTP) microcontrollers. The modular system can support different subsets of PIC16C5X or PIC16CXXX products through the use of interchangeable personality modules, or daughter boards. The emulator is capable of emulating without target application circuitry being present.

14.8 MPLAB ICD In-Circuit Debugger

Microchip's In-Circuit Debugger, MPLAB ICD, is a powerful, low cost, run-time development tool. This tool is based on the FLASH PIC16F87X and can be used to develop for this and other PICmicro microcontrollers from the PIC16CXXX family. The MPLAB ICD utilizes the in-circuit debugging capability built into the PIC16F87X. This feature, along with Microchip's In-Circuit Serial Programming™ protocol, offers cost-effective in-circuit FLASH debugging from the graphical user interface of the MPLAB Integrated Development Environment. This enables a designer to develop and debug source code by watching variables, single-stepping and setting break points. Running at full speed enables testing hardware in real-time.

14.9 PRO MATE II Universal Device Programmer

The PRO MATE II universal device programmer is a full-featured programmer, capable of operating in stand-alone mode, as well as PC-hosted mode. The PRO MATE II device programmer is CE compliant.

The PRO MATE II device programmer has programmable VDD and VPP supplies, which allow it to verify programmed memory at VDD min and VDD max for maximum reliability. It has an LCD display for instructions and error messages, keys to enter commands and a modular detachable socket assembly to support various package types. In stand-alone mode, the PRO MATE II device programmer can read, verify, or program PICmicro devices. It can also set code protection in this mode.

14.10 PICSTART Plus Entry Level Development Programmer

The PICSTART Plus development programmer is an

14.13 PICDEM 3 Low Cost PIC16CXXX Demonstration Board

The PICDEM 3 demonstration board is a simple demonstration board that supports the PIC16C923 and PIC16C924 in the PLCC package. It will also support future 44-pin PLCC microcontrollers with an LCD Module. All the necessary hardware and software is included to run the basic demonstration programs. The user can program the sample microcontrollers provided with the PICDEM 3 demonstration board on a PRO MATE II device programmer, or a PICSTART Plus development programmer with an adapter socket, and easily test firmware. The MPLAB ICE in-circuit emulator may also be used with the PICDEM 3 demonstration board to test firmware. A prototype area has been provided to the user for adding hardware and connecting it to the microcontroller socket(s). Some of the features include a RS-232 interface, push button switches, a potentiometer for simulated analog input, a thermistor and separate headers for connection to an external LCD module and a keypad. Also provided on the PICDEM 3 demonstration board is a LCD panel, with 4 commons and 12 segments, that is capable of displaying time, temperature and day of the week. The PICDEM 3 demonstration board provides an additional RS-232 interface and Windows software for showing the demultiplexed LCD signals on a PC. A simple serial interface allows the user to construct a hardware demultiplexer for the LCD signals.

14.14 PICDEM 17 Demonstration Board

The PICDEM 17 demonstration board is an evaluation board that demonstrates the capabilities of several Microchip microcontrollers, including PIC17C752, PIC17C756A, PIC17C762 and PIC17C766. All necessary hardware is included to run basic demo programs, which are supplied on a 3.5-inch disk. A programmed sample is included and the user may erase it and program it with the other sample programs using the PRO MATE II device programmer, or the PICSTART Plus development programmer, and easily debug and test the sample code. In addition, the PICDEM 17 demonstration board supports downloading of programs to and executing out of external FLASH memory on board. The PICDEM 17 demonstration board is also usable with the MPLAB ICE in-circuit emulator, or the PICMASTER emulator and all of the sample programs can be run and modified using either emulator. Additionally, a generous prototype area is available for user hardware.

14.15 KEELOQ Evaluation and Programming Tools

KEELOQ evaluation and programming tools support Microchip's HCS Secure Data Products. The HCS evaluation kit includes a LCD display to show changing codes, a decoder to decode transmissions and a programming interface to program test transmitters.

TABLE 14-1: DEVELOPMENT TOOLS FROM MICROCHIP

	sloc	T ə	war	floS	S10	Emulat	Depnââeı	ammers	Progr
	MPLAB® Integrated Development Environment	MPLAB® C17 C Compiler	MPLAB® C18 C Compiler	MPASM™ Assembler/ MPLINK™ Object Linker	MPLAB® ICE In-Circuit Emulator	ICEPIC TM In-Circuit Emulator	MPLAB® ICD In-Circuit Debugger	PICSTART® Plus Entry Level	
PIC12CXXX	>			>	` -	>			
PIC14000	>			>	>				
PIC16C5X	>			>	>	>			
PIC16C6X	>			>	>	>	*		
PIC16CXXX	>			>	>	>			
PIC16F62X	>			>	*				
XTO81519	>			>	>	>	*		
PIC16C7XX	>			>	>	>			
PIC16C8X	>			>	>	>			
PIC16F8XX	· ·			`	`		,		
PIC16C9XX	`	•		`	`	`			
PIC17C4X	<i>></i>	,		<i>,</i>	`				
PIC18CXX2	<i>></i> ,		>	>	>				
63CXX SPCXX\				>					
нсеххх				>					
WCKEXXX									
WCP2510									
TABLE 14	-1:	D)E\	/ELC	ארנ	/IEN I	TOOLS	FROM	MIC

NOTES:

15.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

Ambient temperature under bias	
Storage temperature	65°C to +150°C
Voltage on any pin with respect to Vss (except VDD, MCLR, and RA4)	
Voltage on VDD with respect to Vss	0V to +7.5V
Voltage on MCLR with respect to Vss	0V to +13.25V
Voltage on RA4 with respect to Vss	0V to +8.5V
Voltage on VLCD2, VLCD3 with respect to Vss	0V to +10V
Total power dissipation (Note 1)	1.0 W
Maximum current out of Vss pin	300 mA
Maximum current into VDD pin	250 mA
Input clamp current, Iik (VI < 0 or VI > VDD)	± 20 mA
Output clamp current, loκ (Vo < 0 or Vo > VDD)	± 20 mA
Maximum output current sunk by any I/O pin	25 mA
Maximum output current sourced by any I/O pin	25 mA
Maximum current sunk by all Ports combined	200 mA
Maximum current sourced by all Ports combined	200 mA
Note 1: Power dissipation is calculated as follows: PDIS = VDD x {IDD - Σ IOH} + Σ {(VD	OD - VOH) \times IOH} + Σ (VOI \times IOL)

† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

FIGURE 15-1: PIC16C925/926 VOLTAGE-FREQUENCY GRAPH

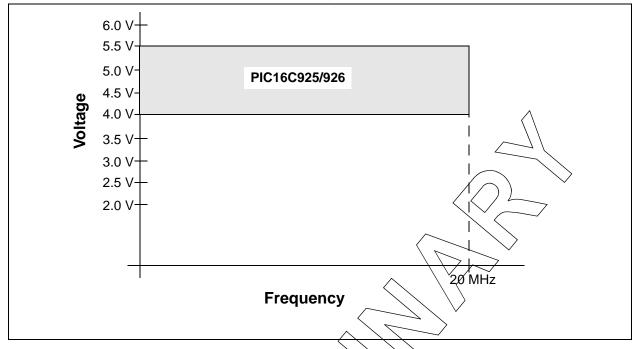
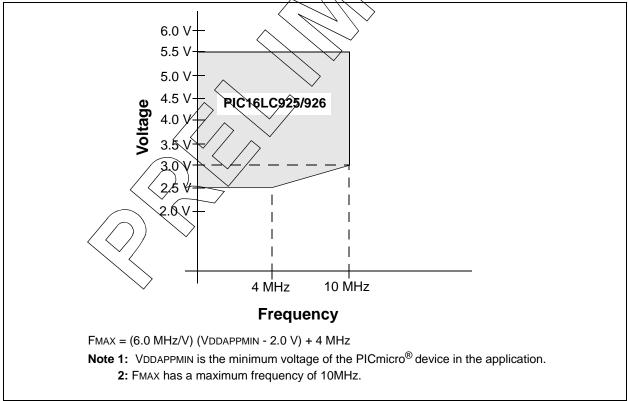


FIGURE 15-2: PIC16LC925/926 VOLTAGE-FREQUENCY GRAPH



DC Characteristics 15.1

PIC16LC925/926

(Commercial, Industrial)

Standard Operating Conditions (unless otherwise stated)

-40°C \leq TA \leq +85°C for industrial Operating temperature

 $0^{\circ}C \leq TA \leq +70^{\circ}C$ for commercial

PIC16C925/926

(Commercial, Industrial)

Standard Operating Conditions (unless otherwise stated)

-40°C ≤ TA ≤ +85°C for industrial 0°C ≤ TA ≤ +70°C for commercial Operating temperature

Param No.

Sym

Characteristic

ise se TTT

Min Typ† Max Units

Conditions

Supply Voltage VDD

D001

D001A

15.1 DC Characteristics (Continued)

	0925/926 mercial, In	ndustrial)	Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}\text{C} \leq \text{Ta} \leq +85^{\circ}\text{C}$ for industrial $0^{\circ}\text{C} \leq \text{Ta} \leq +70^{\circ}\text{C}$ for commercial				
PIC16C9 (Comr	925/926 mercial, In	ndustrial)		ard Op ting ter			ditions (unless otherwise stated) -40°C ≤ TA ≤ +85°C for industrial 0°C ≤ TA ≤ +70°C for commercial
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions
	IPD	Power-down Current (N	ote 3)				
D020		PIC16LC925/926	_	0.9	5	μΑ	VDD = 3.0V
D020		PIC16C925/926	_	1.5	21	μĄ∕	VDD = 4.0V <
		Module Differential Cur	rent (N	ote 5)			
D021	Δ I WDT	Watchdog Timer PIC16LC925/926	_	6.0	20 <	μΑ	VDD = 3.0V
D021		Watchdog Timer PIC16C925/926	_	9.0	25	μÀ	VoD = 4.0V
D022	ΔILCDT1	LCD Voltage Generation with internal RC osc enabled PIC16LC925/926	_ 	36	50))	VDD = 3.0V (Note 7)
D022		LCD Voltage Generation with internal RC osc enabled PIC16C925/926		40	5 <i>8</i>	μΑ	VDD = 4.0V (Note 7)
D022A	∆lbor	Brown out Reset	_ `	100	150	μΑ	BODEN bit set, VDD = 5.0
D024	AILCDT1	LCD Voltage Generation with Timer1 @ 32,768 kHz PIC1/6LC925/926	/>	15	29	μА	VDD = 3.0V (Note 7)
D024		Generation with Timer1 @ 32.768 kHz PIC16C925/926	_	33	60	μА	VDD = 4.0V (Note 7)

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered in SLEEP mode without losing RAM data.

: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption. The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail to rail; all I/O pins tri-stated, pulled to VDD

 $\overline{MCLR} = VDD.$

- 3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD and Vss.
- **4:** For RC osc configuration, current through REXT is not included. The current through the resistor can be estimated by the formula Ir = VDD/2REXT (mA) with REXT in kOhm.
- 5: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.
- 6: PWRT must be enabled for slow ramps.
- 7: ΔILCDT1 and ΔILCDRC includes the current consumed by the LCD Module and the voltage generation circuitry. This does not include current dissipated by the LCD panel.

15.1 DC Characteristics (Continued)

	925/926 mercial, In	dustrial)		Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}\text{C} \leq \text{TA} \leq +85^{\circ}\text{C}$ for industrial $0^{\circ}\text{C} \leq \text{TA} \leq +70^{\circ}\text{C}$ for commercial						
PIC16C9 (Com	925/926 mercial, In	dustrial)		Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial 0°C ≤ TA ≤ +70°C for commercial						
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions			
D025	∆lT1OSC	Timer1 Oscillator PIC16LC925/926	_	_	50	μА	Ved = 3.0V			
D025		Timer1 Oscillator PIC16C925/926	_	_	50	μA	VDD ≠ 4.0V			
D026	ΔIAD	A/D Converter PIC16LC925/926	_	1.0	<	μА	AD on, not converting			
D026		A/D Converter PIC16C925/926	_	1.0		μA	A/D on, not converting			

- Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.
- Note 1: This is the limit to which VDD can be lowered in SLEEP mode without losing RAM data.
 - 2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption. The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave from rail to rail;

all I/O pins tri-stated, pulled to VDD

- MCLR = VDD.

 3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD and Vss.
- 4: For RC osc configuration, current through REXT is not included. The current through the resistor can be estimated by the formula Ir = VDD/2REXT (mA) with REXT in kOhm.
- 5: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base/IDD or IPD measurement.
- 6: PWRT must be enabled for slow ramps.
- 7: \(\Delta \text{ILCDT1} \) and \(\Delta \text{LCDRC} \) includes the current consumed by the LCD Module and the voltage generation circuitry. This does not include current dissipated by the LCD panel.

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15.2 DC Characteristics: PIC16C925/926 (Commercial, Industrial) PIC16LC925/926 (Commercial, Industrial)

DC CHARACTERISTICS Standard Operating Conditions (unless otherwise stated)

Operating temperature $-40^{\circ}\text{C} \leq \text{TA} \leq +85^{\circ}\text{C}$ for industrial $0^{\circ}\text{C} \leq \text{TA} \leq +70^{\circ}\text{C}$ for commercial Operating voltage VDD range as described in DC spec

-		<u> </u>	. `	_		_	·
Param	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
No.							
		Input Low Voltage					
	VIL	I/O ports					
D030		with TTL buffer	Vss		0.15VDD	V	For entire VDD range
			Vss	_	0.8V	V	4.5V ≤ VDD ≤ 5.5V⁄\
D031		with Schmitt Trigger buffer	Vss	_	0.2Vdd	V	[
D032		MCLR, OSC1 (in RC mode)	Vss	_	0.2Vdd	V	
D033		OSC1 (in XT, HS and LP)	Vss	_	0.3VDD	V	(Note 1)
		Input High Voltage					
	VIH	I/O ports					
D040		with TTL buffer	2.0	_	Vdd	V	4.5V ≤ VDD ≤ 5.5V
D040A			0.25VDD	_	Vdd	V	For entire VDQ range
			+ 0.8V				· · ·
D041		with Schmitt Trigger buffer	0.8Vpp		VDD		
D042		MCLR	0.8Vpp		VDD .	V \	
D042A		OSC1 (XT, HS and LP)	0.7Vdd		VDD <	V	(Note 1)
D043		OSC1 (in RC mode)	0.9VDD	_	VDD	///	
D070	IPURB	PORTB Weak Pull-up Current	50	250	400~	μA	VDD = 5V, VPIN = VSS
		Input Leakage Current		<			
		(Notes 2, 3)		\wedge		\sim	
D060	lı∟	I/O ports	_ <	(~	(p.f±	μΑ	Vss ≤ VPIN ≤ VDD, Pin at hi-Z
D061		MCLR, RA4/T0CKI	_	1	±5 ±5	μΑ	Vss ≤ Vpin ≤ Vdd
D063		OSC1		7,	\±5	μΑ	Vss ≤ VPIN ≤ VDD, XT, HS and LP
					/		osc configuration
		Output Low Voltage		\triangleright			
D080	Vol	I/O ports	+/	. —	0.6	V	IOL = 4.0 mA, VDD = 4.5V
D083		OSC2/CLKOUT (RC osc mode)	_~		0.6	V	IOL = 1.6 mA, VDD = 4.5V
		Output High Voltage					
D090	Voн	I/O ports (Note 3)	√p6 - 0.7		_	V	IOH = -3.0 mA, VDD = 4.5V
D092		OSC2/CLKOUT (RC osc mode)	√DD - 0.7	_		V	IOH = -1.3 mA, VDD = 4.5V
		Capacitive Loading Specs on					
		Output Pins					
D100	Cosc ₂	OSC2 pin	_	_	15	pF	In XT, HS and LP modes when
							external clock is used to drive
	_				_		OSC1.
D101		All I/O pins and QSC2 (in RC)	-	_	50	pF	
D102	Св	SCL, SDA in I ² C mode	_	_	400	pF	
D150	VDD	Open Drain High Voltage	_	_	8.5	V	RA4 pin

[†] Data in "Typ" column is at 5 V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: In RC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended that the PIC16C925/926 be driven with external clock in RC mode.

^{2:} The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

^{3:} Negative current is defined as current sourced by the pin.



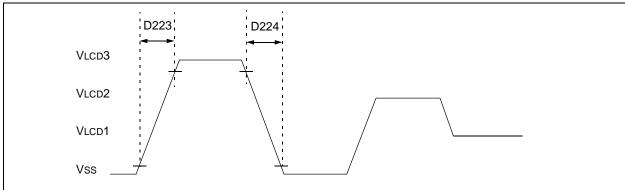


TABLE 15-1: LCD MODULE ELECTRICAL SPECIFICATIONS

Parameter No.	Symbol	Characteristic	Min	Тур†	Max	Units	Conditions
D200	VLCD3	LCD Voltage on pin VLCD3	VDD - 0.3	_	Vss + 7.0	V/	
D201	VLCD2	LCD Voltage on pin VLCD2	Vss - 0.3	_	VLCD3	V.	
D202	VLCD1	LCD Voltage on pin VLCD1	Vss - 0.3	_	VLCD3	V /	7
D220	Voн	Output High Voltage	Max VLCDN - 0.1	_	Max VLCDN	X	COM outputs IOH = 25 μA SEG outputs IOH = 3 μA
D221	VOL	Output Low Voltage	Min VLCDN	_	Min VLCBN + 0.1	V	COM outputs IOL = 25 μ A SEG outputs IOL = 3 μ A
D222	FLCDRC	LCDRC Oscillator Frequency	5	14	22	kHz	VDD = 5V, -40°C to +85°C
D223	TrLCD	Output Rise Time	- <		200	μs	COM outputs Cload = 5,000 pF SEG outputs Cload = 500 pF VDD = 5.0V, T = 25°C
D224	TfLCD	Output Fall Time ⁽¹⁾	TrLCD - 0:05		TrLCD + 0.05 TrLCD	μs	COM outputs Cload = 5,000 pF SEG outputs Cload = 500 pF VDD = 5.0V, T = 25°C

[†] Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: 0 ohm source impedance at VLCD

TABLE 15-2: VLCD CHARGE PUMP ELECTRICAL SPECIFICATIONS

Parameter No.	Symbol	Charact	eristic	Min	Тур	Max	Units	Conditions
D250	IVADJ)	VYCDADJ Regulated Current Output			10	_	μΑ	
D252	∆ WADJ/∆ VDD	VLCDADJ Current VDD Rejection			_	0.1	μA/V	
D265	VVADJ	VLCDADJ Voltage	PIC16 C 925/926	1.0	_	2.3	V	
	\	Limits	PIC16 LC 925/926	1.0		VDD - 0.7V	V	VDD < 3V

Note 1: For design guidance only.

1. TppS2ppS

BUF

HD

DAT

STA

FIGURE 15-4:

CC

ST

15.3 Timing Parameter Symbology

The timing parameter symbols have been created following one of the following formats:

2. TppS		4. Ts (I ² C sp	ecifications only)	
Т				
F	Frequency	Т	Time	
Lowercase	letters (pp) and their meanings:			
рр				
CC	CCP1	osc	OSC1	
ck	CLKOUT	rd	RD	
cs	CS	rw	RD or WR	
di	SDI	sc	SCK	
do	SDO	SS	SS	
dt	Data in	t0	T0CKI	
io	I/O port	t1	T1CKI	
mc	MCLR	wr	WR	
Uppercase	letters and their meanings:			
S				
F	Fall	Р	Period	
Н	High	R	Rise	
I	Invalid (Hi-impedance)	V	Valid	
L	Low	Z	Hi-impedance	
I ² C only				
AA	output access	High	High	

Low

SU

STO

Low

Setup

STOP condition

3. Tcc:st (I²C specifications only)

DATA input hold

START condition

Bus free

Tcc:st (I²C specifications only)

Hold



LOAD CONDITIONS

15.4 Timing Diagrams and Specifications

FIGURE 15-5: EXTERNAL CLOCK TIMING

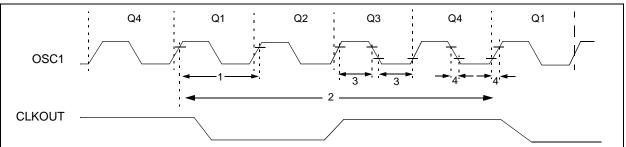


TABLE 15-3: EXTERNAL CLOCK TIMING REQUIREMENTS

Parameter No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions
	Fosc	External CLKIN Frequency	DC	_	4	MHz∠	XT and RC osc mode
		(Note 1)	DC	_	20	MHz	HS osc mode
			DC	_	200	(kHz)	LP osc mode
		Oscillator Frequency	DC	_	4	MHŽ<	RC osc mode
		(Note 1)	0.1	_	4	MHz	XT osc mode
			4	_		VMHz	HS osc mode
			5		200	kHz	LP osc mode
1	Tosc	External CLKIN Period	250	_ <		ns	XT and RC osc mode
		(Note 1)	125	^	17	ns	HS osc mode
			5		Š	μs	LP osc mode
		Oscillator Period	250		\	ns	RC osc mode
		(Note 1)	250		10,000	ns	XT osc mode
			125	$\overline{}$	250	ns	HS osc mode
			5	\Diamond	_	μs	LP osc mode
2	TCY	Instruction Cycle Time (Note 1)	500	1	DC	ns	Tcy = 4/Fosc
3	TosL,	External Clock in (OSC1) High or	50		_	ns	XT oscillator
	TosH	Low Time	2.5	_	_	μs	LP oscillator
			10		_	ns	HS oscillator
4	TosR,	External Clock in (OSC1) Rise or	_	_	25	ns	XT oscillator
	TosF	Fall Time	_	_	50	ns	LP oscillator
			_	_	15	ns	HS oscillator

[†] Data in "Typ" column is at 5 1/25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (TCY) equals four times the input oscillator time-base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min." values with an external clock applied to the OSC1/CLKIN pin.

When an external clock input is used, the "Max." cycle time limit is "DC" (no clock) for all devices.

CLKOUT AND I/O TIMING FIGURE 15-6: Q4 Q1 Q2 Q3 OSC₁ CLKOUT 19 I/O Pin (input) I/O Pin old value new value (output) 20, 21 Note: Refer to Figure 15-4 for load conditions.

TABLE 15-4: CLKOUT AND I/O TIMING REQUIREMENTS

Parameter No.	Symbol	Characte	ristic	Min	Тур†	Max	Units	Conditions
10	TosH2ckL	OSC1↑ to CLKOUT↓		_	75	200	ns	(Note 1)
11	TosH2ckH	OSC1↑ to CLKOUT↑		_	75	200	ns	(Note 1)
12	TckR	CLKOUT rise time		_	35	100	ns	(Note 1)
13	TckF	CLKOUT fall time		_	35	100	ns	(Note 1)
14	TckL2ioV	CLKOUT ↓ to Port out va	alid	_	_	0.5TcY +20	ns	(Note 1)
15	TioV2ckH	Port in valid before CLK	OUT ↑	Tosc + 200	_		ns	(Note 1)
16	TckH2ioI	Port in hold after CLKOL	JT ↑	0 🖯	1	100	ns	(Note 1)
17	TosH2ioV	OSC1↑ (Q1 cycle) to Port out valid		12	50	<u></u>	ns	
18	TosH2iol	OSC1↑ (Q2 cycle) to	PIG16C925/926	100	_	_	ns	
		Port input invalid (I/O in hold time)	PIC 16LC 925X926	200	_	_	ns	
19	TioV2osH	Port input valid to OSC1	#O in setup time)	0	_	_	ns	
20	TioR	Port output rise time	PIC16 C 925/926	_	10	40	ns	
	1 / -		PIC16 LC 925/926	_	_	80	ns	
21	TioF 🗸	Port output fall time	PIC16 C 925/926	_	10	40	ns	
			PIC16 LC 925/926	_	_	80	ns	
22††	Tinp	INT pin high or low time		Tcy	_	_	ns	
23††	Trbp	RB7:RB4 change INT his	gh or low time	Tcy	_	_	ns	

[†] Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Measurements are taken in RC mode where CLKOUT output is 4 x Tosc.

^{††} These parameters are asynchronous events not related to any internal clock edges.

FIGURE 15-7: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING

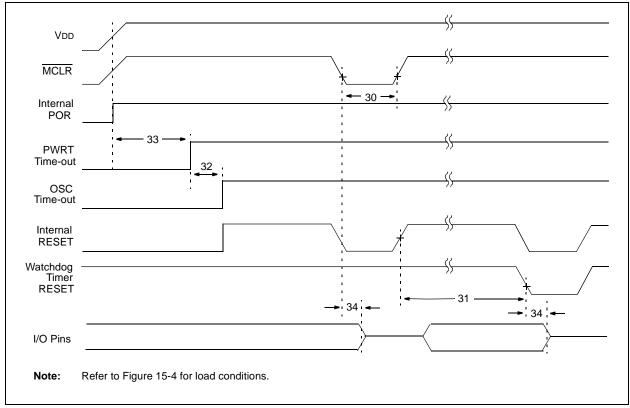


TABLE 15-5: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER REQUIREMENTS

Parameter No.	Symbol	Characteristic	Min	Тур†	Max	Units	Conditions
30	TmcL	MCLR Pulse Width (low)	2	_	_	μs	
31	TWDT	Watchdog Timer Time-out Period (No Prescaler)	7	18	33	ms	VDD = 5V, -40°C to +85°C
32	Tost	Oscillation Start-up Timer Period	_	1024Tosc	_	_	Tosc = OSC1 period
33	TPWRT	Power-up Timer Period	28	72	132	ms	VDD = 5V, -40°C to +85°C
34	Tioz	I/O Hi-impedance from MCLR Low or Watchdog Timer Reset	_	_	2.1	μs	

[†] Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.1

RA4/T0CKI

RC0/T10SO/T1CKI

TMR0 or TMR1

Note: Refer to Figure 15-4 for load conditions.

TABLE 15-6: TIMERO AND TIMER1 EXTERNAL CLOCK REQUIREMENTS

With Prescaler 10	Conditions
Tt0L	Must also meet
With Prescaler 10	parameter 42
Tt0P	Must also meet
With Prescaler Greater of:	parameter 42
Tith	
45 Tt1H T1CKI High Synchronous, Prescaler 1 0.5TcY + 20 — ns Miles Prescaler Prescaler Prescaler 25 — ns Prescaler Dictional Prescaler P	N = prescale value
Time Synchronous, PiC16C925/926 15 — ns pa Asynchronous PIC16C925/926 30 — ns PiC16LC925/926 50 — ns PiC16LC925/926 50 — ns PiC16LC925/926 50 — ns PiC16LC925/926 50 — ns Synchronous, PiC16C925/926 15 — ns Prescaler = PiC16LC925/926 25 — ns Pic16LC925/926 30 — ns Pic16LC925/926 50 — ns Pic16LC925/926 50 — ns Pic16LC925/926 50 — ns Asynchronous PiC16C925/926 30 — ns PiC16LC925/926 50 — ns PiC16LC925/926 50 — ns	(2, 4,, 256)
Prescaler Pic16LG925/926 25	Must also meet
Asynchronous	parameter 47
PIC16LC925/926 50	
Tt1L	
Time Synchronous, PIC16C925/926 15 ns PIC16C925/926 25 ns PIC16C925/926 25 ns PIC16C925/926 30 ns PIC16C925/926 50 ns PIC16C925/926 50 ns PIC16C925/926 Greater of: Period PIC16C925/926 Greater of: N PIC16C925/926 Greater of: PIC16C925	
Prescaler = PIC16LC925/926 25	
2,4,8 Asynchronous PIC16 C 925/926 30 — ns Pa PIC16 LC 925/926 50 — ns Tt1P T1CKI Input Synchronous PIC16 C 925/926 Greater of: Period Period Ns	
PIC16 LC 925/926 50 — ns 47 Tt1P T1CKI Input Synchronous PIC16 C 925/926 Greater of: Period Priod Priod Priod Pic16 C 925/926 PIC16C92	Must also meet parameter 47
47 Tt1P T1CKI Input Synchronous PIC16C925/926 Greater of:	
Period ns N	
1 Period	N = prescale value
	(1, 2, 4, 8)
PIC16 LC 925/926	N = prescale value
1	(1, 2, 4, 8)
Asynchronous PIC16 C 925/926 60 — ns	
PIC16 LC 925/926 100 — ns	
Ft1 Timer1 oscillator input frequency range (oscillator enabled by setting bit T1OSCEN) DC — 200 kHz	
48 TCKEZtmr1 Delay from external clock edge to timer increment 2Tosc — 7Tosc —	

[†] Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 15-9: CAPTURE/COMPARE/PWM TIMINGS

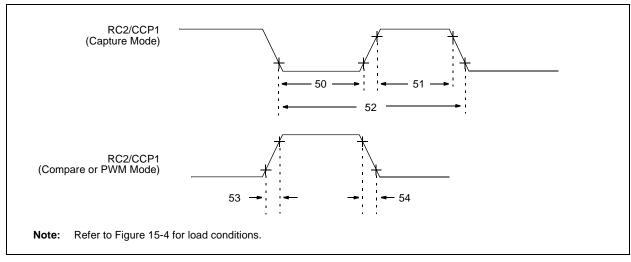
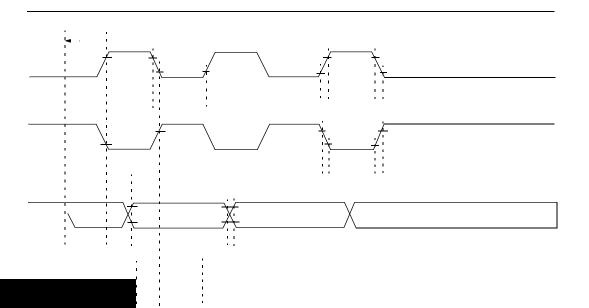


TABLE 15-7: CAPTURE/COMPARE/PWM REQUIREMENTS

Parameter No.	Symbol		Characteristic	:	Min	Тур†	Max	Units	Conditions
50	TccL	Input Low Time	No Prescaler		0.5Tcy + 20	_ <	\exists	ns	
			With Prescaler	PIC16 C 925/926	10	()	14	ns	
				PIC16 LC 925/926	1 20		>-	ns	
51	TccH	Input High Time	No Prescaler	. ^	0,5Tgy + 20	7	_	ns	
			With Prescaler	PIC16@925(926)	10	_	_	ns	
				P(C)164 C925/926\	20	_	_	ns	
52	TccP	Input Period		11/1	3TcY <u> + 40</u> N	_	_		N = prescale value (1,4 or 16)
53	TccR	Output Rise Time		PIC16 C 925/926	_	10	25	ns	
				PIC16 LC 925/926	_	25	45	ns	
54	TccF	Output Fall Time	~	PIC16 C 925/926	_	10	25	ns	
		7		PIC16 LC 925/926	_	25	45	ns	

[†] Data in "Typ" column is at 5 V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 15-10: SPI MASTER MODE TIMING (CKE = 0)



ER MODE TIMING (CKE = 1)

FIGURE 15-12: SPI SLAVE MODE TIMING (CKE = 0)

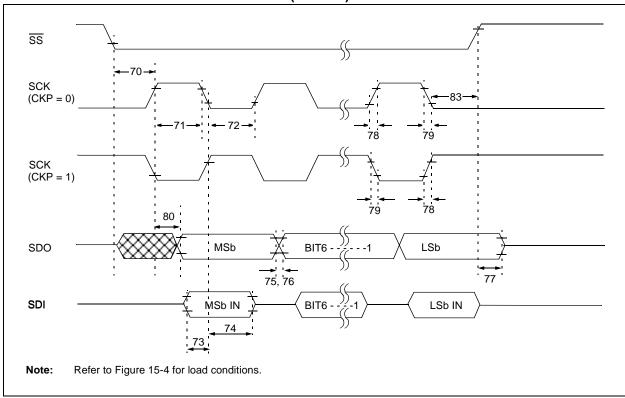


FIGURE 15-13: SPI SLAVE MODE TIMING (CKE = 1)

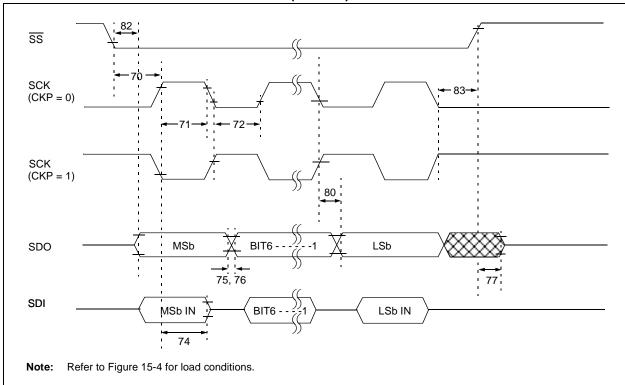


TABLE 15-8: SPI MODE REQUIREMENTS

Param No.	Symbol	Characteristic	:	Min	Typ†	Max	Units	Conditions
70	TssL2scH, TssL2scL	SS↓ to SCK↓ or SCK↑ input		Tcy	_	_	ns	
71	TscH	SCK input high time (Slave	Continuous	1.25Tcy + 30	-	_	ns	
71A		mode)	Single Byte	40	l		ns	
72	TscL	SCK input low time (Slave	Continuous	1.25Tcy + 30	-		ns	
72A		mode)	Single Byte	40	10			
73	TdiV2scH, TdiV2scL	Setup time of SDI data input	to SCK edge	50	1		ns	
74	TscH2diL, TscL2diL	Hold time of SDI data input to	Hold time of SDI data input to SCK edge			_	ns	
75	TdoR	SDO data output rise time	1/4/17 17	_	10	25	ns	
76	TdoF	SDO data output fall time	ПО	_	10	25	ns	
77	TssH2doZ	SS to SDO output hi-impeda	ance	10		50	ns	
78	TscR	SCK output rise time (Master	mode)	_	10	25	ns	
79	TscF	SCK output fall time (Master	mode)	_	10	25	ns	
80	TscH2doV, TscL2doV	SDO data output valid after S	CK edge	_	_	50	ns	
81	TdoV2scH, TdoV2scL	SDO data output setup to SC	K edge	Tcy	_	_	ns	
82	TssL2doV	SDO data output valid after S	S√ edge	_	_	50	ns	
83	TscH2ssH, TscL2ssH	SS ↑ after SCK edge		1.5Tcy + 40	_	_	ns	
84	Tb2b	Delay between consecutive b	ytes	1.5Tcy + 40	_	_	ns	_

[†] Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 15-14: I²C BUS START/STOP BITS TIMING

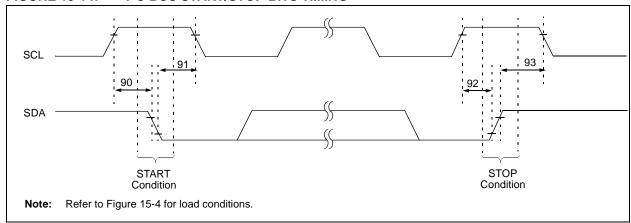


TABLE 15-9: I²C BUS START/STOP BITS REQUIREMENTS

Parameter No.	Symbol	Characteristic	Min	Тур	Max	Units	Conditions
90	Tsu:sta	START condition 100 kHz mode Setup time	4700	\int		ns	Only\relevant for Repeated START condition
91	THD:STA	START condition 100 kHz mode Hold time	4000	A		ns	After this period the first clock pulse is generated
92	Tsu:sto	STOP condition 100 kHz mode Setup time	4700	_		ns	
93	THD:STO	STOP condition 100 kHz mode Hold time	4000	_		ns	

FIGURE 15-15: I²C BUS DATA TIMING

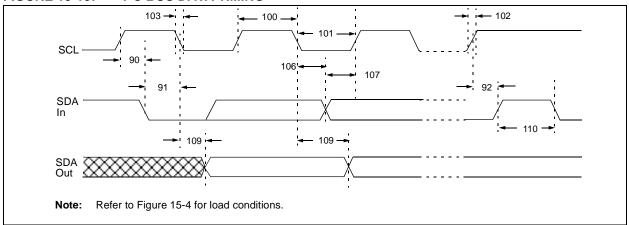


TABLE 15-10: I²C BUS DATA REQUIREMENTS

The stage of time setup time setup time setup figure 100 kHz mode 1.5Tcy 200 ms 1.5 MHz mode 1.5	nameter S	Symbol	Characte	eristic	Min	Max	Units	Conditions
TLOW Clock low time 100 kHz mode 4.7 TLOW Clock low time 100 kHz mode 4.7 TR SDA and SCL rise 100 kHz mode 100 kHz mode 100 kHz mode 100 ns 100 kHz mode 100 kHz mode 100 ns 100 ns 100 kHz mode 100 ns 100	Тн	HIGH	Clock high time	100 kHz mode	4.0	1	μs	Device must operate at a minimum of 1.5 MHz
SSP Module 1.5Tcv TR SDA and SCL rise 100 kHz mode 1.5Tcv T900 ns time 103 TF SDA and SCL fall 100 kHz mode 4.7 μs Only relevant for Restriction START condition 100 kHz mode 4.7 μs Only relevant for Restriction START condition 100 kHz mode 4.0 μs After this period the clock pulse is generated 106 THD:DAT Data input hold 100 kHz mode 0 ns 107 TSU:DAT Data input setup 100 kHz mode 250 ns 109 TAA Output valid from 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs START Late of the clock 100 kHz mode 4.7 μs Late of the clock 100 kHz mode 4.7 μs Late of the clock 100 kHz mode 4.7 μs Late of the clock 100 kHz mode 4.7 μs Late of the clock 100 kHz mode 4.7 μs Late of the clock 100 kHz mode 4.7 μs Late of the clock 100 kHz mode 4.7 μs Late of the clock 100 kHz mode 4.7 μs Late of the clock 100 kHz mode 4.7 μs Late of the clock 100 kHz mode 4.7 μs Late of the clock 100 kHz mode 4.7 μs Late of the clock 100 kHz mode 4.7 μs Late of the clock 100 kHz mode 4.7 μs Late of the clock 100 kHz mode 4.7 μs Late				SSP Module	1.5TcY /	$(\overline{-})$.		
TR SDA and SCL rise time 100 kHz mode time 300 ns 100 kHz mode 4.7 — μs Only relevant for Restriction setup time 100 kHz mode 4.7 — μs Only relevant for Restriction setup time 100 kHz mode 4.0 — μs After this period the clock pulse is generated 106 Thd:DAT Data input hold time 100 kHz mode 0 — ns 100 kHz mode 107 Tsu:DAT Data input setup 100 kHz mode 250 — ns 109 Tsu:STO STOP condition 100 kHz mode 4.7 — μs 100 kHz mode 109 TaA Output valid from 100 kHz mode 100 kHz mode 100 — 100 kHz mode 100 kHz mode 100 — 100 kHz mode 100 MHz mode 100 MHz mode 100 MHz mode 100 kHz mode 100 MHz	Τι	LOW	Clock low time	100 kHz mode	~	\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\	μs	Device must operate at a minimum of 1.5 MHz
time 103 TF SDA and SCL fall time 90 TSU:STA START condition setup time 91 THD:STA START condition hold time 106 THD:DAT Data input hold time 107 TSU:DAT Data input setup time 108 kHz mode 0 — ns 109 TSU:STO STOP condition time 100 kHz mode 0 — ns 100 ms 100 ms 100 kHz mode 0 — ns				SSP Module	1.\$TGY	/		
time 90 TSU:STA START condition setup time 91 THD:STA START condition hold time 100 kHz mode 4.0 — µs After this period the clock pulse is general time 106 THD:DAT Data input hold time 107 TSU:DAT Data input setup time 108 TSU:STO STOP condition setup time 109 TAA Output valid from 100 kHz mode — 3500 ns (Note 1)	TF			100 kHz mode		/1000	ns	
Setup time START condition 100 kHz mode 4.0 — μs After this period the clock pulse is gene 106 ThD:DAT Data input hold 100 kHz mode 0 — ns 107 Tsu:DAT Data input setup 100 kHz mode 250 — ns 109 TaA Output valid from 100 kHz mode 4.7 — μs 109 TaA Output valid from 100 kHz mode — 3500 ns (Note 1) 1	Tr	-		100 kHz mode		300	ns	
hold time 106 THD:DAT Data input hold time 107 TSU:DAT Data input setup time 108 TSU:STO STOP condition setup time 109 TAA Output valid from 100 kHz mode — 3500 ns (Note 1)	Ts				4.7	_	μs	Only relevant for Repeated START condition
time 107 TSU:DAT Data input setup 100 kHz mode 250 — ns 92 TSU:STO STOP condition 100 kHz mode 4.7 — μs 109 TAA Output valid from 100 kHz mode — 3500 ns (Note 1)	Tı				4.0		μs	After this period the first clock pulse is generated
time 92 Tsu:sto STOP condition 100 kHz mode 4.7 — μs setup time 109 Taa Output valid from 100 kHz mode — 3500 ns (Note 1)	Tı			100 kHz mode	0		ns	
setup time 109 TAA Output valid from 100 kHz mode — 3500 ns (Note 1)	Ts		·	100 kHz mode	250		ns	
chock	Ts	SU:STO		100 kHz mode	4.7		μs	
110 TBUF Bus free time 100 kHz mode 4.7 — μs Time the bus must	TA	ĀA	/ / / /	100 kHz mode	_	3500	ns	(Note 1)
before a new trans can start		BUF	Bus free time	100 kHz mode	4.7		μs	Time the bus must be free before a new transmission can start
D102 CB Bus capacitive loading — 400 pF)2 Çı	B	Bus capacitive load	ing	_	400	pF	

Note 1: As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCL to avoid unintended generation of START or STOP conditions.

TABLE 15-11: A/D CONVERTER CHARACTERISTICS: PIC16C925/926 (COMMERCIAL, INDUSTRIAL) PIC16LC925/926 (COMMERCIAL, INDUSTRIAL)

Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions
A01	NR	Resolution	_	_	10-bits	bit	$\begin{aligned} & \text{VREF} = \text{VDD} = 5.12\text{V}, \\ & \text{VSS} \leq \text{VAIN} \leq \text{VREF} \end{aligned}$
A02	EABS	Total Absolute error	_	_	< ± 1	LSb	$\label{eq:VREF} \begin{split} & \text{VREF} = \text{VDD} = 5.12 \text{V}, \\ & \text{VSS} \leq \text{VAIN} \leq \text{VREF} \end{split}$
A03	EIL	Integral linearity error	_	_	< ± 1	LSb	$\label{eq:VREF} \begin{split} & \text{VREF} = \text{VDD} = 5.12 \text{V,} \\ & \text{VSS} \leq \text{VAIN} \leq \end{split}$

FIGURE 15-16: A/D CONVERSION TIMING

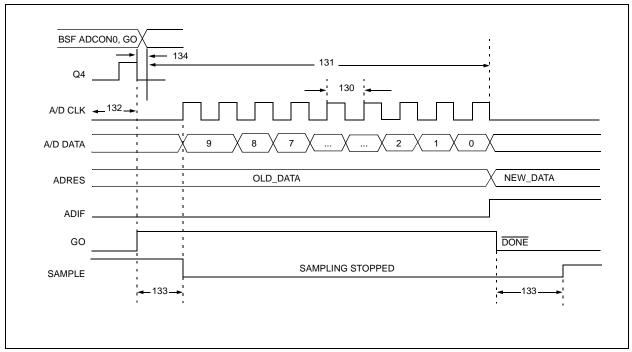


TABLE 15-12: A/D CONVERSION REQUIREMENTS

Param No.	Sym	Characteristic		Min	Тур†	Max	Units	Conditions
130	TAD	A/D clock period	PIC16 C 925/926	1.6	_	_	μs	Tosc based, VREF ≥ 3.0V
			PIC16 LC 925/926	3.0	_	_	μs	Tosc based, VREF ≥ 2.0V
			PIC16 C 925/926	2.0	4.0	6.0	ĮĮS.	A/D RC Mode
			PIC16 LC 925/926	3.0	6.0	9.0	μs	A/D RC Mode
131	TCNV	Conversion time (no (Note 1)	_	_ (12	TAD)	V	
132	TACQ	Acquisition time		(Note-2)\(\)	1	-	µs µs	The minimum time is the amplifier settling time. This may be used if the "new" input voltage has not changed by more than 1 LSb (i.e., 5 mV @ 5.12V) from the last sampled voltage (as stated on CHOLD).
134	Tgo	Q4 to AD clock star		_	Tosc/2	_	_	If the A/D clock source is selected as RC, a time of Tcy is added before the A/D clock starts. This allows the SLEEP instruction to be executed.
135	Tswc	Switching from conv	ert → sample time	1.5	_	_	TAD	

[†] Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: ADRES register may be read on the following TcY cycle.

^{2:} See Section 10.1 for min. conditions.

16.0 DC AND AC CHARACTERISTICS GRAPHS AND TABLES

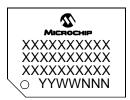
Graphs and Tables are not available at this time.

NOTES:

17.0 PACKAGING INFORMATION

17.1 Package Marking Information

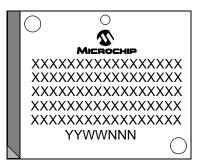
64-Lead TQFP



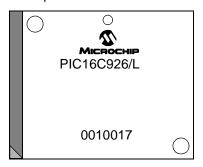
Example



68-Lead PLCC



Example



Legend: XX...X Customer specific information*

YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week '01')

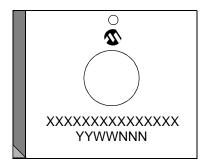
NNN Alphanumeric traceability code

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line thus limiting the number of available characters for customer specific information.

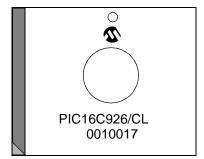
* Standard OTP marking consists of Microchip part number, year code, week code and traceability code. For OTP marking beyond this, certain price adders apply. Please check with your Microchip Sales Office. For QTP devices, any special marking adders are included in QTP price.

Package Marking Information (Continued)

68-Lead CERQUAD Windowed

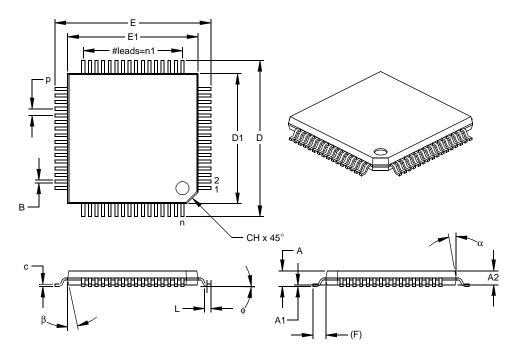


Example



17.2 **Package Details**

64-Lead Plastic Thin Quad Flatpack (PT) 10x10x1 mm Body, 1.0/0.10 mm Lead Form (TQFP)



	INCHES			MILLIMETERS*			
Dimension	Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		64			64	
Pitch	р		.020			0.50	
Pins per Side	n1		16			16	
Overall Height	Α	.039	.043	.047	1.00	1.10	1.20
Molded Package Thickness	A2	.037	.039	.041	0.95	1.00	1.05
Standoff §	A1	.002	.006	.010	0.05	0.15	0.25
Foot Length	L	.018	.024	.030	0.45	0.60	0.75
Footprint (Reference)	(F)		.039			1.00	
Foot Angle	ф	0	3.5	7	0	3.5	7
Overall Width	Е	.463	.472	.482	11.75	12.00	12.25
Overall Length	D	.463	.472	.482	11.75	12.00	12.25
Molded Package Width	E1	.390	.394	.398	9.90	10.00	10.10
Molded Package Length	D1	.390	.394	.398	9.90	10.00	10.10
Lead Thickness	С	.005	.007	.009	0.13	0.18	0.23
Lead Width	В	.007	.009	.011	0.17	0.22	0.27
Pin 1 Corner Chamfer	CH	.025	.035	.045	0.64	0.89	1.14
Mold Draft Angle Top	α	5	10	15	5	10	15
Mold Draft Angle Bottom	β	5	10	15	5	10	15

Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed

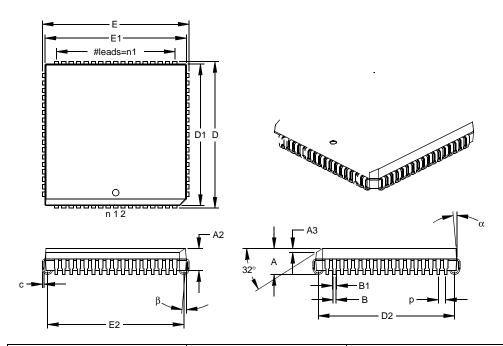
.010" (0.254mm) per side.
JEDEC Equivalent: MS-026

Drawing No. C04-085

Preliminary © 2001 Microchip Technology Inc. DS39544A-page 163

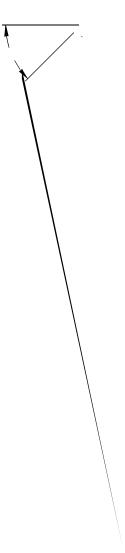
^{*} Controlling Parameter § Significant Characteristic

68-Lead Plastic Leaded Chip Carrier (L) - Square (PLCC)



	INCHES*			MILLIMETERS			
Dimensio	MIN	NOM	MAX	MIN	NOM	MAX	
Number of Pins	n					68	
Pitch	р		.050			1.27	
Pins per Side	n1		17			17	
Overall Height	Α	.165	.173	.180	4.19	4.39	4.57
Molded Package Thickness	A2						
Standoff §	A1	.020			0.51		
Side 1 Chamfer Height	A3	.024	.029	.034	0.61	0.74	0.86
Corner Chamfer 1	CH1	.040	.045	.050	1.02	1.14	1.27
Corner Chamfer (others)	CH2	.000	.005	.010	0.00	0.13	0.25
Overall Width	Е	.985	.990	.995	25.02	25.15	25.27
Overall Length	D	.985	.990	.995	25.02	25.15	25.27
Molded Package Width	E1	.950	.954	.958	24.13	24.23	24.33
Molded Package Length	D1	.950	.954	.958	24.13	24.23	24.33
Footprint Width	E2	.890	.920	.930	22.61	23.37	23.62
Footprint Length	D2	.890	.920	.930	22.61	23.37	23.62
Lead Thickness	С	.008	.011	.013	0.20	0.27	0.33
Upper Lead Width	B1	.026	.029	.032	0.66	0.74	0.81
Lower Lead Width	В	.013	.020	.021	0.33	0.51	0.53
Mold Draft Angle Top	α	0	5	10	0	5	10
Mold Draft Angle Bottom	β	0	5	10	0	5	10

68-Lead Ceramic Leaded (CL) Chip Carrier with Window – Square (CERQUAD)



NOTES:

APPENDIX A: REVISION HISTORY

Version	Date	Description
Α	February 2001	This is a new data sheet.
		However, these devices
		are similar to those
		described in the
		PIC16C923/924 data
		sheet (DS30444).

APPENDIX B: DEVICE

DIFFERENCES

The differences between the devices listed in this data sheet are listed in Table B-1.

TABLE B-1: DEVICE DIFFERENCES

Feature	PIC16C925	PIC16C926
EPROM Program Memory (words)	4K	8K
Data Memory (bytes)	176	336

Note: On 64-pin TQFP, pins RG7 and RE7 are not available.

APPENDIX C: CONVERSION CONSIDERATIONS

Considerations for converting to the devices listed in this data sheet from previous device types are summarized in Table C-1.

TABLE C-1: CONVERSION CONSIDERATIONS

Feature	PIC16C923/ 924	PIC16C925/ 926
Operating Frequency	DC - 8 MHz	DC - 20 MHz
EPROM Program Memory (words)	4K	4K (925) 8K (926)
Data Memory (bytes)	176	176 (925) 336 (926)
A/D Converter Resolution	8-bit (924 only)	10-bit
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Brown-out Reset	No	Yes

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PART NO.	- X	/XX	xxx
Device	Temperature Range	Package	e Pattern
Device	PIC16C92X ⁽¹⁾ , PI VDD range 4. PIC16LC92X ⁽¹⁾ , F VDD range 2.	0V to 5.5 PIC16LC9	iV 92XT ⁽²⁾ ;
Temperature Range	$S = -40^{\circ}C \text{ to}$ - = 0°C to	+85°C +70°C	(Industrial) (Industrial, tape/reel) (Commercial) (Commercial, tape/reel)
Package	CL = Windowe PT = TQFP (Tr L = PLCC		
Pattern	QTP, SQTP, Code (blank otherwise)	or Spec	ial Requirements

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